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THE EFFECTIVENESS OF RANDOM MFSK FREQUENCY-HOPPING ECCM RADIOS AGAINST WORST-CASE PARTIAL-BAND NOISE JAMMING

FINAL REPORT

AUGUST 1986

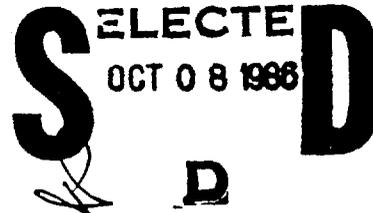
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20. ABSTRACT (Continue on reverse side if necessary and identify by block number) The system analyzed for bit error rate (BER) performance is frequency-hopped (FH) multiple frequency-shift-keying (MFSK) wherein the M symbol frequency assignments are independently and randomly chosen on each hop, and the information transmission is repeated L times (L-hop diversity) for a soft symbol decision. This FH/RMFSK system is intended to counter systematic follow-on jamming. The reported analysis is of the system's BER in worst case partial-band noise jamming (WCPBNJ). To be effective, the receiver must employ nonlinear combining of the L hops; several hop weighting schemes are evaluated with different assumptions about		

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available real-time information on relative powers of signal, noise, and jamming. These include adaptive gain control, clipping, hard-decision, and self-normalizing (nonparametric) schemes. It is shown that a simple, self-normalizing receiver, using no jamming state information or measurements, can perform nearly as well as one using a priori values of received noise-plus-jamming powers for adaptive gain control. It is also demonstrated that a hard-decision receiver (majority logic decoding of the L repetitions) achieves an ECCM effect and is viable if the SNR is high. Although the BER varies with jammer power in much the same way as for conventional FH/MFSK (given the parameters M, L, and the unjammed SNR), including a diversity gain for high SNR, FH/RMFSK in general is more vulnerable to WCPBNJ for M greater than 2. Therefore, it is concluded that implementation of effective diversity schemes is feasible, and that for a binary system the additional complexity of random hopping can be assessed to the additional protection gained against follow-on jamming.-

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TABLE OF CONTENTS

	<u>PAGE</u>
1.0 INTRODUCTION	1-1
1.1 BACKGROUND	1-1
1.1.1 Jamming Strategy Against Frequency-Hopping Radios	1-3
1.1.2 ECCM Waveforms	1-4
1.1.3 Motivation for the Proposed Random Hopping	1-7
1.1.4 Rationale for the Exact Analysis of FH/MFSK.	1-9
1.1.5 Extension of Uncoded Error Analysis to Coded Performance	1-11
1.2 ECCM PROCESSING	1-15
1.2.1 Examples of Receiver Effectiveness Computations for FH/MFSK	1-16
1.2.2 Impact of Random FH/MFSK (FH/RMFSK) on Analysis	1-25
1.3 SUMMARY OF REPORT	
1.3.1 General Description of Work and Approach	1-27
1.3.1.1 Receiver models studied	1-27
1.3.1.2 Jamming model and measure of effectiveness	1-31
1.3.1.3 Organization of report	1-34
1.3.2 Summary of Findings	1-35
2.0 PRELIMINARY ANALYSIS	2-1
2.1 CONDITIONAL PROBABILITY OF ERROR	2-1
2.1.1 Assumed Signals, Noise and Jamming	2-3
2.1.2 Conditional Error Probability Formulation	2-5



Codes

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A-1

TABLE OF CONTENTS (Continued)

	<u>PAGE</u>
2.2 ENUMERATION OF DISTINGUISHABLE JAMMING EVENTS	2-7
2.2.1 Definition of Distinguishable Jamming Events	2-7
2.2.2 Smallest Set of Distinguishable Jamming Events	2-8
2.2.3 Jamming Event Probabilities, Single Hop	2-11
2.2.4 Jamming Event Probabilities for Multiple Hops, Characteristic Function Method	2-13
2.2.5 Jamming Event Probabilities For Multiple Hops Convolution Method	2-16
2.3 TOTAL PROBABILITY OF ERROR	2-20
2.4 FH/RMFSK HARD DECISION ANALYSIS	2-21
2.4.1 Formulation of Error Probability	2-21
2.4.2 Explicit Form of Error Probability	2-23
2.4.2.1 Special case: L=2	2-25
2.4.2.2 Special case: L=3	2-25
2.4.2.3 Special case: L=4	2-26
2.4.2.4 Special case: M=2	2-26
3.0 FH/RMFSK PERFORMANCE USING SQUARE-LAW LINEAR COMBINING RECEIVER	3-1
3.1 ERROR PROBABILITY ANALYSIS	3-1
3.1.1 Distribution of the Decision Statistics	3-2
3.1.2 Formulation of the Conditional Error Probability	3-4
3.1.3 Powers of Non-Signal Channel Probabilities	3-5
3.1.4 Expectation Over Signal Channel PDF	3-11
3.1.5 Special case: L=1 and M=2	3-18

TABLE OF CONTENTS (Continued)

	<u>PAGE</u>
3.2 NUMERICAL RESULTS	3-19
3.2.1 Soft-Decision Receiver	3-19
3.2.2 Hard-Decision Receiver	3-33
4.0 FH/RMFSK PERFORMANCE USING SQUARE-LAW AGC RECEIVER	4-1
4.1 POSSIBLE AGC WEIGHTING SCHEMES FOR FH/RMFSK	4-2
4.1.1 Average AGC Scheme	4-2
4.1.2 Individual Channel AGC Scheme	4-4
4.1.3 Any-Channel-Jammed AGC Scheme	4-5
4.2 ANALYSIS OF FH/RMFSK PERFORMANCE USING INDIVIDUAL CHANNEL AGC SCHEME	4-7
4.2.1 Conditional Probability of Error	4-7
4.2.2 Total Error Probability	4-9
4.3 ANALYSIS OF FH/RMFSK PERFORMANCE USING ANY-CHANNEL- JAMMED AGC SCHEME	4-9
4.3.1 Conditional Probability of Error	4-9
4.3.1.1 Formulations of nonsignal channel probabilities	4-11
4.3.1.2 Formulation of symbol error probability in terms of previous results	4-12
4.3.2 Enumeration and Probabilities for Jamming Events	4-13
4.4 NUMERICAL RESULTS	4-18
4.4.1 Numerical Results for Individual-Channel AGC Receiver	4-18
4.4.2 Numerical Results for Any-Channel-Jammed AGC (ACJ-AGC) Receiver	4-28

TABLE OF CONTENTS (Continued)

	<u>PAGE</u>
5.0 FH/RMFSK PERFORMANCE USING CLIPPER RECEIVER	5-1
5.1 DISTRIBUTIONS OF THE DECISION VARIABLES	5-1
5.1.1 General Form of the pdf	5-1
5.1.2 Non-Signal Channel pdf	5-10
5.1.3 Signal Channel pdf	5-13
5.2 ERROR PROBABILITY FORMULATION	5-15
5.2.1 Conditional Probability of Error	5-15
5.2.1.1 Special case: $L=1$ (one hop/symbol)	5-20
5.2.1.2 Special case: $\underline{g}=0$ (no jamming)	5-22
5.2.2 Total Probability of Error	5-24
5.2.2.1 Choice of clipping threshold	5-25
5.2.2.2 Total error for $L=1$	5-26
5.3 NUMERICAL RESULTS	5-27
5.3.1 The Optimum Threshold Setting	5-28
5.3.2 Probability of Bit Error	5-30
6.0 FH/RMFSK PERFORMANCE USING SQUARE-LAW SELF-NORMALIZATION RECEIVER	6-1
6.1 THE SELF-NORMALIZATION SCHEME	6-1
6.1.1 Single-hop Distribution of Decision Variables	6-2
6.1.2 Alternate Forms	6-5
6.1.2.1 Special case: $M=2$ (binary)	6-6
6.1.2.2 Special case: no jamming	6-6

TABLE OF CONTENTS (Continued)

	<u>PAGE</u>
6.1.3	Conditional pdf's for M=2 6-7
6.1.3.1	Single hop/bit case (L=1) 6-7
6.1.3.2	Two hops/bit case (L=3) 6-8
6.1.4	Conditional pdf's for M=4 6-10
6.1.4.1	Single hop/symbol case (L=1) 6-10
6.1.4.2	Two hop/symbol case (L=2) 6-12
6.2	JAMMING EVENTS AND ERROR PROBABILITY FOR L=2 6-16
6.2.1	Jamming Events and Probabilities for M=2 6-16
6.2.2	Jamming Events and Probabilities for M=4 6-19
6.3	AN ALTERNATE APPROACH FOR M=2 AND L=3 HOPS/SYMBOL 6-23
6.4	NUMERICAL RESULTS. 6-25
7.0	COMPARISONS OF RECEIVER PERFORMANCES 7-1
7.1	RECEIVER PERFORMANCES IN FOLLOW-ON NOISE JAMMING 7-1
7.1.1	Formulation of Follow-on Jamming Analysis: Simple Jammer. 7-1
7.1.2	Formulation of FNJ Analysis: Advanced Jammer. 7-2
7.1.3	Performance of Conventional FH/MFSK in FNJ 7-2
7.1.4	Performance of FH/RMFSK in Follow-on Jamming . . . 7-4
7.1.4.1	Soft-decision receivers 7-7
7.1.4.2	Hard-decision receiver 7-13
7.2	COMPARISONS OF FH/RMFSK RECEIVER PERFORMANCES IN WORST-CASE PARTIAL-BAND NOISE JAMMING (WCPBNJ) 7-20

TABLE OF CONTENTS (Continued)

	<u>PAGE</u>
7.3 COMPARISONS OF FH/RMFSK AND FH/MFSK	7-35
7.3.1 The M=2, L=1 Case	7-35
7.3.2 L=1 with Alphabet Size Varied	7-38
7.3.3 Cases Where L>1 Hop/Symbol	7-41
7.3.3.1 Linear Combining Receiver	7-42
7.3.3.2 AGC receivers	7-42
7.3.3.3 Clipper receiver	7-42
7.3.3.4 Hard-decision receiver	7-50
7.3.3.5 Self-normalizing receiver	7-51
7.4 COMPARISON OF RECEIVER DIVERSITY EFFECTS	7-56
8.0 ECCM RECEIVER IMPLEMENTATION STUDIES	8-1
8.1 IMPLEMENTATION ISSUES AND CONCEPTS	8-1
8.1.1 ECCM Receiver Information Requirements	8-2
8.1.2 Measurement Approaches	8-5
8.1.2.1 A look-ahead measurement scheme	8-6
8.1.2.2 In-band measurement schemes	8-11
8.1.3 Jamming State Decisions	8-15
8.1.3.1 Jammed channel detection	8-16
8.1.3.2 Jammed symbol detection	8-19
8.2 ASSESSMENT OF IMPLEMENTATION EFFECTS	8-22
8.2.1 Methodology for Direct Assessment	8-22
8.2.2 In Search of an Upper Bound: Simplified Measurement Models	8-25

TABLE OF CONTENTS
(Continued)

	<u>PAGE</u>
8.2.3 Example Evaluations of Practical AGC Receivers	8-28
8.2.3.1 Analysis of the PACJ receiver	8-28
8.2.3.2 Analysis of the PIC receiver	8-30
8.3 CONCLUSIONS AND RECOMMENDATIONS	8-36
8.3.1 Knowledge Gained from Study	8-36
8.3.1.1 Ideal receiver performances	8-36
8.3.1.2 Practical receiver performances	8-39
8.3.1.3 RMFSK vs MFSK	8-40
8.3.2 Recommendations	8-41
APPENDIX A: PROBABILITY DENSITY FUNCTIONS FOR SOFT DECISION RECEIVER STATISTICS	A-1
APPENDIX B: COMBINATORIAL RELATIONS FOR JAMMING EVENT ENUMERATION	B-1
B.1 NUMBERS OF ORDERED VECTORS	B-1
B.2 NUMBERS OF PARTITIONED ORDERED VECTORS	B-2
B.3 TOTAL NUMBER OF VECTORS REPRESENTED BY ORDERED VECTORS	B-3
B.4 NUMBERS OF ACJ-AGC JAMMING EVENTS	B-5
B.4.1 Number of $\{\ell_0, \underline{\ell}\}$ events	B-5
B.4.2 Summation over ℓ_0 events	B-7
B.4.3 Summation over $\underline{\ell}$ events	B-7
APPENDIX C: DERIVATION OF ERROR RATE EXPRESSIONS FOR "PRACTICAL ACJ" AND "PRACTICAL IC" RECEIVERS	C-1
C.1 JOINT PDF FOR ONE PAIR OF SAMPLES (PACJ)	C-1
C.2 JOINT PDF FOR SUMS OF TWO SAMPLES (PACJ)	C-3

TABLE OF CONTENTS
(Continued)

	<u>PAGE</u>
C.3 ERROR PROBABILITY FOR L=2 (PACJ)	C-3
C.4 JOINT PDF FOR ONE PAIR OF SAMPLES (PIC)	C-6
C.5 CDF FOR SUM OF TWO NON-SIGNAL VARIABLES (PIC)	C-6
C.6 ERROR PROBABILITY FOR L=2 (PIC)	C-7
APPENDIX D: COMPUTER PROGRAM FOR SQUARE-LAW LINEAR COMBINING RECEIVER	D-1
APPENDIX E: COMPUTER PROGRAM FOR INDIVIDUAL CHANNEL ADAPTIVE GAIN CONTROL RECEIVER	E-1
APPENDIX F: COMPUTER PROGRAM FOR NUMERICAL COMPUTATIONS FOR THE ANY-CHANNEL-JAMMED ADAPTIVE GAIN CONTROL RECEIVER	F-1
APPENDIX G: COMPUTER PROGRAM FOR PLOTTING GRAPHICAL RESULTS FOR THE ANY-CHANNEL-JAMMED ADAPTIVE GAIN CONTROL RECEIVER	G-1
APPENDIX H: COMPUTER PROGRAM FOR CLIPPER RECEIVER WITH M=2 AND L=2	H-1
APPENDIX I: COMPUTER PROGRAM FOR CLIPPER RECEIVER FOR M=4 AND L=2	I-1
APPENDIX J: COMPUTER PROGRAM FOR CLIPPER RECEIVER WITH L=3 HOPS/SYMBOL	J-1
APPENDIX K: COMPUTER PROGRAM FOR SELF-NORMALIZING RECEIVER WITH M=2, L=2	K-1
APPENDIX L: COMPUTER PROGRAM FOR SELF-NORMALIZING RECEIVER WITH M=2, L=3	L-1
APPENDIX M: COMPUTER PROGRAM FOR PRACTICAL ADAPTIVE GAIN CONTROL RECEIVER	M-1
REFERENCES	R-1

LIST OF FIGURES

<u>Figures</u>	<u>Pages</u>
1.0-1 COMMUNICATIONS ECCM STRATEGY EVALUATION SCHEME	1-2
1.1-1 THERMAL NOISE AND PARTIAL-BAND NOISE JAMMING MODEL	1-5
1.1-2 TYPICAL L HOPS/SYMBOL FH/MFSK WAVEFORM PATTERN	1-6
1.1-3 TYPICAL L HOPS/SYMBOL FH/RMFSK WAVEFORM PATTERN	1-8
1.1-4 COMPARISON OF THE EXACT ANALYSES WITH THE UNION BOUNDS FOR THE SQUARE-LAW LINEAR COMBINING RECEIVER	1-12
1.1-5 COMPARISON OF THE EXACT ANALYSES WITH THE UNION BOUNDS FOR THE AGC RECEIVER	1-13
1.1-6 COMPARISON OF THE EXACT ANALYSES WITH THE UNION BOUNDS FOR THE CLIPPER RECEIVER	1-14
1.2-1 FH/MFSK SQUARE-LAW GENERIC RECEIVER	1-18
1.2-2 OPTIMUM JAMMING FRACTION (γ) VS. NUMBER OF HOPS/SYMBOL (L) FOR THE AGC AND CLIPPER FH/MFSK (M=8) RECEIVERS WHEN $E_b/N_0 = 9.09$ dB WITH E_b/N_J AS A PARAMETER (FOR 10^{-5} ERROR RATE WITHOUT JAMMING)	1-21
1.2-3 OPTIMUM PARTIAL-BAND NOISE JAMMING PERFORMANCE OF FH/MFSK (M=8) SQUARE-LAW COMBINING RECEIVERS FOR L=2 HOPS/SYMBOL WHEN $E_b/N_0 = 9.09$ dB (FOR IDEAL MFSK (M=8) CURVE THE ABSCISSA READS E_b/N_0)	1-22
1.2-4 OPTIMUM JAMMING PERFORMANCE OF THE AGC FH/MFSK (M=4) RECEIVER WHEN $E_b/N_0 = 13.16$ dB WITH THE NUMBER OF HOPS/SYMBOL (L) AS A PARAMETER (FOR IDEAL MFSK (M=4) CURVE THE ABSCISSA READS E_b/N_0)	1-23
1.2-5 OPTIMUM JAMMING PERFORMANCE OF THE AGC RECEIVER FOR BFSK/FH WITH THE NUMBER OF HOPS/BIT (L) AS A PARAMETER WHEN THERMAL NOISE IS ABSENT (FOR IDEAL BFSK CURVE THE ABSCISSA READS E_b/N_0)	1-24

LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>	
1.2-6	PROBABILITY OF BIT ERROR VS. E_b/N_j FOR AGC FH/BFSK RECEIVER USING OPTIMUM NUMBER (L) OF HOPS/BIT, FOR DIFFERENT VALUES OF E_b/N_0 IN WORST-CASE PARTIAL-BAND NOISE JAMMING	1-26
1.3-1	SOFT-DECISION RECEIVER FOR FH/RMFSK	1-28
1.3-2	PARTIAL-BAND JAMMING OF FH/RMFSK	1-30
1.3-3	HARD-DECISION RECEIVER FOR FH/RMFSK	1-33
2.1-1	GENERIC RECEIVER FOR FH/RMFSK	2-2
2.2-1	PROGRAM STRUCTURE FOR CALCULATING JAMMING EVENT PROBABILITIES INDIRECTLY	2-19
3.2-1	BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH M=2 AND L=1 HOP/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.3525$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	3-21
3.2-2	BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH M=2 AND L=2 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.3525$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	3-22
3.2-3	BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH M=2 AND L=3 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.3525$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	3-23
3.2-4	BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH M=2 AND L=4 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.3525$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	3-24

LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>
3.2-5 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER AND M=2 WITH NUMBER OF HOPS/BIT AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING	3-25
3.2-6 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH M=4 AND L=1 HOP/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	3-27
3.2-7 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH M=4 AND L=2 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	3-28
3.2-8 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH M=4 AND L=3 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	3-29
3.2-9 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH M=4 AND L=4 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	3-30
3.2-10 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER AND M=4 WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING	3-31
3.2-11 WORST-CASE PARTIAL-BAND NOISE JAMMING PERFORMANCE OF FH/RMFSK RECEIVERS FOR L=1 HOP/SYMBOL AND M = 2, 4, 8 WHEN E_b/N_0 GIVES A 10^{-5} BER WITHOUT JAMMING	3-32

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LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>
3.2-12 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR M=2 IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT $E_b/N_0 = 13.35247$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1)	3-34
3.2-13 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR M=4 IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT $E_b/N_0 = 10.60657$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1)	3-35
3.2-14 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR M=8 IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT $E_b/N_0 = 9.09401$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1)	3-36
3.2-15 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR M=2 IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT $E_b/N_0 = 20$ dB (FOR MINIMIZATION OF THERMAL NOISE EFFECT)	3-38
3.2-16 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR M=4 IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT $E_b/N_0 = 20$ dB (FOR MINIMIZATION OF THERMAL NOISE EFFECT)	3-39
3.2-17 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR M=8 IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT $E_b/N_0 = 2$ - dB (FOR MINIMIZATION OF THERMAL NOISE EFFECT)	3-40

J. S. LEE ASSOCIATES, INC.

LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>
3.2-18 OPTIMUM NUMBER OF HOPPING SLOTS JAMMED (Q_{\max} THAT PRODUCES $P(e)_{\max}$) VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR THE SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR M=2 AT $E_b/N_0 = 13.35247$ dB (FOR 10^{-5} BER WHEN L=1)	3-42
3.2-19 OPTIMUM NUMBER OF HOPPING SLOTS JAMMED (Q_{\max} THAT PRODUCES $P(e)_{\max}$) VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR THE SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR M=4 AT $E_b/N_0 = 10.60657$ dB (FOR 10^{-5} BER WHEN L=1)	3-43
3.2-20 OPTIMUM NUMBER OF HOPPING SLOTS JAMMED (Q_{\max} THAT PRODUCES $P(e)_{\max}$) VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR THE SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR M=8 AT $E_b/N_0 = 9.09401$ dB (FOR 10^{-5} BER WHEN L=1)	3-44
3.2-21 OPTIMUM NUMBER OF HOPPING SLOTS JAMMED (Q_{\max} THAT PRODUCES $P(e)_{\max}$) VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR THE SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR M=2 AT $E_b/N_0 = 20$ dB (FOR MINIMIZATION OF THERMAL NOISE)	3-46
4.1-1 POSSIBLE AGC NORMALIZATION SCHEMES	4-3
4.4-1 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=2 AND L=1 HOP/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-19
4.4-2 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=2 AND L=2 HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-20
4.4-3 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=2 AND L=3 HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-21

J. S. LEE ASSOCIATES, INC.

LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>
4.4-4 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=2 AND L=4 HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-22
4.4-5 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER AND M=2 WITH NUMBER OF HOPS/SYMBOL L AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING	4-23
4.4-6 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=4 AND L=1 HOP/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-24
4.4-7 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=4 AND L=2 HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-25
4.4-8 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=4 AND L=3 HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-26
4.4-9 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=4 AND L=4 HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-27
4.4-10 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER AND M=4 WITH NUMBER OF HOPS/SYMBOL L AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING	4-29

LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>
4.4-11 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=8 AND L=1 HOP/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=9.094011$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-30
4.4-12 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=8 AND L=2 HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=9.094011$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-31
4.4-13 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=8 AND L=3 HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=9.094011$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-32
4.4-14 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH M=8 AND L=4 HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=9.094011$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-33
4.4-15 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER AND M=8 WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING	4-34
4.4-16 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH M=2 AND L=1 HOP/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-36

