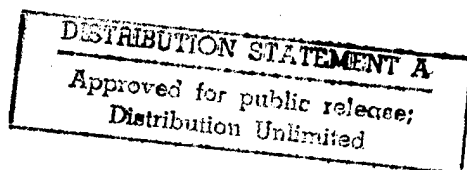


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1 Navy Case No. 78605

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3 ACOUSTIC VECTOR SENSING SONAR SYSTEM

4

5 STATEMENT OF GOVERNMENT INTEREST

6 The invention described herein may be manufactured and used
7 by or for the Government of the United States of America for
8 governmental purposes without the payment of any royalties
9 thereon or therefor.

10

11 BACKGROUND OF THE INVENTION

12 (1) Field of the Invention

13 The present invention relates to the detection of acoustic
14 signals. More specifically, the present invention relates to a
15 sonar system which uses an array of acoustic vector sensors for
16 detecting acoustic signals.

17 (2) Description of the Prior Art

18 The undersea operations of the Navy have shown a shift from
19 deep water, open ocean environments to littoral, shallow water
20 environments. These new environments often have high ambient
21 noise levels and varying bottom conditions which reduce or impair
22 the ability of existing sonar systems to detect, at a reasonable
23 distance, all but low frequency acoustic signals. Thus, there is
24 a need for an improved sonar system to detect acoustic signals in
25 noisy, shallow water environments.

26 A fundamental measure of array performance is directivity.
27 Increases in directivity can provide increased detection range

1 and resolution. Conventional sonar systems use arrays of
2 pressure sensors (hydrophones) to detect acoustic signals. For
3 an array of pressure sensors, gains in directivity are
4 fundamentally limited by the size of the array. Therefore, the
5 only way to increase the directivity of an array of pressure
6 sensors is to increase the size, or aperture, of the array.

7 In conventional submarine sonar arrays, the hydrophones are
8 typically mounted to a steel conditioning plate or similar
9 structure to enhance signal reception (pressure doubling).
10 Increasing directivity by building a large aperture array would
11 require heavy and unyielding signal conditioning plate. Creating
12 a large aperture, low frequency array would require a relatively
13 thick plate to create a rigid boundary condition. Such a large
14 conditioning plate is unfeasible in many applications.
15 Therefore, an improved sonar system, capable of detecting
16 acoustic plane waves in noisy environments would be a welcome
17 addition to the art.

18

19 SUMMARY OF THE INVENTION

20 Accordingly, it is a general purpose and object of the
21 present invention to provide a sonar system which provides
22 improved detection of acoustic signals in noisy environments.

23 Another object of the present invention is the provision of
24 a sonar system having greater directivity than existing submarine
25 sonar arrays.

1 A further object of the present invention is to provide a
2 sonar system which uses an array of acoustic vector sensors for
3 detecting acoustic signals.

4 These and other objects made apparent hereinafter are
5 accomplished with the present invention by providing an array of
6 acoustic vector sensors coupled to a beamformer. The beamformer
7 generates steered acoustic beams using the output from one or
8 more of the triaxial acoustic vector sensors. The beamformer
9 weights the output signals from the sensors by forming the inner
10 product of the velocity components measured at each element
11 location with an element weighting vector, $\vec{w}_n^{(3)}$, delaying the
12 resulting scalar field generated by weighting and then summing
13 the delayed signals.

14 In a preferred embodiment, each element of the array
15 comprises a triaxial velocity sensor co-located with an acoustic
16 pressure sensor. The beamformer generates a weighted output
17 signal for each velocity sensor by forming the inner product of
18 the measured velocity components at each element location with a
19 velocity weighting vector. The weighted output from each
20 velocity sensor is combined with a weighted output from the
21 respective acoustic pressure sensor to generate a weighted
22 element output. The weighted element output signals are delayed
23 to synchronize the phase of the weighted element output signals
24 to that of a weighted signal at a reference array location. The
25 resulting delayed weighted output signals are then summed.

BRIEF DESCRIPTION OF THE DRAWINGS

A more complete understanding of the invention and many of the attendant advantages thereto will be readily appreciated as the same becomes better understood by reference to the following detailed description when considered in conjunction with the accompanying drawings wherein like reference numerals and symbols designate identical or corresponding parts throughout the several views and wherein:

FIG. 1 illustrates the array coordinates and geometry for an equispaced line array of vector sensors;

FIG. 2 shows the angular coordinates used to define the incident direction of an acoustic planewave;

FIG. 3 is a block diagram of a sonar system in accordance with the present invention; and

FIG. 4 is a block diagram of a submarine sonar system in accordance with the present invention.

