

Homogenization of Periodically Varying Coefficients in Electromagnetic Materials

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Abstract

In this paper we employ the periodic unfolding method for simulating the electromagnetic field in a composite material exhibiting heterogeneous microstructures which are described by spatially periodic parameters. We consider cell problems to calculate the effective parameters for a Debye dielectric medium in the cases of circular and square microstructures in two dimensions. We assume that the composite materials are quasi-static in nature, i.e., the wavelength of the electromagnetic field is much larger than the relevant dimensions of the microstructure.

Keywords: Homogenization, Maxwell's equations, Debye dielectric materials, pulsed antenna source microwaves.

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1 Introduction

In this article we study the behaviour of the electromagnetic field in a material presenting heterogeneous microstructures (composite materials) which are described by spatially periodic parameters. We will subject such composite materials to electromagnetic fields generated by currents of varying frequencies. When the period of the structure is small compared to the wavelength, the coefficients in Maxwell's equations oscillate rapidly. These oscillating coefficients are difficult to treat numerically in simulations. Homogenization is a process in which the composite material having a microscopic structure is replaced with an equivalent material having macroscopic, homogeneous properties. In this process of homogenization the rapidly oscillating coefficients are replaced with new effective constant coefficients. The primary objective of homogenization, i.e., of the micro-macro approach, is to replace a system with periodically varying coefficients by a limiting homogeneous system that facilitates computation. The approach that we take here is based on the periodic unfolding method presented in [8, 9, 7]. We first mention other efforts on homogenization of Maxwell's equations.

In [25], a method based on spectral expansions for Maxwell's equations is presented, which utilizes eigenvectors of the curl operators combined with the microscopic description of the material. The homogenized material is represented using mean values of only a few eigenvectors. This method relies on the material being lossless, in which case Maxwell's equations can be associated with a self-adjoint partial differential operator. However, most materials usually have losses due to a small conductivity or dispersive effects, which renders the corresponding operator in Maxwell's equations non-selfadjoint. In [24] the authors use a singular value decomposition for analyzing non-selfadjoint operators that arise in Maxwell's equations. They expand the electromagnetic field in the modes corresponding to the singular values, and show that only the smallest singular values make a significant contribution to the total field when the scale is small. Using this approach they find effective, or homogenized, material parameters for Maxwell's equations when the microscopic scale becomes small compared to the scale induced by the frequencies of the imposed currents. In [12], the authors compare two different homogenization methods for Maxwell's equations in two and three dimensions. The first method is the classical way of determining the homogenized coefficients [9], which consists of solving an elliptic problem in a unit cell. The second method based on spectral expansions is described in [25]. In [16], the author presents an overview of the homogenization of anisotropic materials at fixed frequency using the concept of two-scaled convergence [20, 1]. The homogenized electric and magnetic parameters, the relative permittivity and the relative permeability, respectively, are found by suitable averages of the solution of a local problem in the unit cell. In [13] a homogenization technique for harmonic Maxwell equations in a composite periodic medium is presented. See also [26, 17, 21] for some other homogenization methods for Maxwell's equations.

In this paper we use the periodic unfolding method introduced in [8] in the abstract framework of stationary elliptic equations. The study in this paper considers constitutive laws that take into account bianisotropy, chiral symmetry, thermal and memory effects. The homogenization procedure yields a limit constitutive law different from the original one wherein the convolution operator that accounts for memory effects is replaced by a more complex Hilbert-Schmidt operator. We refer the reader to [6, 5] for the relevant theory. In the following sections, we present the electromagnetic problem that is of interest to us and then set up the corresponding homogenized problem to be solved. A comparison is made between the effective parameters obtained by the

exact homogenization method presented here and those computed by traditional mixture formulae, such as the Maxwell Garnett formula, the Böttcher mixture rule or Bruggeman formula, some of which are based on physical arguments [22].

Our efforts here are motivated by use of electromagnetic interrogating signals (possibly in the Terahertz range) for detection of defects [4, 14] in the insulating foam on the fuel tanks of the NASA space shuttles. Defects in the foam are believed to contribute to the problem of separation of the foam during liftoff, resulting in significant damage to and possibly subsequent destruction of the space vehicle itself. The low density, closed cell foam is a very complex heterogeneous material [19]. It is a polyurethane-type foam composed of five primary substances: polymeric isocyanate, a flame retardant, a surfactant, a blowing agent, and a catalyst. The surfactant controls the surface tension of a liquid and thus cell formation. The blowing agent creates the foam's cellular structure by creating millions of tiny bubbles or foam cells. As a first approximation, we consider here materials with periodic gas filled cells surrounded by a matrix of polyurethane type nonmagnetic material. The dielectric properties (permittivity, conductivity, etc.) vary substantially between the cellular and matrix materials, leading to highly oscillating coefficients in the Maxwell system describing propagation, reflection and dispersion of the electromagnetic fields resulting from the interrogating probes.

2 Maxwell's Equations in a Continuous Medium

We employ Maxwell's equations for a linear and isotropic medium in a form that includes terms for the electric polarization given by

$$\left(\begin{array}{l} \text{(i)} \quad \frac{\partial \mathbf{D}}{\partial t} + \mathbf{J}_c - \nabla \times \mathbf{H} = \mathbf{J}_s \text{ in } (0, T) \times \Omega, \\ \text{(ii)} \quad \frac{\partial \mathbf{B}}{\partial t} + \nabla \times \mathbf{E} = 0 \text{ in } (0, T) \times \Omega, \\ \text{(iii)} \quad \nabla \cdot \mathbf{D} = \rho \text{ in } (0, T) \times \Omega, \\ \text{(iv)} \quad \nabla \cdot \mathbf{B} = 0 \text{ in } (0, T) \times \Omega, \\ \text{(v)} \quad \mathbf{E} \times \mathbf{n} = 0 \text{ on } (0, T) \times \partial\Omega, \\ \text{(vi)} \quad \mathbf{E}(0, \mathbf{x}) = \mathbf{0}, \mathbf{H}(0, \mathbf{x}) = \mathbf{0} \text{ in } \Omega. \end{array} \right. \quad (1)$$

The vector valued functions \mathbf{E} and \mathbf{H} represent the strengths of the electric and magnetic fields, respectively, while \mathbf{D} and \mathbf{B} are the electric and magnetic flux densities, respectively. The conduction current density is denoted by \mathbf{J}_c , while the source current density is given by \mathbf{J}_s . The scalar ρ represents the density of free electric charges unaccounted for in the electric polarization. We assume perfect conducting boundary conditions (1,v), on the boundary $\partial\Omega$, with unit outward normal \mathbf{n} . We also assume zero initial conditions for all the unknown fields. System (1) is completed by constitutive laws that embody the behaviour of the material in response to the electromagnetic fields. These are given in $(0, T) \times \Omega$ in the form

$$\left(\begin{array}{l} \text{(i)} \quad \mathbf{D}(t, \mathbf{x}) = \epsilon_0 \epsilon_r(\mathbf{x}) \mathbf{E}(t, \mathbf{x}) + \mathbf{P}(t, \mathbf{x}), \\ \text{(ii)} \quad \mathbf{B}(t, \mathbf{x}) = \mu_0 \mathbf{H}(t, \mathbf{x}). \end{array} \right. \quad (2)$$

For the media that is of interest to us, we can neglect magnetic effects; we also assume that Ohms's law governs the electric conductivity, i.e.,

$$\mathbf{J}_c(t, \mathbf{x}) = \sigma(\mathbf{x})\mathbf{E}(t, \mathbf{x}) \text{ in } (0, T) \times \Omega. \quad (3)$$

We will modify system (1) and the constitutive laws (2) by performing a change of variables that renders the system in a form that is convenient for analysis and computation. From (1,i) we have

$$\frac{\partial}{\partial t} \left(\mathbf{D} + \int_0^t \mathbf{J}_c(s, \mathbf{x}) \, ds \right) - \nabla \times \mathbf{H} = \mathbf{J}_s \text{ in } (0, T) \times \Omega. \quad (4)$$

Then we define the new variable

$$\tilde{\mathbf{D}}(t, \mathbf{x}) = \mathbf{D}(t, \mathbf{x}) + \int_0^t \mathbf{J}_c(s, \mathbf{x}) \, ds. \quad (5)$$

Using definition (5) in (4) and (1) we obtain the modified system

$$\left(\begin{array}{l} \text{(i)} \quad \frac{\partial \tilde{\mathbf{D}}}{\partial t} - \nabla \times \mathbf{H} = \mathbf{J}_s \text{ in } (0, T) \times \Omega, \\ \text{(ii)} \quad \frac{\partial \mathbf{B}}{\partial t} + \nabla \times \mathbf{E} = 0 \text{ in } (0, T) \times \Omega, \\ \text{(iii)} \quad \nabla \cdot \tilde{\mathbf{D}} = 0 \text{ in } (0, T) \times \Omega, \\ \text{(iv)} \quad \nabla \cdot \mathbf{B} = 0 \text{ in } (0, T) \times \Omega, \\ \text{(v)} \quad \mathbf{E} \times \mathbf{n} = 0 \text{ on } (0, T) \times \partial\Omega, \\ \text{(vi)} \quad \mathbf{E}(0, \mathbf{x}) = \mathbf{0}, \mathbf{H}(0, \mathbf{x}) = \mathbf{0} \text{ in } \Omega. \end{array} \right. \quad (6)$$

We note that equation (6,iii) follows from the continuity equation $\frac{\partial \rho}{\partial t} + \nabla \cdot \mathbf{J}_c = 0$, the assumption that $\rho(0) = 0$, and the assumption that $\nabla \cdot \mathbf{J}_s = 0$ (in the sense of distributions). The modified constitutive law (2,i) after substitution of (3) and (5) becomes

$$\tilde{\mathbf{D}}(t, \mathbf{x}) = \epsilon_0 \epsilon_r(\mathbf{x})\mathbf{E}(t, \mathbf{x}) + \int_0^t \sigma(\mathbf{x})\mathbf{E}(s, \mathbf{x}) \, ds + \mathbf{P}(t, \mathbf{x}). \quad (7)$$

To describe the behaviour of the media's macroscopic electric polarization \mathbf{P} , we employ a general integral representation model in which the polarization explicitly depends on the past history of the electric field. This convolution model is sufficiently general to include microscopic polarization mechanisms such as dipole or orientational polarization as well as ionic and electronic polarization and other frequency dependent polarization mechanisms. The resulting constitutive law can be given in terms of a polarization or displacement susceptibility kernel ν as

$$\mathbf{P}(t, \mathbf{x}) = \int_0^t \nu(t-s, \mathbf{x})\mathbf{E}(s, \mathbf{x}) \, ds. \quad (8)$$

Thus the modified constitutive laws are

$$\left(\begin{array}{l} \text{(i)} \quad \mathbf{D}(t, \mathbf{x}) = \epsilon_0 \epsilon_r(\mathbf{x})\mathbf{E}(t, \mathbf{x}) + \int_0^t \{ \sigma(\mathbf{x}) + \nu(t-s, \mathbf{x}) \} \mathbf{E}(s, \mathbf{x}) \, ds, \\ \text{(ii)} \quad \mathbf{B}(t, \mathbf{x}) = \mu_0 \mathbf{H}(t, \mathbf{x}), \end{array} \right. \quad (9)$$

where, in the above and henceforth we have dropped the \sim symbol over \mathbf{D} , at the same time keeping in mind that $\tilde{\mathbf{D}}$ in definition (5) is the *modified electric flux density*. Let us define the vector of fields

$$\mathbf{u} = (\mathbf{u}_1, \mathbf{u}_2)^T = (\mathbf{E}, \mathbf{H})^T \in W^{1,1}(0, T; H^1(\Omega; \mathbb{R}^6)), \quad (10)$$

and the operator

$$\mathbf{L}\mathbf{u}(t, \mathbf{x}) = \begin{pmatrix} \mathbf{D}(t, \mathbf{x}) \\ \mathbf{B}(t, \mathbf{x}) \end{pmatrix}, \quad (11)$$

which from (9) can be written as:

$$\begin{aligned} \mathbf{L}\mathbf{u}(t, \mathbf{x}) &= \begin{bmatrix} \epsilon_0 \epsilon_r(\mathbf{x}) \mathbf{I}_3 & \mathbf{0}_3 \\ \mathbf{0}_3 & \mu_0 \mathbf{I}_3 \end{bmatrix} \begin{pmatrix} \mathbf{E}(t, \mathbf{x}) \\ \mathbf{H}(t, \mathbf{x}) \end{pmatrix} \\ &+ \int_0^t \left\{ \begin{bmatrix} \sigma(\mathbf{x}) \mathbf{I}_3 & \mathbf{0}_3 \\ \mathbf{0}_3 & \mathbf{0}_3 \end{bmatrix} + \begin{bmatrix} \nu(t-s, \mathbf{x}) \mathbf{I}_3 & \mathbf{0}_3 \\ \mathbf{0}_3 & \mathbf{0}_3 \end{bmatrix} \right\} \begin{pmatrix} \mathbf{E}(s, \mathbf{x}) \\ \mathbf{H}(s, \mathbf{x}) \end{pmatrix} ds. \end{aligned} \quad (12)$$

We label the three 6×6 coefficient matrices in (12) as

$$\mathbf{A}(\mathbf{x}) = \begin{bmatrix} \epsilon_0 \epsilon_r(\mathbf{x}) \mathbf{I}_3 & \mathbf{0}_3 \\ \mathbf{0}_3 & \mu_0 \mathbf{I}_3 \end{bmatrix}, \quad (13a)$$

$$\mathbf{B}(\mathbf{x}) = \begin{bmatrix} \sigma(\mathbf{x}) \mathbf{I}_3 & \mathbf{0}_3 \\ \mathbf{0}_3 & \mathbf{0}_3 \end{bmatrix}, \quad (13b)$$

$$\mathbf{C}(t, \mathbf{x}) = \begin{bmatrix} \nu(t, \mathbf{x}) \mathbf{I}_3 & \mathbf{0}_3 \\ \mathbf{0}_3 & \mathbf{0}_3 \end{bmatrix}, \quad (13c)$$

where, in the above definitions \mathbf{I}_n is an $n \times n$ identity matrix and $\mathbf{0}_n$ is an $n \times n$ matrix of zeros, $n \in \mathbb{N}$. Using these definitions we may rewrite (12) as

$$\mathbf{L}\mathbf{u}(t, \mathbf{x}) = \mathbf{A}(\mathbf{x})\mathbf{u}(t, \mathbf{x}) + \int_0^t \mathbf{B}(\mathbf{x})\mathbf{u}(s, \mathbf{x}) ds + \int_0^t \mathbf{C}(t-s, \mathbf{x})\mathbf{u}(s, \mathbf{x}) ds. \quad (14)$$

Next, we define the Maxwell operator \mathbf{M} as

$$\mathbf{M}\mathbf{u}(t, \mathbf{x}) = \mathbf{M} \begin{pmatrix} \mathbf{E}(t, \mathbf{x}) \\ \mathbf{H}(t, \mathbf{x}) \end{pmatrix} = \begin{pmatrix} \nabla \times \mathbf{H}(t, \mathbf{x}) \\ -\nabla \times \mathbf{E}(t, \mathbf{x}) \end{pmatrix} \quad (15)$$

and the vector \mathbf{J}_s as

$$\mathbf{J}_s(t) = -J_s(t)\mathbf{e}_1 \quad (16)$$

where, $\mathbf{e}_1 = (1, 0, 0, 0, 0, 0)^T \in \mathbb{R}^6$, is a unit basis vector. Thus Maxwell's equation can be rewritten in the form

$$\begin{cases} \text{(i)} & \frac{d}{dt} \mathbf{L}\mathbf{u} = \mathbf{M}\mathbf{u} + \mathbf{J}_s(t) \text{ in } (0, T) \times \Omega, \\ \text{(ii)} & \mathbf{u}(0, \mathbf{x}) = \mathbf{0} \text{ in } \Omega, \\ \text{(iii)} & \mathbf{u}_1(t, \mathbf{x}) \times \mathbf{n}(\mathbf{x}) = \mathbf{0} \text{ on } (0, T) \times \partial\Omega, \end{cases} \quad (17)$$

where \mathbf{L} is the operator associated with the constitutive law (14), and \mathbf{M} is the Maxwell operator (15). Note that the exterior source term \mathbf{J}_s has only one non-zero component.