J. S. LEE ASSOCIATES, INC.

LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>
4.4-17 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH M=2 AND L=2 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-37
4.4-18 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH M=2 AND L=3 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-38
4.4-19 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH M=2 AND L=4 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-39
4.4-20 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER AND M=2 WITH NUMBER OF HOPS/BIT AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL- BAND NOISE JAMMING	4-40
4.4-21 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH M=4 AND L=1 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-41
4.4-22 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH M=4 AND L=2 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-42
4.4-23 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH M=4 AND L=3 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-43

LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>
4.4-24 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH M=4 AND L=4 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-44
4.4-25 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER AND M=4 WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL- BAND NOISE JAMMING	4-45
4.4-26 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER AND M=8 AND L=1 HOP/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=9.094011$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-46
4.4-27 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER AND M=8 AND L=2 HOPS/ SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=9.094011$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN L=1) WITH JAMMING FRACTION γ AS A PARAMETER	4-47
4.4-28 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER AND M=8 WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL- BAND NOISE JAMMING	4-48
5.0-1 SOFT-DECISION FH/RMFSK RECEIVER WITH CLIPPERS	5-2
5.1-1 EFFECTS OF CLIPPING ON pdf	5-4
5.3-1 PERFORMANCE OF CLIPPER RECEIVER FOR FH/RMFSK WITH M=2, L=2 HOPS/ SYMBOL, AND $E_b/N_0=13.35247$ dB (FOR $P_b(e)=10^{-5}$ IN THE ABSENCE OF JAMMING WHEN L=1)	5-32

J. S. LEE ASSOCIATES, INC.

LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>
5.3-2 PERFORMANCE OF CLIPPER RECEIVER FOR FH/RMFSK WHEN L=2 HOPS/ SYMBOL WITH M AS A PARAMETER AND E_b/N_0 CORRESPONDING TO $P_b(e)=10^{-5}$ IN THE ABSENCE OF JAMMING (WHEN L=1 HOP/SYMBOL) . . .	5-33
6.1-1 REGION CONTAINING DECISION VARIABLES	6-13
6.1-2 TWO- AND THREE-DIMENSIONAL CONVOLUTIONS	6-14
6.2-1 VOLUME OF INTEGRATION FOR CORRECT SYMBOL DECISION	6-22
6.4-1 PERFORMANCE OF SELF-NORMALIZING RECEIVER FOR FH/RMFSK WITH M=2, L=2 HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=13.352471$ dB (FOR $P_b(e)=10^{-5}$ IN THE ABSENCE OF JAMMING WHEN L=1)	6-26
6.4-2 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR SELF-NORMALIZING RECEIVER IN WORST-CASE PARTIAL- BAND NOISE JAMMING FOR M=2, L=2 AT $E_b/N_0=20$ dB	6-27
6.4-3 PERFORMANCE OF SELF-NORMALIZING RECEIVER FOR FH/RMFSK WITH M=2, L=3 HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0=13.35247$ dB (FOR $P_b(e)=10^{-5}$ IN THE ABSENCE OF JAMMING WHEN L=1)	6-29
6.4-4 PERFORMANCE OF SELF-NORMALIZING RECEIVER FOR FH/RMFSK WITH M=2, AND $E_b/N_0=13.35247$ dB WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER	6-30
7.1-1 THERMAL NOISE AND FOLLOW-ON JAMMER MODELS	7-3
7.1-2 FOLLOW-ON JAMMING PERFORMANCE OF LCP AND ACJ-AGC RECEIVERS FOR FH/RMFSK WITH L=1,2,6 AND M=2,32 WHEN E_b/N_0 YIELDS A 10^{-5} BER FOR L=1	7-5
7.1-3 ADVANCED FOLLOW-ON NOISE JAMMING PERFORMANCE OF FH/RMFSK AND FH/MFSK FOR L=1 HOP/SYMBOL AND UNJAMMED BER = 10^{-5}	7-6
7.1-4 PERFORMANCE OF FH/RMFSK RECEIVERS IN SIMPLE FOLLOW-ON NOISE JAMMING VS. EFFECTIVE BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR L=1 HOP/SYMBOL AND M=2,4,8 WHEN E_b/N_0 YIELDS A 10^{-5} BER WITHOUT JAMMING	7-10

LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>
7.1-5 PERFORMANCE OF FH/RMFSK RECEIVERS IN SIMPLE FOLLOW-ON NOISE JAMMING VS. EFFECTIVE BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR L=2 HOPS/SYMBOL AND M=2,4,8 WHEN E_b/N_0 YIELDS A 10^{-5} BER WITHOUT JAMMING	7-12
7.1-6 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNJ VS. EFFECTIVE E_b/N_J FOR M=2 AND E_b/N_0 SUCH THAT UNJAMMED BER IS 10^{-5} WHEN L=1	7-14
7.1-7 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNJ VS. EFFECTIVE E_b/N_J FOR M=4 AND E_b/N_0 SUCH THAT UNJAMMED BER IS 10^{-5} WHEN L=1	7-15
7.1-8 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNJ VS. EFFECTIVE E_b/N_J FOR M=8 AND E_b/N_0 SUCH THAT UNJAMMED BER IS 10^{-5} WHEN L=1	7-16
7.1-9 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNJ VS. EFFECTIVE E_b/N_J FOR M=2 AND $E_b/N_0=20$ dB	7-17
7.1-10 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNJ VS. EFFECTIVE E_b/N_J FOR M=4 AND $E_b/N_0=20$ db	7-18
7.1-11 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNJ VS. EFFECTIVE E_b/N_J FOR M=8 AND $E_b/N_0=20$ dB	7-19
7.2-1 COMPARISON OF DIFFERENT RECEIVERS FOR FH/RMFSK WITH M=2 AND L=2 HOPS/SYMBOL WHEN $E_b/N_0=13.35247$ dB	7-22
7.2-2 COMPARISON OF DIFFERENT RECEIVERS FOR FH/RMFSK WITH M=2 AND L=2 HOPS/SYMBOL WHEN $E_b/N_0=13.35247$ dB	7-23
7.2-3 COMPARISON OF DIFFERENT RECEIVERS FOR FH/RMFSK WITH M=4 AND L=2 HOPS/SYMBOL WHEN $E_b/N_0=10.60657$ dB	7-24
7.2-4 COMPARISON OF DIFFERENT RECEIVERS FOR FH/RMFSK WITH M=4 AND L=3 HOPS/SYMBOL WHEN $E_b/N_0=10.60657$ dB	7-25
7.2-5 COMPARISON OF DIFFERENT RECEIVERS FOR FH/RMFSK WITH M=2 AND L=2 HOPS/SYMBOL WHEN $E_b/N_0=20$ dB	7-30

J. S. LEE ASSOCIATES, INC.

LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>
7.2-6 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR IC-AGC RECEIVER IN WORST-CASE PARTIAL-BAND NOISE FOR M=2 AND L=2 WHEN $E_b/N_0=20$ dB	7-32
7.2-7 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR SELF-NORMALIZING RECEIVER IN WORST-CASE PARTIAL-BAND NOISE JAMMING FOR M=2, L=2 AT $E_b/N_0=20$ dB	7-33
7.2-8 WORST-CASE PARTIAL-BAND NOISE JAMMING PERFORMANCE OF THE SELF-NORMALIZING FH/BFSK RECEIVER WHEN THERMAL NOISE IS ABSENT, WITH THE NUMBER OF HOPS/BIT (L) AS A PARAMETER	7-34
7.3-1 COMPARISON OF PERFORMANCE OF ALL RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH M=2 AND L=1 WHEN $E_b/N_0=13.35247$ dB	7-36
7.3-2 PERFORMANCE OF INDIVIDUAL CHANNEL AGC RECEIVER FOR FH/RMFSK WHEN L=1 HOP/SYMBOL WITH M AS A PARAMETER AND E_b/N_0 CORRESPONDING TO $P_b(e)=10^{-5}$ IN THE ABSENCE OF JAMMING	7-39
7.3-3 WORST-CASE PARTIAL-BAND NOISE JAMMING PERFORMANCE OF FH/RMFSK RECEIVERS FOR L=1 HOP/SYMBOL AND M=2,4,8 WHEN E_b/N_0 GIVES A 10^{-5} BER WITHOUT JAMMING	7-40
7.3-4 COMPARISON OF SQUARE-LAW LINEAR COMBINING RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH M=2 AND L=2 HOPS/SYMBOL WHEN $E_b/N_0=13.35247$ dB	7-43
7.3-5 COMPARISON OF SQUARE-LAW LINEAR COMBINING RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH M=4 AND L=2 HOPS/SYMBOL WHEN $E_b/N_0=10.60657$ dB	7-44
7.3-6 COMPARISON OF AGC RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH M=2 AND L=2 HOPS/SYMBOL WHEN $E_b/N_0=13.35247$ dB	7-45
7.3-7 COMPARISON OF AGC RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH M=2 AND L=3 HOPS/SYMBOL WHEN $E_b/N_0=13.35247$ dB	7-46
7.3-8 COMPARISON OF AGC RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH M=4 AND L=2 HOPS/SYMBOL WHEN $E_b/N_0=10.60657$ dB	7-47

J. S. LEE ASSOCIATES, INC.

LIST OF FIGURES (continued)

<u>Figures</u>	<u>Pages</u>
7.3-9 COMPARISON OF AGC RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH M=4 AND L=3 HOPS/SYMBOL WHEN $E_b/N_0 = 10.60657$ dB	7-48
7.3-10 PERFORMANCE OF CLIPPER RECEIVER FOR FH/RMFSK WHEN L=2 HOPS/ SYMBOL WITH M AS A PARAMETER AND E_b/N_0 CORRESPONDING TO $P_b(e) = 10^{-5}$ IN THE ABSENCE OF JAMMING (WHEN L=1 HOP/ SYMBOL)	7-49
7.3-11 ERROR PERFORMANCE FOR HARD SYMBOL DECISION FH/RMFSK RECEIVER IN WORST-CASE PARTIAL-BAND NOISE JAMMING WHEN L=3 HOPS/ SYMBOL; M=2,4,8; AND E_b/N_0 GIVES 10^{-5} ERROR WITHOUT JAMMING WHEN L=1	7-52
7.3-12 WORST-CASE PARTIAL-BAND NOISE JAMMING PERFORMANCES OF FH/RMFSK AND FH/MFSK USING THE SELF-NORMALIZING RECEIVER, FOR M=2 AND L=2	7-53
7.4-1 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER AND M = 2 WITH NUMBER OF HOPS/BIT AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING	7-57
7.4-2 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER AND M=2 WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING	7-59
7.4-3 OPTIMUM DIVERSITY PERFORMANCE OF INDIVIDUAL CHANNEL AGC RECEIVER FOR FH/RMFSK WITH M=2 AND E_b/N_0 AS A PARAMETER	7-60
7.4-4 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER AND M=2 WITH NUMBER OF HOPS/BIT AS A PARAMETER IN PRESENCE OF WORST- CASE PARTIAL-BAND NOISE JAMMING	7-62
7.4-5 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR SELF-NORMALIZING RECEIVER IN WORST- CASE PARTIAL-BAND NOISE FOR M=2 AND $E_b/N_0=13.35$ dB, WITH THE NUMBER OF HOPS/BIT (L) VARIED	7-63

J. S. LEE ASSOCIATES, INC.

LIST OF FIGURES (continued)

<u>Figures</u>		<u>Pages</u>
7.4-6	OPTIMUM DIVERSITY PERFORMANCE OF HARD DECISION RECEIVER FOR FH/RMFSK WITH M=2	7-64
8.1-1	ACQUISITION OF JAMMING STATE DATA BY A "LOOK-AHEAD" SCHEME	8-7
8.1-2	USE OF LOOK-AHEAD MEASUREMENTS TO DERIVE ECCM WEIGHTS	8-8
8.1-3	POSSIBLE USE OF MULTIPLE-HOP/MULTIPLE CHANNEL SAMPLES TO IMPROVE MEASUREMENTS	8-10
8.1-4	CONCEPTUAL PROCESSING OF IN-BAND MEASUREMENTS TO IMPLEMENT AGC RECEIVER	8-13
8.1-5	PROCESSING OF IN-BAND MEASUREMENTS TO IMPLEMENT CLIPPER RECEIVER	8-14
8.2-1	PRACTICAL ACJ-AGC (PACJ) RECEIVER FOR FH/RMFSK WHEN M=2	8-29
8.2-2	BER PERFORMANCE FOR PACJ FH/RMFSK RECEIVER IN PARTIAL-BAND NOISE JAMMING FOR M=2, L=2, AND $E_b/N_0 = 13.35$ dB	8-31
8.2-3	BER PERFORMANCE FOR PACJ FH/RMFSK RECEIVER IN PARTIAL-BAND NOISE JAMMING FOR M=2, L=2, AND $E_b/N_0 = 20$ dB	8-32
8.2-4	PRACTICAL IC-AGC (PIC) RECEIVER FOR FH/RMFSK WHEN M=2	8-33
8.2-5	PERFORMANCE OF THE PRACTICAL IC-AGC FH/RMFSK RECEIVER IN FULL-BAND NOISE JAMMING FOR M=L=2	8-35
8.2-6	PERFORMANCE OF THE PRACTICAL IC-AGC FH/RMFSK RECEIVER USING A TWO-SAMPLE NOISE ESTIMATE IN FULL-BAND NOISE JAMMING FOR M=L=2	8-37

J. S. LEE ASSOCIATES, INC.

LIST OF TABLES

<u>TABLES</u>		<u>PAGE</u>
1.2-1	DESCRIPTIONS OF THE RECEIVERS	1-19
1.3-1	RECEIVER PROCESSING FUNCTIONS STUDIED	1-32
2.2-1	NUMBER OF JAMMING EVENTS	2-8
2.2-2	MINIMUM NUMBER OF DISTINGUISHABLE EVENTS	2-9
2.2-3	JAMMING EVENT PROBABILITIES FOR M=2, L=1 TO 4	2-17
2.2-4	PROBABILITIES OF $\underline{\ell}$ JAMMING EVENTS FOR M = 4	2-18
3.1-1	POLYNOMIALS FOR $F_L(\alpha; \ell)$	3-10
3.1-2	COEFFICIENTS FOR EQUATION (3.1-29)	3-12
3.1-3	EXPRESSION FOR CONDITIONAL ERROR PROBABILITY	3-13
4.3-1	PROBABILITIES OF $(\ell_0, \underline{\ell})$ JAMMING EVENTS FOR M = 4	4-17
5.1-1	NON-SIGNAL CHANNEL PROBABILITY DENSITY FUNCTIONS, L = 1 TO 3	5-12
5.1-2	NON-SIGNAL CHANNEL CUMULATIVE DISTRIBUTION FUNCTIONS, L=1 TO 3	5-14
5.1-3	SIGNAL CHANNEL PROBABILITY DENSITY FUNCTIONS, L = 1,2	5-16
5.1-4	SIGNAL CHANNEL PROBABILITY DENSITY FUNCTIONS, L = 3	5-17
5.3-1	OPTIMUM NORMALIZED CLIPPING THRESHOLD $\eta_0/2\sigma_N^2$	5-29
6.1-1	L=1 PROBABILITY DENSITIES FOR M=4	6-11
6.2-1	JAMMING EVENTS AND PROBABILITIES FOR M=2, L=2	6-17
6.2-2	JAMMING EVENTS AND PROBABILITIES FOR M=4, L=2	6-20
8.1-1	SUMMARY OF ECCM TECHNIQUES FOR THE RECEIVERS STUDIED	8-3

J. S. LEE ASSOCIATES, INC.

THE EFFECTIVENESS OF RANDOM MFSK FREQUENCY HOPPING ECCM RADIOS AGAINST WORST-CASE PARTIAL-BAND NOISE JAMMING

1.0 INTRODUCTION

The purpose of this study is to provide the Army a direct comparison of the uncoded bit error rate (BER) performance of several receiver anti-jam processing schemes with varying degrees of implementation complexity, under the same conditions of system noise and jamming. In this manner the engineering cost of complex anti-jam receiver designs can be weighed against their effectiveness, as illustrated in Figure 1.0-1. In what follows we discuss the issues surrounding the work and summarize our effort.

1.1 BACKGROUND

In the Electronic Warfare (EW) environment, where a "battle" is waged between the communicating party and the party that is engaged in the pursuit of disrupting the communicator's link, strategy plays an important and fundamental role for the opposing parties. To the communicating party, the opponent's Electronic Support Measures (ESM) and Electronic Countermeasures (ECM) pose as threats. ESM involves essentially activities for spectrum surveillance and direction finding by passive means, whereas ECM involves activities for the purpose of victimizing the communicator's link. Jamming is an active measure of accomplishing ECM objectives. It is, therefore,

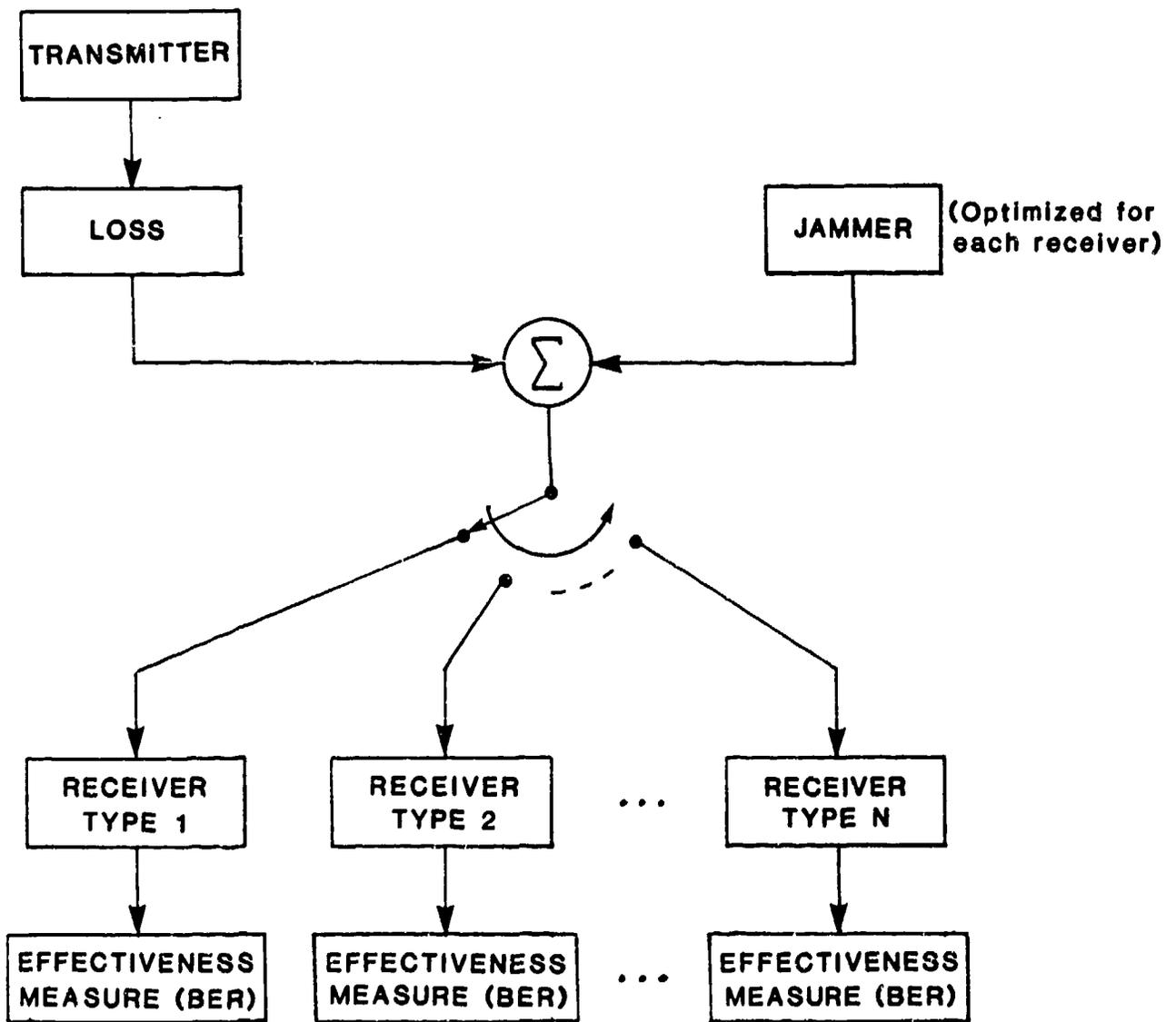


FIGURE 1.0-1 COMMUNICATIONS ECCM STRATEGY EVALUATION SCHEME

easy to recognize that fixed-frequency radios are very much vulnerable to ESM and ECM attacks. Communication systems that are designed to counter or mitigate the effects of ESM or ECM attacks are termed Electronic Counter-countermeasures (ECCM) radios, or jam-resistant communication systems.

In principle, there exist many different schemes which can provide the communicator with jam-resistant radio capabilities; Direct Sequence (DS) spread-spectrum and Frequency-Hopping (FH) spread-spectrum systems are two generic schemes. While the DS/SS system requires phase coherence over the system's wide operational bandwidth in its implementation, the FH/SS system does not. The fact that most of the tactical ECCM radios are of FH/SS type is based not only on this reason, but also on the fact that the attainable "processing gain" is achieved with less complexity and cost.

1.1.1 Jamming Strategy Against Frequency-Hopping Radios

ECCM radio designs are based on the desire to suppress the total jamming power by an amount equal to the processing gain, defined as the ratio of FH system bandwidth to the receiver noise bandwidth. The difference between the processing gain in dB and the SNR in dB required for traffic demodulation is the (anti-jam) margin that the communicator can use to tolerate an excess of jammer power over signal power at the system front end. The intelligent jammer, however, does not spread his power over the entire system bandwidth, so that the definition and effects of processing gain will not apply.

The jammer may employ a partial-band noise jamming strategy, in which the available jammer power is placed in a fraction (γ) of the radio

system bandwidth, as illustrated in Figure 1.1-1. Assuming that total power is fixed, there is an optimum value of γ which achieves the most effective tradeoff of the probability of jamming and the probability of error when jammed, thus achieving a maximum overall error rate for the given amount of jammer power.

The jammer may, if it is feasible, concentrate his power further if he can intercept the hopping signal in real time and immediately broadcast a strong burst of noise in the frequencies near the signal (follow-on jamming). This type of jamming can be successful against FH systems in which a conventional narrowband communications signal is slowly hopped by simple translation in frequency.