DESCRIPTION OF THE PREFERRED EMBODIMENT

Referring to FIG. 1, there is shown an equispaced line array 10 of multiaxis vector sensors 12 mounted on a baffle 14. The orientation of the coordinate system $\bar{x} = \{x, y, z\}$ and the mutually perpendicular unit vectors $\{\bar{n}_x, \bar{n}_y, \bar{n}_z\}$ is shown in FIG. 1 wherein the x-axis is aligned with the line array of multiaxis vector sensors 12 and the y-axis is normal to baffle 14 which coincides with the xz-plane. Thus, sensors 12 lie along the x-axis within the xz-plane and are bounded by a semi-infinite acoustic medium

1 for $y > 0$. In developing a model for the beamformed response of
 2 vector sensors, it will be assumed that baffle 14 provides an
 3 ideal boundary. That is, baffle 14 is entirely pressure release
 4 for each orthogonal acoustic velocity component. It is
 5 understood that for some applications such a baffle may not be
 6 realizable; however, such an assumption allows for the comparison
 7 of an array of vector sensor to an array of scalar pressure
 8 sensors based on ideal conditions.

9 Acoustic particle velocities are defined as small amplitude
 10 motions of constant volume fluid particles, each having unvarying
 11 fluid properties, about an equilibrium position. Time harmonic
 12 acoustic planewaves, which have wavefronts propagating in the
 13 direction of an acoustic wavevector $-\vec{k}$ (by convention the
 14 acoustic wavevector is defined to point from the origin to the
 15 source), can be characterized by acoustic particle velocity.
 16 With the orientation and geometry of FIG. 1, the vector
 17 components of the acoustic velocity are given by:

$$18 \quad v_x(\vec{r}, \omega) = V_x e^{i(\vec{k} \cdot \vec{r} + \omega t + \phi_x)} \quad (1)$$

$$19 \quad v_y(\vec{r}, \omega) = V_y e^{i(\vec{k} \cdot \vec{r} + \omega t + \phi_y)} \quad (2)$$

20 and

$$21 \quad v_z(\vec{r}, \omega) = V_z e^{i(\vec{k} \cdot \vec{r} + \omega t + \phi_z)} \quad (3)$$

22 where $\|\vec{k}\| = \omega/c$ is the acoustic wavenumber, ω is the circular
 23 frequency, c is the sound speed, $\vec{r} = [x, y, z]$ is the position
 24 vector, and V_x , V_y , V_z are the amplitudes of the component

1 velocities in the x, y, z directions, respectively. The complete
2 vector field may be written as:

$$3 \quad \vec{v}^{(3)}(\vec{r}, \omega) = v_x(\vec{r}, \omega) \vec{n}_x + v_y(\vec{r}, \omega) \vec{n}_y + v_z(\vec{r}, \omega) \vec{n}_z \quad (4)$$

4 where the superscript (3) denotes the dimension of the velocity
5 vector with uniaxis vector sensors having a dimension of 1,
6 biaxial sensors having a dimension of 2, and triaxial sensors
7 having a dimension of 3. The elements of an uniaxis vector
8 sensing array measure only one of the velocity components v_x , v_y ,
9 v_z while multiaxis arrays measure two or more the vector
10 components.

11 Each orthogonal component measured by vector sensor 12 has
12 an amplitude response or element sensitivity which is
13 proportional to the cosine of the angle between the planewave
14 incidence arrival direction and the orthogonal component being
15 measured. That is, for a triaxial vector sensor 12, the measured
16 amplitude V_x of velocity vector component v_x is proportional to
17 the cosine of the angle between the arrival direction of the
18 planewave and the x-unit normal \vec{n}_x . Similarly, the amplitude V_y
19 of velocity component v_y is proportional to the cosine of the
20 angle between the arrival direction and the y-unit normal \vec{n}_y and
21 the amplitude V_z of component v_z is proportional to the cosine of
22 the angle between the arrival direction and the z-unit normal \vec{n}_z .
23 The cosine angular dependence for the amplitude response is
24 derived from elementary solutions to the wave equation at the
25 boundary of an ideal pressure release surface.