We assume that the structure that occupies the domain Ω entails periodic microstructures leading to matrices \mathbf{A} , \mathbf{B} and \mathbf{C} with spatially oscillatory coefficients. Specifically, we will assume that ϵ_r , σ and ν are rapidly oscillating spatial functions.

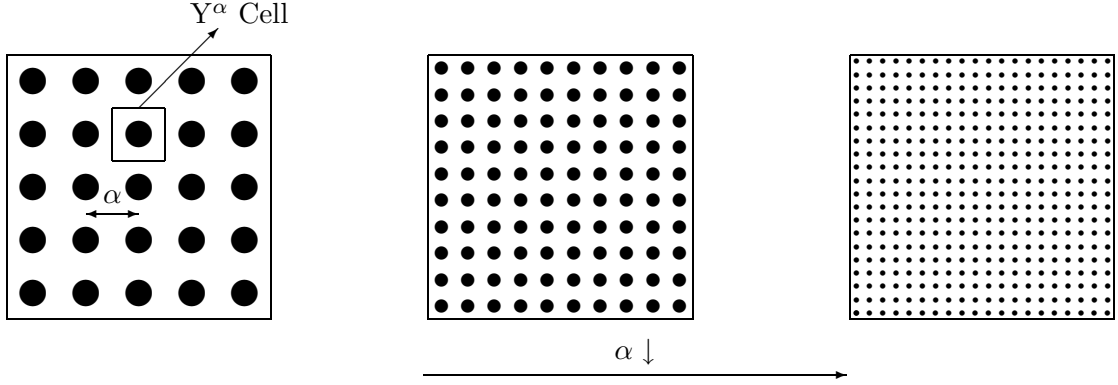


Figure 1: Periodic composite material presenting a circular microstructure with periodicity α . The figure shows α decreasing from left to right.

3 The Homogenized Problem

The theory presented in this section is based on results from [6]. We denote by Y^α the reference cell of the periodic structure that occupies Ω (see Figure 1). The construction of the homogenized problem involves solving for the *corrector* subterms $\bar{\mathbf{w}}_k^A \in H_{\text{per}}^1(Y; \mathbb{R}^2)$, $\bar{\mathbf{w}}_k \in W^{1,1}(0, T; H_{\text{per}}^1(Y; \mathbb{R}^2))$ and $\bar{\mathbf{w}}_k^0 \in W^{2,1}(0, T; H_{\text{per}}^1(Y; \mathbb{R}^2))$, solutions to the corrector equations

$$\left\{ \begin{array}{l} \text{(i)} \int_Y \mathbf{A}(\mathbf{y}) \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k^A \cdot \nabla_{\mathbf{y}} \bar{\mathbf{v}}(\mathbf{y}) \, d\mathbf{y} = - \int_Y \mathbf{A}(\mathbf{y}) \mathbf{e}_k \cdot \nabla_{\mathbf{y}} \bar{\mathbf{v}}(\mathbf{y}) \, d\mathbf{y}, \\ \text{(ii)} \int_Y \left\{ \mathbf{A}(\mathbf{y}) \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k(t, \mathbf{y}) + \int_0^t \{ \mathbf{B}(\mathbf{y}) + \mathbf{C}(t-s, \mathbf{y}) \} \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k(s, \mathbf{y}) \, ds \right\} \cdot \nabla_{\mathbf{y}} \bar{\mathbf{v}}(\mathbf{y}) \, d\mathbf{y} \\ \quad = - \int_Y \{ \mathbf{B}(\mathbf{y}) + \mathbf{C}(t, \mathbf{y}) \} \{ \mathbf{e}_k + \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k^A \} \cdot \nabla_{\mathbf{y}} \bar{\mathbf{v}}(\mathbf{y}) \, d\mathbf{y}, \\ \text{(iii)} \int_Y \left\{ \mathbf{A}(\mathbf{y}) \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k^0(t, \mathbf{y}) + \int_0^t \{ \mathbf{B}(\mathbf{y}) + \mathbf{C}(t-s, \mathbf{y}) \} \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k^0(s, \mathbf{y}) \, ds \right\} \cdot \nabla_{\mathbf{y}} \bar{\mathbf{v}}(\mathbf{y}) \, d\mathbf{y} \\ \quad = - \int_Y \mathbf{A}(\mathbf{y}) \mathbf{e}_k \cdot \nabla_{\mathbf{y}} \bar{\mathbf{v}}(\mathbf{y}) \, d\mathbf{y}, \end{array} \right. \quad (18)$$

for all $\bar{\mathbf{v}} \in H_{\text{per}}^1(Y; \mathbb{R}^2)$ and $k \in \{1, \dots, 6\}$. Here, $H_{\text{per}}^1(Y)$ denotes the space of periodic functions with vanishing mean value. Note that if we decompose $\bar{\mathbf{v}}$ into $c_1 \bar{\mathbf{v}}_1 + c_2 \bar{\mathbf{v}}_2$ where $\bar{\mathbf{v}}_1 = [v_1, 0]$ and $\bar{\mathbf{v}}_2 = [0, v_2]$, then each set of equations above decouple into an equation for $\bar{\mathbf{w}}_{k,1}$ and one for $\bar{\mathbf{w}}_{k,2}$.

For example, equation (18,i) becomes

$$\begin{cases} \text{(ia)} & \int_Y \mathbf{A}_{11}(\mathbf{y}) \nabla_{\mathbf{y}} \bar{\mathbf{w}}_{k,1}^A \cdot \nabla_{\mathbf{y}} \bar{\mathbf{v}}_1(\mathbf{y}) \, d\mathbf{y} = - \int_Y \mathbf{A}_{11}(\mathbf{y}) \mathbf{e}_{k,1} \cdot \nabla_{\mathbf{y}} \bar{\mathbf{v}}_1(\mathbf{y}) \, d\mathbf{y} \\ \text{(ib)} & \int_Y \mathbf{A}_{22}(\mathbf{y}) \nabla_{\mathbf{y}} \bar{\mathbf{w}}_{k,2}^A \cdot \nabla_{\mathbf{y}} \bar{\mathbf{v}}_2(\mathbf{y}) \, d\mathbf{y} = - \int_Y \mathbf{A}_{22}(\mathbf{y}) \mathbf{e}_{k,2} \cdot \nabla_{\mathbf{y}} \bar{\mathbf{v}}_2(\mathbf{y}) \, d\mathbf{y}. \end{cases} \quad (19)$$

where \mathbf{A}_{11} denotes the first 3×3 block of \mathbf{A} and $\mathbf{e}_{k,1} = [1, 1, 1, 0, 0, 0]^T$, and similarly \mathbf{A}_{22} denotes the last 3×3 block of \mathbf{A} with $\mathbf{e}_{k,2} = [0, 0, 0, 1, 1, 1]^T$.

The first corrector term $\bar{\mathbf{u}}$, from the two-scale expansion (see [6])

$$\mathbf{u}^\alpha = \mathbf{u}(\mathbf{x}) + \nabla_{\mathbf{y}} \bar{\mathbf{u}}(\mathbf{x}, \mathbf{y}) + \dots, \quad \mathbf{x} \in \Omega, \mathbf{y} \in Y, \quad (20)$$

is given as

$$\bar{\mathbf{u}}(t, \mathbf{x}, \mathbf{y}) = \bar{\mathbf{w}}_k^A(\mathbf{y}) u_k(t, \mathbf{x}) + \int_0^t \bar{\mathbf{w}}_k(t-s, \mathbf{y}) u_k(s, \mathbf{x}) \, ds + \bar{\mathbf{w}}_k^0(t, \mathbf{y}) u_k^0(\mathbf{x}). \quad (21)$$

with $\bar{\mathbf{u}} \in W^{2,1}(0, T; H_{\text{per}}^1(Y; \mathbb{R}^2))$, where we have considered the decompositions $\mathbf{u}(t, \mathbf{x}) = u_k(t, \mathbf{x}) \mathbf{e}_k$ and $\mathbf{u}^0(\mathbf{x}) = u_k^0(\mathbf{x}) \mathbf{e}_k$, $k = 1, \dots, 6$. Rewriting (21) in matrix form we have

$$\bar{\mathbf{u}}(t, \mathbf{x}, \mathbf{y}) = \bar{\mathbf{w}}^A(\mathbf{y}) \mathbf{u}(t, \mathbf{x}) + \int_0^t \bar{\mathbf{w}}(t-s, \mathbf{y}) \mathbf{u}(s, \mathbf{x}) \, ds + \bar{\mathbf{w}}^0(t, \mathbf{y}) \mathbf{u}^0(\mathbf{x}). \quad (22)$$

where $\bar{\mathbf{w}}^A \in \mathbb{R}^{2 \times 6}$ with columns $\{\bar{\mathbf{w}}_k^A\}_{k=1}^6$. Similarly $\bar{\mathbf{w}}^0, \bar{\mathbf{w}} \in \mathbb{R}^{2 \times 6}$. The expansion (20) can be written as

$$E_{x_1}^\alpha = E_{x_1} + \partial_{y_1} \bar{u}_1(\mathbf{x}, \mathbf{y}) + \dots \quad (23a)$$

$$E_{x_2}^\alpha = E_{x_2} + \partial_{y_2} \bar{u}_1(\mathbf{x}, \mathbf{y}) + \dots \quad (23b)$$

$$E_{x_3}^\alpha = E_{x_3} + \partial_{y_3} \bar{u}_1(\mathbf{x}, \mathbf{y}) + \dots \quad (23c)$$

$$H_{x_1}^\alpha = H_{x_1} + \partial_{y_1} \bar{u}_2(\mathbf{x}, \mathbf{y}) + \dots \quad (23d)$$

$$H_{x_2}^\alpha = H_{x_2} + \partial_{y_2} \bar{u}_2(\mathbf{x}, \mathbf{y}) + \dots \quad (23e)$$

$$H_{x_3}^\alpha = H_{x_3} + \partial_{y_3} \bar{u}_2(\mathbf{x}, \mathbf{y}) + \dots \quad (23f)$$

We now define a new operator

$$\mathcal{L}\mathbf{u}(t, \mathbf{x}) = \mathcal{A}\mathbf{u}(t, \mathbf{x}) + \mathcal{B} \int_0^t \mathbf{u}(s, \mathbf{x}) \, ds + \int_0^t \mathcal{C}(t-s) \mathbf{u}(s, \mathbf{x}) \, ds, \quad (24)$$

where the 6×6 matrices \mathcal{A}, \mathcal{B} and \mathcal{C} are computed using the solutions of system (18) as follows

$$\begin{cases} \text{(i)} & \mathcal{A}_k = \int_Y \mathbf{A}(\mathbf{y}) \{ \mathbf{e}_k + \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k^A(\mathbf{y}) \} \, d\mathbf{y}, \\ \text{(ii)} & \mathcal{B}_k = \int_Y \mathbf{B}(\mathbf{y}) \{ \mathbf{e}_k + \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k^A(\mathbf{y}) \} \, d\mathbf{y}, \\ \text{(iii)} & \mathcal{C}_k(t) = \int_Y \mathbf{C}(t, \mathbf{y}) \{ \mathbf{e}_k + \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k^A(\mathbf{y}) \} \, d\mathbf{y} + \int_Y \mathbf{A}(\mathbf{y}) \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k(t, \mathbf{y}) \, d\mathbf{y} \\ & \quad + \int_Y \int_0^t \{ \mathbf{B}(\mathbf{y}) + \mathbf{C}(t-s, \mathbf{y}) \} \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k(s, \mathbf{y}) \, ds \, d\mathbf{y}, \end{cases} \quad (25)$$

for $k = 1, 2, \dots, 6$, and where $\mathcal{A}_k, \mathcal{B}_k, \mathcal{C}_k$ are the k th columns of the homogenized matrices \mathcal{A}, \mathcal{B} , and \mathcal{C} , respectively. In the homogenized problem, the electromagnetic field \mathbf{u} is the solution of the system

$$\begin{cases} \text{(i)} & \frac{d}{dt} \mathcal{L}\mathbf{u} = \mathbf{M}\mathbf{u} + \mathbf{J}_s \text{ in } (0, T) \times \Omega, \\ \text{(ii)} & \mathbf{u}(0, \mathbf{x}) = \mathbf{0} \text{ in } \Omega, \\ \text{(iii)} & \mathbf{u}_1(t) \times \mathbf{n} = \mathbf{0} \text{ on } (0, T) \times \partial\Omega, \end{cases} \quad (26)$$

where \mathbf{J}_s is as defined in (16), \mathbf{M} is as defined in (15), and \mathcal{L} is as defined in (24). We note that if the initial conditions are nonzero, then there is a supplementary source term \mathcal{J}^0 that should be introduced in the right side of (26,i), which is given to be

$$\mathcal{J}^0(t, \mathbf{x}) = u_k^0(\mathbf{x}) \frac{d}{dt} \left\{ \int_Y \left(\mathbf{A} \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k^0(t) + \int_0^t (\mathbf{B} + \mathbf{C}(t-s)) \nabla_{\mathbf{y}} \bar{\mathbf{w}}_k^0(s) \, ds \right) \right\}, \quad (27)$$

$k = 1, \dots, 6$. See [6] for details.

