# **METAMATERIAL CATHODES IN MULTI-CAVITY MAGNETRONS** Andrey D. Andreev<sup>\*</sup> and Kyle J. Hendricks

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#### Abstract

The paper presents results of computer experiments performed for research to enhance operational parameters of multicavity magnetrons through using the metal-thin-wire (MTW) and the metal-split-ring (MSR) metamaterial structures as a cold (non-thermionic) cathode. It is demonstrated by particle-in-cell (PIC) simulations of a generic 10-vane non-relativistic ultra-high frequency (UHF) magnetron that the output microwave power and the start-up time of this magnetron are improved by replacing either a solid or a transparent cathode by a bulk MTW metamaterial structure formed by an array of individual wires directed parallel to the axis of the magnetron (rodded cathode).

### I. INTRODUCTION

The artificial MTW and the MSR structures possess metamaterial properties, i.e., negative- $\varepsilon$ /positive- $\mu$  and positive- $\varepsilon$ /negative- $\mu$ , respectively, when the incident time-varying electric field  $E_1$  oscillates along the wires of the MTW structure, or when the incident time-varying magnetic  $H_1$  field oscillates perpendicularly to the rings of the MSR structure [1], [2]. In both cases, the incident  $E_1$ and  $H_1$  fields induce electric currents flowing along the individual metal wires (rings). The oscillations of the induced currents generate the artificial electric or magnetic dipole moments within the bulk of a metamaterial structure, which either constructively or destructively interact with the incident fields. Both MTW and MSR structures may be considered as quasi-metamaterials (transparent) structures when the incident time-varying electric field  $E_1$ is orthogonal to the individual wires and freely penetrates inside of those structures.

The multi-cavity magnetron operates in one of the transverse-electrical-like (TE-like) cavity modes of the magnetron resonant cavity (magnetron diode) formed by the cathode emitting electrons and the anode consisting of a number of resonators [3], [4]. The resonant cavity slows down the induced electromagnetic wave azimuthally traveling between the cathode and the anode. The spatial distribution of the time-varying electromagnetic field  $(E_1 \times H_1)$ , induced within the resonant cavity and associated with the traveling electromagnetic wave, is characterized by the circumferential  $E_{1\phi}$  and the radial  $E_{1\rho}$  components of the electric field and the axial  $H_{1z}$  component of the magnetic field. The number of  $E_{1\phi}$  phase variations in the azimuthal direction is determined by the operating frequency of the magnetron (mode of the magnetron operation) and the anode geometry (number of cavities). The number of  $E_{1\rho}$  and  $H_{1z}$  phase variation along the radius and the axis of the resonant system are usually minimized

to one. This allows the magnetron to operate in the lowerfrequency *TE-like* mode (at the given number of  $E_{1\phi}$  circumferential variations) and prevents axial and radial mode competition. The external electric field  $E_0$  in the magnetron diode is directed radially from the anode to the cathode. The external magnetic field  $H_0$  is directed axially. The crossed  $E_0 \times H_0$  fields force electrons emitted from the cathode to drift in the azimuthal direction. The multicavity magnetron starts to generate (induce)  $E_1 \times H_1$  fields and produce microwaves (output power) when the drift velocity of the emitted electron is approximately equal to the phase velocity of  $E_{1\phi}$ .

For the multi-cavity magnetron operating in the TE-like mode, the bulk metamaterial cathode may be designed in the form of a single/double/multi-helix structure (Fig. 1a), the individual elements (turns) of which are oriented parallel to  $E_{1\phi}$  and perpendicular to  $H_{1z}$ . The proper choosing of the helix pitch may allow implementing the negative- $\varepsilon$ /negative- $\mu$  properties of the metamaterial cathode within the desired microwave frequency range to enhance and improve the magnetron operational parameters. The metamaterial cathode may be also designed in the form of the rodded medium whose individual elements (rods) are placed either radially (Fig. 1b) (parallel to  $E_{1\rho}$  and perpendicular to both  $E_{1\phi}$  and  $H_{1z}$  of longitudinally (Fig. 1c) (perpendicular to both  $E_{1\rho}$  and  $E_{1\phi}$  and parallel to  $H_{1z}$ ). In the latter case (Fig. 1c), the rodded medium is essentially a quasi-metamaterial structure.



Fig. 1. The different cathodes: a) multi-helix structure with metal wires parallel to  $E_{1\phi}$  and the metal rings perpendicular to  $H_{1z}$ . b) multi-point structure with metal wires parallel to  $E_{1\rho}$  (multipoint cathode), c) multi-rod structure with metal wires parallel to  $H_{1z}$  (rodded cathode), d) standard helical cathode with nonemitting central rod.