1 Referring now to FIG. 2, there is shown the angular
2 coordinate system adopted to describe the present invention. In
3 FIG. 2, wavevector \vec{k} with standard Cartesian components $[k_x, k_y, k_z]$
4 indicates the orientation of the arrival of an incident acoustic
5 planewave. The angle between the planewave incidence arrival
6 direction and orthogonal component k_y is illustrated in FIG. 2 as
7 conical angle Θ_y . In FIG. 2, wavevector \vec{k} and, thus, the arrival
8 direction of the acoustic planewave is defined by angular
9 coordinates (θ, ϕ) . Recalling the orientation shown in FIG. 1, the
10 unit normals for components $[k_x, k_y, k_z]$ are equivalent to the unit
11 normals $\{\vec{n}_x, \vec{n}_y, \vec{n}_z\}$ and wavevector \vec{k} can be defined by:

$$12 \quad k_x = \|\vec{k}\| \sin(\theta) \cos(\phi) \quad (5)$$

$$13 \quad k_y = \|\vec{k}\| \cos(\theta) \cos(\phi) \quad (6)$$

14 and

$$15 \quad k_z = \|\vec{k}\| \sin(\phi) . \quad (7)$$

16 The angle formed between wavevector \vec{k} and any unit normal
17 $\{\vec{n}_x, \vec{n}_y, \vec{n}_z\}$ can be determined via the dot product of the wavevector
18 and the selected normal. Thus, the cosine angular dependence
19 yields an amplitude response given by:

$$20 \quad V_j(\theta, \phi) = \begin{cases} 2 \cos(\Theta_j) & \text{for } -90^\circ < \theta, \phi < 90^\circ \text{ and } j = x, y, z \\ 0 & \text{otherwise} \end{cases} \quad (8)$$

21 where

$$\begin{aligned}
& \cos(\Theta_x) = \cos(\phi) \sin(\theta), \\
& \cos(\Theta_y) = \cos(\phi) \cos(\theta), \text{ and} \\
& \cos(\Theta_z) = \sin(\phi)
\end{aligned} \tag{9}$$

2 The factor of 2 in equation (8) is due to the assumption of an
 3 ideal boundary thereby doubling the signal amplitude for all
 4 angles of incidence. Given the amplitude response given by
 5 equations (8) and (9), the response of a baffled triaxial vector
 6 sensor can be expressed as:

$$\bar{v}^{(3)}(\theta, \phi, \bar{r}_o) = 2 \left\{ \cos(\Theta_x) e^{ik_x x_o} \bar{n}_x + \cos(\Theta_y) e^{ik_y y_o} \bar{n}_y + \cos(\Theta_z) e^{ik_z z_o} \bar{n}_z \right\} \tag{10}$$

8 wherein the dependence on the azimuth and elevation angles
 9 explicitly indicated. Computing the element response for a
 10 triaxial vector sensor using equation (10) reveals that element
 11 response is uniform for all elevation and azimuth arrival angles.
 12 For reference, it is noted that the amplitude or angular
 13 sensitivity of a pressure sensor mounted on an ideal (perfectly
 14 rigid) baffle is given by:

$$P(\theta, \phi) = \begin{cases} 2 & \text{for } -90^\circ < \theta, \phi < 90^\circ \\ 0 & \text{otherwise} \end{cases} \tag{11}$$

16 Thus, as is apparent to those skilled in the art, in magnitude a
 17 triaxial vector sensor yields an element response equivalent to a
 18 scalar pressure sensor.

19 Having determined the response of a baffled triaxial vector
 20 sensor, it is possible to evaluate different approaches for
 21 beamforming an array of vector sensors. It should be noted that
 22 a single multiaxis vector sensor can be steered in the direction
 23 of an incident signal. That is, it is possible to form a broad

1 acoustic beam, which is steered in the direction of the incident
 2 wavevector \vec{k} , using a single triaxial vector sensor. For a
 3 vector sensor located at the origin, the sensor response given by
 4 equation (10) can be written as:

$$5 \quad \vec{v}^{(3)}(\theta, \phi, \vec{0}) = 2 \left\{ \cos(\Theta_x) \vec{n}_x + \cos(\Theta_y) \vec{n}_y + \cos(\Theta_z) \vec{n}_z \right\} \quad (12)$$

6 The steered sensor response for a triaxial vector sensor is
 7 obtained by taking the dot product of the normalized velocity
 8 field vector at the angle at which the element is steered (θ_s, ϕ_s)
 9 and the velocity vector at arrival angle (θ, ϕ) which yields:

$$10 \quad g^{(3)}(\theta_s, \phi_s, \theta, \phi) = \frac{\vec{v}^{(3)}(\theta_s, \phi_s, \vec{0}) \cdot \vec{v}^{(3)}(\theta, \phi, \vec{0})}{\|\vec{v}^{(3)}(\theta_s, \phi_s, \vec{0})\| \|\vec{v}^{(3)}(\theta, \phi, \vec{0})\|} \quad (13)$$

$$= \cos(\phi_s) \cos(\phi) \cos(\theta - \theta_s) + \sin(\phi_s) \sin(\phi)$$

11 The product $g^{(3)}(\theta_s, \phi_s, \theta, \phi)$ reaches a maximum of unity when
 12 $(\theta_s, \phi_s) = (\theta, \phi)$. Similarly, it is possible to steer a biaxial vector
 13 sensor; however, a biaxial vector sensor can only be steered in
 14 one angle. For a biaxial sensor measuring the components v_x and
 15 v_y , the steered element response is given by:

$$16 \quad g^{(2)}(\theta_s, \theta) = \cos(\theta - \theta_s) \quad (14)$$

17 Computing the amplitude response and the associated beam
 18 pattern (log of the squared amplitude response) reveals that the
 19 resolution of the arrival angle is limited due to the broad beam
 20 that is generated. However, assuming that the steering errors at
 21 each sensor are uncorrelated, the accuracy of the estimation of
 22 the signal arrival angle obtained from steering an individual