4 Reduction to Two Spatial Dimensions

We now assume our problem possesses uniformity in the spatial direction y (see Figure 2 for a schematic of the computational domain). In this case Maxwell's equations decouple into two different modes, the *transverse electric* (TE) and *transverse magnetic* (TM) modes. Here, we are interested in the TE_y mode. The TE_y mode involves the components E_x, E_z for the electric field and the component H_y of the magnetic field. Let $\mathbf{x}, \mathbf{y} \in \mathbb{R}^3$ with $\mathbf{x} = (x_1, x_2, x_3)$ and $\mathbf{y} = (y_1, y_2, y_3)$. We will use $\mathbf{x} \in \mathbb{R}^3$ for points on the macro scale, and $\mathbf{y} \in \mathbb{R}^3$ for points on the micro scale (reference cell). Since we assume uniformity in the x_2 -direction, we may take all derivatives with respect to x_2 (or y_2) to be zero. Then equation (17,i) can be written in scalar form as

$$\begin{bmatrix} \partial_t D_{x_1} \\ \partial_t D_{x_2} \\ \partial_t D_{x_3} \\ \partial_t B_{x_1} \\ \partial_t B_{x_2} \\ \partial_t B_{x_3} \end{bmatrix} = \begin{bmatrix} \partial_{x_3} H_{x_2} - J_s \\ \partial_{x_3} H_{x_1} - \partial_{x_1} H_{x_3} \\ \partial_{x_1} H_{x_2} \\ -\partial_{x_3} E_{x_2} \\ -\partial_{x_3} E_{x_1} + \partial_{x_1} E_{x_3} \\ -\partial_{x_1} E_{x_2} \end{bmatrix}, \quad (28)$$

with $\mathbf{L}\mathbf{u} = (\mathbf{D}, \mathbf{B})^T$. Recall here that \mathbf{D} is the modified electric flux density, where we have dropped the \sim notation. We may decouple system (28) into the TE mode,

$$\begin{bmatrix} \partial_t D_{x_1} \\ \partial_t B_{x_2} \\ \partial_t D_{x_3} \end{bmatrix} = \begin{bmatrix} \partial_{x_3} H_{x_2} - J_s \\ -\partial_{x_3} E_{x_1} + \partial_{x_1} E_{x_3} \\ \partial_{x_1} H_{x_2} \end{bmatrix}, \quad (29)$$

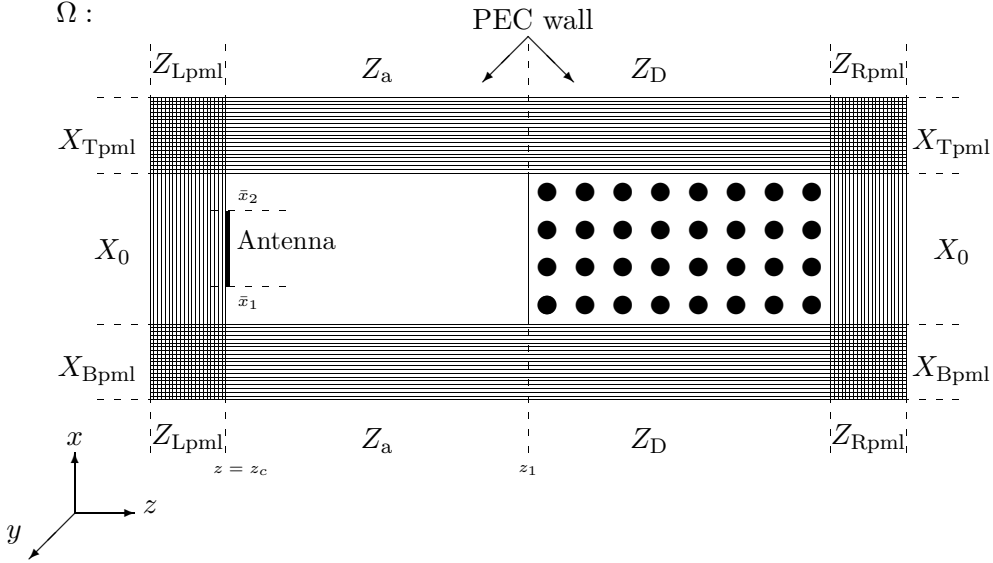


Figure 2: The computational domain with surrounding perfectly matched absorbing layers (PML). The figure shows an antenna in the region $[\bar{x}_1, \bar{x}_2] \times z_c$, $\bar{x}_1, \bar{x}_2 \in X_0$, and the composite material displaying periodic circular microstructures in the region $X_0 \times Z_D$. The PML layers are terminated by perfectly conducting boundary conditions (PEC walls).

and the TM mode

$$\begin{bmatrix} \partial_t B_{x_1} \\ \partial_t D_{x_2} \\ \partial_t B_{x_3} \end{bmatrix} = \begin{bmatrix} -\partial_{x_3} E_{x_2} \\ \partial_{x_3} H_{x_1} - \partial_{x_1} H_{x_3} \\ -\partial_{x_1} E_{x_2} \end{bmatrix}. \quad (30)$$

We assume that our pulse is polarized to only have an x_1 -component. In this case the component that is of interest in our problem is the E_{x_1} component.

5 Homogenization Model in Two Dimensions

In a similar manner to the three dimensional case, we may construct matrices \mathbf{A}^{TE} , \mathbf{B}^{TE} , and \mathbf{C}^{TE} that represent the constitutive relations in two dimensions. Thus the constitutive matrices are

$$\mathbf{A}^{\text{TE}} = \begin{bmatrix} \mathbf{A}_{11}^{\text{TE}} & 0 \\ 0 & \mu_0 \end{bmatrix}; \quad \mathbf{B}^{\text{TE}} = \begin{bmatrix} \mathbf{B}_{11}^{\text{TE}} & 0 \\ 0 & 0 \end{bmatrix}; \quad \mathbf{C}^{\text{TE}} = \begin{bmatrix} \mathbf{C}_{11}^{\text{TE}} & 0 \\ 0 & 0 \end{bmatrix}; \quad (31)$$

$$\mathbf{A}_{11}^{\text{TE}}(\mathbf{x}) = \begin{bmatrix} \epsilon_0 \epsilon_r(\mathbf{x}) & 0 \\ 0 & \epsilon_0 \epsilon_r(\mathbf{x}) \end{bmatrix}; \quad \mathbf{B}_{11}^{\text{TE}}(\mathbf{x}) = \begin{bmatrix} \sigma(\mathbf{x}) & 0 \\ 0 & \sigma(\mathbf{x}) \end{bmatrix}; \quad \mathbf{C}_{11}^{\text{TE}}(t, \mathbf{x}) = \begin{bmatrix} \nu(t, \mathbf{x}) & 0 \\ 0 & \nu(t, \mathbf{x}) \end{bmatrix}. \quad (32)$$

The homogenized solution for the TE mode is obtained from the formal asymptotic expansion

(23) as

$$E_{x_1}^\alpha = E_{x_1} + \partial_{y_1} \bar{u}_1(\mathbf{x}, \mathbf{y}) + \dots \quad (33a)$$

$$E_{x_3}^\alpha = E_{x_3} + \partial_{y_3} \bar{u}_1(\mathbf{x}, \mathbf{y}) + \dots \quad (33b)$$

$$H_{x_2}^\alpha = H_{x_2} + \partial_{y_2} \bar{u}_2(\mathbf{x}, \mathbf{y}) + \dots \quad (33c)$$

Also, since we are assuming uniformity in the x_2 direction, we set the term $\partial_{y_2} \bar{u}_2(\mathbf{x}, \mathbf{y})$ to zero. So (33) becomes

$$E_{x_1}^\alpha = E_{x_1} + \partial_{y_1} \bar{u}_1(\mathbf{x}, \mathbf{y}) + \dots \quad (34a)$$

$$E_{x_3}^\alpha = E_{x_3} + \partial_{y_3} \bar{u}_1(\mathbf{x}, \mathbf{y}) + \dots \quad (34b)$$

$$H_{x_2}^\alpha = H_{x_2}. \quad (34c)$$

Hence the homogenized electric field for the TE mode is

$$\mathbf{E}^\alpha = \mathbf{E} + \nabla_{\mathbf{y}} \bar{u}_1(\mathbf{x}, \mathbf{y}) + \dots, \quad (35)$$

where the gradient operator in this case is $\nabla_{\mathbf{y}} = (\partial_{y_1}, \partial_{y_3})^T$. Therefore we only need to solve for $\bar{u}_1(\mathbf{x}, \mathbf{y})$, which in turn only depends on the first component of $\bar{\mathbf{w}}_k^A$, $\bar{\mathbf{w}}_k^0$, and $\bar{\mathbf{w}}_k$, for $k = 1, 2$; we will refer to these as \bar{w}_k^A , \bar{w}_k^0 , and \bar{w}_k , respectively.

Let us again denote by Y the reference cell of the periodic structure that occupies $\Omega \subset \mathbb{R}^2$. The construction of the two-dimensional homogenized problem involves solving for the *corrector* subterms $\bar{w}_k^A \in H_{\text{per}}^1(Y; \mathbb{R})$, $\bar{w}_k \in W^{1,1}(0, T; H_{\text{per}}^1(Y; \mathbb{R}))$ and $\bar{w}_k^0 \in W^{2,1}(0, T; H_{\text{per}}^1(Y; \mathbb{R}))$, solutions to the corrector equations

$$\left\{ \begin{array}{l} \text{(i)} \int_Y \mathbf{A}_{11}^{\text{TE}}(\mathbf{y}) \nabla_{\mathbf{y}} \bar{w}_k^A \cdot \nabla_{\mathbf{y}} \bar{v}(\mathbf{y}) \, d\mathbf{y} = - \int_Y \mathbf{A}_{11}^{\text{TE}}(\mathbf{y}) \mathbf{e}_k \cdot \nabla_{\mathbf{y}} \bar{v}(\mathbf{y}) \, d\mathbf{y}, \\ \text{(ii)} \int_Y \mathbf{A}_{11}^{\text{TE}}(\mathbf{y}) \nabla_{\mathbf{y}} \bar{w}_k(t, \mathbf{y}) \cdot \nabla_{\mathbf{y}} \bar{v}(\mathbf{y}) \, d\mathbf{y} \\ \quad + \int_Y \int_0^t \{ \mathbf{B}_{11}^{\text{TE}}(\mathbf{y}) + \mathbf{C}_{11}^{\text{TE}}(t-s, \mathbf{y}) \} \nabla_{\mathbf{y}} \bar{w}_k(s, \mathbf{y}) \, ds \cdot \nabla_{\mathbf{y}} \bar{v}(\mathbf{y}) \, d\mathbf{y} \\ \quad = - \int_Y \{ \mathbf{B}_{11}^{\text{TE}}(\mathbf{y}) + \mathbf{C}_{11}^{\text{TE}}(t, \mathbf{y}) \} \{ \mathbf{e}_k + \nabla_{\mathbf{y}} \bar{w}_k^A \} \cdot \nabla_{\mathbf{y}} \bar{v}(\mathbf{y}) \, d\mathbf{y}, \\ \text{(iii)} \int_Y \mathbf{A}_{11}^{\text{TE}}(\mathbf{y}) \nabla_{\mathbf{y}} \bar{w}_k^0(t, \mathbf{y}) \cdot \nabla_{\mathbf{y}} \bar{v}(\mathbf{y}) \, d\mathbf{y} \\ \quad + \int_Y \int_0^t \{ \mathbf{B}_{11}^{\text{TE}}(\mathbf{y}) + \mathbf{C}_{11}^{\text{TE}}(t-s, \mathbf{y}) \} \nabla_{\mathbf{y}} \bar{w}_k^0(s, \mathbf{y}) \, ds \cdot \nabla_{\mathbf{y}} \bar{v}(\mathbf{y}) \, d\mathbf{y} \\ \quad = - \int_Y \mathbf{A}_{11}^{\text{TE}}(\mathbf{y}) \mathbf{e}_k \cdot \nabla_{\mathbf{y}} \bar{v}(\mathbf{y}) \, d\mathbf{y}, \end{array} \right. \quad (36)$$

for all $\bar{v} \in H_{\text{per}}^1(Y; \mathbb{R})$, $k = 1, 2$ and $\mathbf{e}_1 = [1, 0]^T$, $\mathbf{e}_2 = [0, 1]^T$. Since the initial conditions that we have chosen are zero, we will not need to calculate the corrector \bar{w}_k^0 . We will only need to solve for \bar{w}_k^A and \bar{w}_k . Once we have solved for these corrector terms, we can then construct the homogenized

matrices from

$$\left\{ \begin{array}{l} \text{(i)} \quad (\mathcal{A}_{11}^{\text{TE}})_k = \int_Y \mathbf{A}_{11}^{\text{TE}}(\mathbf{y}) \{ \mathbf{e}_k + \nabla_{\mathbf{y}} \bar{w}_k^A(\mathbf{y}) \} d\mathbf{y}, \\ \text{(ii)} \quad (\mathcal{B}_{11}^{\text{TE}})_k = \int_Y \mathbf{B}_{11}^{\text{TE}}(\mathbf{y}) \{ \mathbf{e}_k + \nabla_{\mathbf{y}} \bar{w}_k^A(\mathbf{y}) \} d\mathbf{y}, \\ \text{(iii)} \quad (\mathcal{C}_{11}^{\text{TE}})_k(t) = \int_Y \mathbf{C}_{11}^{\text{TE}}(t, \mathbf{y}) \{ \mathbf{e}_k + \nabla_{\mathbf{y}} \bar{w}_k^A(\mathbf{y}) \} d\mathbf{y} + \int_Y \mathbf{A}_{11}^{\text{TE}}(\mathbf{y}) \nabla_{\mathbf{y}} \bar{w}_k(t, \mathbf{y}) d\mathbf{y} \\ \quad \quad \quad + \int_Y \int_0^t \{ \mathbf{B}_{11}^{\text{TE}}(\mathbf{y}) + \mathbf{C}_{11}^{\text{TE}}(t-s, \mathbf{y}) \} \nabla_{\mathbf{y}} \bar{w}_k(s, \mathbf{y}) ds d\mathbf{y}, \end{array} \right. \quad (37)$$

where \mathbf{e}_k , $k = 1, 2$ are the basis vectors in \mathbb{R}^2 , $(\mathcal{A}_{11}^{\text{TE}})_k$, $(\mathcal{B}_{11}^{\text{TE}})_k$, and $(\mathcal{C}_{11}^{\text{TE}})_k$ are the k th columns of the matrices $\mathcal{A}_{11}^{\text{TE}}$, $\mathcal{B}_{11}^{\text{TE}}$, and $\mathcal{C}_{11}^{\text{TE}}$, respectively, and the homogenized matrices are given by

$$\mathcal{A}^{\text{TE}} = \begin{bmatrix} \mathcal{A}_{11}^{\text{TE}} & 0 \\ 0 & \mu_0 \end{bmatrix}; \quad \mathcal{B}^{\text{TE}} = \begin{bmatrix} \mathcal{B}_{11}^{\text{TE}} & 0 \\ 0 & 0 \end{bmatrix}; \quad \mathcal{C}^{\text{TE}} = \begin{bmatrix} \mathcal{C}_{11}^{\text{TE}} & 0 \\ 0 & 0 \end{bmatrix}. \quad (38)$$

The corresponding system of equations in the TE mode are

$$\left\{ \begin{array}{l} \text{(i)} \quad \frac{d}{dt} \mathcal{L}^{\text{TE}} \mathbf{v} = \mathbf{M}^{\text{TE}} \mathbf{v} + \mathbf{J}_s^{\text{TE}} \text{ in } (0, T) \times \Omega, \\ \text{(ii)} \quad \mathbf{v}(0, \mathbf{x}) = \mathbf{0} \text{ in } \Omega, \\ \text{(iii)} \quad \mathbf{v}_3 n_{x_1} - \mathbf{v}_1 n_{x_3} = 0 \text{ on } (0, T) \times \partial\Omega, \end{array} \right. \quad (39)$$

where $\mathbf{v} = (E_{x_1}, H_{x_2}, E_{x_3})^T$, $\mathbf{n} = (n_{x_1}, n_{x_3})^T$ is the unit outward normal vector to $\partial\Omega$, the operator \mathcal{L}^{TE} is defined as

$$\mathcal{L}^{\text{TE}} \mathbf{u}(t, \mathbf{x}) = \mathcal{A}^{\text{TE}} \mathbf{u}(t, \mathbf{x}) + \mathcal{B}^{\text{TE}} \int_0^t \mathbf{u}(s, \mathbf{x}) ds + \int_0^t \mathcal{C}^{\text{TE}}(t-s) \mathbf{u}(s, \mathbf{x}) ds, \quad (40)$$

and \mathbf{M}^{TE} is the two-dimensional curl operator

$$\begin{pmatrix} 0 & -\partial_{x_3} & 0 \\ \partial_{x_3} & 0 & -\partial_{x_1} \\ 0 & \partial_{x_1} & 0 \end{pmatrix}, \quad (41)$$

and $\mathbf{J}_s^{\text{TE}}(t) = -J_s \mathbf{e}_1$, $\mathbf{e}_1 = [1, 0]^T$.