1.1.2 ECCM Waveforms

Along with using hopped signals, the communicator can exercise an additional degree of freedom by employing a low energy density waveform to minimize interceptions by a potential jammer. Such a waveform is FH/MFSK using a number of hops per symbol (L), a kind of repetition code or diversity [5] to permit transmission at lower power and/or to combat fading. The conventional form of FH/MFSK is illustrated in Figure 1.1-2; once a symbol has been chosen for a given interval T_s , a conventional MFSK signal is generated and randomly hopped (translated) L times at a rate $R_H = L/T_s = 1/\tau$ before a new symbol is keyed. Because the M possible symbol frequencies are adjacent, this waveform is vulnerable to follow-on repeat jamming, unless the hopping rate can be made very high.

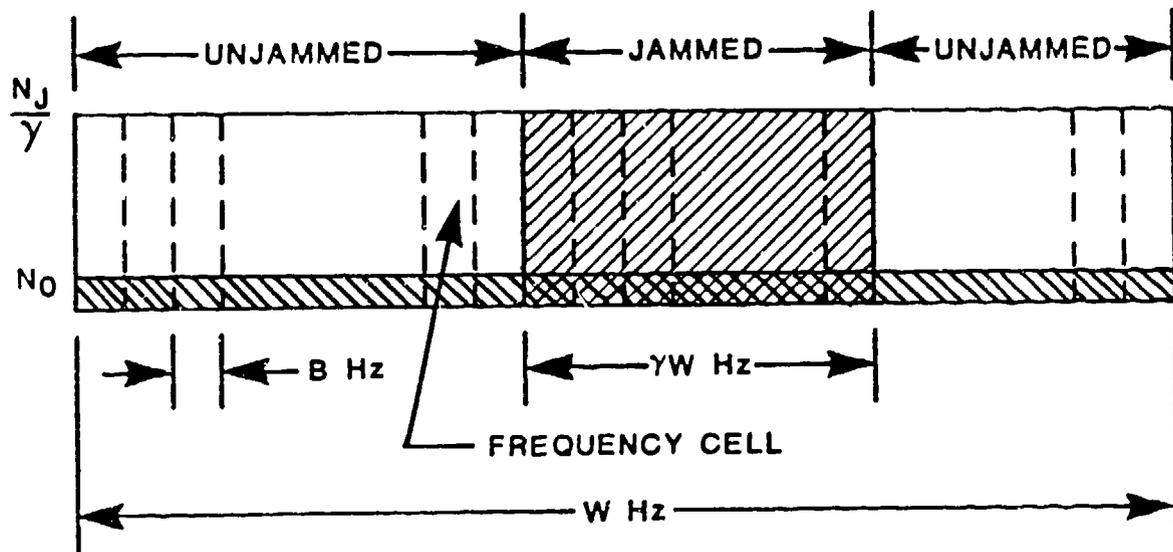


FIGURE 1.1-1 THERMAL NOISE AND PARTIAL-BAND NOISE JAMMING MODEL

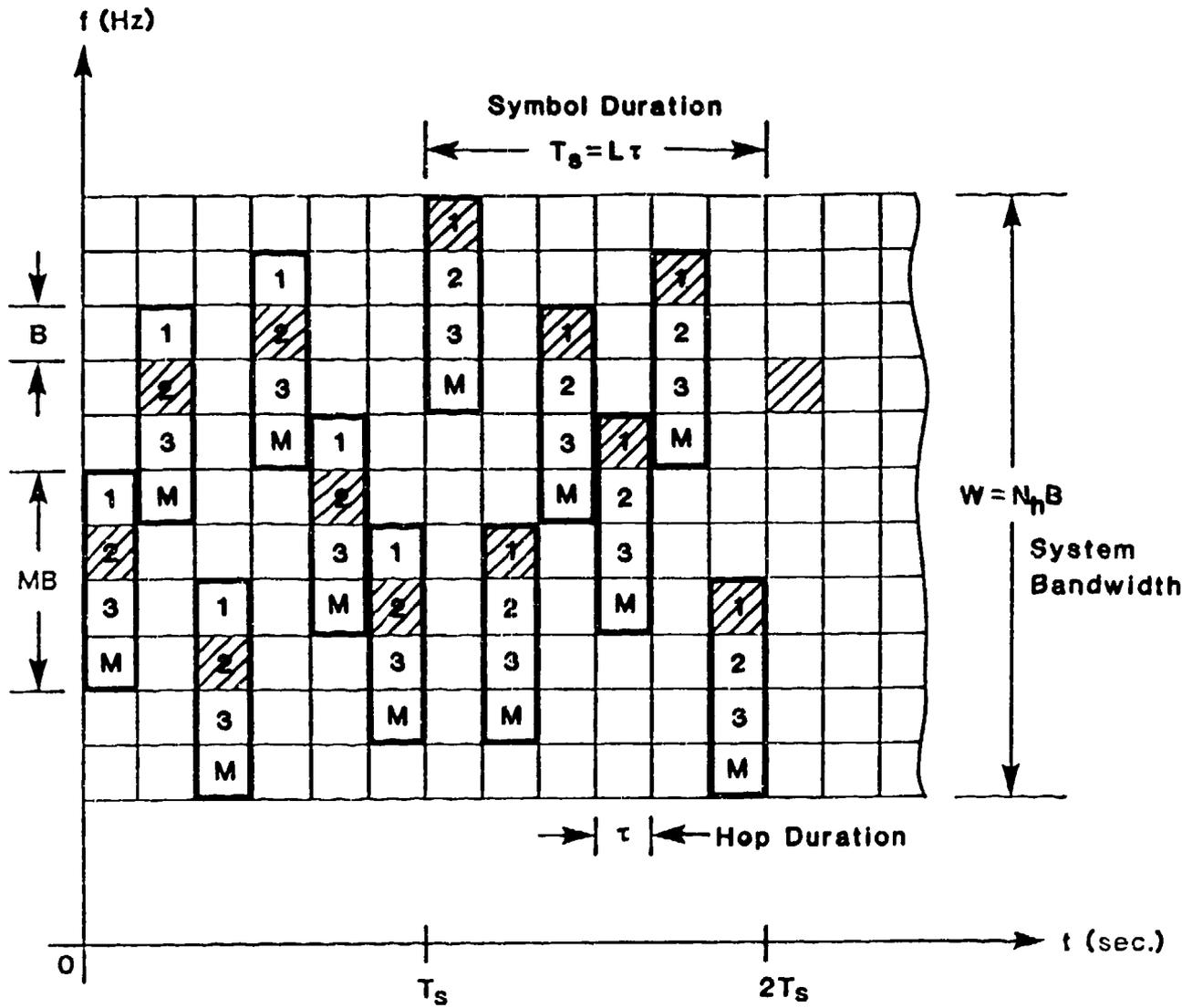


FIGURE 1.1-2 TYPICAL L HOPS/SYMBOL FH/MFSK WAVEFORM PATTERN

If the individual symbol frequencies of the MFSK symbol are assigned randomly on a per hop basis, as illustrated in Figure 1.1-3, then repeat jamming is less likely to produce an error since the symbol frequencies (at RF) are no longer adjacent [6,7]. (We shall refer to this waveform as FH/RMFSK with L hops per symbol.) It has been shown [6] that, for the special case of one hop/bit binary systems ($M=2$) and very little system or thermal noise ($E_b/N_0 = 30$ dB), the two forms of FH/BFSK achieve the same performance in optimum partial-band noise jamming. This suggests that the random MFSK waveform will perhaps be a better choice for $L > 1$ and $M > 2$, although it has not been established as a fact that its performance in partial-band noise is always equal to that of the conventional system, while offering additional protection against follow-on jamming.

The FH/RMFSK implementation is more complex, and the study in this report will permit the cost of this additional complexity to be weighed against its performance, compared to that of the more conventional FH/MFSK as calculated by LAI [1].

1.1.3 Motivation for the Proposed Random Hopping

As we have stated above, the proposed FH/RMFSK waveform is less vulnerable to follow-on or repeat jamming than is a conventional FH/MFSK systems with M contiguous signalling frequencies. In addition, the FH/RMFSK waveform is less vulnerable to tone jamming, since the randomized selection of the M frequencies reduces the amount of structure in the signal. This makes it

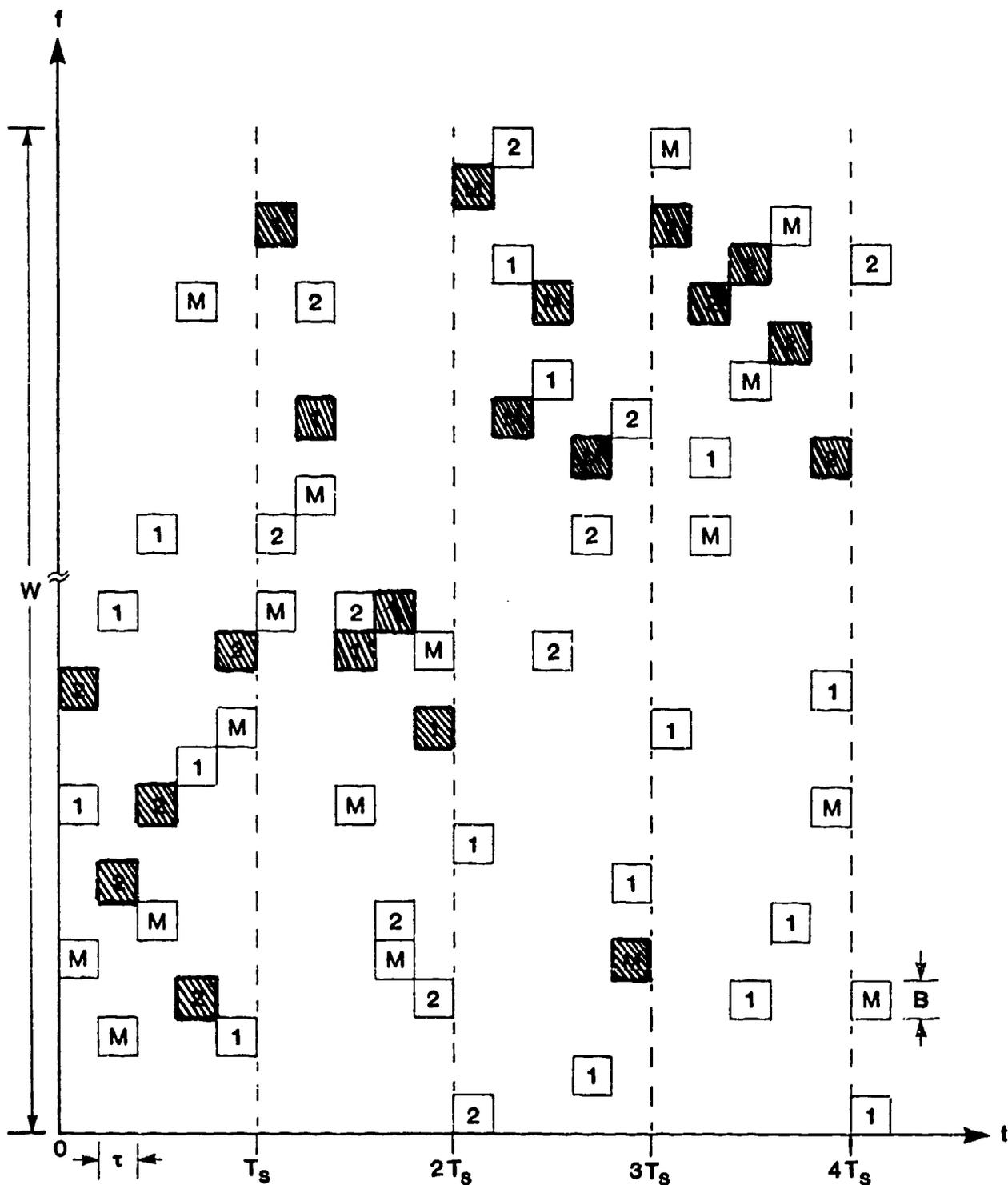


FIGURE 1.1-3 TYPICAL L HOPS/SYMBOL FH/RMFSK WAVEFORM PATTERN

much more difficult for a tone jammer to implement an optimum jamming strategy consisting of one jamming tone per M-ary symbol: the lack of structure in the hopping gives the jammer no features to exploit to insure a tone hits the symbol; thus the jammer is forced to divide his available power up into more tones, resulting in a lessened effect on the communications link when it hops into a jammed slot.

Therefore, the motivation for considering the use of FH/RMFSK as an LPI anti-jam communications system design is based upon a desire to lessen or reduce vulnerability to certain more sophisticated jamming threats, such as follow-on jamming and tone jamming. However, before the system can be considered a viable design candidate, its performance under the less sophisticated jamming, namely partial-band noise jamming, must be known.

1.1.4 Rationale for the Exact Analysis of FH/MFSK System Performance In Partial-Band Noise Jamming

It is known that the advantage of an M-ary orthogonal modulation system rests on the fact that the scheme requires less energy per data bit transmission than other available modulation schemes. Cost-effective implementation (efficient non-coherent detection) is another reason in selecting M-ary FSK waveforms by designers of ECCM radios. Recently, Hughes Aircraft Company has conducted studies for U.S. Army CECOM on feasibility of AJ/LPI ECCM techniques, employing L-hops per symbol FH/MFSK [8].

Exact knowledge of performance measures and vulnerability of L-hops per symbol FH/MFSK SS systems has not been available until recently,

J. S. LEE ASSOCIATES, INC.

and that of FH/RMFSK is yet to be determined. Workers in this field previously held the view that M-ary system performance measures could be estimated once the performance measures of the binary systems are available. This view was based on the conventional wisdom of applying the "union bound". As is well known, once we know the binary system performance, that of the M-ary system can be approximated by the union bound given by

$$P_M(e;E_s) \leq (M-1) P_2(e;E_s) \quad (1.1-1)$$

where $P_2(e;E_s)$ is the probability of error for the binary system, using any pair of symbols from the set $\{s_1(t), s_2(t), \dots, s_M(t)\}$, and $P_M(e;E_s)$ is the M-ary system probability of error, where E_s is the symbol energy.

The workers have also invoked the well-known relationship between the bit error probability and the symbol (K-bit word) error probability for the M-ary orthogonal system. That is,

$$P_b(e;E_b) = \frac{M}{2(M-1)} P_M(e;E_s). \quad (1.1-2)$$

where $P_b(e;E_b)$ denotes bit error probability and E_b is the energy per bit.

By putting equation (1.1-1) into equation (1.1-2), we obtain the "union bound"-based approximate-performance measure of the bit error probability of an MFSK systems, given by

$$P_b(e;E_b) \leq \frac{M}{2} P_2(e;E_s) = 2^{K-1} P_2(e;KE_b) \quad (1.1-3)$$

where

$$K = \log_2 M. \quad (1.1-4)$$

In principle, one can use equation (1.1-3) to assess the performance of an M-ary orthogonal system. Our discovery, however, did not support this generality when the communication channel is of the non-exponential type. In an attempt to obtain an approximate performance measure of 2-hops per symbol FH/MFSK system under partial-band noise jamming environment for M=4, 8, 16, and 32, we have used equation (1.1-3) in applying the binary results, as shown in Figures 1.1-4 to 1.1-6, to three different receiver schemes. A surprising result is that as M is increased, bit error probability as given by the union bound is worsened for all three receivers! Interpretation of this result is that one needs to expend more energy per bit in the higher-order-message-alphabet orthogonal system, a result that is not supportable even on the basis of intuition; and is, indeed, contrary to the exact results shown in the figures.

The above paragraph is to point out that the "union bound" cannot be used when one considers non-Gaussian channels such as partial-band noise, as experienced by a FH/MFSK system. These channels are inverse linear channels, and they do not allow the union bounding techniques to be applicable in assessing M-ary system performances. Thus, one can conclude that exact analysis is necessary.

1.1.5 Extension of Uncoded Error Analysis to Coded Performance

While error-control coding is quite likely to be used by the communicator to counter any jamming effects, the analysis of total system performance may be usefully divided into two parts: uncoded performance and