1 sensor can be improved by individually steering additional
 2 sensors. This ability to steer individual sensors can be applied
 3 to a line array, such as is shown in FIG. 1, to provide
 4 resolution of both the azimuth and elevation arrival angles.
 5 Therefore, one possible array beamforming approach is to steer
 6 each vector sensor in the direction of the estimated arrival
 7 angle and beamform the resulting responses conventionally.

8 The process of passive sonar conventional beamforming is
 9 independent of the type of sensor used in an array. A beamformer
 10 is designed to add, in phase, the sensor outputs resulting from
 11 acoustic fields which are incident upon the array. The sensor
 12 outputs are time-delayed to synchronize the phase of acoustic
 13 signals received at each of the sensors. The time delayed
 14 signals are weighted (spatially shaded) to control the beamwidth
 15 or sidelobe levels and then summed. Thus, one possible approach
 16 to beamforming an array of vector sensors would be to delay,
 17 weight and sum the velocity components from each vector sensor
 18 separately. Assuming an array of triaxial vector sensors, $\bar{v}_n^{(3)}$,
 19 for $1 \leq n \leq N$, for the ideal baffle configuration this approach
 20 yields a summed velocity response given by:

$$\begin{aligned}
 \bar{v}_N^{(3)}(\theta, \phi) = & 2 \sin(\theta) \cos(\phi) \sum_{n=1}^N w_n^{(x)} e^{i(\bar{k} - \bar{k}_s) \cdot \bar{r}_n} \bar{n}_x + 2 \cos(\theta) \cos(\phi) \sum_{n=1}^N w_n^{(y)} e^{i(\bar{k} - \bar{k}_s) \cdot \bar{r}_n} \bar{n}_y \\
 & + 2 \sin(\phi) \sum_{n=1}^N w_n^{(z)} e^{i(\bar{k} - \bar{k}_s) \cdot \bar{r}_n} \bar{n}_z
 \end{aligned} \tag{15}$$

22 wherein the amplitude shading coefficients for each orthogonal
 23 direction, $w_n^{(x)}, w_n^{(y)}, w_n^{(z)}$ for $1 \leq n \leq N$, are arbitrary; wavevector \bar{k} ,

1 defined by equations (5), (6) and (7), corresponds to the
 2 arriving acoustic plane wave; and \vec{k}_s is the wavevector to which
 3 the array is steered. It is noted that the exponential functions
 4 in equation (15) may be written more explicitly as:

$$5 \quad e^{i(\vec{k}-\vec{k}_s)\cdot\vec{r}_n} = \exp\left[i\frac{\omega}{c}\left\{x_n\alpha(\theta,\phi) + y_n\beta(\theta,\phi) + z_n\gamma(\theta,\phi)\right\}\right] \quad (16)$$

6 where

$$\begin{aligned} \alpha(\theta,\phi) &= \cos(\phi)\sin(\theta) - \cos(\phi_s)\sin(\theta_s), \\ \beta(\theta,\phi) &= \cos(\phi)\cos(\theta) - \cos(\phi_s)\cos(\theta_s), \text{ and} \\ \gamma(\theta,\phi) &= \sin(\phi) - \sin(\phi_s) \end{aligned} \quad (17)$$

8 One approach as a means to process the vector components, which
 9 should be avoided, is to take the norm (or length of the weighted
 10 sum of all the vector components) of $\vec{v}_N^{(3)}$ in equation (15). The
 11 norm can be written as:

$$12 \quad B^{(3)}(\theta,\phi) = \sqrt{\left(\sum_{n=1}^N w_n^{(x)} v_{xn}\right)^2 + \left(\sum_{n=1}^N w_n^{(y)} v_{yn}\right)^2 + \left(\sum_{n=1}^N w_n^{(z)} v_{zn}\right)^2} \quad (18)$$

13 where $v_{xn} = V_x \exp\{i(\vec{k} - \vec{k}_s)\cdot\vec{r}\}$, $v_{yn} = V_y \exp\{i(\vec{k} - \vec{k}_s)\cdot\vec{r}\}$ and $v_{zn} = V_z \exp\{i(\vec{k} - \vec{k}_s)\cdot\vec{r}\}$. It
 14 is clear from equation (18) that this approach results in taking
 15 the square-root of the sum of the squared vector velocity
 16 components. This is a nonlinear *ad hoc* processing technique that
 17 should be avoided.