6 Models for Polarization

The constitutive law in (8) is sufficiently general to include models based on differential equations and systems of differential equations or delay differential equations whose solutions can be expressed through fundamental solutions (in general variation-of-parameters representations)—see [2] for details. A number of known polarization laws can be readily treated.

1. The choice of the kernel function

$$\nu(t, \mathbf{x}) = \frac{\epsilon_0(\epsilon_s(\mathbf{x}) - \epsilon_\infty(\mathbf{x}))}{\tau(\mathbf{x})} e^{-t/\tau(\mathbf{x})}, \quad \mathbf{x} \in X_0 \times Z_D \quad (42)$$

in the dielectric $X_0 \times Z_D$ (see Figure 2) corresponds to the differential equation of the *Debye Model* for *orientational* or *dipolar polarization* given by

$$\tau \mathbf{P}_t + \mathbf{P} = \epsilon_0(\epsilon_s - \epsilon_\infty)\mathbf{E}, \quad (43)$$

$$\mathbf{D}(t, \mathbf{x}) = \epsilon_0\epsilon_\infty(t, \mathbf{x})\mathbf{E}(t, \mathbf{x}) + \mathbf{P}(t, \mathbf{x}). \quad (44)$$

Here ϵ_s is the static relative permittivity. The presence of instantaneous polarization is accounted for in this case by the coefficient ϵ_∞ in the electric flux equation. That is, $\epsilon_r = \epsilon_\infty$ in $X_0 \times Z_D$, $\epsilon_r = 1$ in air. The remainder of the electric polarization is seen to be a decaying exponential with relaxation parameter τ , driven by the electric field, less the part included in the instantaneous polarization. This model was first proposed by Debye [11] to model the behaviour of materials that possess permanent dipole moments. The magnitude of the polarization term \mathbf{P} represents the degree of alignment of these individual moments. The choice of coefficients in (43) gives a physical interpretation to ϵ_s and ϵ_∞ as the relative permittivities of the medium in the limit of the static field and very high frequencies, respectively. In the static case, we have $\mathbf{P}_t = 0$, so that $\mathbf{P} = \epsilon_0(\epsilon_s - \epsilon_\infty)\mathbf{E}$ and $\mathbf{D} = \epsilon_s\epsilon_0\mathbf{E}$. For very high frequencies, $\tau\mathbf{P}_t$ dominates \mathbf{P} so that $\mathbf{P} \approx 0$ and $\mathbf{D} = \epsilon_\infty\epsilon_0\mathbf{E}$.

2. The general model also includes the *Lorentz model* for *electronic* polarization which, in differential form, is represented with the second order equation:

$$\mathbf{P}_{tt} + \frac{1}{\tau}\mathbf{P}_t + \omega_0^2\mathbf{P} = \epsilon_0\omega_p^2\mathbf{E}, \quad (45)$$

$$\mathbf{D} = \epsilon_\infty\epsilon_0\mathbf{E} + \mathbf{P}. \quad (46)$$

In (45), ω_p is called the *plasma frequency* and is defined to be

$$\omega_p = \omega_0\sqrt{\epsilon_s - \epsilon_\infty}, \quad (47)$$

where ω_0 is the resonance frequency. A simple variation of constants solution yields the correct kernel function

$$\nu(t) = \frac{\epsilon_0\omega_p^2}{\nu_0}e^{-t/2\tau}\sin(\nu_0 t), \quad (48)$$

$$\nu_0 = \sqrt{\omega_0^2 - \frac{1}{4\tau^2}}. \quad (49)$$

3. For more complex dielectric materials, a simple Debye or Lorentz polarization model is often not adequate to characterize the dispersive behaviour of the material. One can then turn to combinations of Debye, Lorentz, or even more general n th order mechanisms as well as Cole-Cole type (fractional order derivatives) models. We again refer the reader to [2, 10] for details.

7 Numerical Discretization for the Cell Problem

7.1 Spatial discretization via finite elements

We divide the reference cell Y into elementary rectangles, and consider \mathcal{T}_h to be a uniform mesh with elements $\{K\}$ of edge length h . We define the finite dimensional space

$$\mathbf{V}_h = \{v_h \mid v_h \in C^0(\bar{Y}), v_h|_K \in Q_1 \text{ for all } K \in \mathcal{T}_h\}, \quad (50)$$

which approximates $H^1(Y)$. In (50), the space Q_1 is defined as $Q_1 = P_{11}$, where, for $k_1, k_2 \in \mathbb{N} \cup \{0\}$

$$P_{k_1 k_2} = \{p(x_1, x_2) \mid p(x_1, x_2) = \sum_{0 \leq i \leq k_1} \sum_{0 \leq j \leq k_2} a_{ij} x_1^i x_2^j, a_{ij} \in \mathbb{R}\}. \quad (51)$$

Thus, P_{11} is the space of the bilinear functions of two variables, and \mathbf{V}_h is the space of continuous piecewise bilinear functions. We now consider the subspace

$$\mathbf{V}_{\text{per},h} = \{v_h \mid v_h \in \mathbf{V}_h, v_h(y_1, 0) = v_h(y_1, 1) \text{ and } v_h(0, y_2) = v_h(1, y_2) \text{ for all } y_1, y_2 \in [0, 1]\}, \quad (52)$$

of \mathbf{V}_h , where $Y = [0, 1] \times [0, 1]$.

Thus the spatially discrete problem is to find $\bar{w}_{k,h}^A \in \mathbf{V}_{\text{per},h}$, $\bar{w}_{k,h} \in W^{1,1}(0, T; \mathbf{V}_{\text{per},h})$ and $\bar{w}_{k,h}^0 \in W^{2,1}(0, T; \mathbf{V}_{\text{per},h})$, solutions to the corrector equations

$$\left(\begin{array}{l} \text{(i)} \quad \epsilon_0 \int_Y \epsilon_{r,h}(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y} = -\epsilon_0 \int_Y \epsilon_{r,h}(\mathbf{y}) \mathbf{e}_k \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y}, \\ \text{(ii)} \quad \epsilon_0 \int_Y \epsilon_{r,h}(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}(t, \mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y} \\ \quad + \int_Y \int_0^t \{\sigma_h(\mathbf{y}) + \nu_h(t-s, \mathbf{y})\} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}(s, \mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, ds \, d\mathbf{y} \\ \quad = - \int_Y \{\sigma_h(\mathbf{y}) + \nu_h(t, \mathbf{y})\} \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A \right\} \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y}, \\ \text{(iii)} \quad \epsilon_0 \int_Y \epsilon_{r,h}(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^0(t, \mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y} \\ \quad + \int_Y \int_0^t \{\sigma_h(\mathbf{y}) + \nu_h(t-s, \mathbf{y})\} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^0(s, \mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, ds \, d\mathbf{y} \\ \quad = -\epsilon_0 \int_Y \epsilon_{r,h}(\mathbf{y}) \mathbf{e}_k \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y}, \end{array} \right. \quad (53)$$

for all $\bar{v}_h \in \mathbf{V}_{\text{per},h}$ and $k = 1, 2$, and $\mathbf{e}_1 = [1, 0]^T$, $\mathbf{e}_2 = [0, 1]^T$, and $\nabla_{\mathbf{y}}^h$ is a discrete approximation to the gradient. In (53), we have used the definitions (32) of the matrices $\mathbf{A}_{11}^{\text{TE}}$, $\mathbf{B}_{11}^{\text{TE}}$ and $\mathbf{C}_{11}^{\text{TE}}$.

7.2 Time discretization via a recursive convolution approach

Since the susceptibility kernel $\nu(t, \mathbf{x})$, is exponential in nature for many materials of interest, we can use recursion to compute the discretized time convolution of the susceptibility kernel with the

electric field in the corrector subproblems, e.g., (53,ii) and (53,iii). A similar approach, known as the recursive convolution (RC) method has been used to compute the discrete convolution terms that appear in Maxwell's equations [18, 15]. Let $t_n = n\Delta t$, for some timestep Δt and let \mathbf{V}^n denote the time component $\mathbf{V}(n\Delta t)$ for any vector field \mathbf{V} . Thus, the equation (53,ii) can be discretized in time as

$$\begin{aligned} & \epsilon_0 \int_Y \epsilon_{r,h}(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} \\ & + \int_Y \int_0^{n\Delta t} \{\sigma_h(\mathbf{y}) + \nu_h(n\Delta t - s, \mathbf{y})\} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}(s, \mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) ds d\mathbf{y} \\ & = - \int_Y \{\sigma_h(\mathbf{y}) + \nu_h(n\Delta t, \mathbf{y})\} \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A \right\} \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y}. \end{aligned} \quad (54)$$

We will assume that all field components are constant over each time interval of length Δt . Thus, assuming that all field vectors are zero for $t < 0$ we have

$$\begin{aligned} & \epsilon_0 \int_Y \epsilon_{r,h}(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} \\ & + \int_Y \sum_{m=0}^{n-1} \int_{m\Delta t}^{(m+1)\Delta t} \{\sigma_h(\mathbf{y}) + \nu_h(n\Delta t - s, \mathbf{y})\} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) ds d\mathbf{y} \\ & = - \int_Y \{\sigma_h(\mathbf{y}) + \nu_h(n\Delta t, \mathbf{y})\} \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A \right\} \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y}. \end{aligned} \quad (55)$$

Interchanging the space and time integral terms and rearranging we obtain

$$\begin{aligned} & \epsilon_0 \int_Y \epsilon_{r,h}(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} \\ & + \sum_{m=0}^{n-1} \int_Y \left(\int_{m\Delta t}^{(m+1)\Delta t} \nu_h(n\Delta t - s, \mathbf{y}) ds \right) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} \\ & + \sum_{m=0}^{n-1} \int_Y \sigma_h(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} \\ & = - \int_Y \{\sigma_h(\mathbf{y}) + \nu_h(n\Delta t, \mathbf{y})\} \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A \right\} \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y}. \end{aligned} \quad (56)$$

Let us define

$$\nu_m^n(\mathbf{y}) = \int_{m\Delta t}^{(m+1)\Delta t} \nu_h(n\Delta t - s, \mathbf{y}) ds. \quad (57)$$

Using definition (57) in (56) we have

$$\begin{aligned} & \epsilon_0 \int_Y \epsilon_{r,h}(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} + \sum_{m=0}^{n-1} \int_Y \nu_m^n \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} \\ & + \sum_{m=0}^{n-1} \Delta t \int_Y \sigma_h(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} \\ & = - \int_Y \{\sigma_h(\mathbf{y}) + \nu_h(n\Delta t, \mathbf{y})\} \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A \right\} \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y}. \end{aligned} \quad (58)$$

Upon rearranging, we have

$$\begin{aligned}
& \epsilon_0 \int_Y \{ \epsilon_{r,h}(\mathbf{y}) + \Delta t \sigma_h(\mathbf{y}) + \nu_{n-1}^n \} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} \\
& + \sum_{m=0}^{n-2} \int_Y \nu_m^n \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} + \sum_{m=0}^{n-2} \Delta t \int_Y \sigma_h(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} \quad (59) \\
& = - \int_Y \{ \sigma_h(\mathbf{y}) + \nu_h(n\Delta t, \mathbf{y}) \} \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A \right\} \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y}.
\end{aligned}$$

In equation (59) all the terms can be computed with the knowledge of just the solution at time $t_n = n\Delta t$, i.e., $\nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n$, except the discrete convolution term involving ν_m^n for $m = 0, \dots, n-2$. We now show that the discrete convolution of all previous $\nabla_{\mathbf{y}}^h \bar{w}_{k,h}^m$ field values and the discrete susceptibility function can be reduced to recursive updating of a single vector on each element in the finite element mesh, which involves a matrix vector multiplication at each time step.

7.3 Recursive convolution for Debye polarization

We consider the case of Debye polarization in the remainder of this paper. In this case the spatially discrete susceptibility function is defined as

$$\nu_h(t, \mathbf{y}) = \frac{\epsilon_0(\epsilon_{s,h}(\mathbf{y}) - \epsilon_{\infty,h}(\mathbf{y}))}{\tau_h(\mathbf{y})} e^{-t/\tau_h(\mathbf{y})}, \quad \mathbf{y} \in Y, t > 0, \quad (60)$$

and from (57) and (60) the function ν_m^n is defined as

$$\nu_m^n(\mathbf{y}) = \int_{m\Delta t}^{(m+1)\Delta t} \frac{\epsilon_0(\epsilon_{s,h}(\mathbf{y}) - \epsilon_{\infty,h}(\mathbf{y}))}{\tau_h(\mathbf{y})} e^{-(t_n-s)/\tau_h(\mathbf{y})} ds, \quad \mathbf{y} \in Y. \quad (61)$$

From (59) we define the summation

$$\Psi_K^n = \sum_{m=0}^{n-2} \int_K \nu_m^n \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y}, \quad K \in \mathcal{T}_h, \quad (62)$$

where ν_m^n is defined in (61). Equation (62) constitutes the time discrete convolution of the susceptibility function ν_h and all field values of $\nabla_{\mathbf{y}}^h \bar{w}_{k,h}^m$ up to the n th time step.

Then for the $n+1$ st step we find

$$\begin{aligned}
\Psi_K^{n+1} &= \sum_{m=0}^{n-1} \int_K \nu_m^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} \\
&= \sum_{m=0}^{n-2} \int_K \nu_m^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y} + \int_K \nu_{n-1}^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) d\mathbf{y}.
\end{aligned} \quad (63)$$

From (61) we can derive the identity

$$\nu_m^{n+1}(\mathbf{y}) = \nu_m^n(\mathbf{y}) e^{-\Delta t/\tau_h(\mathbf{y})}. \quad (64)$$

Using (64) in (63) we obtain

$$\begin{aligned}\Psi_K^{n+1} &= \sum_{m=0}^{n-2} \int_K \nu_m^n e^{-\Delta t/\tau_h(\mathbf{y})} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y} \\ &\quad + \int_K \nu_{n-1}^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y}.\end{aligned}\tag{65}$$

In the spatial discretization of the reference cell Y , we will assume that the functions spatially dependent $\tau_h(\mathbf{y})$, $\epsilon_{s,h}(\mathbf{y})$, $\epsilon_{\infty,h}(\mathbf{y})$, and $\sigma_h(\mathbf{y})$ are constant on each element K of the finite element mesh \mathcal{T}_h . With this assumption, for every element $K \in \mathcal{T}_h$, we have

$$\begin{aligned}\Psi_K^{n+1} &= e^{-\Delta t/\tau_h(K)} \sum_{m=0}^{n-2} \int_K \nu_m^n \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y} \\ &\quad + \int_K \nu_{n-1}^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y}.\end{aligned}\tag{66}$$

Combining (66) and (62), we obtain the recursion

$$\Psi_K^{n+1} = e^{-\Delta t/\tau_h(K)} \Psi_K^n + \int_K \nu_{n-1}^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y},\tag{67}$$

where $\nu_{n-1}^{n+1}(K)$, from the definition (61), can be calculated to be

$$\nu_{n-1}^{n+1}(K) = \epsilon_0(\epsilon_{s,h}(K) - \epsilon_{\infty,h}(K)) e^{(-2n\Delta t)/\tau_h(K)} (e^{\Delta t/\tau_h(K)} - 1).\tag{68}$$

The finite element function Ψ_h^n which is defined to be Ψ_K^n on element K in the triangulation \mathcal{T}_h can be calculated as

$$\begin{aligned}\Psi_h^{n+1} &= \sum_{m=0}^{n-1} \int_Y \nu_m^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y}, \\ &= \sum_{K \in \mathcal{T}_h} \sum_{m=0}^{n-1} \int_K \nu_m^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y}, \\ &= \sum_{K \in \mathcal{T}_h} \left(e^{-\Delta t/\tau_h(K)} \Psi_K^n + \int_K \nu_{n-1}^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y} \right).\end{aligned}\tag{69}$$

Thus we have the recursion

$$\Psi_h^{n+1} = \sum_{K \in \mathcal{T}_h} e^{-\Delta t/\tau_h(K)} \Psi_K^n + \int_Y \nu_{n-1}^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^n(\mathbf{y}) \cdot \nabla_{\mathbf{y}}^h \bar{v}_h(\mathbf{y}) \, d\mathbf{y}.\tag{70}$$

We note that given the value Ψ_K^n for every element $K \in \mathcal{T}_h$ at $t_n = n\Delta t$, we can calculate Ψ_h^{n+1} by equation (70) which only requires the knowledge of the solution vector $\bar{w}_{k,h}^n$ at the n th step! The solutions $\bar{w}_{k,h}^m$ for $m < n$ are not needed to be stored in memory for this calculation. We can treat the discrete convolution in (53,iii) in the same manner as above.