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Results of the PIC simulations of a generic 10-vane non-relativistic continuous-wave (CW) UHF magnetron with a quasi-metamaterial multi-rod or rodded cathode (Fig. 1c) are described below. Most important geometrical dimensions of the magnetron are similar to the commercially available CWM-75/100L and L-4031 tubes of the L-3 Communications, Electron Devices [5].

#### **II. SIMULATION GEOMETRY AND PROCEDURE**

The simulations are performed using ICEPIC code developed and maintained by the *AFRL/RDHE* [6]-[8]. ICEPIC is a fully relativistic, three-dimensional (3D), variable-mesh PIC code capable of simulating the interdisciplinary physics of charge-particle beams, high-power microwaves (HPM), and plasmas in complex geometries of modern microwave vacuum electronic devices (MVEDs) [8]. The code incorporates numerous advances in parallel computing and performs automatic partitioning using a single domain decomposition algorithm, asynchronous message passing, and dynamic load balancing [7]. ICEPIC simulates from the first principles (Maxwell's equations and Lorentz's forces law) the relativistic electrodynamics and charged-particle dynamics within the resonant structures of different MVEDs [6].

The simulation model of the magnetron is built within the Cartesian *x-y-z* geometry using simulation cells with grid size of  $0.5 \times 0.5 \times 0.5$  mm<sup>3</sup> [9]. It consists of: (i) the cathode block with the input port, (ii) the 10-vane anode block surrounding the emitting part of the cathode block, (iii) the "double-ring-strapping" system coupling the alternative resonators of the anode block, and (iv) three output electrodes connecting three vanes of the anode block with the appropriate output ports.

The anode block consists of ten vanes, N=10, forming a cylindrical slow-wave structure (SWS). There are two grooves, one on each end of the anode block. The double-ring-strapping system is located within the grooves on each end of the anode block. Each strap electrically connects alternate vanes and passes over adjacent vanes of the anode block. Three cylindrical electrodes connect the first, the fourth, and the sixth vanes of the anode block with the output ports. The cathode block consists of the non-emitting central electrode connecting the input port with the emitting part of the cathode, two non-emitting end caps preventing emitted electrons from leaking out the interaction space, and the cathode itself (Fig. 2).

The simulations of the magnetron are performed at input voltage  $V_{in}$ =45 kV and magnetic flux density  $B_0$ =0.49 T. The rise time of  $V_{in}$  is 10 ns [10] at simulation time of 2000 ns. This allows one to consider  $V_{in}$  as the "steadystate" input parameter.  $B_0$  is axially directed, uniform, and constant within the simulation volume and during the simulation time.

In the course of the simulations,  $V_{in}$  is monitored (as well as the input current  $I_{in}$  and the input power  $P_{in}$ ) at the input port. The output power  $P_{out}$  is monitored (as well as the output *rf* voltage  $V_{out}$ ) at the output ports. The resona-

tor voltage  $V_{\rm res}$  is monitored inside the first resonator of the anode block. The anode current  $I_a$  is monitored at the plane located on a small distance (one or two simulation cells) from the anode surface. All monitored parameters are smoothed over 100,000 consecutive time steps and averaged over their steady-state phases (when the monitored parameter is not changed in time) in order to get the steady-state characteristics of the magnetron operation.

The entire simulation procedure develops as follows. Once  $V_{in}$  is applied to the input port, the space-charge limited (SCL) algorithm of the ICEPIC code [11] begins to produce electrons at a small distance from the cathode (one or two grid cells). These originated electrons are extracted from the cathode and accelerated by  $E_0$ , deflected by  $H_0=\mu_0B_0$  ( $\mu_0$  is permeability of vacuum), and move within the interaction space in accordance with the given distributions of  $E_0 \times H_0$  and the electron spacecharge  $q_e$  accumulated within the interaction space.

The accelerated electrons give rise to  $I_{in}$  flowing along the central electrode connecting the input port with the cathode. After some amount of time, determined by the loaded Q (quality factor [4]) of the magnetron resonant cavity, the magnetron spokes begin to grow and the magnetron starts to operate. After formation of the magnetron spokes, both  $I_a$  and  $P_{out}$  increase and go into their steady-state phases. The electrons begin to move within the interaction space in accordance with the given distributions of  $E_0 \times H_0$ ,  $E_1 \times H_1$ , and  $q_e$ .