18 Another approach to array beamforming, as previously
 19 indicated, is to steer each sensor (array element) toward the
 20 source and beamform the responses of the steered vector sensors
 21 conventionally. In this approach each vector sensor in equation
 22 (15) is steered towards the source and the squared magnitude of

1 the resulting expression is then obtained. That is, form the
 2 inner product of the summed velocities $\bar{v}_N^{(3)}$ with a steered vector
 3 of unit amplitude and take the magnitude-squared of the product.
 4 If it is assumed that the weighting of each component of every
 5 vector sensor is identical, i.e., $w_n^{(x)} = w_n^{(y)} = w_n^{(z)} = w_n$, and for line
 6 array 10 of FIG. 1, $\bar{r} = [x_n, 0, 0]$, this approach yields:

$$\begin{aligned}
 B^{(3)}(\theta, \phi) &= \left\| \bar{v}_n^{(3)}(\theta, \phi) \cdot \bar{v}_n^{(3)}(\theta_s, \phi_s) \right\|^2 \\
 &= 4g^{(3)}(\theta_s, \phi_s, \theta, \phi)^2 \left| \sum_{n=1}^N w_n \exp\left(i \frac{\omega}{c} x_n \alpha(\theta, \phi)\right) \right|^2
 \end{aligned} \tag{19}$$

8 Similarly, for a biaxial vector sensor the approach yields:

$$B^{(2)}(\theta, \phi) = 4 \left\{ g^{(2)}(\theta_s, \theta) \cos(\phi) \right\}^2 \left| \sum_{n=1}^N w_n \exp\left(i \frac{\omega}{c} x_n \alpha(\theta, \phi)\right) \right|^2 \tag{20}$$

10 There is no equivalent expression for an array of uniaxis
 11 vector sensors since the elements of such an array cannot be
 12 steered in azimuth or elevation. Therefore, the beam response
 13 for an array of uniaxis vector sensors, each measuring the vector
 14 component v_y , would be written as:

$$B^{(1)}(\theta, \phi) = \left\| \bar{v}_n^{(1)}(\theta, \phi) \right\|^2 = 4 \cos^2(\theta) \cos^2(\phi) \left| \sum_{n=1}^N w_n \exp\left(i \frac{\omega}{c} x_n \alpha(\theta, \phi)\right) \right|^2 \tag{21}$$

16 Although this approach provides a linear processing
 17 technique to beamforming, it may be limited in its capability.
 18 The restriction that weighting values be essentially the
 19 direction cosines and constant from element to element disallows
 20 emphasis of some element outputs over others and imposes the same
 21 relative weighting to the three velocity components.

1 A more general linear processing technique to beamforming an
 2 array of vector sensors would form the inner product of the
 3 velocity components at each element location with an arbitrary
 4 element weighting vector, $\bar{w}_n^{(3)}$. The weighting vector $\bar{w}_n^{(3)}$ may
 5 vary the weighting of each velocity component at each element.
 6 After forming the inner product, the resulting scalar field may
 7 then be delayed and summed. Thus, in general, the response is
 8 given by:

$$9 \qquad B^{(3)}(\theta, \phi) = \left| \sum_{n=1}^N \bar{w}_n^{(3)} \cdot \bar{v}_n^{(3)} e^{i(\bar{k} - \bar{k}_s) \cdot \bar{r}_n} \right|^2 \qquad (22)$$

10 and for a line array:

$$11 \qquad B_L^{(3)}(\theta, \phi) = \left| \sum_{n=1}^N \left(w_n^{(x)} V_x + w_n^{(y)} V_y + w_n^{(z)} V_z \right) \exp\left(i \frac{\omega}{c} x_n \alpha(\theta, \phi)\right) \right|^2 \qquad (23)$$

12 As can be seen from equations (22) and (23), the result of this
 13 approach is a linearly-weighted combination of all three measured
 14 velocity components. That is, this approach yields the square of
 15 the sum of the weighted velocity components.