The calculation of the discrete homogenized matrix \mathcal{C}_h^{TE} also requires the computation of a time integral which at time $t_n = n\Delta t$ requires the knowledge of the solutions $\bar{w}_{k,h}^m$ at previous times $m\Delta t, m = 1, \dots, n$. Thus, we again try to obtain a recursive formula that will aid in the computation of these time integrals. Once we have solved for the discrete corrector terms, we can then construct the discrete homogenized matrices from

$$\left\{ \begin{array}{l} \text{(i)} \quad (\mathcal{A}_{11,h}^{\text{TE}})_k = \int_Y \mathbf{A}_{11,h}^{\text{TE}}(\mathbf{y}) \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A(\mathbf{y}) \right\} d\mathbf{y}, \\ \text{(ii)} \quad (\mathcal{B}_{11,h}^{\text{TE}})_k = \int_Y \mathbf{B}_{11,h}^{\text{TE}}(\mathbf{y}) \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A(\mathbf{y}) \right\} d\mathbf{y}, \\ \text{(iii)} \quad (\mathcal{C}_{11,h}^{\text{TE}})_k(t) = \int_Y \mathbf{C}_{11,h}^{\text{TE}}(t, \mathbf{y}) \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A(\mathbf{y}) \right\} d\mathbf{y} + \int_Y \mathbf{A}_{11,h}^{\text{TE}}(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}(t, \mathbf{y}) d\mathbf{y} \\ \quad + \int_Y \int_0^t \left\{ \mathbf{B}_{11,h}^{\text{TE}}(\mathbf{y}) + \mathbf{C}_{11,h}^{\text{TE}}(t-s, \mathbf{y}) \right\} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}(s, \mathbf{y}) ds d\mathbf{y}, \end{array} \right. \quad (71)$$

where, as in (37), $\mathbf{e}_k, k = 1, 2$ are the basis vectors in \mathbb{R}^2 , $(\mathcal{A}_{11,h}^{\text{TE}})_k, (\mathcal{B}_{11,h}^{\text{TE}})_k$, and $(\mathcal{C}_{11,h}^{\text{TE}})_k$ are the k th columns of the discretized matrices $\mathcal{A}_{11,h}^{\text{TE}}, \mathcal{B}_{11,h}^{\text{TE}}$, and $\mathcal{C}_{11,h}^{\text{TE}}$, respectively, and the discrete homogenized matrices are calculated as

$$\mathcal{A}_h^{\text{TE}} = \begin{bmatrix} \mathcal{A}_{11,h}^{\text{TE}} & 0 \\ 0 & \mu_0 \end{bmatrix}; \quad \mathcal{B}_h^{\text{TE}} = \begin{bmatrix} \mathcal{B}_{11,h}^{\text{TE}} & 0 \\ 0 & 0 \end{bmatrix}; \quad \mathcal{C}_h^{\text{TE}} = \begin{bmatrix} \mathcal{C}_{11,h}^{\text{TE}} & 0 \\ 0 & 0 \end{bmatrix}. \quad (72)$$

Using the definitions of the discrete matrices we rewrite (71) as

$$\left\{ \begin{array}{l} \text{(i)} \quad (\mathcal{A}_{11,h}^{\text{TE}})_k = \epsilon_0 \int_Y \epsilon_{\infty,h}(\mathbf{y}) \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A(\mathbf{y}) \right\} d\mathbf{y}, \\ \text{(ii)} \quad (\mathcal{B}_{11,h}^{\text{TE}})_k = \int_Y \sigma_h(\mathbf{y}) \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A(\mathbf{y}) \right\} d\mathbf{y}, \\ \text{(iii)} \quad (\mathcal{C}_{11,h}^{\text{TE}})_k(t_n) = \int_Y \nu_h(t, \mathbf{y}) \left\{ \mathbf{e}_k + \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^A(\mathbf{y}) \right\} d\mathbf{y} + \epsilon_0 \int_Y \epsilon_{\infty,h}(\mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}(t, \mathbf{y}) d\mathbf{y} \\ \quad + \int_Y \int_0^{n\Delta t} \left\{ \sigma_h(\mathbf{y}) + \nu_h(n\Delta t - s, \mathbf{y}) \right\} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}(s, \mathbf{y}) ds d\mathbf{y}. \end{array} \right. \quad (73)$$

The term

$$\int_Y \int_0^{n\Delta t} \nu_h(n\Delta t - s, \mathbf{y}) \nabla_{\mathbf{y}}^h \bar{w}_{k,h}(s, \mathbf{y}) ds d\mathbf{y}, \quad (74)$$

in equation (73,iii) can be computed using a recursive formula that does not require knowledge of the solutions $\bar{w}_{k,h}^m$ at times $m\Delta t, m = 1, \dots, n$. To derive a recursive formula, which again need to use properties of the susceptibility function ν_h . We define the term in (74) to be

$$T_h^n = \sum_{m=0}^{n-1} \int_Y \nu_m^n \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(s, \mathbf{y}) d\mathbf{y}. \quad (75)$$

where ν_m^n is defined in (61). We then have

$$\begin{aligned} T_h^{n+1} &= \sum_{m=0}^n \int_Y \nu_m^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(s, \mathbf{y}) d\mathbf{y} \\ &= \sum_{m=0}^{n-1} \int_Y \nu_m^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(s, \mathbf{y}) d\mathbf{y} + \int_Y \nu_n^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{n+1}(s, \mathbf{y}) d\mathbf{y}. \end{aligned} \quad (76)$$

Using (64) in (76) we obtain the recursion

$$\begin{aligned} T_h^{n+1} &= \sum_{m=0}^{n-1} \int_Y e^{-\Delta t/\tau_h(\mathbf{y})} \nu_m^n \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(s, \mathbf{y}) \, d\mathbf{y} + \int_Y \nu_n^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{n+1}(s, \mathbf{y}) \, d\mathbf{y} \\ &= \sum_{K \in \mathcal{T}_h} e^{-\Delta t/\tau(K)} T_h^n(K) + \int_Y \nu_n^{n+1} \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{n+1}(s, \mathbf{y}) \, d\mathbf{y}, \end{aligned} \quad (77)$$

where

$$T_h^n(K) = \sum_{m=0}^{n-1} \int_K \nu_m^n \nabla_{\mathbf{y}}^h \bar{w}_{k,h}^{m+1}(s, \mathbf{y}) \, d\mathbf{y}. \quad (78)$$

8 Numerical Examples

In this section we present results of initial computations based on the above discretization schemes that we have carried out for both circular and square microstructures. In these computations we have used a conjugate gradient method to solve the resulting linear systems that arise after discretizing our model in space and time.

8.1 Examples with varying relative permittivity

In the following examples we choose the value of the relative permittivity to be

$$\epsilon_r(\mathbf{x}) = \epsilon_\infty(\mathbf{x}) = \begin{cases} \epsilon_i = 1.003, & \text{if } \mathbf{x} \in S, \\ \epsilon_e = 2.7, & \text{if } \mathbf{x} \in Y/\bar{S}, \end{cases} \quad (79)$$

where S is the microstructure that is enclosed inside the reference cell. These values are taken from experimental measurements for air and polyurethane material which are the primary components of the insulating foam described in the introduction.

8.1.1 Circular microstructures

We first consider the example of a composite material which possesses circular microstructures in two-dimensions. We will solve a cell problem in the reference cell $Y = [0, 1] \times [0, 1]$, in which the relative permittivity is given in equation (79) where S is the circular microstructure enclosed inside the reference cell, as depicted in Figure 3. In this test case we will assume that $\epsilon_s = \epsilon_\infty$, $\sigma = 0$ and $\tau = 3.16 \times 10^{-8}$. Hence the parameters σ and τ are constant over the entire dielectric material. Since $\epsilon_\infty = \epsilon_s$, we have that $\nu(t) = 0$ for all time t . Thus in this example, the model possesses instant polarization but does not have a hysteretic term in the polarization. Our numerical simulation is performed on a 51×51 nodes mesh grid. We define the inclusion volume fraction f as the ratio

$$f = \frac{\text{area of inclusion}}{\text{area of domain } Y} = \text{area of inclusion}. \quad (80)$$

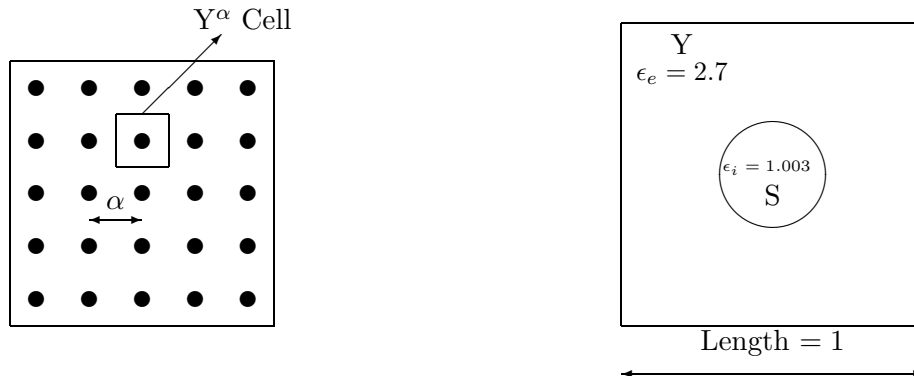


Figure 3: (left) Periodic composite material presenting a circular microstructure with periodicity α . (right) The reference cell $Y = [0, 1] \times [0, 1]$.

For $f = 0.5$ the homogenized matrix \mathcal{A}_{11}^{TE} is

$$\mathcal{A}_{11}^{TE} = \epsilon_0 \begin{bmatrix} 1.68009 & -2.92 \times 10^{-4} \\ -2.92 \times 10^{-4} & 1.68009 \end{bmatrix}, \quad (81)$$

and the homogenized matrices \mathcal{B}_{11}^{TE} and \mathcal{C}_{11}^{TE} are the zero matrices.

In the numerical solution for the homogenized model we have approximated the circular microstructure in a staircase fashion. In Table 1 we present the homogenized relative permittivities for different inclusion volume fractions f , and different refinements of the mesh grid that is imposed on the reference cell Y . Here h denotes the mesh step size. In this table we note that the lower and upper off-diagonal entries, $(\mathcal{A}_{11}^{TE})_{\text{lod}}$, and $(\mathcal{A}_{11}^{TE})_{\text{uod}}$, respectively, of the homogenized matrix \mathcal{A}_{11}^{TE} decrease at least as fast as $\mathcal{O}(h)$ as the mesh grid is refined. The error in the off-diagonal terms is probably due to the inaccurate representation of the circular microstructure by a staircase approximation. Hence \mathcal{A}_{11}^{TE} is approximately diagonal with identical diagonal entries. For the case of $f = 0.5$ we have

$$\mathcal{A}_{11}^{TE} \approx 1.68 \epsilon_0 \begin{bmatrix} 1.0 & 0 \\ 0 & 1.0 \end{bmatrix}. \quad (82)$$

Thus the homogenized or effective relative permittivity for $f = 0.5$ is $\epsilon_r = 1.68$.

In Figure 4 we plot the relative effective permittivity versus the inclusion volume fraction for the method discussed in this paper and compare it to different theoretical mixture models, namely the Maxwell-Garnett formula, the Bruggeman mixture rule, as well as the weighted average of the different relative permittivities. (We refer the reader to [22] for a discussion of the various theoretical mixing formulas.) We note how close our predicted values of the effective relative permittivity are to those predicted by these mixing rules. The prediction of the effective relative permittivity of the

f	f_{sc}	h	$(\mathcal{A}_{11}^{TE})_d$	$(\mathcal{A}_{11}^{TE})_{lod}$	$(\mathcal{A}_{11}^{TE})_{uod}$
0.7854	0.78	0.02	1.23068	5.92955×10^{-3}	5.92955×10^{-3}
	0.78	0.01	1.23186	1.86948×10^{-3}	1.86948×10^{-3}
	0.785	0.005	1.23347	1.16484×10^{-3}	1.16484×10^{-3}
	0.785	0.0025	1.23767	5.41779×10^{-4}	5.41779×10^{-4}
0.5027	0.502	0.02	1.68009	-2.92023×10^{-4}	-2.92023×10^{-4}
	0.502	0.01	1.68299	-7.31937×10^{-5}	-7.31937×10^{-5}
	0.502	0.005	1.68509	-1.94965×10^{-5}	-1.94965×10^{-5}
	0.502	0.0025	1.68523	-5.02621×10^{-6}	-5.02621×10^{-6}
0.7124	0.7124	0.02	1.34434	-8.69546×10^{-19}	-2.42287×10^{-17}
	0.7129	0.01	1.34789	-3.21656×10^{-18}	-1.73589×10^{-17}
	0.7124	0.005	1.35081	2.19197×10^{-16}	5.59808×10^{-16}
	0.7125	0.0025	1.35216	-1.96101×10^{-18}	5.536×10^{-16}
0.1257	0.1252	0.02	2.40606	-6.64092×10^{-4}	-6.64092×10^{-4}
	0.1253	0.01	2.40637	-1.72749×10^{-4}	-1.72749×10^{-4}
	0.1255	0.005	2.40609	-4.39996×10^{-5}	-4.39996×10^{-5}
	0.1255	0.0025	2.40626	-1.15057×10^{-5}	-1.15057×10^{-5}

Table 1: The effective relative permittivities for different volume fractions with a circular microstructure. The table lists the diagonal entries, $(\mathcal{A}_{11}^{TE})_d$, of the homogenized matrix \mathcal{A}_{11}^{TE} as well as the lower and upper off-diagonal entries $(\mathcal{A}_{11}^{TE})_{lod}$ and $(\mathcal{A}_{11}^{TE})_{uod}$, respectively, for different levels of refinement of the mesh grid on the reference cell Y . Here f is the volume fraction and f_{sc} is the computed volume fraction using the staircase approximation.