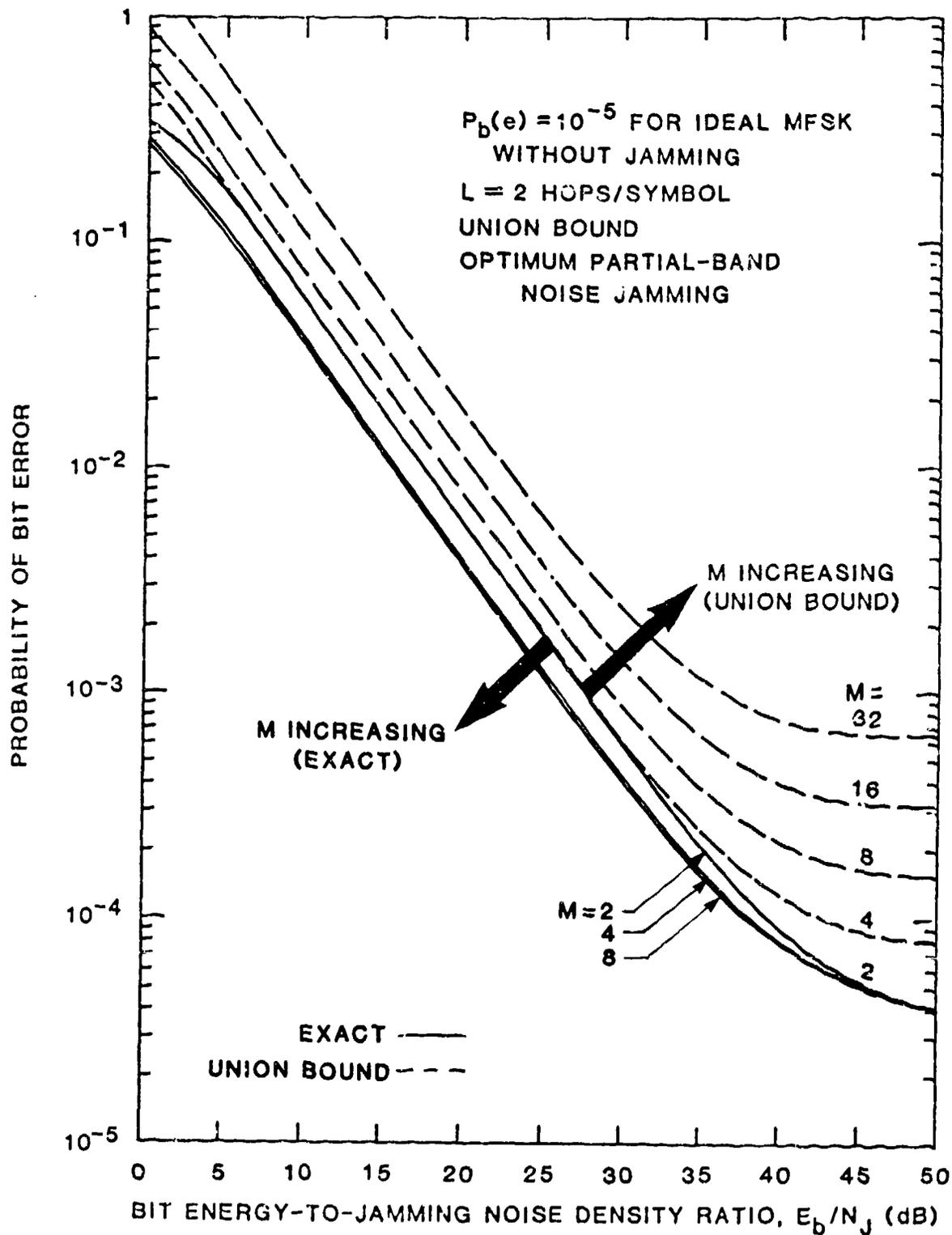


FIGURE 1.1-4 COMPARISON OF THE EXACT ANALYSES WITH THE UNION BOUNDS FOR THE SQUARE-LAW LINEAR COMBINING RECEIVER

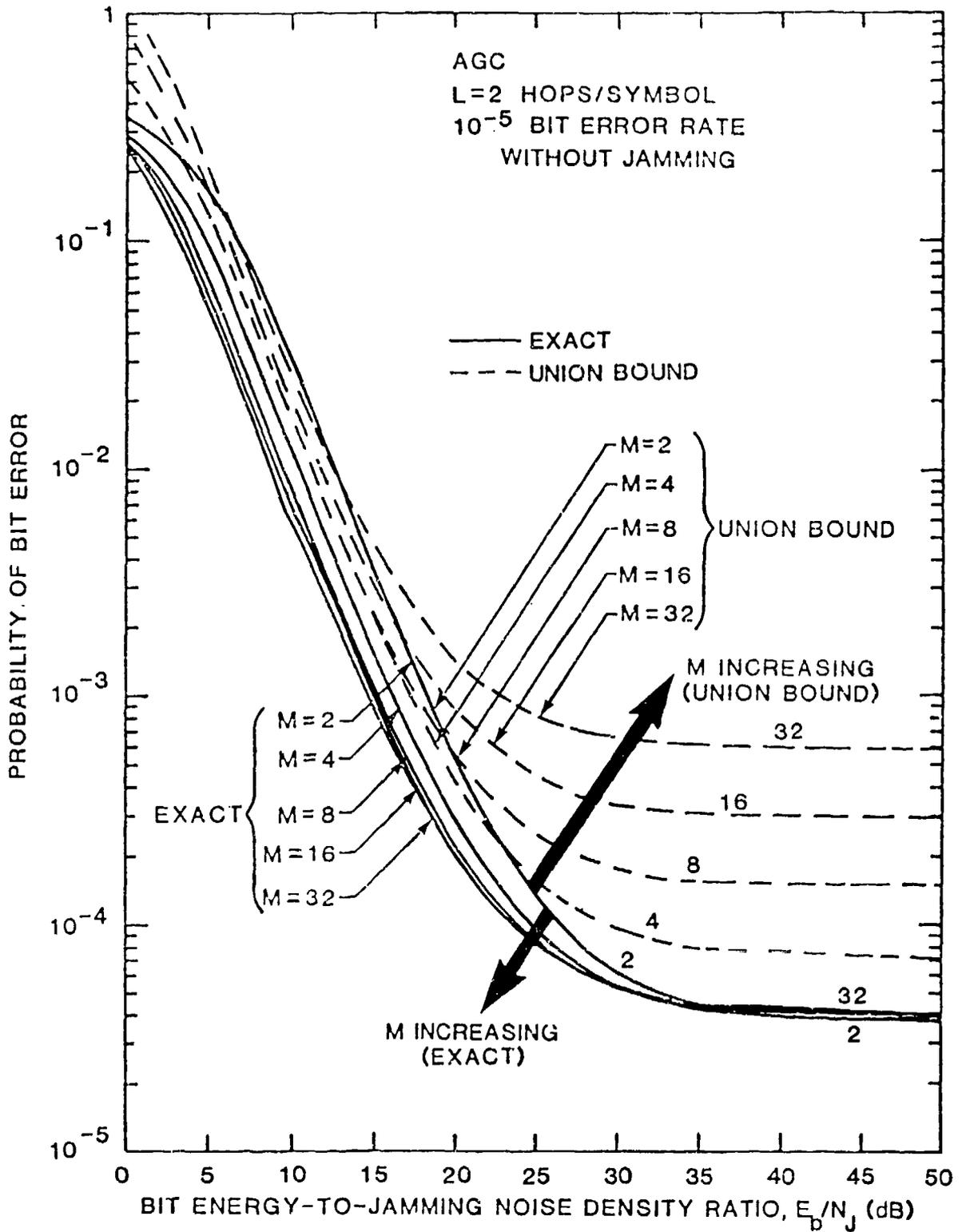


FIGURE 1.1-5 COMPARISON OF THE EXACT ANALYSES WITH THE UNION BOUNDS FOR THE AGC RECEIVER

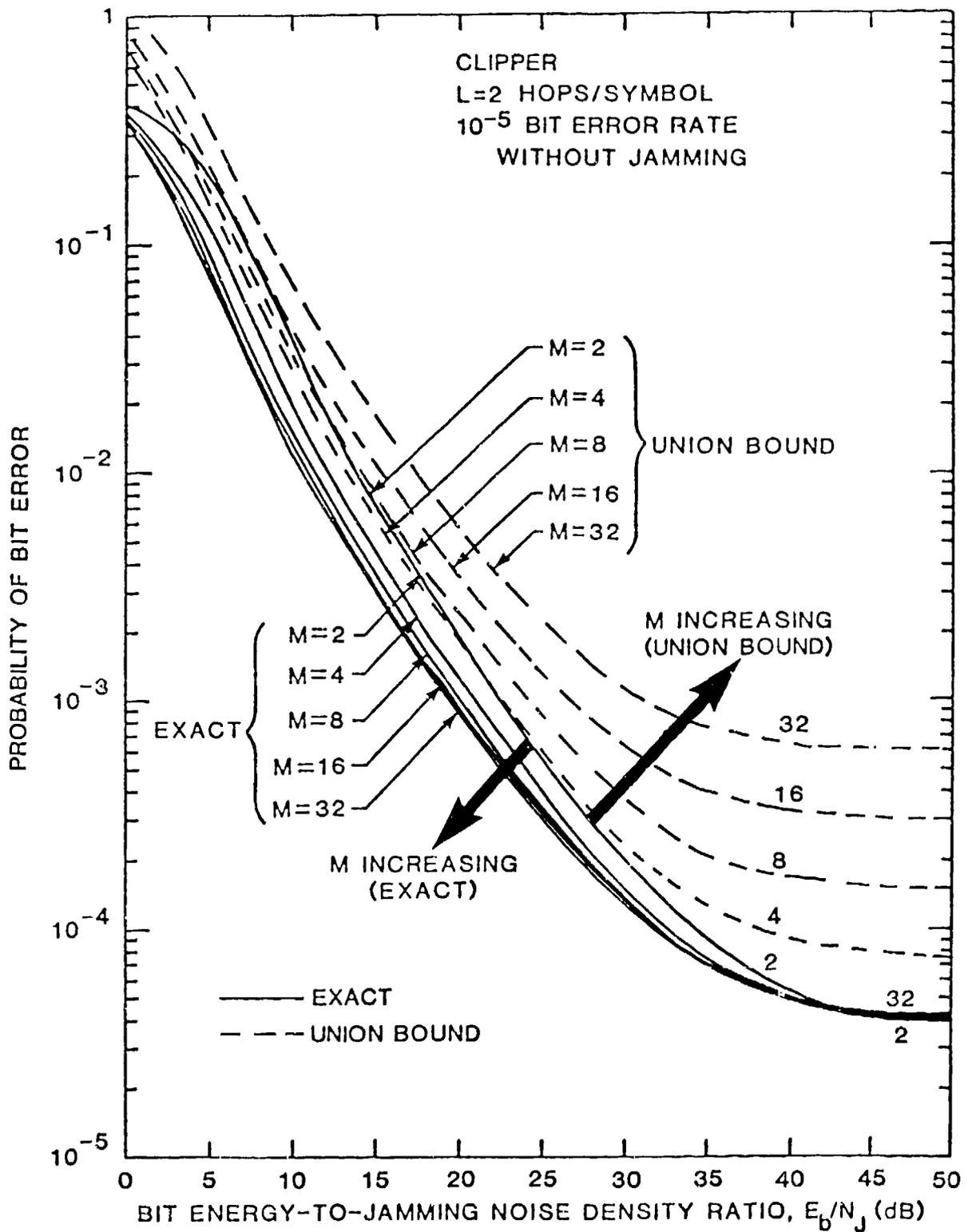


FIGURE 1.1-6 COMPARISON OF EXACT ANALYSES WITH THE UNION BOUNDS FOR THE CLIPPER RECEIVER

enhanced performance using coding. For comparison studies of anti-jam demodulation schemes such as we are proposing, it is sufficient to consider uncoded performance, since the coded system performance is proportional to the uncoded.

For example, for code words using n channel symbols the probability of word error for a bounded-distance decoding algorithm is [9]

$$P_W = \sum_{i=t+1}^n \binom{n}{i} P_S^i (1 - P_S)^{n-i} \quad (1.1-5)$$

where P_S is the uncoded performance in terms of symbol errors and t is the number of correctable errors. This word error probability can be translated into an equivalent information bit error probability by a formula appropriate to the particular coding and decoding algorithms.

1.2 ECCM PROCESSING

Once the FH/MFSK or FH/RMFSK waveform has been dehopped at the intended receiver, the L hops constituting the MFSK symbol can be combined in several ways. It has been shown [10] that the conventional method of summing up the (non-coherent) L hop energies, although effective against fading, produces a BER which increases with L against optimum partial-band noise jamming. Therefore, a number of non-linear combining schemes have been studied, based on weighting the dehopped and envelope-detected hops in some fashion to discriminate against those hops which have been jammed [1, 11, 12].

Using these nonlinear combining schemes it has been shown for FH/MFSK that the use of $L > 1$ hops per symbol can be understood as providing a kind of diversity improvement against the jamming, depending on the system noise level.

It has not been determined how a FH/RMFSK waveform with L hops per symbol will perform against optimum partial-band noise, whether using conventional or nonlinear soft-decision combining of the hops.*

1.2.1 Examples of Receiver Effectiveness Computations for FH/MFSK

Under contract to the Office of Naval Research, LAI has studied in great detail the uncoded performances of frequency hopped BFSK and MFSK communication systems under optimum partial-band noise jamming [1, 10, 11, 13, 14, 15]. The focus of these efforts has been to determine both the optimum partial-band jamming strategy and the most effective anti-jam receiver processing schemes for this type of modulation, using exact analyses which include the system's thermal noise. One of the chief results of our work has been the discovery that conclusions drawn from previous, approximate studies neglecting thermal noise are not strictly valid. It had been commonly asserted that the use of multiple hops per symbol in FH/MFSK systems provides a diversity gain improvement against optimum partial-band jamming in much the same way that it does against the effect of fading on the signal. We have been able to show that this improvement does not exist for the conventional (linear combining) receiver, and we have demonstrated quantitatively

*In a recent paper [16], FH/RMFSK performance with L hops per symbol has been shown for a receiver using hard decisions. The binary hard-decision case was also analyzed in [17].

that a limited improvement holds for certain nonlinear hop combining receiver processing schemes ("metrics"), as a function of the system's thermal noise.

A generic model of FH/MFSK square-law receivers is given in Figure 1.2-1. Among the processing schemes, represented by the function $f_k(\cdot)$ in the figure prior to the accumulation of soft decision statistics $\{z_m\}$, are those listed in Table 1.2-1. The performance of the conventional, linear combining receiver in optimum partial-band noise jamming was calculated directly and compared to that of the three nonlinear combining receivers. For the calculation, the bit error probabilities were expressed by

$$P_b(e) = \sum_{\ell=0}^L p_{\ell} P_b(e|\ell), \quad (1.2-1)$$

where p_{ℓ} is the probability that ℓ out of L hops constituting a given symbol are jammed, and $P_b(e|\ell)$ is the bit error probability given that ℓ hops are jammed. For conventional FH/MFSK, we have assumed that

$$p_{\ell} = \binom{L}{\ell} \gamma^{\ell} (1-\gamma)^{L-\ell} \quad (1.2-2)$$

based on all of the M symbol frequency slots being jammed on a given hop, with probability γ (the fraction of the system bandwidth which is jammed), or none of them being jammed, with the probability $1-\gamma$.

For each receiver type and values of E_b/N_0 and E_b/N_J , the maximum bit error probability was found as a function of γ , the partial-band jamming

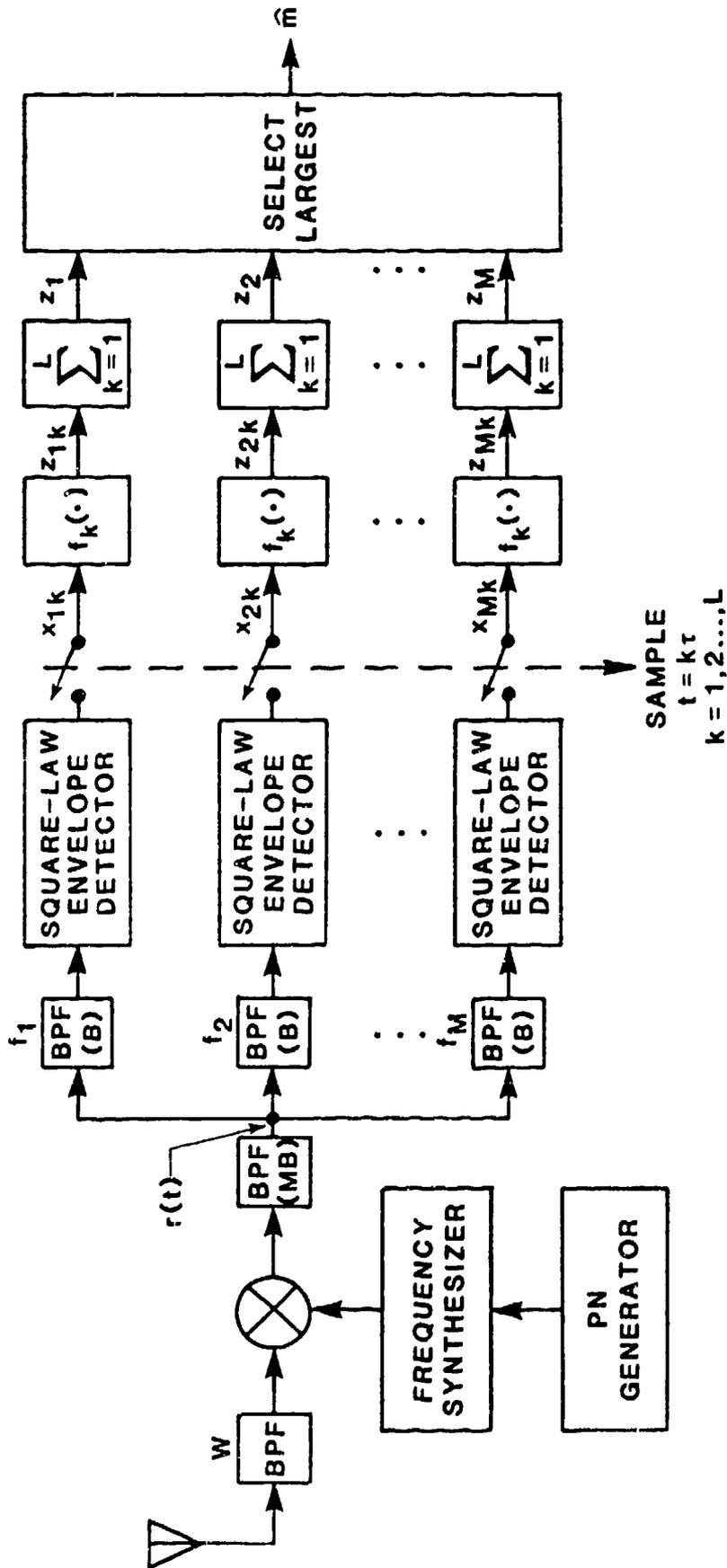


FIGURE 1.2-1 FH/MFSK SQUARE-LAW GENERIC RECEIVER

TABLE 1.2-1
 DESCRIPTIONS OF THE RECEIVERS

RECEIVER TYPE	SPECIFICATION OF $z_{ik} = f_k(x_{ik}), i=1,2,\dots,M$	REMARKS
LINEAR COMBINING RECEIVER	$z_{ik} = x_{ik}$	Direct Connection (Linear Combining)
CLIPPER RECEIVER	$z_{ik} = \begin{cases} x_{ik}, & x_{ik} \leq n \\ n, & x_{ik} > n \end{cases}$	Soft Limiter (Nonlinear Combining)
AGC RECEIVER	$z_{ik} = x_{ik} / \sigma_k^2$ $\sigma_k^2 = \begin{cases} \sigma_N^2, & \text{if not jammed} \\ \sigma_N^2 + \sigma_J^2, & \text{if jammed} \end{cases}$ <p>($\sigma_k^2 = \text{measured}$)</p>	Adaptive Gain Control (Nonlinear Combining)
SELF-NORMALIZING RECEIVER	$z_{ik} = \frac{x_{ik}}{\sum_{i=1}^M x_{ik}}$	Practical Realization of AGC Using In-Band Measurements

fraction. These calculations revealed significant differences among the receiver types in the optimum value of γ as well as in the bit error probability. For example, in Figure 1.2-2, we show that for $M=8$ that the jammer's optimum γ is much more sensitive to the value of L , the number of hops per MFSK symbol, for the AGC receiver than for the clipper receiver. Therefore the jammer must have more accurate information on the modulation parameters in order to be as effective as possible against the AGC receiver.

Another typical result is the comparison shown in Figure 1.2-3, also for $M=8$, and for $L=2$ hops per symbol. We see that the (ideal) AGC form of ECCM receiver processing is significantly better at combatting the effects of the jamming, and that the clipper receiver also improves the BER, but not as much.

Figure 1.2-4 shows the effect of increasing E_b/N_0 so as to provide a lower bit error probability in the absence of jamming for the AGC receiver with $M=4$ and L as a parameter. We see that under these conditions, the optimum choice of L includes higher values of the number of hops per symbol before increased noncoherent combining loss dominates and forces a choice of a lower value of L .

Figure 1.2-5 illustrates the performance as $E_b/N_0 \rightarrow \infty$, i.e. no thermal noise, for FH/BFSK (i.e. $M=2$). We see that in the absence of thermal noise, the optimum value of L increases without limit as E_b/N_0 increases. A similar result holds for the case of $M>2$.

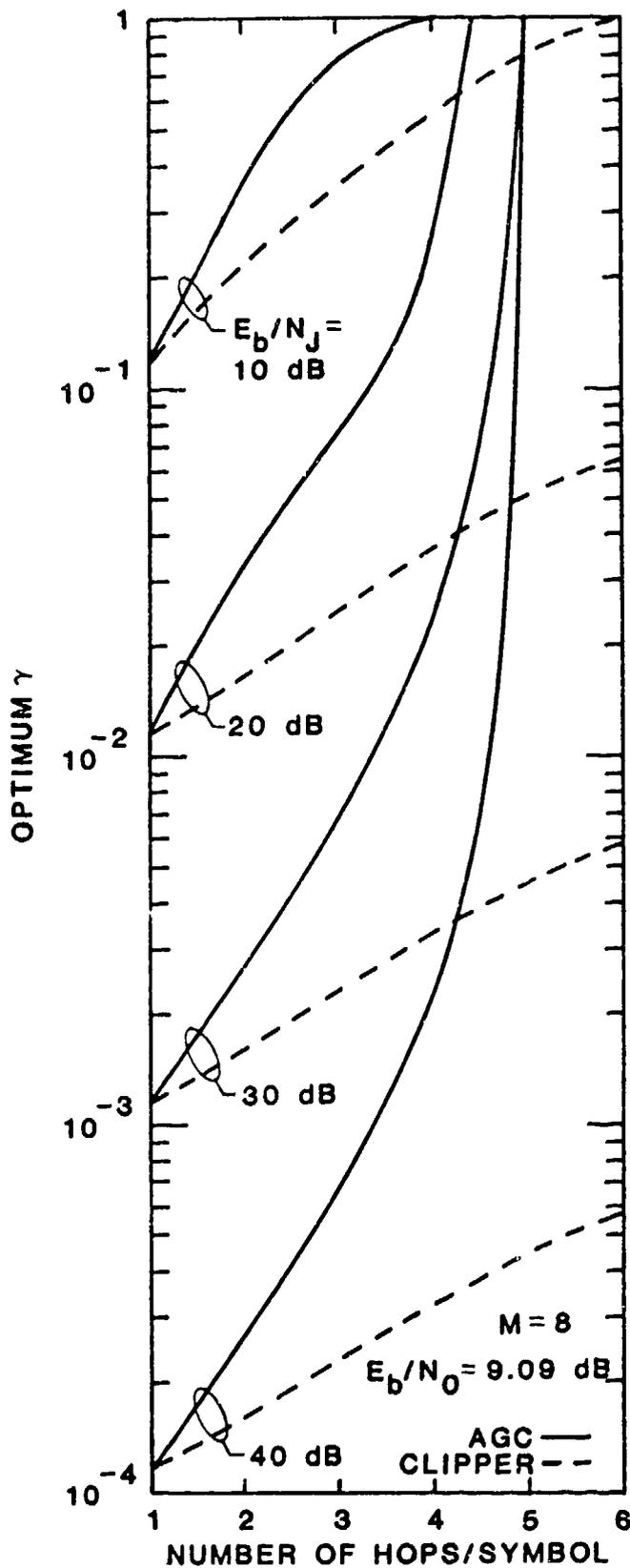


FIGURE 1.2-2 OPTIMUM JAMMING FRACTION (γ) VS. NUMBER OF HOPS/SYMBOL (L) FOR THE AGC AND CLIPPER FH/MFSK ($M=8$) RECEIVERS WHEN $E_b/N_0 = 9.09$ dB WITH E_b/N_J AS A PARAMETER (FOR 10^{-5} ERROR RATE WITHOUT JAMMING)

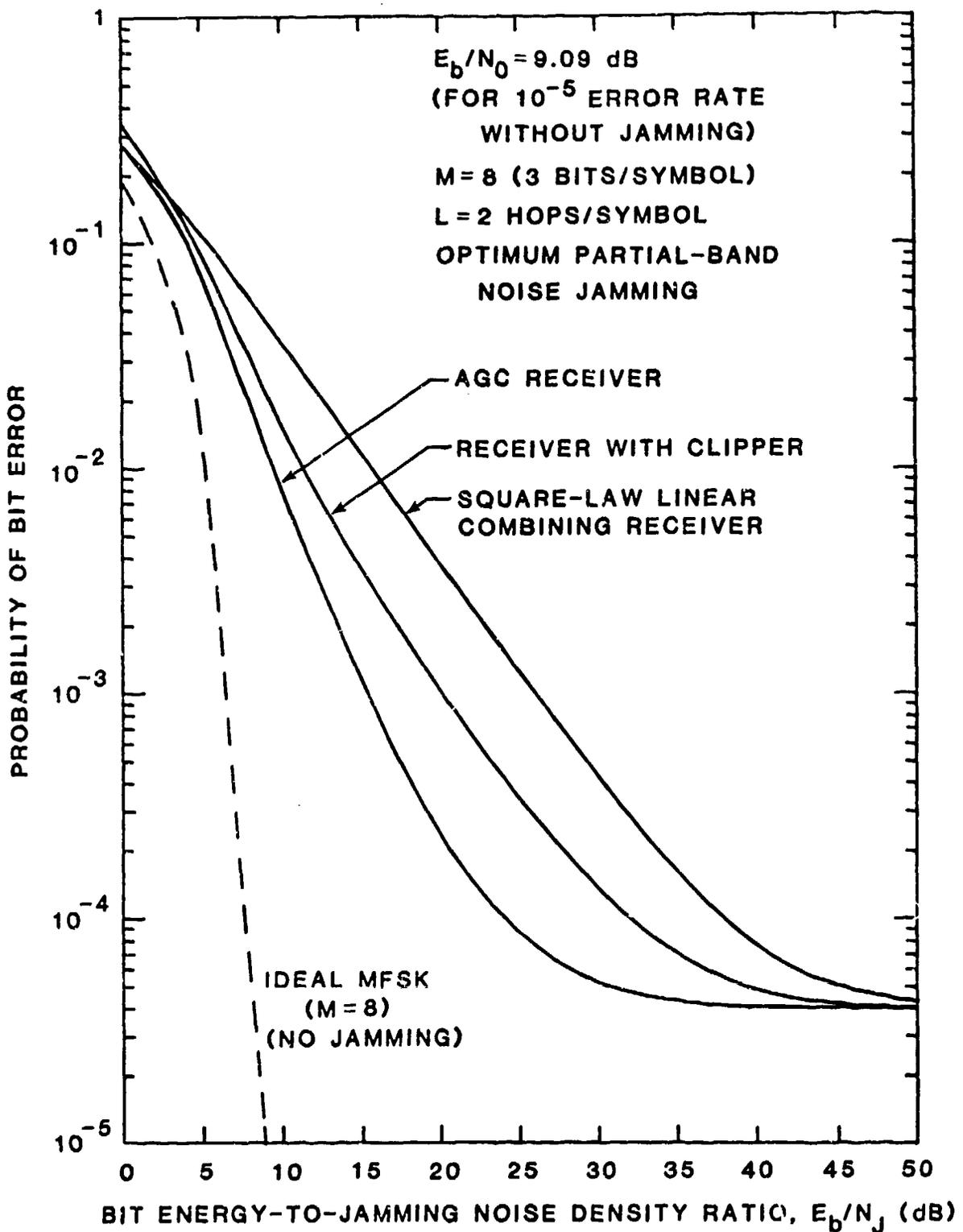


FIGURE 1.2-3 OPTIMUM PARTIAL-BAND NOISE JAMMING PERFORMANCE OF FH/MFSK ($M=8$) SQUARE-LAW COMBINING RECEIVERS FOR $L=2$ HOPS/SYMBOL WHEN $E_b/N_0=9.09$ dB (FOR IDEAL MFSK ($M=8$) CURVE THE ABSCISSA READS E_b/N_0)