16 A more general processing approach, likely realizable only
 17 in a free-field environment, would be to augment an array of
 18 vector sensors to include a measurement of acoustic pressure.
 19 That is, using an array in which each element within the array
 20 provides a measure acoustic pressure and the three orthogonal
 21 vector components of particle velocity. Processing all four
 22 measured quantities (pressure and three velocity components) from
 23 each element location yields:

$$B^{(pr)}(\theta, \phi) = \left| \sum_{n=1}^N \left(\vec{w}_n^{(3)} \cdot \vec{v}_n^{(3)} + w_p P_n \right) e^{i(\vec{k} - \vec{k}_s) \cdot \vec{r}_n} \right|^2 \quad (24)$$

As can be seen from equation (24), there are 4N degrees of freedom to manipulate in order to maximize some measure of array performance, such as signal to noise ratio. These 4N degrees of freedom are the N pressure weights $\{w_p\}$ and the N velocity weights for each orthogonal direction, $\{w_n^{(x)}\}$, $\{w_n^{(y)}\}$, $\{w_n^{(z)}\}$ for $1 \leq n \leq N$. This increased number of degrees of freedom can provide for improved performance relative to processing vector sensors alone (3N degrees of freedom). Likewise, the increased number of degrees of freedom available from processing pressure and three velocity components (as well as that from processing three velocity components) can provide for considerable improved performance relative to processing scalar pressure sensors alone.

Referring to FIG. 3, there is shown a block diagram of an embodiment of a sonar system in accordance with the present invention. The sonar system of FIG. 3 comprises an array 20 of acoustic vector sensors 22 each which may or may not be mounted on a baffle 24. Each acoustic vector sensor 22 is responsive to acoustic signals present at the sensor and generates an electrical signal proportional to the received acoustic signal as output 26 which is coupled to beamformer 28. Beamformer 28 operates to select the output signal 26 from one or more of the sensors 22 selected as elements for a desired array configuration and generates an acoustic beam using the selected output signals.

1 Unlike the scalar pressure sensors used in existing
2 submarine sonar arrays, acoustic vector sensors 22 measure the
3 amplitude and phase of acoustic particle motion in a given
4 direction. Vector sensors 22, generally referred to as velocity
5 sensors, measure a vector quantity such as acoustic particle
6 acceleration, particle velocity, or particle displacement.
7 Although accelerometers, velocity sensors, and displacement
8 sensors can each be used to measure acoustic particle motion, for
9 a submarine sonar system, a velocity sensor is preferred as it
10 provides a relatively flat signal to electronic noise ratio over
11 the frequency range of interest.

12 A uniaxis acoustic vector sensor measures one Cartesian
13 component of the acoustic field vector. Similarly, a biaxial
14 vector sensor measures two orthogonal components of the acoustic
15 field vector and a triaxial vector sensor measures all three
16 orthogonal components. Array 20 can be comprised entirely of
17 uniaxis, biaxial, or triaxial vector sensors 22 or the array can
18 be comprised of a combination of uniaxis and multiaxis vector
19 sensors 22. However, as presently understood, an array 20 of
20 triaxial vector sensors 22 measuring three components of acoustic
21 particle motion at the same relative position provides greater
22 gains in directivity and system performance than an array of
23 biaxial or uniaxis vector sensors.

24 For use as a submarine sonar system, the vector sensors 22
25 can be arranged in any desired configuration. In operation,
26 vector sensors 22 measure one or more vector components of
27 acoustic particle motion to generate a vector output signal 26

1 for each sensor within array 20. The output signals 26 are
2 coupled to beamformer 28. Beamformer 28 generates steered
3 acoustic beams using the output signal 26 from one or more of the
4 acoustic vector sensors 22. Beamformer 28 weights the output
5 signals 26 from the sensors by forming the inner product of the
6 velocity components measured at each element location with an
7 element weighting vector, $\vec{w}_n^{(3)}$. Beamformer 28 delays the
8 resulting scalar field generated by weighting and then sums the
9 delayed signals to generate a steered acoustic beam.

10 Referring to FIG. 4, there is shown a block diagram of an
11 embodiment of a submarine sonar system in accordance with the
12 present invention. In FIG. 4 there is shown a cutaway view of a
13 submarine 30 revealing bow array 32. Bow array 32, which
14 corresponds to the bow line array of the BQQ-5 hull array or the
15 conformal bow array of the BSY-1 system, comprises a framework 34
16 supporting a plurality of acoustic elements 36. Framework 34
17 extends around spherical sonar dome 38 of submarine 30.

18 Acoustic elements 36 each comprise a multiaxis vector sensor
19 co-located with a scalar pressure sensor. That is, the elements
20 of the array measure acoustic pressure and the three orthogonal
21 vector components of particle velocity at the same spatial
22 location. Acoustic elements 36 can be used in a one-for-one
23 replacement of the individual pressure sensors in the bow line
24 array segments of the BQQ-5 hull array or the pressure sensors in
25 the conformal bow array of the BSY-1 system. Elements 36 can be
26 mounted in a manner similar to that for the existing pressure

1 sensors. However, the multiaxis vector sensors of elements 36,
2 which measure particle motion, are sensitive to vibrations and
3 must be mounted in a manner such that they are optimally isolated
4 from self-noise vibration. Any vibration isolation device or
5 method known in the art can be used to provide the desired
6 vibration isolation when mounting sensors 36.