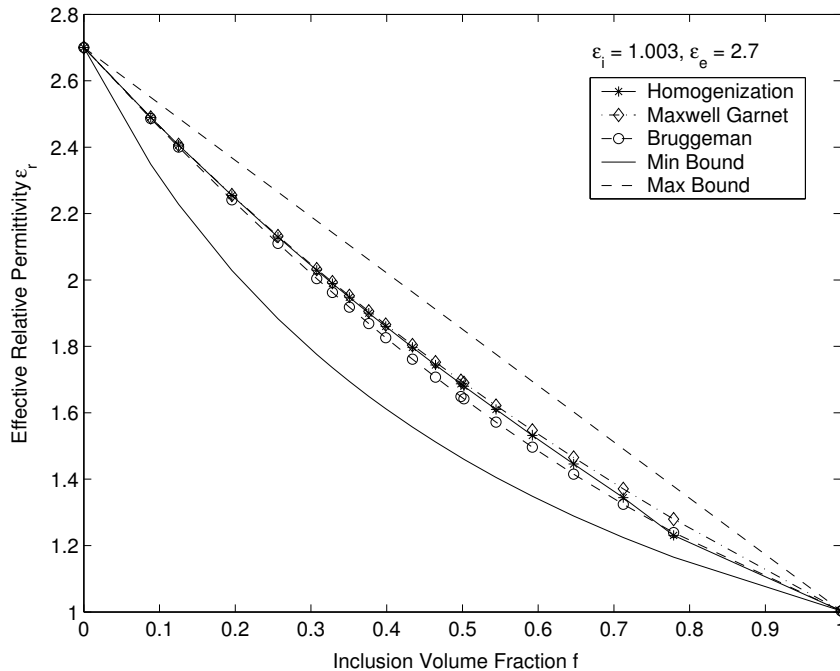


Figure 4: Effective relative permittivity distribution for random mixtures with circular microstructures.

composite mixture ϵ_{eff} according to these formulas is given by

$$\epsilon_{\text{eff}} = \epsilon_e + 2f\epsilon_e \frac{\epsilon_i - \epsilon_e}{\epsilon_i + \epsilon_e - f(\epsilon_i - \epsilon_e)} \quad (\text{Maxwell-Garnett}), \quad (83)$$

$$(1 - f) \frac{\epsilon_e - \epsilon_{\text{eff}}}{\epsilon_e + \epsilon_{\text{eff}}} + f \frac{\epsilon_i - \epsilon_{\text{eff}}}{\epsilon_i + \epsilon_{\text{eff}}} = 0 \quad (\text{Bruggeman}). \quad (84)$$

These formulas hold for randomly distributed circular inclusions of permittivity ϵ_i in a homogeneous environment of permittivity ϵ_e . The inclusions occupy a volume fraction f of the homogeneous medium. These different mixing models predict different effective permittivity values for a given mixture. There are also bounds that limit the range of the predictions. These bounds are the Weiner bounds given by

$$\epsilon_{\text{eff,max}} = f\epsilon_i + (1 - f)\epsilon_e \quad (\text{weighted average}), \quad (85)$$

$$\epsilon_{\text{eff,min}} = \frac{\epsilon_i \epsilon_e}{f\epsilon_e + (1 - f)\epsilon_i}. \quad (86)$$

These bounds hold for all values of ϵ_i and ϵ_e . Other mixing models such as power-law models, the Lichtenecker formula, etc., also exist in the literature. We again refer the reader to [23, 22] for further details. In Figure 5 we plot the solution vectors \bar{w}_1^A and \bar{w}_2^A , for $f = 0.5$, over the reference cell Y . In Figure 6 we plot a topview of the corresponding solution vectors in the domain Y .

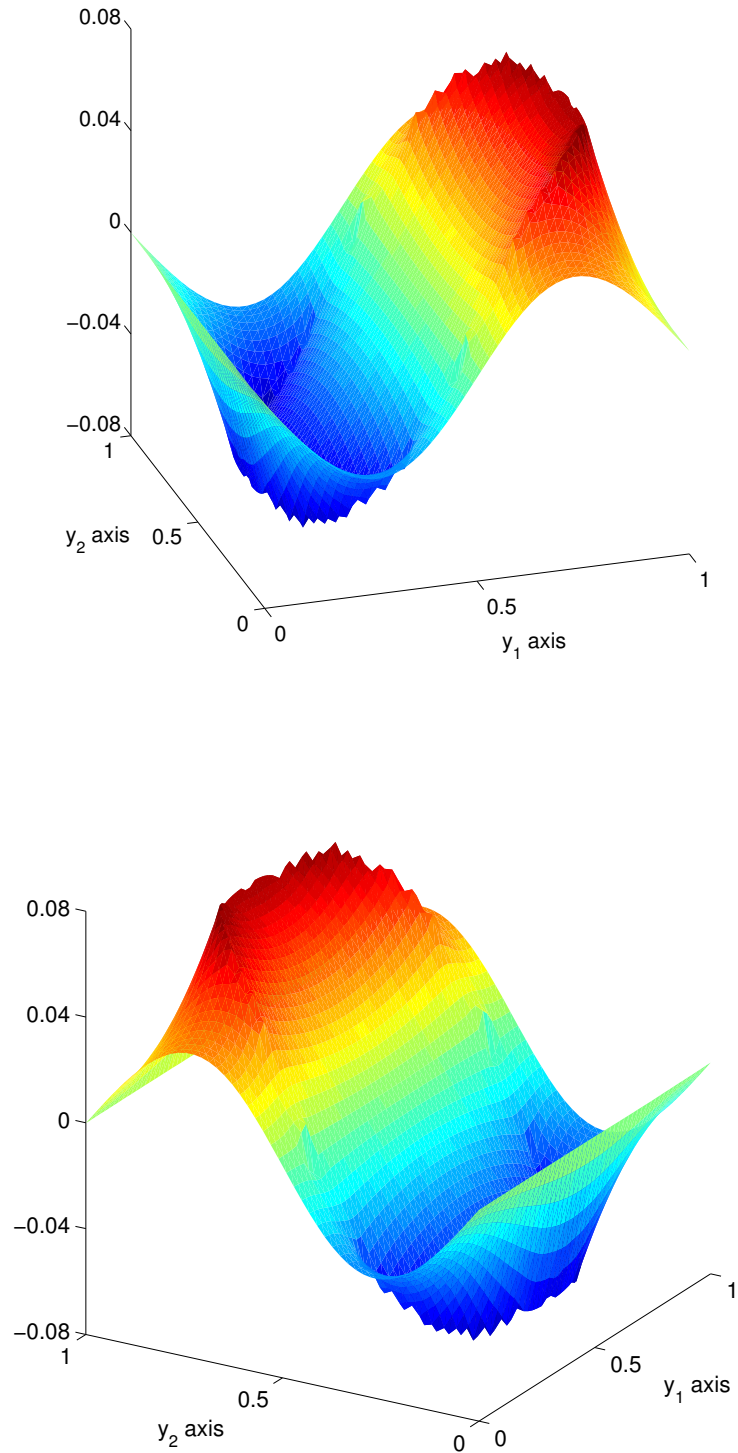


Figure 5: Plot of the solution vectors \bar{w}_1^A (top) and \bar{w}_2^A (bottom) for an inclusion fraction volume $f = 0.5$ on a 50×50 cells mesh grid. The homogenized value of ϵ_r for this case is ≈ 1.68

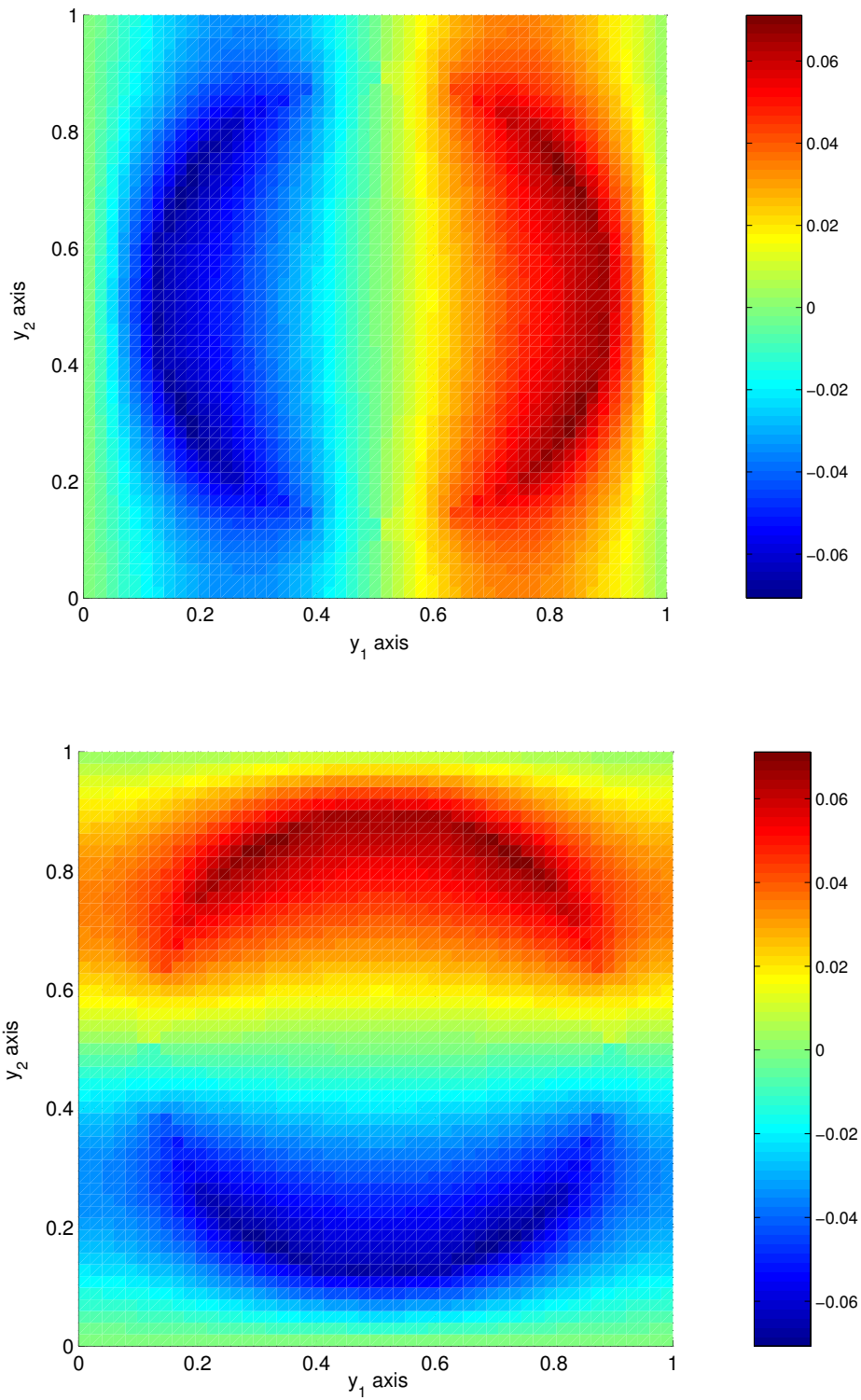


Figure 6: Top view of the solution vectors \bar{w}_1^A (top) and \bar{w}_2^A (bottom)

8.1.2 Square microstructures

We next replaced the circular microstructures in two-dimensions with square microstructures and repeated the experiment performed in Section 8.1.1. We solved a cell problem in the reference cell $Y = [0, 1] \times [0, 1]$ in which the value of the relative permittivity is given in (79) where S is now a square microstructure that is enclosed inside the reference cell. In this test case, as before, we will assume that $\epsilon_s = \epsilon_\infty$, $\sigma = 0$ and $\tau = 3.16 \times 10^{-8}$. Thus, the parameters σ and τ are constant over the entire dielectric material. Since $\epsilon_\infty = \epsilon_s$, we again have that $\nu(t) = 0$ for all time t . Our

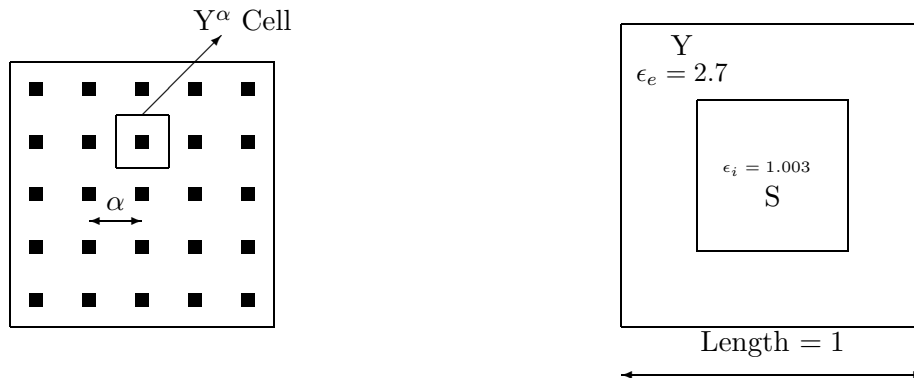


Figure 7: (left) Periodic composite material presenting a square microstructure with periodicity α . (right) The reference cell $Y = [0, 1] \times [0, 1]$.

numerical simulation is performed on a 51×51 nodes mesh grid. For $f = 0.52$ the homogenized matrix \mathcal{A}_{11}^{TE} is

$$\mathcal{A}_{11}^{TE} = \epsilon_0 \begin{bmatrix} 1.65057 & -3.04533 \times 10^{-18} \\ -1.8576 \times 10^{-18} & 1.65057 \end{bmatrix}, \quad (87)$$

and the homogenized matrices \mathcal{B}_{11}^{TE} and \mathcal{C}_{11}^{TE} are the zero matrices. In this case the sides of the square are aligned with the mesh grid on the reference cell Y , so we do not have the staircase error that was observed in the case of circular microstructures. Again \mathcal{A}_{11}^{TE} is approximately diagonal with identical diagonal entries. For the case of $f = 0.52$ we have

$$\mathcal{A}_{11}^{TE} \approx 1.65 \epsilon_0 \begin{bmatrix} 1.0 & 0 \\ 0 & 1.0 \end{bmatrix}. \quad (88)$$

Thus the homogenized or effective relative permittivity for $f = 0.52$ is $\epsilon_r = 1.65$. In Figure 8 we plot the effective relative permittivity for different volume fractions and compare these values to those obtained for circular microstructures. We note here how close the effective relative permittivity values are for circular and square microstructures of the same volume fraction. In Figure 9 we plot the solution vectors \bar{w}_1^A and \bar{w}_2^A , for $f = 0.52$, over the reference cell Y . In Figure 10 we plot a topview of the corresponding solution vectors in the domain Y .

8.2 Example with varying relaxation times

8.2.1 Circular microstructures

We now consider an example of a composite material which possesses circular microstructures in which the value of the relaxation time τ is given to be

$$\tau(\mathbf{x}) = \begin{cases} \tau_i = 1.58 \times 10^{-8}, & \text{if } \mathbf{x} \in S, \\ \tau_e = 3.16 \times 10^{-8}, & \text{if } \mathbf{x} \in Y/\bar{S}. \end{cases} \quad (89)$$

In this test case, we will assume that the other parameters namely, ϵ_s , σ and ϵ_∞ are constant over the entire dielectric material with $\epsilon_\infty = 5.5$, $\epsilon_s = 78.2$ and $\sigma = 1.0 \times 10^{-5}$.