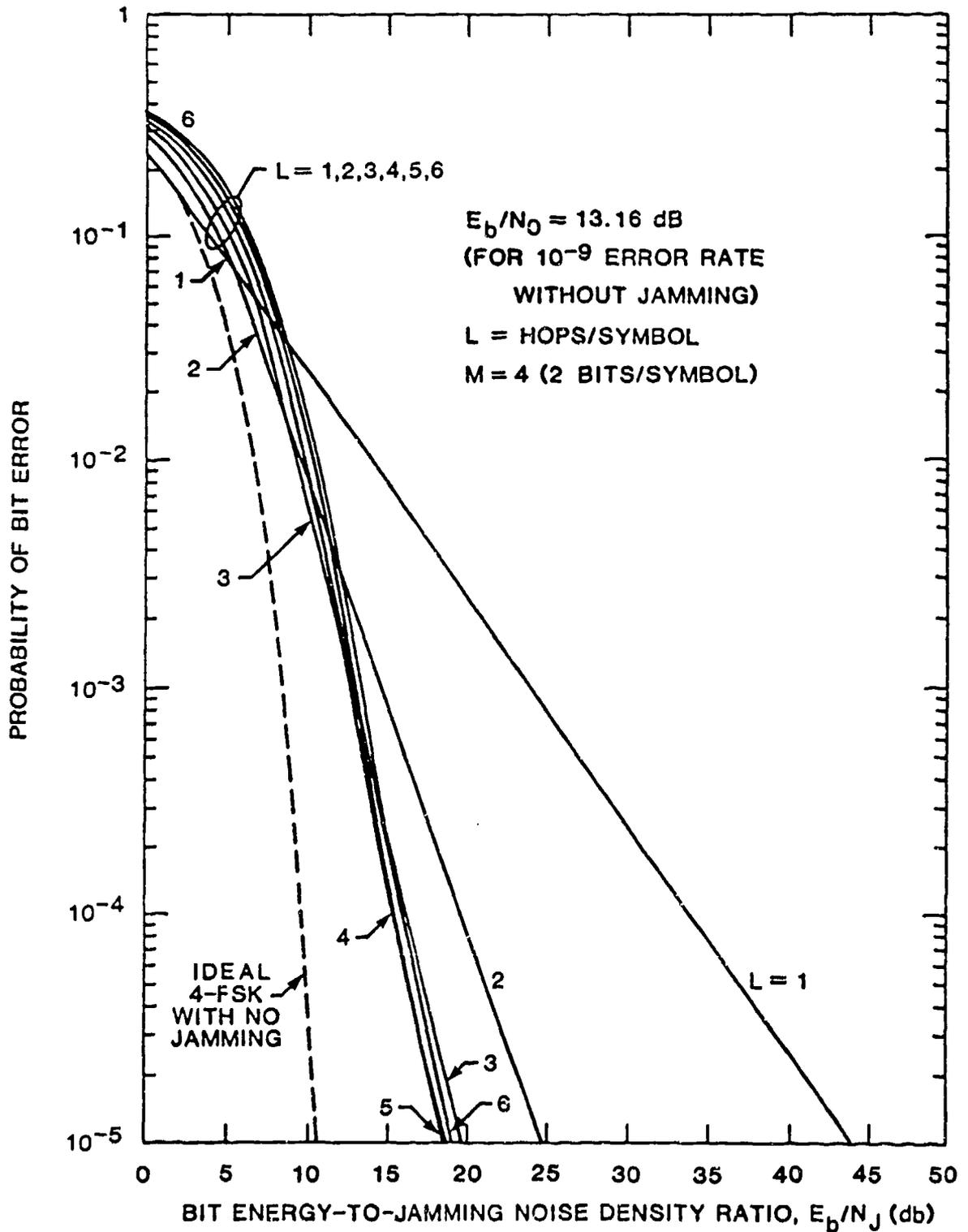


FIGURE 1.2-4 OPTIMUM JAMMING PERFORMANCE OF THE AGC FH/MFSK ($M=4$) RECEIVER WHEN $E_b/N_0 = 13.16 \text{ dB}$ WITH THE NUMBER OF HOPS/SYMBOL (L) AS A PARAMETER (FOR IDEAL MFSK ($M=4$) CURVE THE ABSCISSA READS E_b/N_0)

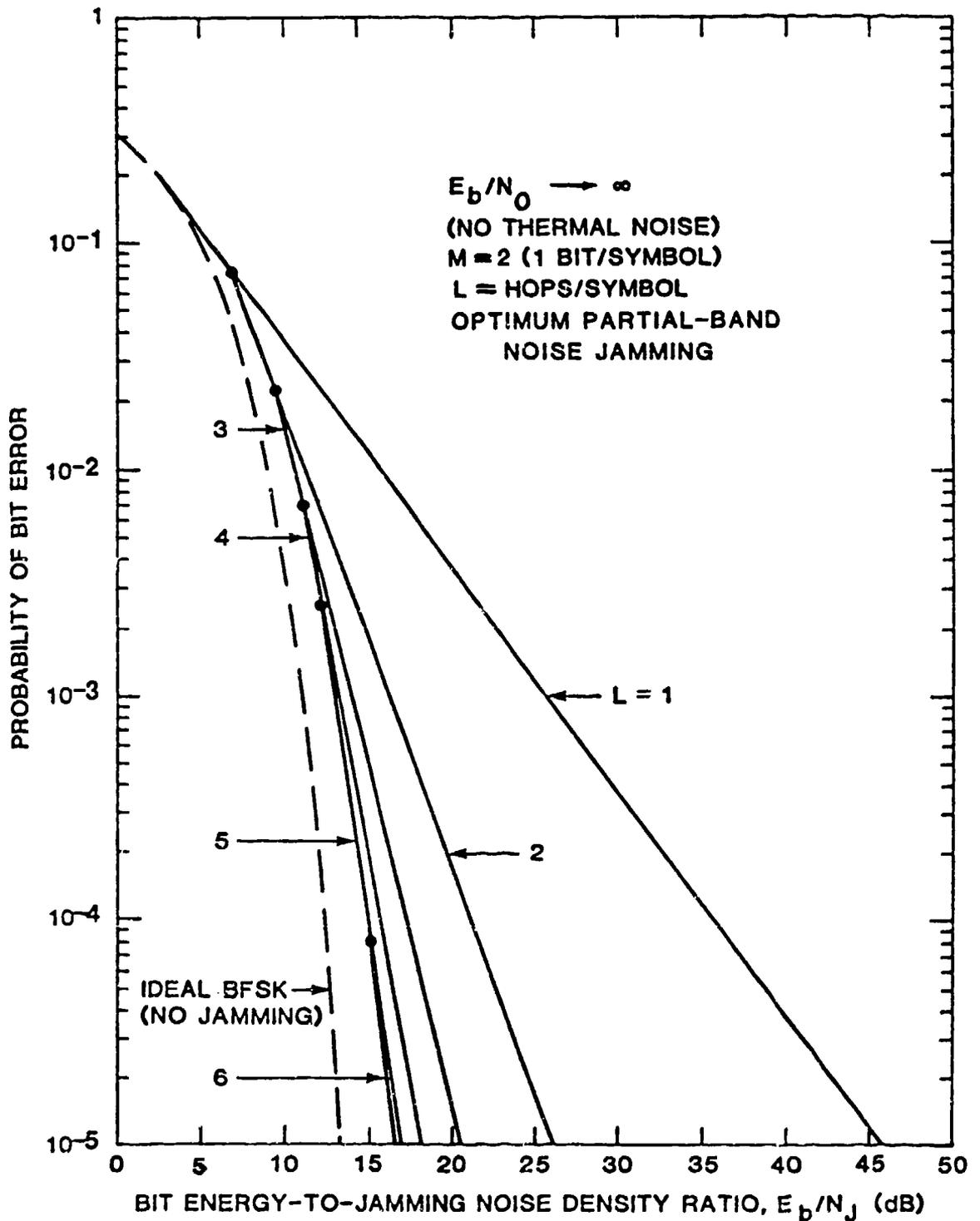


FIGURE 1.2-5 OPTIMUM JAMMING PERFORMANCE OF THE AGC RECEIVER FOR BFSK/FH WITH THE NUMBER OF HOPS/BIT (L) AS A PARAMETER WHEN THERMAL NOISE IS ABSENT (FOR IDEAL BFSK CURVE THE ABSCISSA READS E_b/N_0)

Our exact calculations permit the construction of composite curves such as illustrated in Figure 1.2-6, in which the performance of the AGC receiver processing scheme for FH/BFSK (M=2) is shown for the optimum L values at different thermal noise levels. It is seen that for $E_b/N_0 > 15$ dB, the use of the proper number of hops per bit enables the communication systems to recover the performance of unjammed BFSK to within 3 dB of SNR. The results for $E_b/N_0 < 15$ dB are very sensitive to thermal noise, and had not been predicted by other workers, who ignored thermal noise.

1.2.2 Impact of Random FH/MFSK (FH/RMFSK) on Analysis

Evaluation of the BER performance of ECCM receiver processing schemes becomes significantly more complex for $M > 2$ and $L > 1$ when the MFSK symbol frequency assignments are not contiguous but each randomly chosen to be anywhere in the hopping band. The complexity consists in there being many more jamming events than those reflected in equation (1.2-1), since now on each hop there can be from 0 to M of the dehopped symbol frequency slots jammed on a given hop (rather than 0 or M). The probability of bit error expression accordingly must be generalized, giving

$$P(e) = \sum_{\ell_1=0}^L \sum_{\ell_2=0}^L \cdots \sum_{\ell_M=0}^L \Pr(\ell_1, \ell_2, \dots, \ell_M) P(e|\ell_1, \ell_2, \dots, \ell_M) \quad (1.2-3)$$

in which the number of jammed hops in each symbol channel is explicitly enumerated and accounted for in the conditional probability of error calculations.

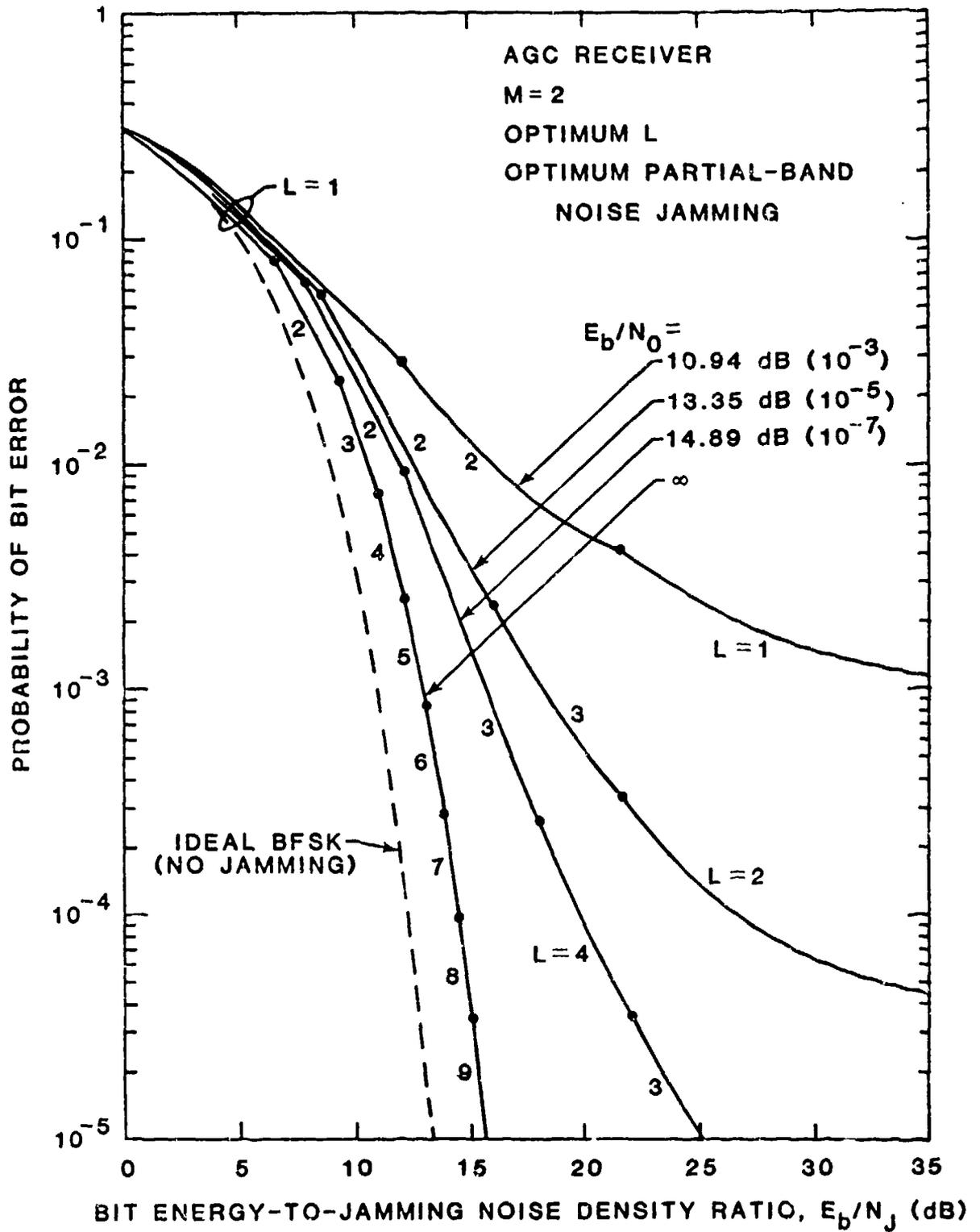


FIGURE 1.2-6 PROBABILITY OF BIT ERROR VS. E_b/N_j FOR AGC FH/BFSK RECEIVER USING OPTIMUM NUMBER (L) OF HOPS/BIT, FOR DIFFERENT VALUES OF E_b/N_0 IN WORST-CASE PARTIAL-BAND NOISE JAMMING

From equation (1.2-3) it is apparent that up to $(L + 1)^M$ jamming events may be distinguished, if it can be assumed that the symbol decision is affected only by the total numbers of jammed hops $\{\ell_m\}$ in the M dehopped channels, rather than by the individual hop patterns. The sheer number of events can therefore become the major factor influencing the magnitude of the receiver effectiveness evaluation task in terms of computational effort. LAI has had experience in the computation of similar expressions in the connection with the evaluation of tone jamming effects on FH/MFSK systems [1].

1.3 SUMMARY OF REPORT

In this section, we will first give a general description of the work. We then summarize the report organization and major findings.

1.3.1 General Description of Work and Approach.

In Sections 1.1 and 1.2 we discussed the fundamental issues concerning ECCM systems and ECCM processing. Now, we treat the more specific ECCM system which we have studied, namely FH/RMFSK in the presence of partial-band noise jamming.

1.3.1.1 Receiver models studied.

A generic soft-decision receiver structure for an FH/RMFSK waveform is shown in Figure 1.3-1. The incoming waveform is dehopped by mixing it separately with M hopping local oscillators controlled by replicas of the M possible hopping sequences available for transmission by the transmitter.

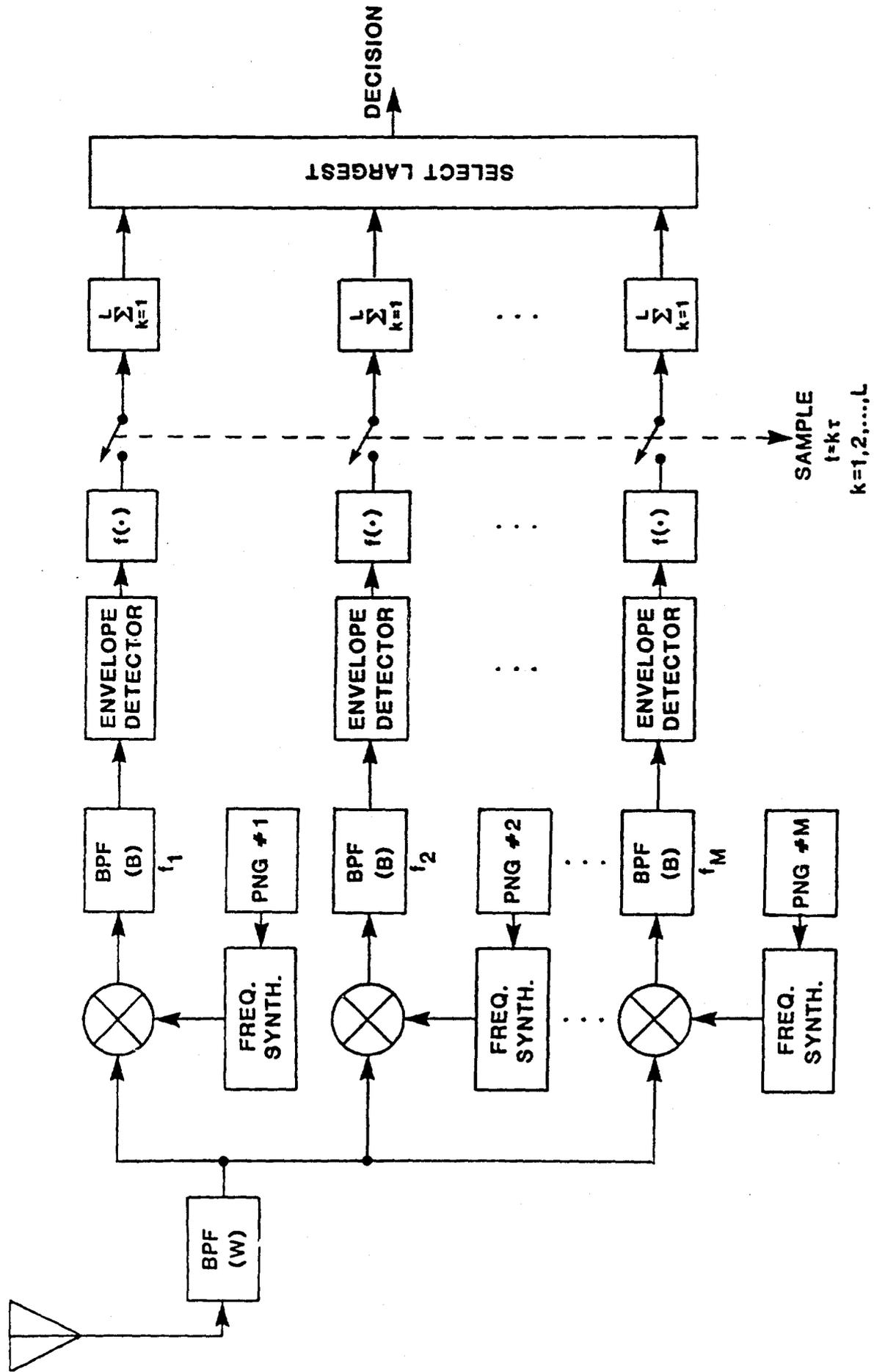


FIGURE 1.3-1 SOFT-DECISION RECEIVER FOR FH/RMFSK

Thermal noise with power spectral density N_0 is present over the entire bandwidth W . A fraction, γ , of the band is jammed by bandlimited white Gaussian noise of power spectral density N_J/γ , where $N_J \triangleq J/W$ with J being the total jammer power. The jamming fraction, γ , is constrained to the range $0 < \gamma \leq 1$.

The relation between the jammed bandwidth γW and the FH/RMFSK waveform is illustrated in Figure 1.3-2. On any given hop, anywhere from 0 to M of the possible signalling frequencies may have hopped into the jammed portion of the band; thus a multitude of jamming events may occur. Let the L hops for a given symbol be referred to individually by the index k ($k = 1, 2, \dots, L$). The jamming events for the k th hop can be described in terms of which of the M symbol frequencies are jammed, and which are not. In general there are 2^M possibilities for a given hop, which we may specify by the indicator vector

$$\underline{v}_k = (v_{1k}, v_{2k}, \dots, v_{Mk}) \quad (1.3-1)$$

where

$$v_{mk} = \begin{cases} 1 & \text{if symbol slot } m \text{ is jammed on hop } k \\ 0 & \text{if not;} \end{cases} \quad m = 1, 2, \dots, M; \quad k = 1, 2, \dots, L. \quad (1.3-2)$$

For the L hops comprising a symbol, there are 2^{ML} possible jamming events, and these can be specified individually by the $M \times L$ indicator matrix $[v] \equiv [v_{mk}]$.