7 In operation, bow array 32 is bathed in seawater at the same
8 pressure as the surrounding sea to minimize interference with
9 acoustic transmission. Elements 36 measure one or more vector
10 components of acoustic particle motion and the acoustic pressure
11 to generate an element output signal for each element 36. The
12 element output signals are coupled to a beamformer (not shown)
13 located within submarine 30. The beamformer generates a weighted
14 output for each acoustic element 36 by combining a weighted
15 output of the scalar pressure sensor with the scalar field
16 generated by forming the inner product of the vector components
17 measured at each sensor location with a vector component
18 weighting vector. The weighted output signals are delayed to
19 synchronize the phase of the weighted output signal to that of a
20 weighted signal at a reference array location. The resulting
21 delayed weighted output signals are then summed.

22 It will be understood that various changes in the details,
23 materials, steps and arrangement of parts, which have been herein
24 described and illustrated in order to explain the nature of the
25 invention, may be made by those skilled in the art within the
26 principle and scope of the invention

27

1 Navy Case No. 78605

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3 ACOUSTIC VECTOR SENSING SONAR SYSTEM

4

5 ABSTRACT OF THE DISCLOSURE

6 A system for the detection of acoustic signals utilizing an
7 array of acoustic sensors coupled to a beamformer which generates
8 a steered acoustic beam using the output from one or more of the
9 acoustic sensors within the array. The acoustic sensors include
10 a multiaxis vector sensor co-located with a scalar acoustic
11 pressure sensor. The beamformer generates a weighted output for
12 each acoustic sensor by combining the weighted output of the
13 scalar pressure sensor with the scalar field generated by forming
14 the inner product of the vector components measured at each
15 sensor location with a vector component weighting vector. The
16 weighted output signals are delayed to synchronize the phase of
17 the weighted output signal to that of a weighted signal at a
18 reference array location. The resulting delayed weighted output
19 signals are then summed.