In Figure 11 we plot the results of the cell problem. This figure plots the effective dielectric response function (DRF) as a function of time. We compare the homogenized DRF with four different cases. The low frequency case corresponds to one in which the relaxation time τ is given to be the weighted average of τ_i and τ_e

$$\tau_{\text{low}} = f\tau_i + (1 - f)\tau_e. \quad (90)$$

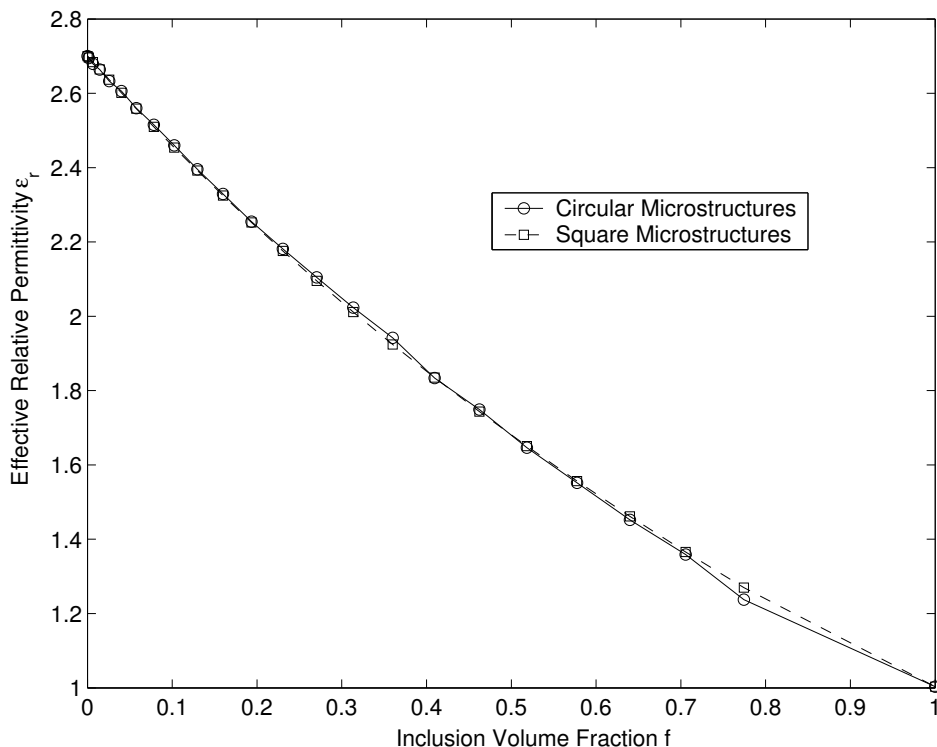


Figure 8: Comparison of the effective relative permittivity distribution for square and circular microstructures

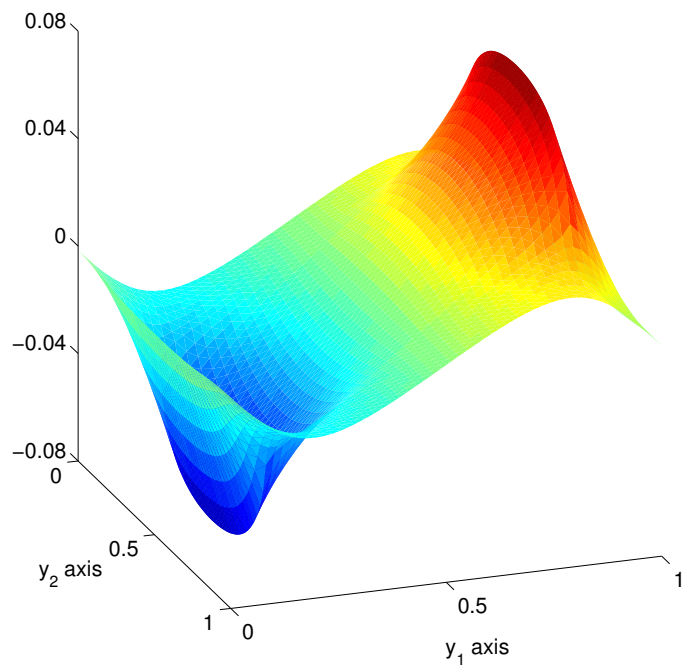
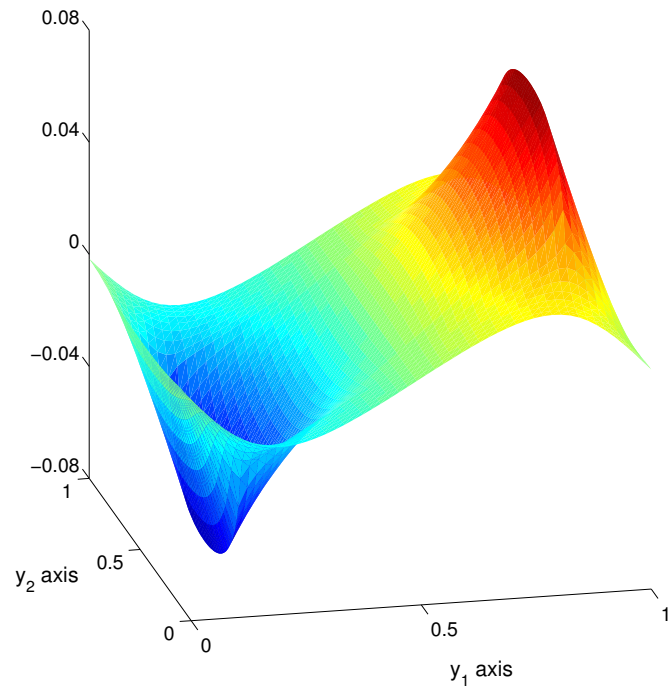


Figure 9: Plot of the solution vectors \bar{w}_1^A (top) and \bar{w}_2^A (bottom) for a square inclusion fraction volume $f = 0.52$ on a 50×50 cells mesh grid. The homogenized value of ϵ_r for this case is ≈ 1.65

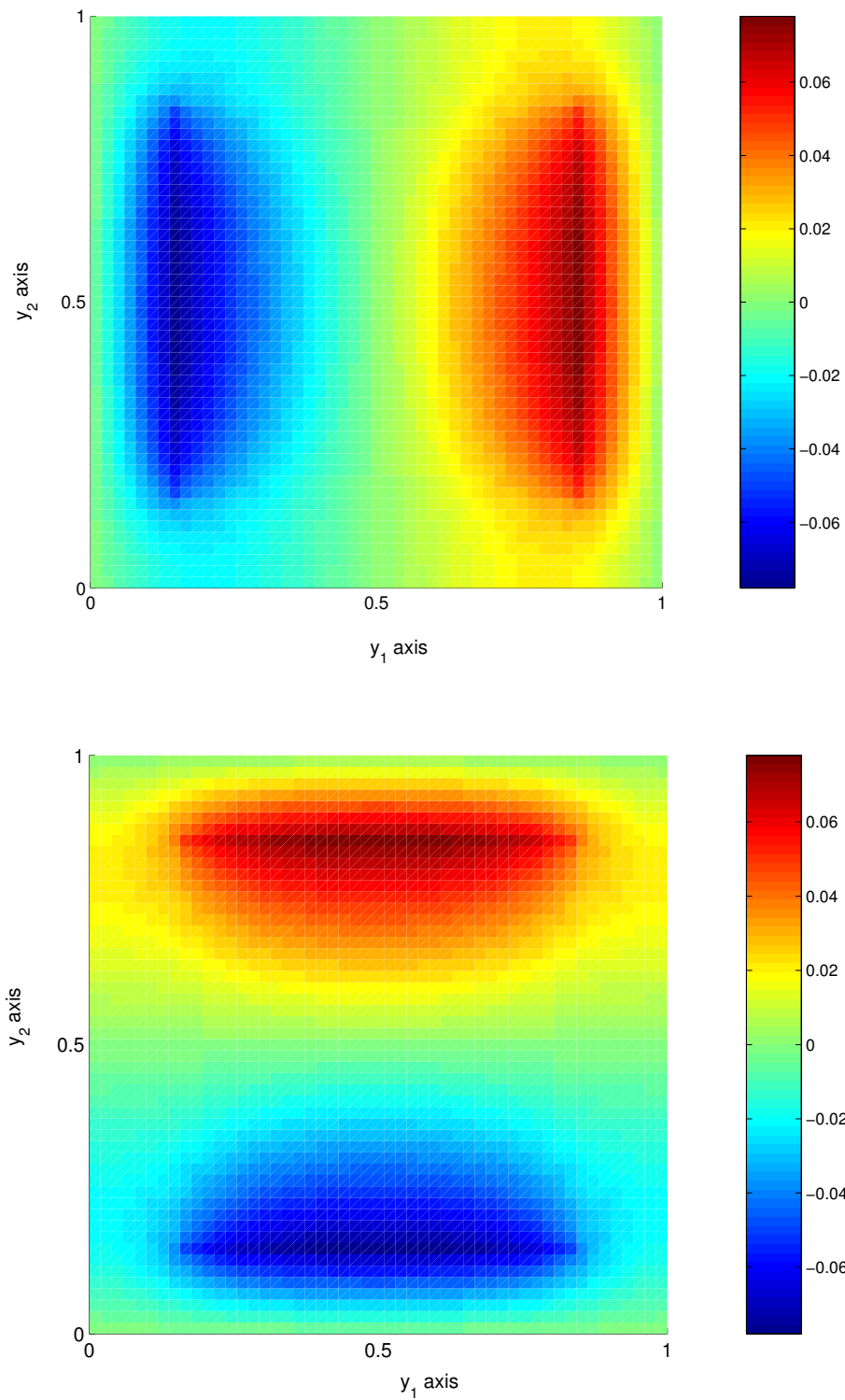


Figure 10: Top view of the solution vectors \bar{w}_1^A (top) and \bar{w}_2^A (bottom) for square microstructures.

The high frequency case corresponds to the relaxation time

$$\tau_{\text{high}} = \left(\frac{f}{\tau_i} + \frac{1-f}{\tau_e} \right)^{-1}. \quad (91)$$

The low and high frequency averages for the relaxation time τ were observed in numerical experiments that were performed in [3], in which a probabilistic approach is taken to an electromagnetic interrogation problem of a dielectric medium that is a mixture of two Debye media with different relaxation times τ . The low frequency average corresponds to the average relaxation time that is observed when the composite media is interrogated by electromagnetic pulses at frequencies of 10^6 Hz, whereas the high frequency average was observed at higher frequencies of $10^9 - 10^{11}$ Hz (frequencies are considered low here if the corresponding angular frequency $\omega < 1/\tau$). In Figure 11 we also compare the homogenized DRF to the weighted average of the corresponding DRF's, i.e.,

$$\nu_{\text{wa}}(t) = f\nu_i + (1-f)\nu_e, \quad (92)$$

where

$$\begin{aligned} \text{(i)} \quad \nu_i(t) &= \frac{\epsilon_0(\epsilon_s - \epsilon_\infty)}{\tau_i} e^{-t/\tau_i}, \\ \text{(ii)} \quad \nu_e(t) &= \frac{\epsilon_0(\epsilon_s - \epsilon_\infty)}{\tau_e} e^{-t/\tau_e}, \end{aligned} \quad (93)$$

as well as to another average that we call the inverse weighted average, which is the inverse of the weighted averages of the inverses of the two DRF's, i.e.,

$$\nu_{\text{Iwa}}(t) = \left(\frac{f}{\nu_i} + \frac{(1-f)}{\nu_e} \right)^{-1}, \quad (94)$$

where ν_i and ν_e are defined in (93,ii), and (93,iii), respectively. Clearly Figure 11 demonstrates good agreement between the homogenized DRF and the other approximations. Finally in Figure 12 we plot the homogenized DRF as a function of the inclusion volume fraction at times $t = 0.556$ ns and $t = 27.78$ ns. For $t \ll \min(\tau_i, \tau_e)$ one can show that $\nu_{\text{Iwa}}(t) \approx \nu(t; \tau_{\text{low}})$ as is clearly seen in the top plot of this figure. Similarly we have $\nu_{\text{wa}}(t) \approx \nu(t; \tau_{\text{high}})$. We observe that for all values of f the plot of the homogenized DRF lies between the other approximations for both small and large times.

8.2.2 Square microstructures

We next repeated the above numerical calculation for square microstructures in which the value of the relaxation time τ is given to be

$$\tau(\mathbf{x}) = \begin{cases} \tau_i = 1.58 \times 10^{-8}, & \text{if } \mathbf{x} \in S, \\ \tau_e = 3.16 \times 10^{-8}, & \text{if } \mathbf{x} \in Y/\bar{S}. \end{cases} \quad (95)$$

As before we have assumed that the other parameters namely, ϵ_s , σ and ϵ_∞ are constant over the entire dielectric material with $\epsilon_\infty = 5.5$, $\epsilon_s = 78.2$ and $\sigma = 1.0 \times 10^{-5}$. In Figure 13 we plot the homogenized DRF as a function of the inclusion volume fraction at time $t = 0.556$ ns (top) and time $t = 27.78$ ns (bottom). In Figure 14 we plot the homogenized DRF as a function of time, for $f = 0.52$. In both Figures 13 and 14 we observe that the difference between the values of the DRF for square versus circular microstructures is of the order of $10^{-4} - 10^{-5}$.

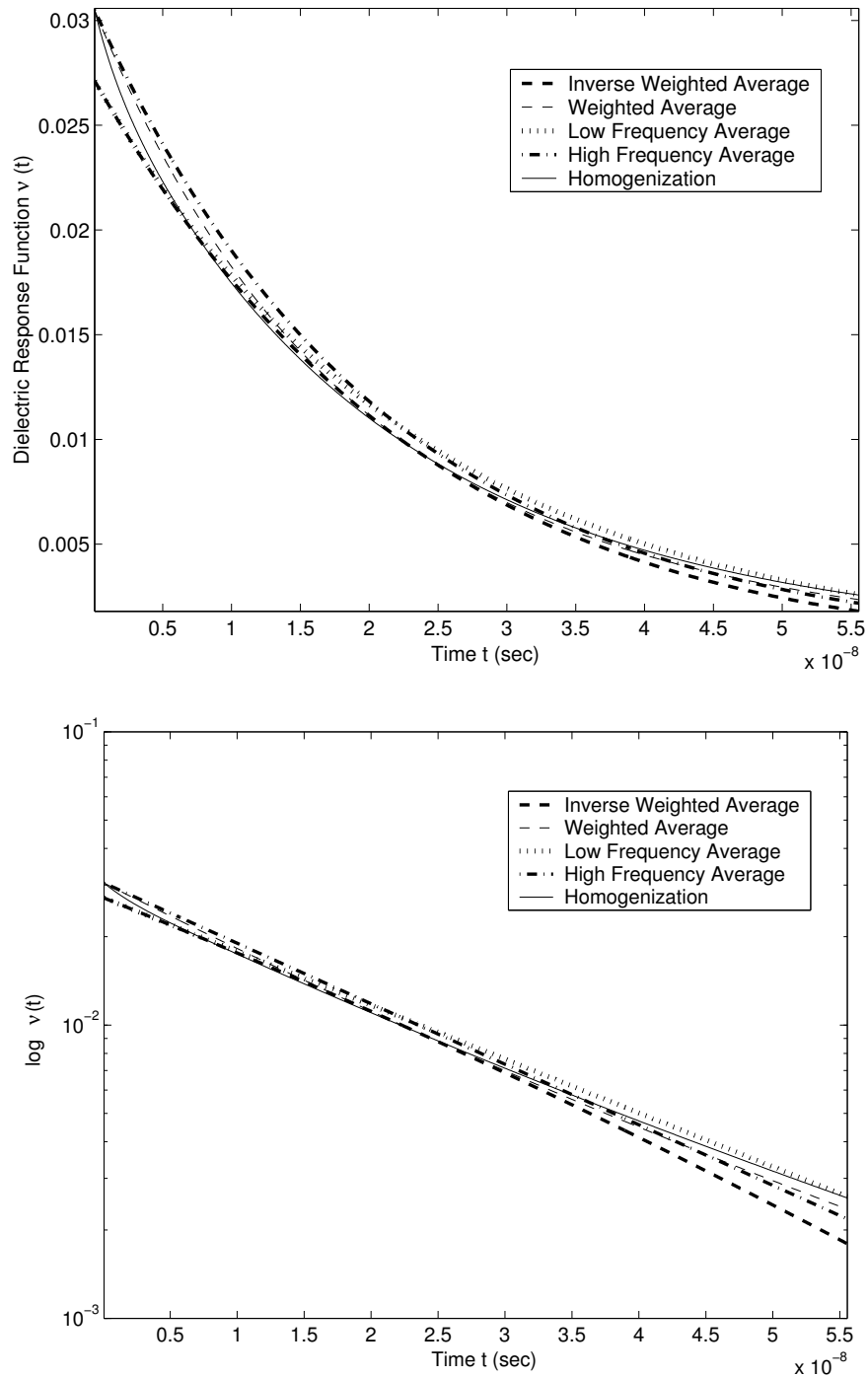


Figure 11: (top) The effective dielectric response function (DRF) $\nu(t)$ as a function of time, for volume fraction $f = 0.5$. We compare the homogenized DRF with the DRF that is a weighted average of the corresponding DRF's, as well averages that are observed in other experiments. (bottom) Semilog plot of the DRF's.