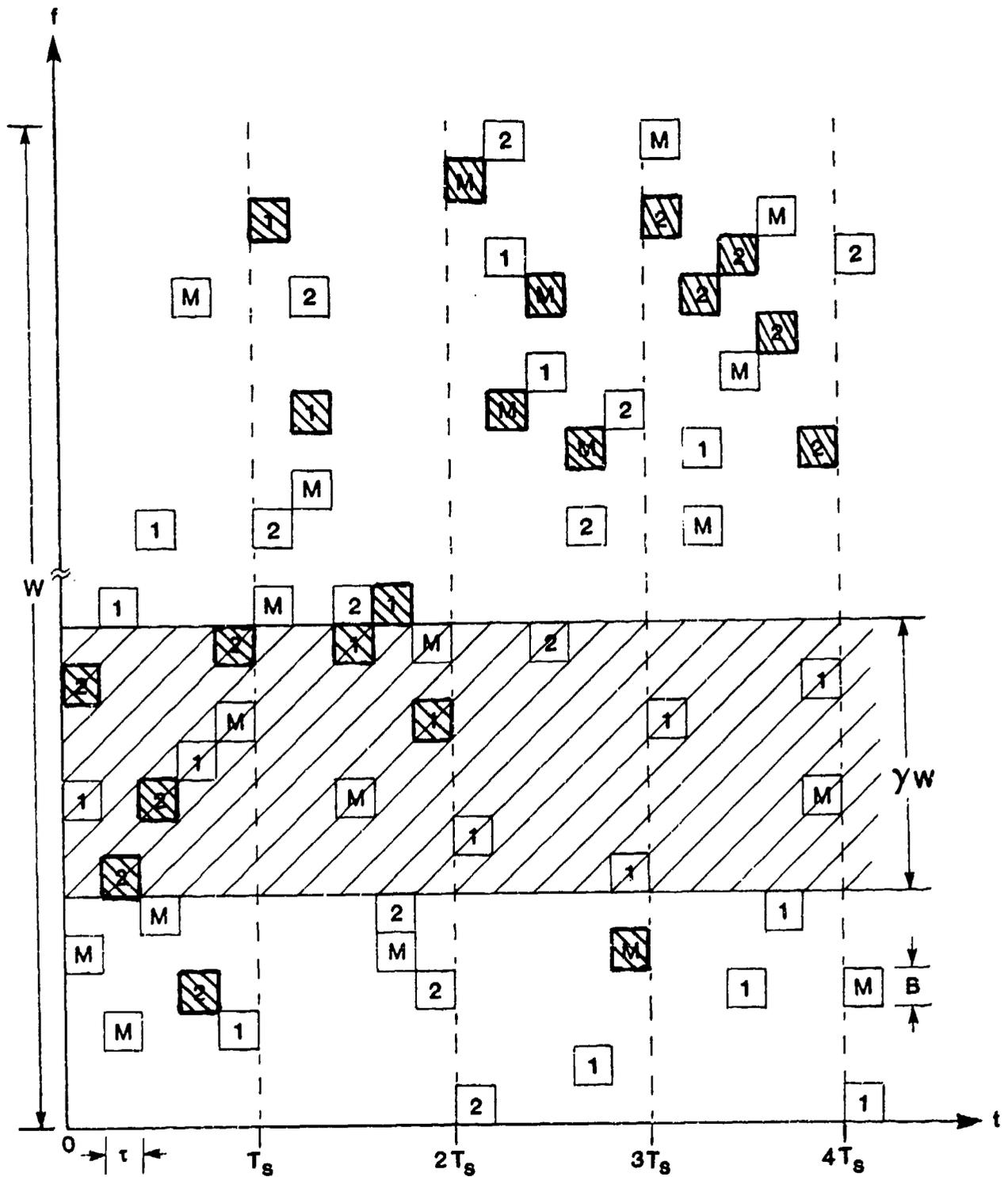


FIGURE 1.3-2 PARTIAL-BAND JAMMING OF FH/RMFSK

Each dehopped channel, corresponding to one of the M possible symbols, is then passed through a bandpass filter of width B Hz and the filter output is envelope detected. The output of each linear envelope detector is subjected to a function $f(\cdot)$; the form of this function defines the particular receiver structure.* Table 1.3-1 gives the forms of $f(\cdot)$ for the several receiver structures we include in the study. The modified envelopes are sampled once per hop and the samples in each channel are summed over the L hops comprising a symbol. The largest of these sums is selected and the index identifying the channel in which it occurred is outputted as the symbol decision.

As an alternative to the soft-decision receiver scheme described above, we may also consider the hard-decision receiver structure which is shown in Figure 1.3-3. The processing in this hard-decision receiver is identical with the soft-decision receiver up to the outputs of the samplers. In the hard-decision receiver, unlike the soft-decision receiver, the samples are not summed; rather, a symbol decision is made each hopping interval, giving a sequence of L decisions. These L decisions may be considered as a noise-corrupted received code-word in an M -ary repetition code wherein the transmitted symbol is repeated L times; thus the sequence is fed into an L -hop M -ary repetition code decoder which delivers the final decision as to which symbol was transmitted.

1.3.1.2 Jamming model and measure of effectiveness.

The partial-band noise jamming model was shown in Figure 1.1-2.

*As long as $f(\cdot)$ is a memoryless transformation, the order of applying $f(\cdot)$ and the sampling may be interchanged without altering the receiver's performance.

TABLE 1.3-1
RECEIVER PROCESSING FUNCTIONS STUDIED

RECEIVER TYPE	$f(\cdot)$
Square-Law Linear Combining	$f(x_i) = x_i^2$
Square-Law with Clipper	$f(x_i) = \begin{cases} x_i^2, & x_i < \sqrt{n} \\ n, & x_i \geq \sqrt{n} \end{cases}$
Square-Law AGC	$f(x_i) = x_i^2 / \sigma_i^2$
Self-normalizing receiver	$f(x_i) = \frac{x_i^2}{\sum_{j=1}^M x_j^2}$

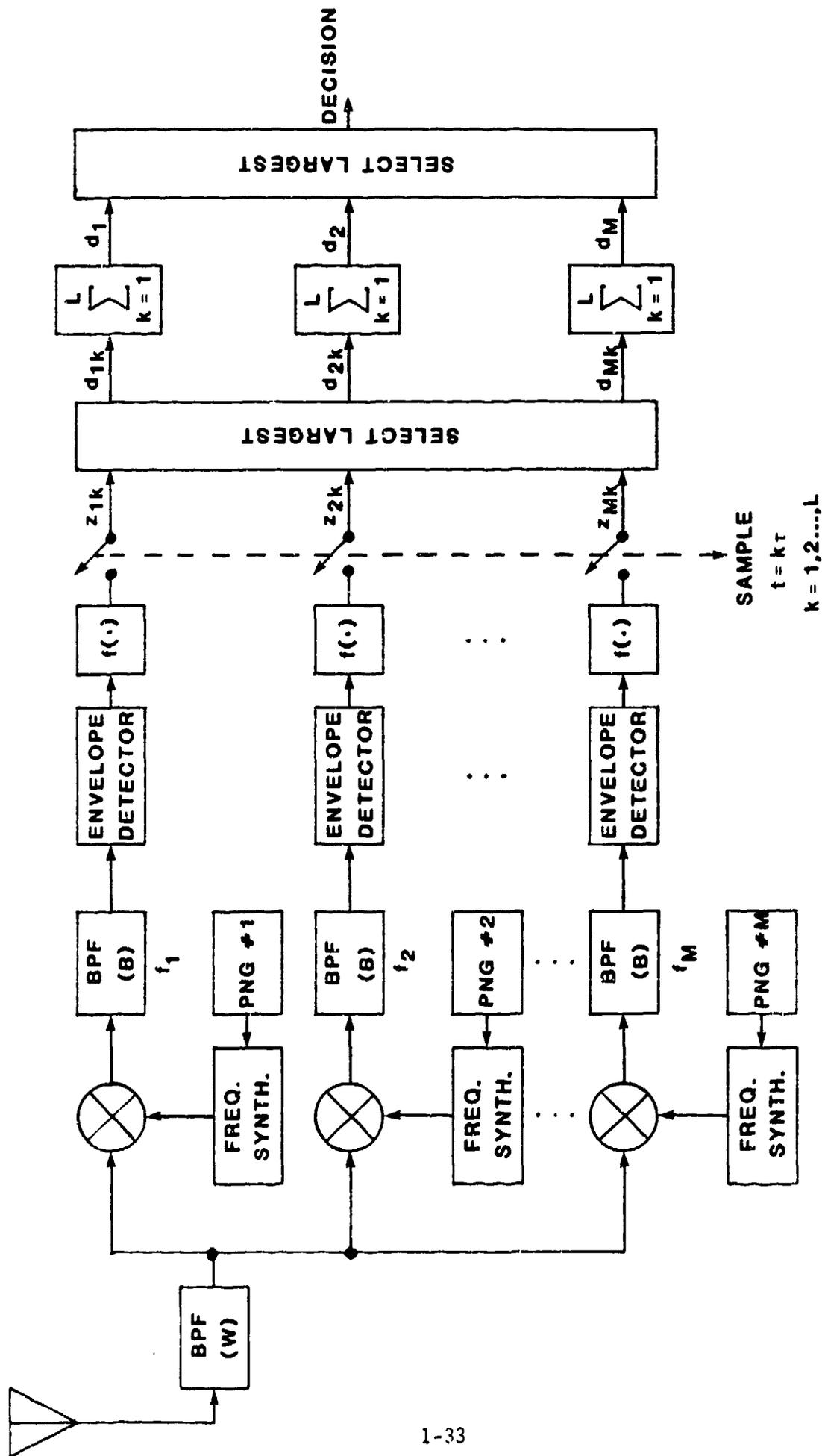


FIGURE 1.3-3 HARD-DECISION RECEIVER FOR FH/RMFSK

The measure of the effectiveness of the jammer is the degradation of the communicator's bit error probability inflicted by the presence of the jamming. Since the bit error probability $P_b(e)$ will depend upon the jamming event, we must average the error probability over the jamming events. Thus, the measure of effectiveness of the jamming is

$$P_b(e; E_b/N_0, E_b/N_j, \gamma, M, L) = \sum_{[v]} P_b(e; E_b/N_0, E_b/N_j, \gamma, M, L|[v])\pi_L[v]$$

where $\pi_L[v]$ is the probability of jamming event $[v]$ occurring over the L hops of the M -ary symbol. Thus the required analysis may be divided into two parts: determination of $\pi_L[v]$ and determination of $P_b(e; E_b/N_0, E_b/N_j, \gamma, M, L|[v])$. These two parts can then be combined to perform the final optimization, namely finding the receiver performance under the optimum jamming fraction γ , $\max_{\gamma} P_b(e; \gamma)$.

1.3.1.3 Organization of report.

In Section 2 we address parts of the analysis considered preliminary or containing aspects common to the several receiver types. This material includes enumeration of jamming events and analysis of their probabilities, as well as an analysis of the hard-decision receiver.

Sections 3 to 6 are devoted, respectively, to analysis and numerical results for the worst-case partial-band noise jamming error performances of FH/RMFSK using the linear combining receiver, the adaptive gain control (AGC) type receivers, the clipper receiver, and the self-normalizing receiver.

Section 7 first provides analysis and results for the performance of FH/RMFSK in follow-on noise jamming, then comparisons of RMFSK receivers

with regard to their overall relative error performance, their performances in both RMFSK and MFSK, and their success in using diversity (multiple hops per symbol) to mitigate jamming effects.

Section 8 considers issues related to implementation of the FH/RMFSK receivers, including a discussion of possible measurement approaches to support the ECCM weighting schemes, and an assessment of the effect of using practical measurements (instead of a priori information) on the system performance. Conclusions and recommendations growing out of our study are included in Section 8 also.

1.3.2 Summary of Findings.

Here we only briefly cite the more significant findings from our study; more detailed information is contained throughout the report.

The overall significance of the work we have accomplished may be described as follows: For the first time, the expected performance of an FH/RMFSK system using L hops per symbol and soft decisions, in both thermal noise and worst-case partial-band jamming noise, has been derived and calculated. Moreover, we have demonstrated through direct analysis and calculation of bit error rate (BER) that the performances of certain practical soft-decision ECCM receivers (using no a priori or side information) are quite acceptable, being very close to those for idealized receivers (using a priori information on received noise and jamming conditions). While we have shown that the hard-decision receiver does implement a form of ECCM processing (not previously shown) against PBNJ in the most simple manner, it cannot be

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considered a viable alternative unless the system's unjammed SNR is quite high.

Specifically, we find that:

(a) Generally random frequency hopping MFSK is more vulnerable to partial-band noise jamming than is conventional FH/MFSK for $M > 2$ and $L > 1$. However, for certain diversity weighting schemes the increased vulnerability is small enough to justify saying that the two hopping systems achieve comparable performance for $M=2$ or 4. For one combining scheme studied, the self-normalizing receiver, FH/RMFSK performs better than FH/MFSK for $M=2$.

(b) A diversity effect for L hops/symbol is observed for RMFSK using nonlinear hop combining, in the same manner as for MFSK and subject to the same condition that thermal noise is relatively small.

(c) Using optimum diversity values, if thermal noise is negligible ($E_b/N_0 \geq 20$ dB), FH/RMFSK with ideal nonlinear combining can exhibit a nearly exponential dependence upon E_b/N_J , as opposed to an inverse linear one for no diversity; the jamming then is limited to inflicting about a 4 dB loss in system performance. However, this effect is very sensitive to the amount of thermal noise present, since the jammed BER cannot be better than the unjammed error, and the use of diversity tends to degrade the no-jamming performance due to noncoherent combining losses.

(d) Simple nonlinear combining receivers can duplicate the ideal receiver optimum diversity performance with about a one-dB loss when $E_b/N_0 > E_b/N_J$; the hard decision receiver can approach to within 2 dB with a sufficiently high E_b/N_0 .

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2.0 PRELIMINARY ANALYSES

The general approach to be followed in obtaining the probability of error for the multi-hops/symbol FH/RMFSK communications system under partial band noise jamming is to expand the total probability of error in terms of individual jamming events:

$$\begin{aligned} P(e) &= \sum_{\text{jamming events}} P(e, \text{jamming event}) \\ &= \sum_{\text{jamming events}} \Pr(\text{jamming event}) P(e|\text{jamming event}), \end{aligned}$$

where $P(e|\text{jamming event})$ is the probability of error conditioned upon the occurrence of a particular jamming event. In Section 2.1, we consider a general formulation for this conditional probability, and in Section 2.2 the jamming events and their probabilities are developed. The computational procedures necessary for efficient evaluation of the error probability are discussed in Section 2.3. These analyses and procedures are applied to specific receiver structures beginning in Section 3.

2.1 CONDITIONAL PROBABILITY OF ERROR

The generic form of the receiver to be analyzed for reception of FH/RMFSK is shown in Figure 2.1-1. In effect it is an M-channel receiver with IF frequencies f_1, f_2, \dots, f_M ; M pseudorandom sequence generators, assumed to be in synchronism with the transmitter, command the frequency synthesizers used to tune the M channels to B-Hz wide frequency slots, one of which will be occupied by signal energy on a given hop. The message information is

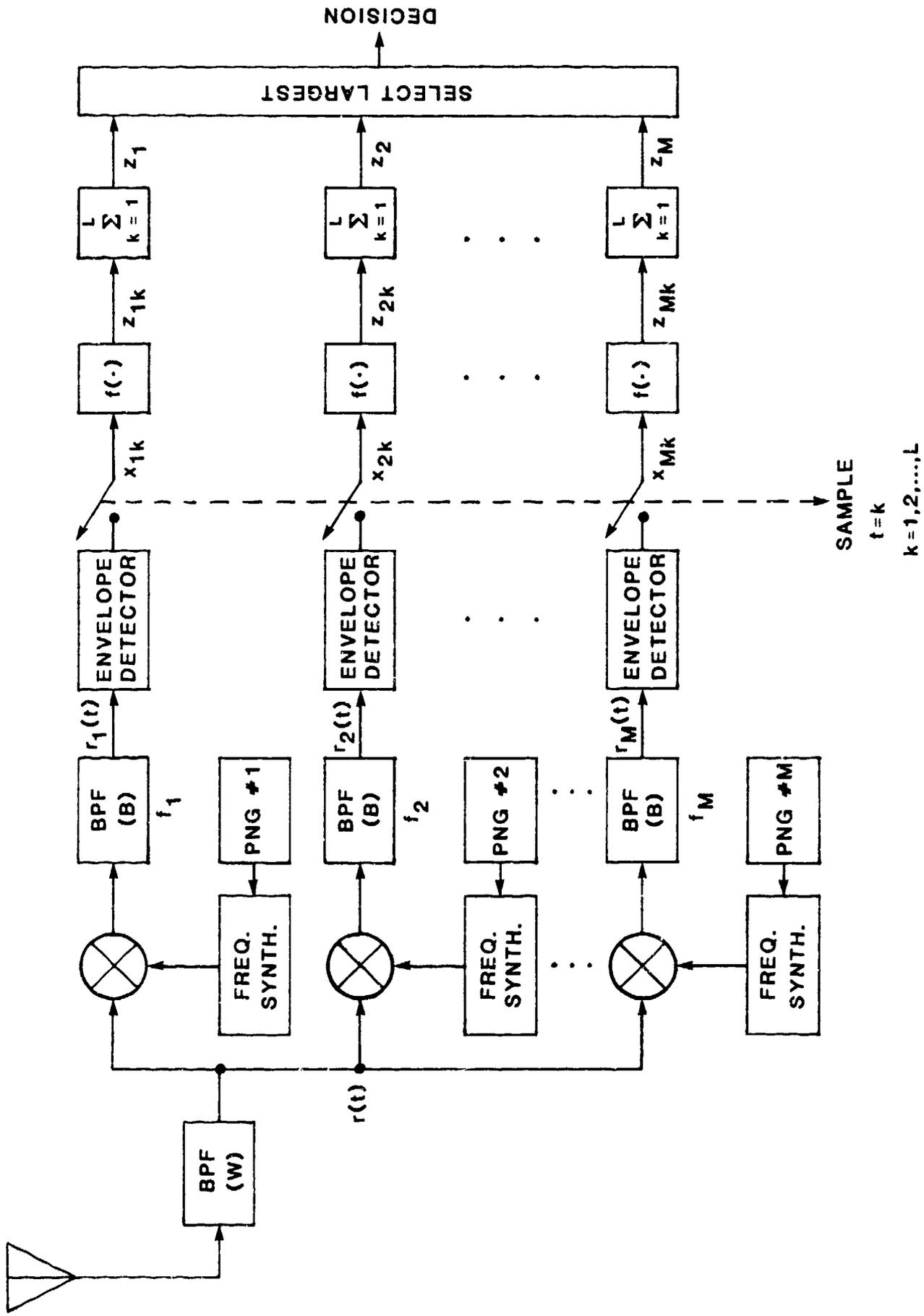


FIGURE 2.1-1 GENERIC RECEIVER FOR FH/RMFSK

conveyed through selection of the (randomly-shifted) IF (symbol) frequency for transmission of signal energy. In order to determine at which symbol frequency the signal is present, each of the IF waveforms $r_1(t), r_2(t), \dots, r_M(t)$ is first subjected to envelope detection, then sampled before processing through memoryless, possibly nonlinear devices with transfer functions $f(\cdot)$. The outputs of these devices are the per-hop decision statistics $\{z_{mk}\}$, which are accumulated to form the final decision statistics

$$z_m = \sum_{k=1}^L z_{mk}, \quad m = 1, 2, \dots, M. \quad (2.1-1)$$

In the following subsections, we formulate the probability of error associated with the decision performed by the FH/RMFSK receiver, conditioned upon the possible jamming events.

2.1.1 Assumed Signals, Noise and Jamming.

After dehopping, the received signal is assumed equally likely to be present in any one of the M channels for the entire symbol period $T_s = L\tau$, where τ is the hop period and L is the number of hops per MFSK symbol. Without loss of generality, we assume that the signal with power S is in channel 1, or

$$s(t) = \sqrt{2S} \cos(\omega_1 t + \theta_k), \quad (k-1)\tau < t \leq k\tau, \quad k = 1, 2, \dots, L, \quad (2.1-2)$$

where θ_k is an arbitrary carrier phase and $\omega_1 = 2\pi f_1$.

Thermal noise is considered also to be present in each channel, and is assumed to be zero-mean narrowband Gaussian noise with variance $\sigma_N^2 = N_0 B$, where $N_0/2$ is the (two-sided) noise power spectral density and B is the bandwidth of each channel. Thus for no jamming the samples of the M envelope detector outputs on the k th hop are the variables

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$$x_{1k} = \left[\left(\sqrt{2S} \cos\theta_k + n_{c1k} \right)^2 + \left(\sqrt{2S} \sin\theta_k + n_{s1k} \right)^2 \right]^{1/2} \quad (2.1-3a)$$

and

$$x_{mk} = \left(n_{cmk}^2 + n_{smk}^2 \right)^{1/2}, \quad m = 2, 3, \dots, M, \quad (2.1-3b)$$

where n_{cmk} , n_{smk} , $m = 1, 2, \dots, M$; $k = 1, 2, \dots, L$, are the independent noise quadrature components in the channels at the sample times $t_k = k\tau$, with

$$E(n_{cmk}^2) = E(n_{smk}^2) = \sigma_N^2 = N_0B, \quad \text{for all } m, k. \quad (2.1-4)$$

Because the MFSK symbol slots are hopped independently, none, some, or all of the dehopped channels can be jammed on an individual hop. The possible combinations of such events and their probabilities are discussed in Section 2.2.

When jamming noise is present in a channel, it is assumed to be zero-mean narrowband Gaussian noise with variance $\sigma_J^2 = N_J B / \gamma$, where $N_J / 2$ is the (two-sided) noise power spectral density averaged over the system bandwidth; and γ is the fraction of this bandwidth which is jammed. That is,

$$N_J = \frac{J}{W}, \quad (2.1-5)$$

where J is the total jammer power and W is the system bandwidth. When the channels are jammed on the k -th hop, the combination of jamming and thermal noise produces the detector output samples

$$x_{1k} = \left[\left(\sqrt{2S} \cos\theta_k + n_{c1k} + j_{c1k} \right)^2 + \left(\sqrt{2S} \sin\theta_k + n_{s1k} + j_{s1k} \right)^2 \right]^{1/2} \quad (2.1-6a)$$

$$x_{mk} = \left[\left(n_{cmk} + j_{cmk} \right)^2 + \left(n_{smk} + j_{smk} \right)^2 \right]^{1/2}, \quad m = 2, 3, \dots, M, \quad (2.1-6b)$$

where j_{cmk} , j_{smk} , $i = 1, 2, \dots, M$; $k = 1, 2, \dots, L$, are the independent jamming noise quadrature components in the channels at the sample times, with

$$E(j_{cmk}^2) = E(j_{smk}^2) = \sigma_J^2 = N_J B / \gamma, \text{ for all } m, k, \quad (2.1-7)$$

and γ is the fraction of the system bandwidth which is jammed.

In a summary way, we can express the detector output samples by

$$x_{1k} = \sigma_{1k} \left[\left(\sqrt{\frac{2S}{\sigma_{1k}^2}} \cos \theta_k + \mu_{c1k} \right)^2 + \left(\sqrt{\frac{2S}{\sigma_{1k}^2}} \sin \theta_k + \mu_{s1k} \right)^2 \right]^{1/2} \quad (2.1-8a)$$

$$x_{mk} = \sigma_{mk} \left(\mu_{cmk}^2 + \mu_{smk}^2 \right)^{1/2}, \quad m = 2, 3, \dots, M, \quad (2.1-8b)$$

where μ_{cmk} and μ_{smk} are independent, unit-variance, zero-mean Gaussian random variables, and for channel m on hop k ,

$$\sigma_{mk}^2 = \begin{cases} \sigma_N^2 = N_0 B, & \text{unjammed} \\ \sigma_T^2 = \sigma_N^2 + \sigma_J^2 = (N_0 + N_J / \gamma) B, & \text{jammed} \end{cases} \quad (2.1-9a)$$

or, more compactly,

$$\sigma_{mk}^2 = \sigma_N^2 + v_{mk} \sigma_J^2. \quad (2.1-9b)$$

In this last equation $v_{mk} = 1$ if channel m is jammed on hop k , and $v_{mk} = 0$ if not. Thus x_1 is σ_{1k} times a Rician random variable with SNR

$$\rho_k = S / \sigma_{1k}^2, \quad (2.1-10)$$

and x_{mk} , $m > 1$, is σ_{mk} times a Rayleigh random variable.