Results of the simulations of the magnetron having the rodded cathode (Fig. 1c, Fig. 2c) are compared with the results obtained during PIC simulations of the same magnetron having the solid cylindrical (Fig. 2a) and the transparent [12], [13] (Fig. 2b) cathodes. All cathodes of the simulation model of the magnetron (Fig. 2) are supposed to operate in the explosive electron emission (EEE) mode [14], which automatically assumes the space-charge limited (SCL) mechanism of the electron extraction from the cathode into the interaction space. Note, however, that the standard high-power industrial-heating UHF CWM-75/100L and L-4031 MHz magnetrons have the helical thermionic cathode (with the central back-current electrode) (Fig. 1d) working in the thermal electron-emission (TEE) mode [5].

## **III. SIMULATION RESULTS**

The particle plot distributions within the interaction space of the magnetron having three different cathodes (Fig. 2) (solid, transparent, and rodded) are shown in Fig. 3 (~10 ns of simulation time), Fig. 4 (~100 ns), and Fig. 5 (~1000 ns). It can be seen in Fig. 3-Fig. 5 that the magnetron begins to oscillate in the  $\pi$ -mode with either cathode. The  $\pi$ -mode is recognized by five magnetron spokes corresponding to the *TE*<sub>51</sub> mode of *E*<sub>1</sub>×*H*<sub>1</sub> distribution within the interaction space.

There are, however, very significant differences in the amount of the anode current  $I_a$  and the output power  $P_{out}$  when the magnetron operates with the different cathodes.



Fig. 2. Cathodes used in simulations: a) solid cathode, b) transparent cathode, c) rodded (quasi-metamaterial) cathode (Fig. 1c).



Fig. 3. Particle plots at 10 ns of simulation time: a) solid cathode, b) transparent cathode, c) rodded (quasi-metamaterial) cathode.



а









Fig. 5. Particle plots at 1000 ns of simulation time: a) solid cathode, b) transparent cathode, c) rodded (quasi-metamaterial) cathode.





c

c

с

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The differences can be seen in Fig. 7 that the steadystate  $I_a$  increases from ~7.7 A in the magnetron having solid cathode, to ~16.7 A in the magnetron having transparent cathode, and up to ~23.1 A in the magnetron having rodded cathode. In the exact accordance with  $I_a$  increase [15], [16], the steady-state  $P_{out}$  increases from ~311 kW in the magnetron having solid cathode, to ~621 kW in the magnetron having transparent cathode, and up to ~820 kW in the magnetron having rodded cathode (Fig. 8).

Detailed analysis of the obtained results show that both  $I_a$  (Fig. 7) and  $P_{out}$  (Fig. 8) increase in direct proportion to the square of the resonator voltage  $V_{res}$  (Fig. 9), which is directly proportional to the induced time-varying electric field amplitude  $E_1$ . The frequency of the magnetron operation  $f_{\pi}$  is determined by the frequency of  $E_1$  oscillations. Frequency  $f_{\pi}$  remains the same with all cathodes (Fig. 10).

The numerical relations between  $I_a$ ,  $P_{out}$  and  $V_{res}$  demonstrate that both  $I_a$  and  $P_{out}$  increase in direct proportion to the square of  $V_{res}$ . This allows one to suppose that the direct cause of both  $I_a$  and  $P_{out}$  increase is nothing but the  $E_1$  increase within the interaction space. The smoothed traces of  $I_a$  and  $P_{out}$  are shown in Fig. 11 and Fig. 12, correspondently. The envelope of  $V_{res}$  is shown in Fig. 13. All traces demonstrate how the cathode construction (Fig. 2) affects the magnetron operation.



Fig. 11. Smoothed traces of the anode current  $I_a$ .



Fig. 12. Smoothed traces of the output power  $P_{out}$ .



Fig. 13. Envelope of the resonator voltage  $V_{\rm res}$ .

## **IV. CONCLUSION**

PIC simulations of a generic 10-vane UHF magnetron with the solid, the transparent, and the quasi-metamaterial (rodded) cathodes (Fig. 2) are performed. Results of the simulation showed that the use of the rodded cathode allows significantly increasing the output microwave power (Fig. 12) by the appropriate increase of the anode current (Fig. 11) of the magnetron. The anode current increase is caused by appropriate increase of the resonator voltage (Fig. 13)) within the interaction space. The latter effect is the direct result of the magnetron resonant cavity.

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