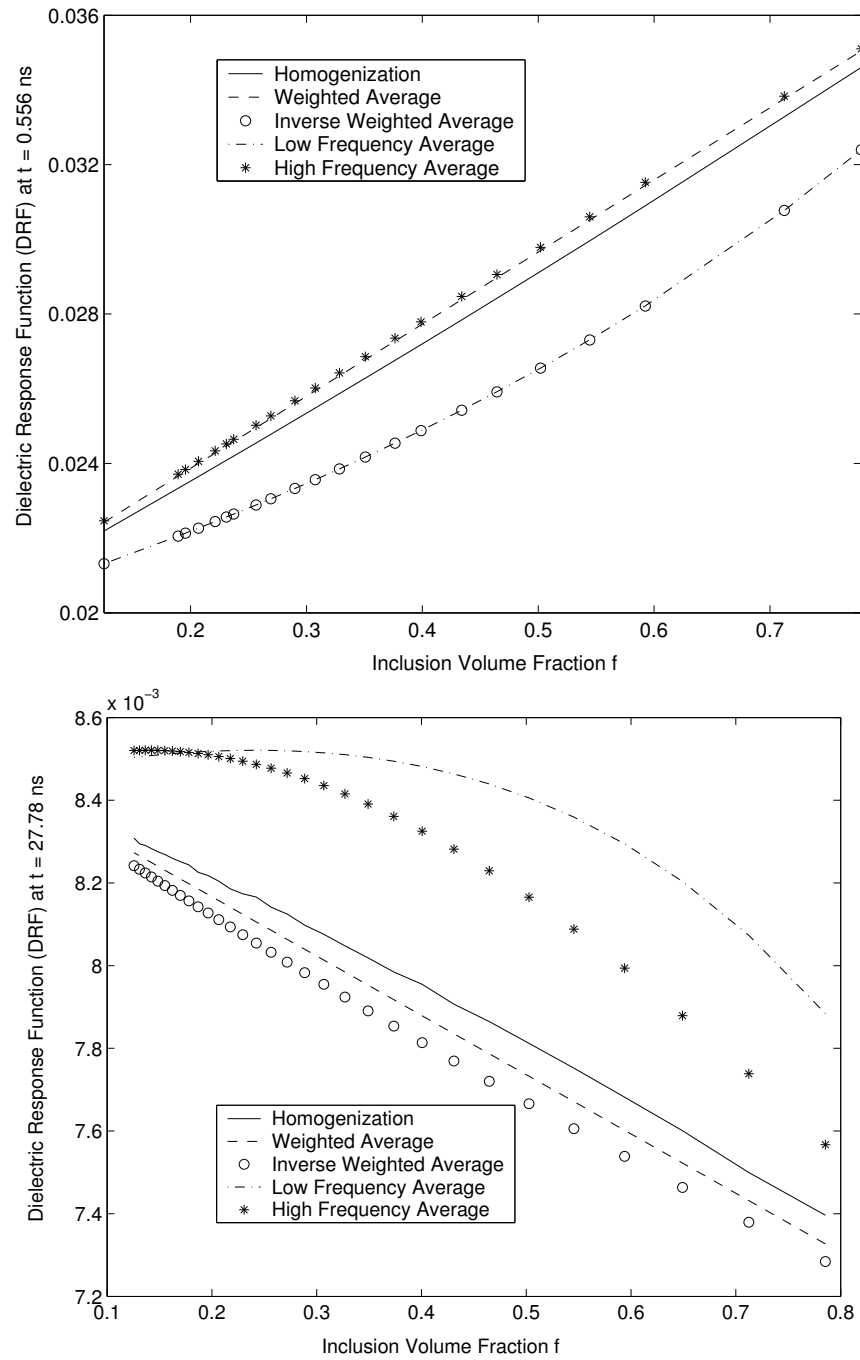


Figure 12: (top) The effective dielectric response function (DRF) $\nu(t)$ as a function of the inclusion volume fraction for $t = 0.556$ ns and (bottom) $t = 27.78$ ns.

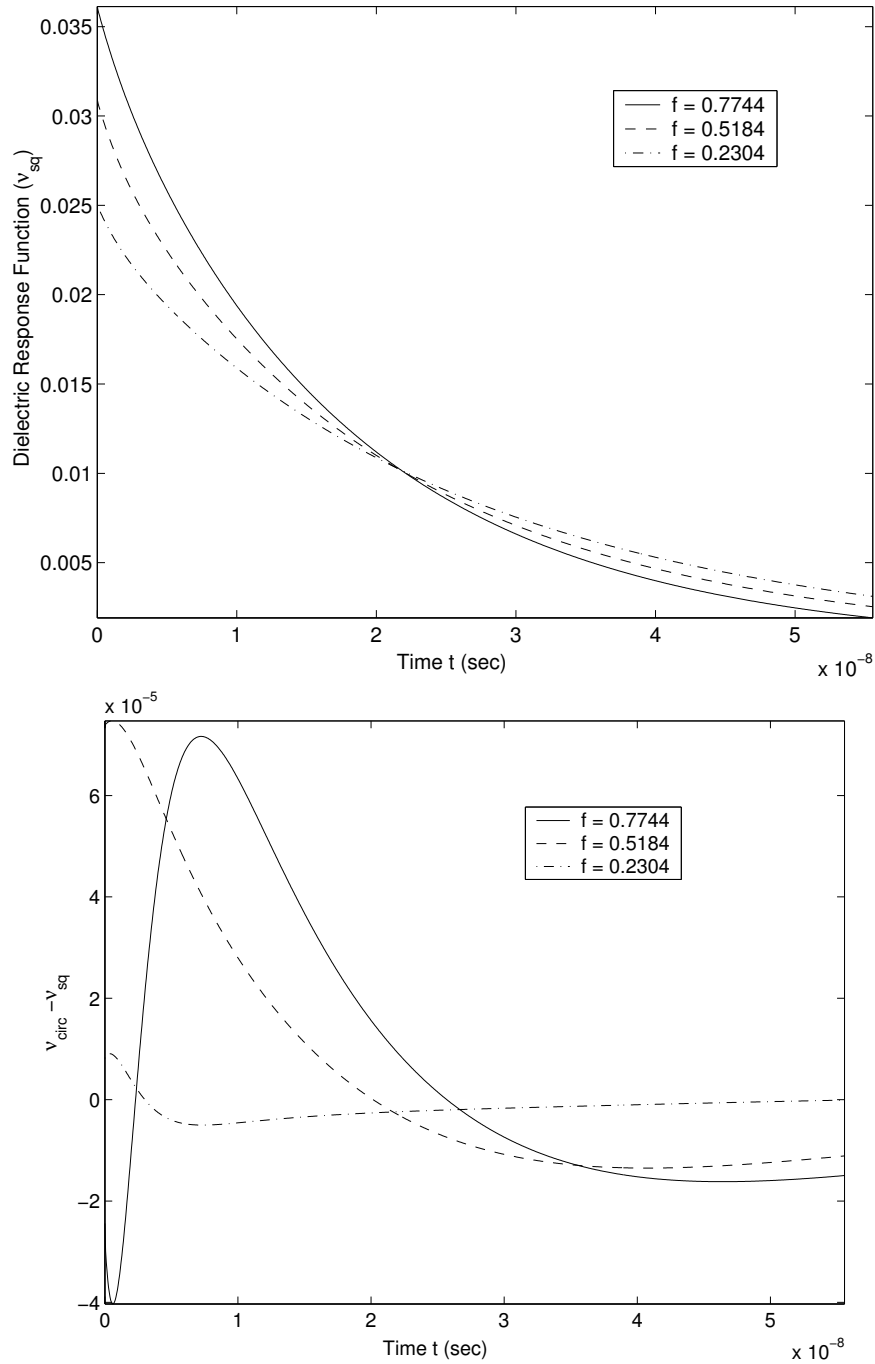


Figure 13: (top) The effective dielectric response function (DRF) $\nu(t)$ in the case of square microstructures for three different values of the volume fraction f . (bottom) The difference in the effective DRF between square and circular microstructures. In all three cases the differences are of the order of 10^{-5} , while we note from the top plot that the functions themselves are of magnitude 10^{-2} .

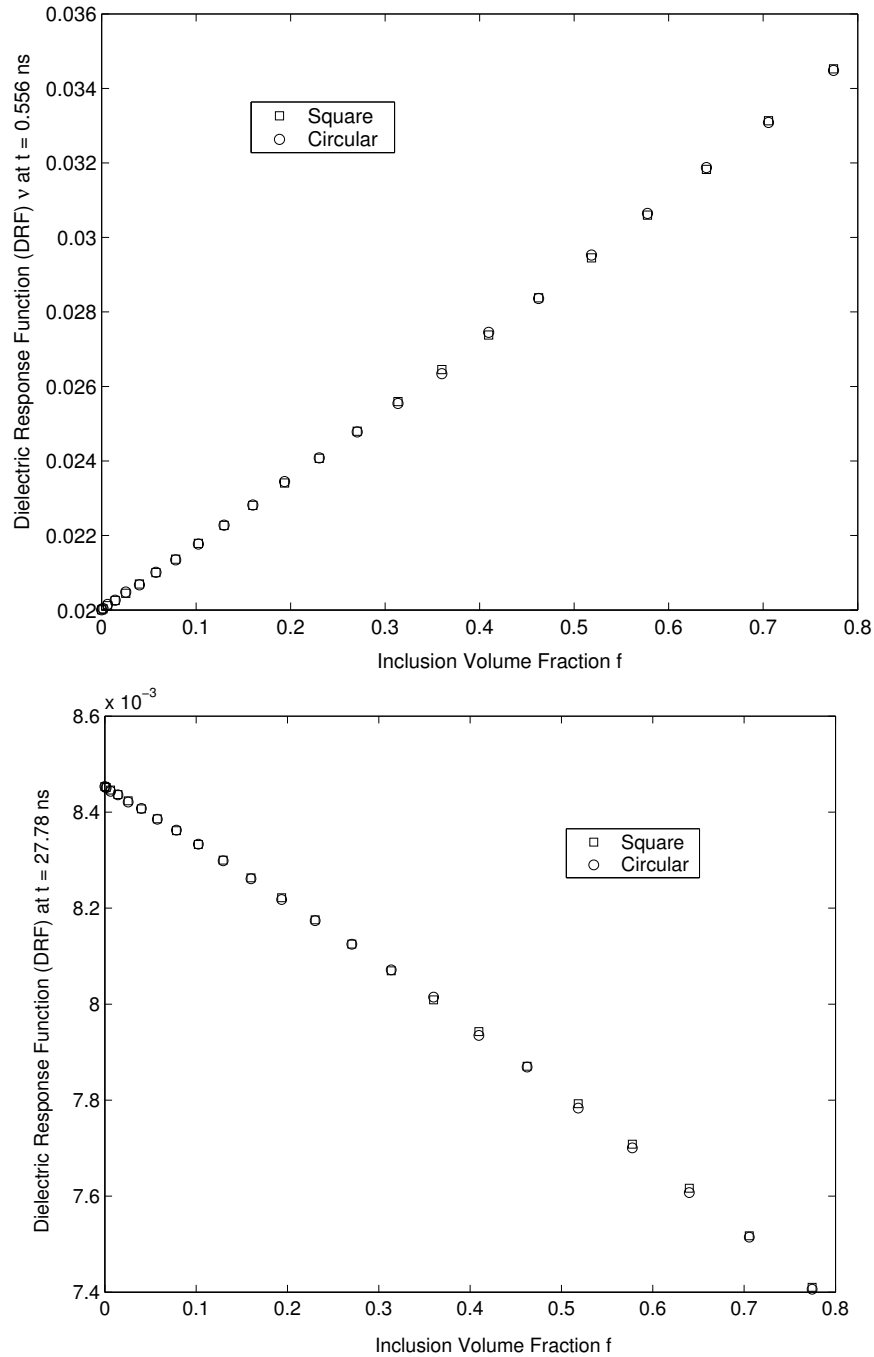


Figure 14: (top) The effective dielectric response function (DRF) $\nu(t)$ as a function of the inclusion volume fraction f for $t = 0.556$ ns and (bottom) $t = 27.78$ ns. The difference between the values of the DRF for the square and the circular microstructures are on the order of $10^{-4} - 10^{-5}$.

9 Conclusions and Future Directions

Taking into account the large deformations of the corrector functions (see Figures 5, 6, 9, and 10) we can expect that the correction of the electric and magnetic fields as given by (20) will be of crucial interest. The results of the cell problem with circular microstructures for calculating effective relative permittivities agree well with the mixing formulae results that are available in the literature. However, there are several advantages to taking a homogenization approach to obtain effective dielectric parameters. First, mixing formulae are not available for general microstructure geometries. Close examination of the materials of long term interest to us reveal that the foam consists of hexagonal-like cells in a microstructure configuration that is 95% gas surrounded by a matrix of polyurethane which has an estimated relative permittivity of 2.7. A homogenization approach will permit us to ascertain the sensitivity of the effective permittivities with respect to the micro-geometry. While we have used a staircase method to approximate the circular microstructure in the cell problem, the approach also permits better approximations (e.g., a fictitious domain approach) for the circular (and other shape) microstructures in the cell problem.

A second advantage of a homogenization approach is the flexibility it affords in assumptions about material polarization laws. Here in proof-of-concept calculations we have used a Debye medium for polarization but could with equal ease investigate the cases of Lorentz media, composite laws, and other higher order dispersive media. Moreover, as we investigate other mechanisms, it may be computationally advantageous to treat directly the polarization law as a side constraint. That is, the hysteretic term of the polarization is most generally represented by an integral term which involves a dielectric response function. Our numerical approximation to this approach involved a recursive convolution method to approximate the integral term. Our current efforts entail an approach in which the homogenized model involves an ordinary differential equation for the hysteretic part of the polarization term. This approach should facilitate numerical approximation of the solution of the electromagnetic interrogation problem for a wide class of assumed polarization mechanisms.

The initial results of the cell problems for the calculation of the dielectric response function compare well with the results that were obtained in [3], where a probabilistic approach is taken to an electromagnetic interrogation problem in a dielectric medium that is a mixture of two Debye media with different relaxation times. Each approach has its own conceptual, theoretical and computational advantages that merit further comparisons.

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