2.1.2 Conditional Error Probability Formulation

Assuming equally likely M -ary symbols, we may write the conditional symbol error probability as

$$P_s(e|[v]) = P_s(e|[v], m_1 \text{ transmitted}) \quad (2.1-11)$$

in which $[v]$ is a matrix describing the jamming event for L hops, with elements v_{mk} .

For M a power of two ($M=2^K$), the conditional bit error probability is obtained from the conditional symbol error probability using the relation

$$P_b(e|[v]) = \frac{M/2}{M-1} P_s(e|[v]). \quad (2.1-12)$$

Since for $M > 2$ there are many error events but only one correct decision, it is convenient to write the conditional symbol error probability in terms of the probability of a correct decision as

$$\begin{aligned} P_s(e|[v]) &= 1 - P_s(c|m_1, [v]) \\ &= 1 - \Pr\{z_2 < z_1, z_3 < z_1, \dots, z_M < z_1\}. \end{aligned} \quad (2.1-13)$$

In terms of the pdf's for the statistics, this becomes

$$P_s(e|[v]) = 1 - \int_0^\infty d\beta_1 \int_0^{\beta_1} d\beta_2 \dots \int_0^{\beta_1} d\beta_M p_{\underline{z}}(\beta_1, \beta_2, \dots, \beta_M|[v]); \quad (2.1-14)$$

if the decision statistics are independent, then

$$P_s(e|[v]) = 1 - \int_0^\infty d\beta p_{z_1}(\beta|[v]) \prod_{m=2}^M \int_0^\beta d\alpha_m p_{z_m}(\alpha_m|[v]). \quad (2.1-15)$$

For certain receiver structures, the probability distributions of the individual channel statistics $\{z_m\}$ are mutually independent. This relationship causes the conditional probability of error to depend only on the number of hops jammed in each of the M channels, rather than on specific patterns, and greatly reduces the number of distinguishable jamming events.

2.2 ENUMERATION OF DISTINGUISHABLE JAMMING EVENTS

Examination of the conditional error probability expression reveals that the same conditional error will occur for several different values of the fundamental jamming event matrix, $[v]$. Therefore, in terms of error probability, there is a number of distinguishable jamming events which is smaller than the 2^{ML} possible values of the $[v]$ matrix. It is important to identify and enumerate these distinguishable jamming events in order to take advantage of the savings in computation which will result.

In this section we first identify and enumerate the distinguishable jamming events, then investigate methods for calculating their probabilities.

2.2.1 Definition of Distinguishable Jamming Events.

The conditional error probability, as shown in Section 2.1, is a function of given values of the $[v]$ matrix elements v_{mk} , where $m=1$ to M (the number of MFSK channels) and $k=1$ to L (the number of hops per symbol). Often this function can be written

$$\begin{aligned}
 P(e|[v]) &= P(e|v_{11}, v_{12}, \dots, v_{1L}; \dots; v_{M1}, \dots, v_{ML}) \\
 &= f\left(\sum_{k=1}^L v_{1k}, \sum_{k=1}^L v_{2k}, \dots, \sum_{k=1}^L v_{mk}\right).
 \end{aligned}
 \tag{2.2-1}$$

Thus, if we define the row sums

$$\lambda_m \triangleq \sum_{k=1}^L v_{mk},
 \tag{2.2-2}$$

the conditional $P(e)$ is a function only of these sums*, which are to be interpreted as the number of hops jammed in the respective channels. This fact can be expressed by the relation

$$P(e|[v]) = f(\ell_1, \ell_2, \ell_3, \dots, \ell_M) \\ \equiv f(\underline{\ell}) , \quad (2.2-3)$$

where $\underline{\ell}$ is the vector of ℓ_m components.

Since each ℓ_m can take integer values from 0 to L, there $(L+1)^M$ possible jamming events described by the vector $\underline{\ell}$. This is a considerable savings in numbers of jamming events, as illustrated by Table 2.2-1.

TABLE 2.2-1
NUMBER OF JAMMING EVENTS

<u>M</u>	<u>L</u>	<u># {[v] events, 2^{ML}}</u>	<u># {$\underline{\ell}$ events, $(L+1)^M$}</u>
2	1	4	4
	2	16	9
	3	64	16
	4	256	25
4	1	16	16
	2	256	81
	3	4,096	256
	4	65,536	625
8	1	256	256
	2	65,536	6,561
	3	16,777,216	65,536
	4	4,294,967,296	390,625

2.2.2 Smallest Set of Distinguishable Jamming Events.

A further reduction in the number of distinguishable jamming events

*An exception to this condition results for an ECCM processing scheme studied in Section 4.

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results from noticing that permutation of the non-signal channel quantities (l_2 to l_M) does not affect the conditional $P(e)$. That is,

$$\begin{aligned}
 f(\underline{l}) &= f(l_1, l_2, l_3, \dots, l_M) \\
 &= f(l_1, l_M, l_3, l_4, \dots, l_2) \\
 &= f(l_1, l_5, l_{M-1}, l_2, \dots, l_3) \\
 &= \dots \text{ etc.}
 \end{aligned}
 \tag{2.2-4}$$

Thus we can restrict our attention to just one permutation of the set of values $\underline{l} = (l_1, l_2, \dots, l_M)$. A convenient way to represent the permutations is the ordered set of numbers

$$\underline{l}' \triangleq \{l_1, l_2, l_3, \dots, l_M : l_2 \leq l_3 < \dots \leq l_M\}.
 \tag{2.2-5}$$

There are, from Appendix B.3, ..

$$\sum_{l_1=0}^L \sum_{l_M=0}^M \sum_{l_{M-1}=0}^{l_M} \dots \sum_{l_3=0}^{l_4} \sum_{l_2=0}^{l_3} = (L+1) \binom{M-1+L}{M-1}
 \tag{2.2-6}$$

such ordered \underline{l}' vectors, which represent the minimum number of distinguishable events. Example values are given in Table 2.2-2.

TABLE 2.2-2
MINIMUM NUMBER OF DISTINGUISHABLE EVENTS

<u>M</u>	<u>L</u>	<u>#{\underline{l}' events}</u>
2	1	4
	2	9
	3	16
	4	25
4	1	8
	2	30
	3	80
	4	175
8	1	16
	2	108
	3	480
	4	1,650

Each of the distinguishable jamming events represents a certain number of events with identical jamming effects. The number of $\underline{\ell}$ vectors thereby represented by a particular ordered vector $\underline{\ell}'$ is

$$\#(\underline{\ell} \leftrightarrow \underline{\ell}') = \binom{M-1}{n_0, n_1, \dots, n_L} \quad (2.2-7a)$$

$$= \frac{(M-1)!}{n_0! n_1! \dots n_L!} \quad (2.2-7b)$$

where

n_ℓ = number of ℓ_m which equal ℓ ; $\ell=0,1,\dots,L$; $m > 1$

and we have

$$\sum_{\ell=0}^L n_\ell = M-1. \quad (2.2-7d)$$

For example, for $M=8$ and $L=6$, the number of jamming event vectors $\underline{\ell}$ represented by the ordered vector $\underline{\ell}' = (\ell_1; 0, 0, 2, 3, 3, 4, 5)$ is

$$\binom{7}{2, 0, 1, 2, 1, 1, 0} = \frac{7!}{2! 0! 1! 2! 1! 1! 0!} = 1260. \quad (2.2-8)$$

As a check on this enumeration, we find that the total number of $\underline{\ell}$ jamming events is given by

$$\begin{aligned} \#(\underline{\ell}) &= \sum_{\ell_1=0}^L \sum_{\ell_M=0}^L \sum_{\ell_{M-1}=0}^{\ell_M} \sum_{\ell_3=0}^{\ell_4} \sum_{\ell_2=0}^{\ell_3} \binom{M-1}{n_0, n_1, \dots, n_L} \\ &= (L+1) \sum_{\ell_M=0}^L \dots \sum_{\ell_2=0}^{\ell_3} \binom{M-1}{n_0, n_1, \dots, n_L}. \end{aligned} \quad (2.2-9)$$

It can be shown (see Appendix B) that the summation in (2.2-9) is equal to $(L+1)^{M-1}$. Thus the total number of \underline{l} vectors computed by (2.2-9) is $(L+1)^M$, which agrees with our previous enumeration.

2.2.3 Jamming Event Probabilities, Single Hop.

Given a jamming event described by the vector \underline{l} , what is the probability of the event under the random hopping scheme and partial-band noise jamming? To find this answer, we first consider the case of one hop per MFSK symbol, or $L=1$.

The jammer spectrum is assumed to be flat, with one-sided power spectral density $J/\gamma W$, where γ is the fraction of the system bandwidth occupied by the jammer. There are $N=W/B$ possible symbol frequency slots, and it is assumed that M of these slots are assigned randomly to the MFSK symbol on each hop. At the same time the number of slots containing jamming power is

$$q \triangleq \gamma N, \quad (2.2-10)$$

assumed to be an integer. That is, $\gamma = q/N$, with q an integer.

The probability that n of the M symbol slots are jammed on a given hop is

$$\begin{aligned} \pi_n &= \frac{q}{N} \cdot \frac{q-1}{N-1} \cdots \frac{q-n+1}{N-n+1} \cdot \frac{N-q}{N-n} \cdots \frac{N-q-M+n+1}{N-M+1} \\ &= \frac{\binom{N-M}{q-n}}{\binom{N}{q}}, \quad n=0,1,2,\dots,\min(q,M). \end{aligned} \quad (2.2-11)$$

Note that the probability π_n is valid for several different jamming events since

$$n_k = \sum_{m=1}^M v_{mk} \quad (2.2-12)$$

on the k-th hop. In fact there are $\binom{M}{n}$ jamming events for $L=1$ which have probability π_n . Thus

$$\sum_{n=0}^L \binom{M}{n} \pi_n = 1, \quad (2.2-13)$$

as required.

In terms of distinguishable jamming events, we differentiate between whether the signal channel is jammed ($v_{1k} = 1$) or not ($v_{1k} = 0$), and describe single-hop jamming events by the pair of numbers (v_{1k}, r_k) with

$$r_k \triangleq \sum_{m=2}^M v_{mk}, \quad (2.2-14)$$

the number of non-signal channels jammed. We have

$$\Pr\{v_{1k}, r_k\} = \pi(v_{1k} + r_k) \equiv \pi_n, \quad n = v_{1k} + r_k \quad (2.2-15)$$

and

$$\sum_{v_{1k}=0}^1 \sum_{r_k=0}^{M-1} \binom{M-1}{r_k} \pi(v_{1k} + r_k) = 1. \quad (2.2-16)$$

2.2.4 Jamming Event Probabilities For Multiple Hops, Characteristic Function Method.

For $L > 1$ hops per symbol, as discussed in Sections 2.2.1 and 2.2.2, the distinguishable jamming events are described by the vectors $\underline{\ell} = (\ell_1, \ell_2, \dots, \ell_M)$. We note that

$$\underline{\ell} = \sum_{k=1}^L \underline{v}_k \quad (2.2-17)$$

where $\underline{v}_k = (v_{1k}, v_{2k}, \dots, v_{mk})$ is the vector whose elements are the k -th column of the matrix $[v]$ of fundamental jamming events. Since the hopping pattern is assumed to be independent from hop to hop, we may treat $\underline{\ell}$ as the sum of identically distributed discrete random vectors, and find the probabilities of the $\underline{\ell}$ jamming events from the characteristic function of the \underline{v} jamming events.

The characteristic function of any one of the random vectors \underline{v}_k is given by

$$\begin{aligned} \phi_{\underline{v}_k}(\underline{\mu}; M) &= E\{\exp[j\underline{v}_k \cdot \underline{\mu}]\} \\ &= E\{\exp[j\mu_1 v_{1k} + j\mu_2 v_{2k} + \dots + j\mu_M v_{Mk}]\} \\ &= \pi_0 + \pi_1(e^{j\mu_1} + e^{j\mu_2} + \dots + e^{j\mu_M}) \\ &\quad + \pi_2(e^{j\mu_1 + j\mu_2} + \dots + e^{j\mu_{M-1} + j\mu_M}) \\ &\quad + \dots \\ &\quad + \pi_M e^{j\mu_1 + j\mu_2 + \dots + j\mu_M} \end{aligned} \quad (2.2-18)$$

For example, for $M=2$ and 4 the characteristic functions for one hop are

$$\phi_{\underline{v}}(\underline{\mu};2) = \pi_0 + \pi_1 e^{j\mu_1} + \pi_1 e^{j\mu_2} + \pi_2 e^{j\mu_1+j\mu_2} \quad (2.2-19)$$

and

$$\begin{aligned} \phi_{\underline{v}}(\underline{\mu};4) = & \pi_0 + \pi_1 e^{j\mu_1} + \pi_1 e^{j\mu_2} + \pi_1 e^{j\mu_3} + \pi_1 e^{j\mu_4} \\ & + \pi_2 e^{j\mu_1+j\mu_2} + \pi_2 e^{j\mu_1+j\mu_3} + \pi_2 e^{j\mu_1+j\mu_4} \\ & + \pi_2 e^{j\mu_2+j\mu_3} + \pi_2 e^{j\mu_2+j\mu_4} + \pi_2 e^{j\mu_3+j\mu_4} \\ & + \pi_3 e^{j\mu_1+j\mu_2+j\mu_3} + \pi_3 e^{j\mu_1+j\mu_2+j\mu_4} \\ & + \pi_3 e^{j\mu_1+j\mu_3+j\mu_4} + \pi_3 e^{j\mu_2+j\mu_3+j\mu_4} \\ & + \pi_4 e^{j\mu_1+j\mu_2+j\mu_3+j\mu_4} \end{aligned} \quad (2.2-20)$$

In this characteristic function, there are 2^M terms, one for each of the events described by \underline{v}_k .

The characteristic function for \underline{x} is simply that of \underline{v}_k , raised to the L -th power:

$$\phi_{\underline{x}}(\underline{\mu};M) = [\phi_{\underline{v}}(\underline{\mu};M)]^L \quad (2.2-21)$$

For example, for $M=2$,

$$\begin{aligned} \phi_{\underline{x}}(\underline{\mu};2) &= [\phi_{\underline{v}}(\underline{\mu};2)]^L \\ &= \sum \binom{L}{n_0, n_1, n_2, n_3} \pi_0^{n_0} \pi_1^{n_1+n_2} \pi_2^{n_3} \\ &\quad \times \exp\{j(n_1+n_3)\mu_1 + j(n_2+n_3)\mu_2\}, \end{aligned} \quad (2.2-22a)$$

where the summation is over all (n_0, n_1, n_2, n_3) such that

$$\sum_{q=0}^{2^M-1} n_q = L . \quad (2.2-22b)$$

The discrete probability density function (pdf) for $\underline{\ell}$ is the inverse Fourier transform of $\phi_{\underline{\ell}}(\underline{\mu}; M)$. Again, for $M=2$, the pdf $p(\underline{\ell}; M, L)$ for $\underline{\ell}$ is

$$p(\underline{\ell}; 2, L) = \sum \binom{L}{n_0, n_1, n_2, n_3} \pi_0^{n_0} \pi_1^{n_1+n_2} \pi_2^{n_3} \times \delta(n_1+n_3-\ell_1) \delta(n_2+n_3-\ell_2), \quad (2.2-23)$$

which can be used to find the individual $\underline{\ell}$ vector probabilities

$$\Pr\{\underline{\ell}; 2, L\} = \sum_{n=0}^L \binom{L}{n, L-\ell_2-n, L-\ell_1-n, \ell_1+\ell_2+n-L} \times \pi_0^n \pi_1^{2L-\ell_1-\ell_2-2n} \pi_2^{\ell_1+\ell_2-L+n} . \quad (2.2-24)$$

In (2.2-24) it is realized that the combinatorial factor is zero if any of its parameters is negative. As an example for $L=3$ and $M=2$,

$$\begin{aligned} \Pr\{\ell_1 = 1, \ell_2 = 2; 2, 3\} &= \binom{3}{0, 1, 2, 0} \pi_1^3 + \binom{3}{1, 0, 1, 1} \pi_0 \pi_1 \pi_2 \\ &= 3\pi_1^3 + 6\pi_0 \pi_1 \pi_2 . \end{aligned} \quad (2.2-25)$$

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An aid to checking our calculations is the fact that each $\underline{\ell}$ event corresponds to

$$\binom{L}{\ell_1} \binom{L}{\ell_2} \cdots \binom{L}{\ell_M} \quad (2.2-26)$$

[v] matrix events. In (2.2-25) therefore, there are $\binom{3}{1} \binom{3}{2} = 9$ terms.

A complete table of M=2 jamming event probabilities for L=1 to 4 is given in Table 2.2-3, and an equation for computing these probabilities for M=4 is included as Table 2.2-4.

2.2.5 Jamming Event Probabilities For Multiple Hops, Convolution Method.

Since $\underline{\ell}$ is the sum of L independent random event vectors \underline{v}_k , the pdf for $\underline{\ell}$ is the L-fold convolution of the pdf for \underline{v}_k :

$$p(\underline{\ell}; M, L) = p(\underline{v}_1; M) * p(\underline{v}_2; M) * \dots * p(\underline{v}_L; M), \quad (2.2-27)$$

or

$$\Pr\{\underline{\ell}; M, L\} = \sum_{\underline{v}_1} \sum_{\underline{v}_2} \cdots \sum_{\underline{v}_L} \Pr\{\underline{v}_1; M\} \dots \Pr\{\underline{v}_L; M\} \delta(\underline{\ell} - \sum_{k=1}^L \underline{v}_k). \quad (2.2-28)$$

Figure 2.2-1 illustrates a programming approach for calculating this equation indirectly.

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TABLE 2.2-3

JAMMING EVENT PROBABILITIES FOR M=2, L=1 TO 4

$\underline{\ell}$	<u>L=1</u>	<u>L=2</u>	<u>L=3</u>	<u>L=4</u>
0,0	π_0	π_0^2	π_0^3	π_0^4
0,1	π_1	$2\pi_0\pi_1$	$3\pi_0^2\pi_1$	$4\pi_0^3\pi_1$
1,0	π_1	$2\pi_0\pi_1$	$3\pi_0^2\pi_1$	$4\pi_0^3\pi_1$
1,1	π_2	$2\pi_0\pi_2 + 2\pi_1^2$	$3\pi_0^2\pi_2 + 6\pi_0\pi_1^2$	$4\pi_0^3\pi_2 + 12\pi_0^2\pi_1^2$
0,2		π_1^2	$3\pi_0\pi_1^2$	$6\pi_0^2\pi_1^2$
1,2		$2\pi_1\pi_2$	$6\pi_0\pi_1\pi_2 + 3\pi_1^3$	$12\pi_0^2\pi_1\pi_2 + 12\pi_0\pi_1^3$
2,0		π_1^2	$3\pi_0\pi_1^2$	$6\pi_0^2\pi_1^2$
2,1		$2\pi_1\pi_2$	$6\pi_0\pi_1\pi_2 + 3\pi_1^3$	$12\pi_0^2\pi_1\pi_2 + 12\pi_0\pi_1^3$
2,2		π_2^2	$3\pi_0\pi_2^2 + 6\pi_1^2\pi_2$	$6\pi_0^2\pi_2^2 + 24\pi_0\pi_1^2\pi_2 + 6\pi_1^4$
0,3			π_1^3	$4\pi_0\pi_1^3$
1,3			$3\pi_1^2\pi_2$	$12\pi_0\pi_1^2\pi_2 + 4\pi_1^4$
2,3			$3\pi_1\pi_2^2$	$12\pi_0\pi_1\pi_2^2 + 12\pi_1^3\pi_2$
3,0			π_1^3	$4\pi_0\pi_1^3$
3,1			$3\pi_1^2\pi_2$	$12\pi_0\pi_1^2\pi_2 + 4\pi_1^4$
3,2			$3\pi_1\pi_2^2$	$12\pi_0\pi_1\pi_2^2 + 12\pi_1^3\pi_2$
3,3			π_2^3	$4\pi_0\pi_2^3 + 12\pi_1^2\pi_2^2$
0,4				π_1^4
1,4				$4\pi_1^3\pi_2$
2,4				$6\pi_1^2\pi_2^2$
3,4				$4\pi_1\pi_2^3$
4,0				π_1^4
4,1				$4\pi_1^3\pi_2$
4,2				$6\pi_1^2\pi_2^2$
4,3				$4\pi_1\pi_2^3$
4,4				π_2^4

$$\Pr(\underline{\ell}) = \sum_{n=0}^L (n, L-\ell_2, n, L-\ell_1, n, \ell_1 + \ell_2 + n - L) \pi_0^n \pi_1^{2L-\ell_1-\ell_2-n} \pi_2^{\ell_1 + \ell_2 + n - L}$$

TABLE 2.2-4

PROBABILITIES OF $\bar{\lambda}$ JAMMING EVENTS FOR M = 4

$$\Pr(\bar{\lambda}; 4, L) = \sum_{n_0=0}^L \sum_{n_1=1}^L \dots \sum_{n_{15}=0}^L \frac{L!}{n_0! n_1! \dots n_{15}!} \pi_0^{n_0} \pi_1^{n_1+n_2+n_4+n_8} \pi_2^{n_3+n_5+n_6+n_9+n_{10}+n_{12}} \pi_3^{n_7+n_{11}+n_{13}+n_{14}} \pi_4^{n_{15}}$$