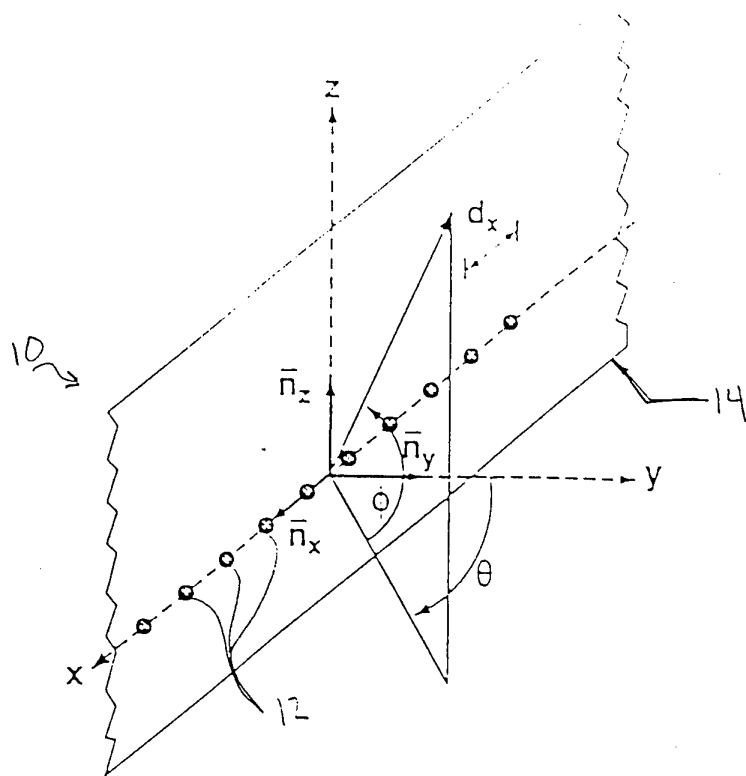


FIG. 1

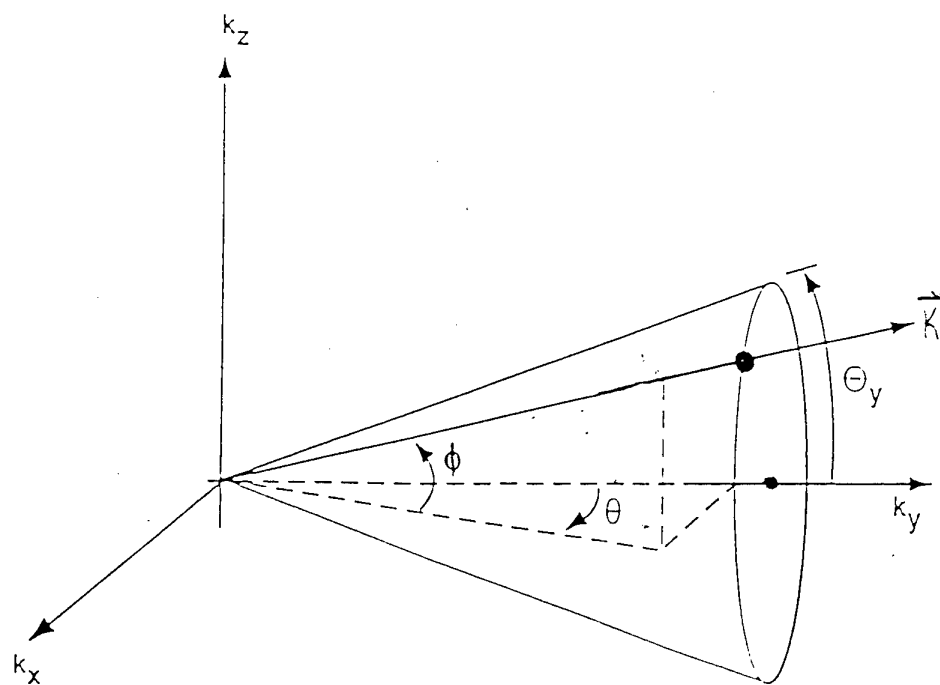


FIG. 2

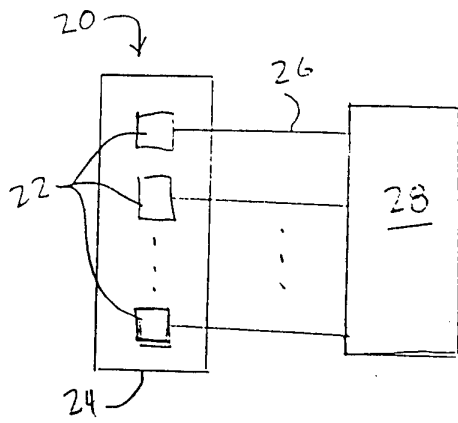


FIG. 3.

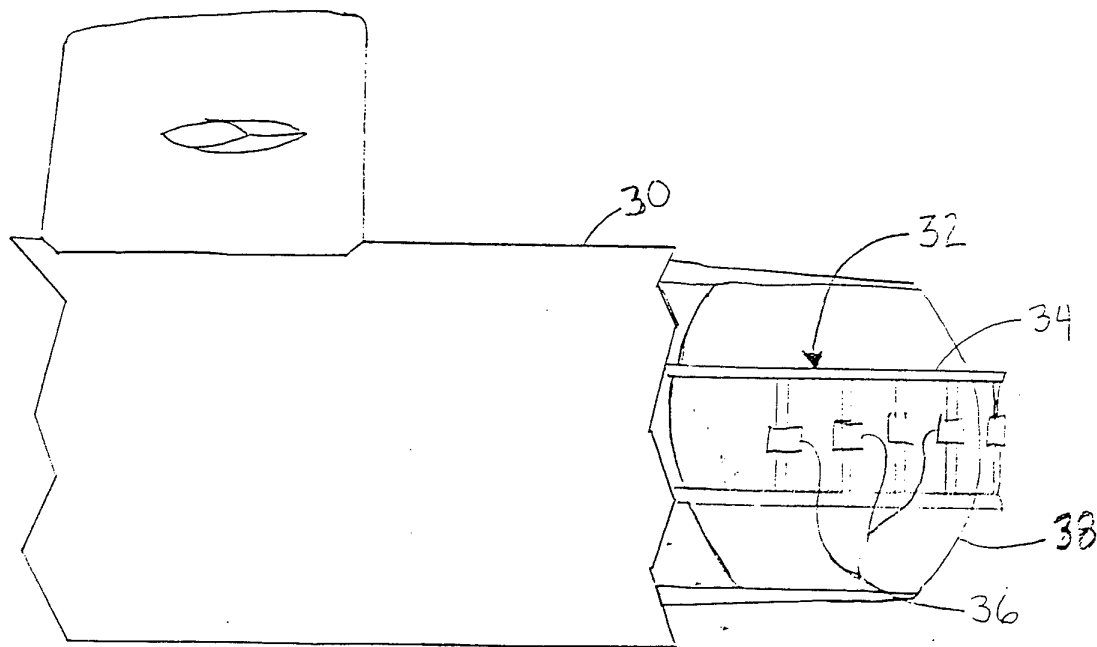


FIG. 4