CONSTRAINTS

$$\text{CONSTRAINTS : } n_1 + n_3 + n_5 + n_7 + n_9 + n_{11} + n_{13} + n_{15} = \lambda_1$$

$$n_2 + n_3 + n_6 + n_7 + n_{10} + n_{11} + n_{14} + n_{15} = \lambda_2$$

$$n_4 + n_5 + n_6 + n_7 + n_{12} + n_{13} + n_{14} + n_{15} = \lambda_3$$

$$n_8 + n_9 + n_{10} + n_{11} + n_{12} + n_{13} + n_{14} + n_{15} = \lambda_4$$

$$\sum_{i=0}^{15} n_i = L$$

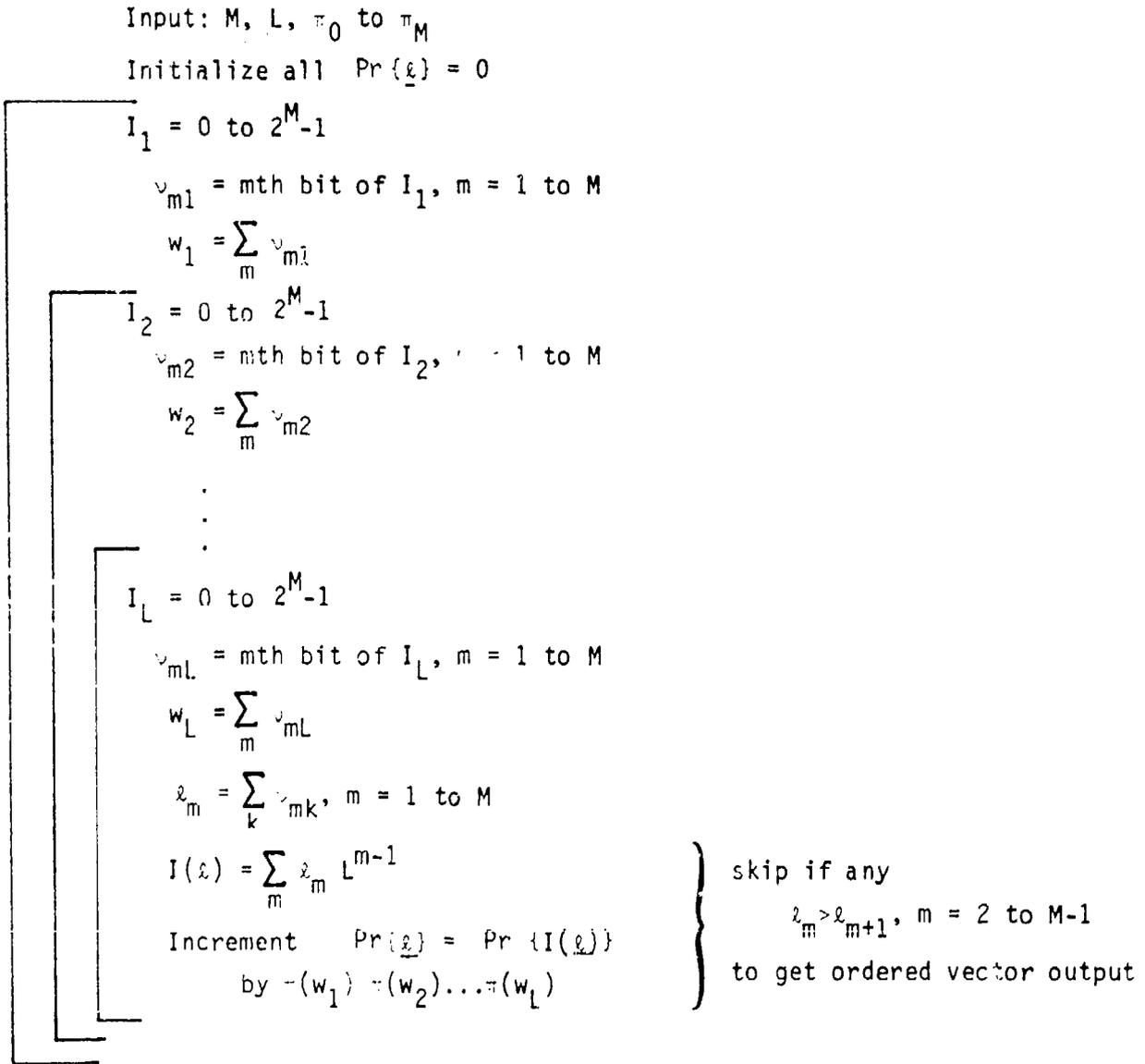


FIGURE 2.2-1 PROGRAM STRUCTURE FOR CALCULATING JAMMING EVENT PROBABILITIES INDIRECTLY

This can also be done iteratively, using the fact that

$$\begin{aligned} \Pr\{\underline{\ell}; M, L\} &= \sum_{\underline{\alpha}} \sum_{\underline{v}_L} \Pr\{\underline{\alpha}; M, L-1\} \Pr\{\underline{v}_L; M\} \delta(\underline{\ell} - \underline{\alpha} - \underline{v}_L) \\ &= \sum_{\underline{v}_L} \Pr\{\underline{\ell} - \underline{v}_L; M, L-1\} \Pr\{\underline{v}_L; M\}. \end{aligned} \quad (2.2-29)$$

To accomplish the vector additions needed for the convolution, we may encode the $M \times 1$ vectors $\underline{\ell}$ and \underline{v}_L into a number using the form

$$I(\underline{\ell}) = \ell_1 + b\ell_2 + b^2\ell_3 + \dots + b^{M-1}\ell_M, \quad (2.2-30a)$$

where $b \geq L$ is an integer base number, supporting the relationship

$$I(\underline{\ell}_1 + \underline{\ell}_2) = I(\underline{\ell}_1) + I(\underline{\ell}_2). \quad (2.2-30b)$$

In this manner the convolution in (2.2-29) can be done using

$$\begin{aligned} \Pr\{\underline{\ell}; M, L\} &= \sum_{I(\underline{\alpha})} \sum_{I(\underline{v}_L)} \Pr\{I(\underline{\alpha}); M, L-1\} \Pr\{I(\underline{v}_L); M\} \\ &\quad \times \delta[I(\underline{\ell}) - I(\underline{\alpha}) - I(\underline{v}_L)]. \\ &= \sum_{I(\underline{v}_L)} \Pr\{I(\underline{\ell}) - I(\underline{v}_L); M, L-1\} \Pr\{I(\underline{v}_L); M\}. \end{aligned} \quad (2.2-31)$$

2.3 TOTAL PROBABILITY OF ERROR

In terms of the conditional probability of symbol error, given a jamming event defined by the vector $\underline{\ell}$, and the probabilities of the jamming events, we now can write the total probability of error as

$$\begin{aligned} P_s(e) &= \sum_{\underline{\ell}} \Pr\{\underline{\ell}\} P_s(e|\underline{\ell}) \\ &= \sum_{\underline{\ell}'} \binom{M-1}{n_0, n_1, \dots, n_L} \Pr\{\underline{\ell}'\} P_s(e|\underline{\ell}'). \end{aligned} \quad (2.3-1)$$

This formulation utilizes the fact that \underline{z} vectors formed by permutations of the nonsignal channel elements $\{z_m, m>1\}$ of the ordered \underline{z} vector are equiprobable, and that the conditional error probability is the same for each permutation.

2.4 FH/RMFSK HARD DECISION ANALYSIS

In addition to studying the performance of L hops/symbols soft decision receivers for various FH/RMFSK combining schemes, we shall calculate the performance when L M-ary hard decisions are combined to produce a final symbol decision. This configuration is in itself a form of ECCM processing, as will be shown in later sections.

2.4.1 Formulation of Error Probability.

Under an M-ary hard decision approach, shown previously as Figure 1.3-3, on the kth hop the decision variables $\{z_{mk}\}$ are compared to find the largest; the signal is assumed to be present in the channel with the largest decision variable. The per hop or "hard" symbol decision can be thought of as selecting one of M vectors $\{\underline{D}_1, \underline{D}_2, \dots, \underline{D}_M\}$, where

$$\underline{D}_m = (D_{1m}, D_{2m}, \dots, D_{Mm}) \quad (2.4-1a)$$

with

$$D_{im} = \begin{cases} 1, & i=m \\ 0, & i \neq m \end{cases} \quad m=1,2,\dots,M. \quad (2.4-1b)$$

The hard symbol decision on the kth hop then can be expressed as the vector

$$\underline{d}_k = (d_{1k}, d_{2k}, \dots, d_{Mk}) \triangleq \underline{D}_{m^*} \quad (2.4-2a)$$

where m^* is chosen such that

$$z_{m^*k} = \max_m(z_{mk}). \quad (2.4-2b)$$

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The components of the per-hop decision vector \underline{d}_k are accumulated over the L hops of the symbol to produce the final, discrete decision variables

$$d_m = \sum_{k=1}^L d_{mk} ; m = 1, 2, \dots, M; \quad (2.4-3a)$$

or, in vector notation, the final discrete-valued decision vector

$$\underline{d} = \sum_{k=1}^L \underline{d}_k . \quad (2.4-3b)$$

The error probability can be formulated as

$$\begin{aligned} P_S(e) &= 1 - P_S(\text{correct decision} \equiv C) \\ &= 1 - \sum_{\underline{l}} P_S(C|\underline{l}) \Pr\{\underline{l}\} \\ &= 1 - \sum_{\underline{l}} \Pr\{\underline{l}\} \sum_{\underline{n} \in \Omega_C} \Pr\{\underline{d} = \underline{n}|\underline{l}\}, \end{aligned} \quad (2.4-4)$$

where $\underline{l} = (l_1, l_2, \dots, l_M)$ describes the jamming events, and Ω_C is the set of decision vectors which produce a correct decision.

Since the components of \underline{d} are discrete-valued, there exists the possibility of a tie between the signal channel's final decision variable value and that of one or more non-signal channels. Thus the error expression (2.4-4) should be modified to

$$P_S(e) = 1 - \sum_{\underline{l}} \Pr\{\underline{l}\} \sum_{\underline{n} \in \Omega_C} h(\underline{n}) \Pr\{\underline{d} = \underline{n}|\underline{l}\}, \quad (2.4-5a)$$

where $h(\underline{n}) = (\#\text{channels equal to maximum})^{-1}$, (2.4-5b)

assuming that a randomized decision is made when there is a tie. For example, if three channels (including the signal channel) are equal to the maximum value, then $h(\underline{n}) = 1/3$.

2.4.2 Explicit Form of Error Probability.

Interchanging the summations in the probability of a correct symbol decision gives

$$P_S(c) = \sum_{\underline{n} \in \Omega_c} h(\underline{n}) \sum_{\underline{\ell}} \Pr\{\underline{\ell}\} \Pr\{\underline{d} = \underline{n} | \underline{\ell}\} . \quad (2.4-6)$$

Now, since the jamming event vector $\underline{\ell}$ is related to the jamming event matrix $[\underline{v}]$ by

$$\underline{\ell} = \sum_{k=1}^L \underline{v}_k , \quad (2.4-7)$$

where \underline{v}_k is the k th column of $[\underline{v}]$, the summation over $\underline{\ell}$ can be replaced by

$$\begin{aligned} & \sum_{\underline{v}_1} \sum_{\underline{v}_2} \dots \sum_{\underline{v}_L} \Pr\{[\underline{v}]\} \Pr\{\underline{d} = \underline{n} | [\underline{v}]\} \\ &= \sum_{\underline{v}_1} \Pr\{\underline{v}_1\} \sum_{\underline{v}_2} \Pr\{\underline{v}_2\} \dots \sum_{\underline{v}_L} \Pr\{\underline{v}_L\} \Pr\{\underline{d} = \underline{n} | \underline{v}_1, \underline{v}_2, \dots, \underline{v}_L\} . \end{aligned} \quad (2.4-8)$$

In this expression we use the fact that the $\{\underline{v}_k\}$ are statistically independent. It is also true that the individual hop decisions $\{\underline{d}_k\}$ are independent, so we can expand (2.4-8) further to obtain

$$\begin{aligned} & \sum_{\underline{v}_1} \Pr\{\underline{v}_1\} \Pr\{\underline{d}_1 | \underline{v}_1\} \sum_{\underline{v}_2} \Pr\{\underline{v}_2\} \Pr\{\underline{d}_2 | \underline{v}_2\} \\ & \dots \sum_{\underline{v}_L} \Pr\{\underline{v}_L\} \Pr\{\underline{d}_L | \underline{v}_L\} \delta(\sum_k \underline{d}_k, \underline{n}) , \end{aligned} \quad (2.4-9)$$

where $\delta(\underline{a}, \underline{b})$ is a vector version of the Kronecker delta function:

$$\delta(\underline{a}, \underline{b}) \triangleq \begin{cases} 1 & \text{if } \underline{a} = \underline{b} \\ 0 & \text{if } \underline{a} \neq \underline{b} . \end{cases} \quad (2.4-10)$$

Recognizing that the sums over the individual $\{\underline{v}_k\}$ are simply averages, we can write

$$P_s(e) = 1 - \sum_{\underline{n} \in \Omega_c} h(\underline{n}) \Pr\{\underline{d}_1\} \Pr\{\underline{d}_2\} \dots \Pr\{\underline{d}_L\} \delta(\underline{d}, \underline{n}), \quad (2.4-11)$$

where $\Pr\{\underline{d}_k\}$ for the k th hop is the average of the (discrete) probability distribution for the hop decision over the jamming events for the k th hop, \underline{v}_k . Assuming without loss of generality that the first channel is the signal channel, these averages are

$$\Pr\{\underline{d}_k = D_1\} \equiv p = 1 - P_s(e; \gamma, \frac{1}{L} \cdot \frac{E_s}{N_0}, \frac{1}{L} \cdot \frac{E_s}{N_J}, L=1) \triangleq 1 - P_1 \quad (2.4-12a)$$

$$\begin{aligned} \Pr\{\underline{d}_k = D_m\} \equiv q &= (1-p)/(M-1) \\ &= P_1/(M-1), \quad m \geq 2. \end{aligned} \quad (2.4-12b)$$

Finally, using the function

$$H(\underline{n}) = \begin{cases} 1 & \text{if } \underline{n} \in \Omega_c \\ 0 & \text{if } \underline{n} \in \Omega'_c \end{cases} \quad (2.4-13)$$

as a "mask" for the correct decisions, we can write (2.4-11) more explicitly as

$$\begin{aligned}
 P_s(e) &= 1 - \underbrace{\sum_{n_1=0}^L \sum_{n_2=0}^L \dots \sum_{n_M=0}^L}_{\sum_i n_i = L} H(\underline{n}) h(\underline{n}) \left(n_1, n_2, \dots, n_M \right) p^{n_1} q^{L-n_1} \\
 &= 1 - \sum_{n_1=0}^L \sum_{n_2=0}^{n_1} \dots \sum_{n_M=0}^{n_1} h(\underline{n}) \left(n_1, n_2, \dots, n_M \right) p^{n_1} q^{L-n_1} \delta(\sum_i n_i, L).
 \end{aligned}
 \tag{2.4-14}$$

2.4.2.1 Special case: $L = 2$.

For $L = 2$, (2.4-14) reduces to

$$\begin{aligned}
 P_s(e; \gamma, \frac{E_s}{N_0}, \frac{E_s}{N_J}, L=2) &= 1 - p^2 - (M-1)pq \\
 &= 1 - p^2 - p P_1 = P_1 \\
 &= P_s(e; \gamma, \frac{1}{2} \frac{E_s}{N_0}, \frac{1}{2} \frac{E_s}{N_J}, L=1);
 \end{aligned}
 \tag{2.4-15}$$

that is, the hard decision receiver is uniformly 3 dB worse for $L = 2$ than for $L = 1$, for any value of M .

2.4.2.2 Special case: $L = 3$.

For $L = 3$, (2.4-14) reduces to

$$P_s(e; \gamma, \frac{E_s}{N_0}, \frac{E_s}{N_J}, L=3) = \frac{1}{M-1} p_1^2 (2M - 1 - M P_1),
 \tag{2.4-16a}$$

with

$$P_1 = P_s(e; \gamma, \frac{1}{3} \frac{E_s}{N_0}, \frac{1}{3} \frac{E_s}{N_0}, L=1).
 \tag{2.4-16b}$$

2.4.2.3 Special case: $L = 4$.

For $L = 4$, (2.4-14) reduces to

$$P_s(e; \gamma, \frac{E_s}{N_0}, \frac{E_s}{N_J}, L = 3) = \frac{1}{M-1} P_1^2 \left[3 + \frac{P_1}{M-1} (3M^2 - 9M + 4) - 2M \cdot \frac{M-2}{M-1} P_1^2 \right] \quad (2.4-17a)$$

with

$$P_1 = P_s(e; \gamma, \frac{1}{4} \frac{E_s}{N_0}, \frac{1}{4} \frac{E_s}{N_J}, L=1). \quad (2.4-17b)$$

Note that for $M = 2$, (2.4-16a) and (2.4-17a) both give $P_s = P_1^2(3-2P_1)$; this implies that the $L = 4$ hard decision performance is uniformly $10 \log_{10} \left(\frac{4}{3}\right) = 1.25$ dB worse than that for $L = 3$ when $M = 2$.

2.4.2.4 Special case: $M = 2$.

For $M = 2$, (2.3-14) reduces to

$$P_b(e) = \begin{cases} 1 - \sum_{n_1=\frac{L}{2}+1}^L \binom{L}{n_1} p^{n_1} q^{L-n_1} - \frac{1}{2} \binom{L}{L/2} p^{L/2} q^{L/2}, & L \text{ even} \\ 1 - \sum_{n_1=(L+1)/2}^L \binom{L}{n_1} p^{n_1} q^{L-n_1}, & L \text{ odd.} \end{cases} \quad (2.4-18)$$

3.0 FH/RMFSK PERFORMANCE USING SQUARE-LAW LINEAR COMBINING RECEIVER

In this section we consider the case of the generic receiver shown in Figure 2.2-1 when the envelope samples are processed using the function

$$f(x_{mk}) = x_{mk}^2 \equiv z_{mk}. \quad (3.0-1)$$

That is, the decision statistics $\{z_m\}$ are the unweighted linear combinations or sums of samples of the squared envelopes in each channel over multiple (L) hops. For non-hopping systems, this receiver is known to give good performance when the signal is subject to Rayleigh fading, L being the order of diversity which can be chosen to optimize performance for a given SNR.

3.1 ERROR PROBABILITY ANALYSIS

In Section 2.2 it was shown that the envelope samples $\{x_{mk}\}$ are σ_{mk} times a Rician random variable for the signal channel ($m=1$) and σ_{mk} times a Rayleigh random variable for the non-signal channels ($m>1$), where the value of σ_{mk} depends upon whether the channel is jammed or not. Therefore, for the square-law linear combining FH/RMFSK receiver, the hop decision statistics, which are the squares of the envelope samples, are σ_{mk}^2 times chi-squared random variables with two degrees of freedom. For the signal channel the noncentrality parameter is

$$\lambda_k = 2\rho_k = 2S/\sigma_{1k}^2. \quad (3.1-1)$$

The probability density function (pdf) for the signal channel samples is

$$p_{z_{1k}}(z; \rho_k, \sigma_{1k}^2) = \frac{1}{2\sigma_{1k}^2} \exp \left\{ -\rho_k - \frac{z}{2\sigma_{1k}^2} \right\} I_0 \left(\sqrt{2\rho_k z / \sigma_{1k}^2} \right), \quad (3.1-2)$$

while that for the non-signal channel samples is

$$P_{z_{mk}}(x; \sigma_{mk}^2) = \frac{1}{2\sigma_{mk}^2} \exp \left\{ -\frac{x}{2\sigma_{mk}^2} \right\}, \quad m > 1. \quad (3.1-3)$$

3.1.1 Distribution of the Decision Statistics.

The accumulated decision statistics $\{z_m\}$ can be expressed as

$$z_1 = \sum_{k=1}^L \sigma_{1k}^2 \chi^2(2; \lambda_k) \quad (3.1-4a)$$

and

$$z_m = \sum_{k=1}^L \sigma_{mk}^2 \chi^2(2) \quad (3.1-4b)$$

where $\chi^2(n)$ denotes a chi-squared random variable with n degrees of freedom and $\chi^2(n; \lambda)$ denotes a noncentral chi-squared random variable with n degrees of freedom and a noncentrality parameter λ . It is well known that sums of equally weighted chi-squared variables yield chi-squared variables:

$$\sum_{k=1}^L \chi^2(n_k; \lambda_k) = \chi^2 \left(\sum_{k=1}^L n_k; \sum_{k=1}^L \lambda_k \right). \quad (3.1-5)$$

This fact can be applied to (3.1-4) by recalling that for a given jamming event, ℓ_m out of L hops in a given channel are jammed. Thus

$$z_1 = \sigma_N^2 \chi^2[2(L-\ell_1); 2(L-\ell_1)S/\sigma_N^2] + \sigma_T^2 \chi^2(2\ell_1; 2\ell_1 S/\sigma_T^2) \quad (3.1-6a)$$

and

$$z_m = \sigma_N^2 \chi^2[2(L-\ell_m)] + \sigma_T^2 \chi^2(2\ell_m), \quad m > 2. \quad (3.1-6b)$$

In Appendix A it is shown that the pdf for the normalized variable

$$u_i = z_i/\sigma_N^2 \text{ is}$$

$$p_{u_1}(\alpha) = \begin{cases} p_{\chi^2}(\alpha; 2L, 2L\rho_N) & \ell_1 = 0; & (3.1-7a) \\ \frac{1}{K} p_{\chi^2}(\alpha/K; 2L, 2L\rho_T), & \ell_1 = L; & (3.1-7b) \\ \sum_{n=0}^{\infty} c_n p_{\chi^2}[\alpha; 2L + 2n, 2(L-\ell_1)\rho_N], & & (3.1-7c) \\ & & 0 < \ell_1 < L; \end{cases}$$

using

$$\rho_N \equiv S/\sigma_N^2, \rho_T \equiv S/\sigma_T^2, K = \sigma_T^2/\sigma_N^2 \quad (3.1-7d)$$

and where

$$c_n = e^{-\ell_1 \rho_T} \left(\frac{K-1}{K}\right)^n \left(\frac{1}{K}\right)^{\ell_1} L_n^{\ell_1-1} \left[\frac{-\ell_1 \rho_T}{K-1} \right], \ell_1 \geq 1. \quad (3.1-7e)$$

In (3.1-7e), the function $L_n^a(x)$ is the generalized Laguerre polynomial.

For the non-signal channels, substitution of $\rho_N = \rho_T = 0$ in (3.1-7) yields, for $u_m = z_m/\sigma_N^2$ ($m > 1$),

$$p_{u_m}(\alpha) = \begin{cases} p_{\chi^2}(\alpha; 2L), \ell_m = 0; & (3.1-8a) \\ \frac{1}{K} p_{\chi^2}(\alpha/K; 2L), \ell_m = L; & (3.1-8b) \\ \sum_{n=0}^{\infty} b_n p_{\chi^2}(\alpha; 2L+2n), 0 < \ell_m < L; & (3.1-8c) \end{cases}$$

where

$$b_n = \left(\frac{K-1}{K}\right)^n \left(\frac{1}{K}\right)^{\ell_m} \binom{n+\ell_m-1}{n}. \quad (3.1-8d)$$

In (3.1-7) and (3.1-8) the chi-squared pdf's are, for $2n$ degrees of freedom,

$$p_{\chi^2}(x; 2n, 2\rho) = \frac{1}{2} \exp\left\{-\frac{x}{2} - \rho\right\} \left(\frac{x}{2\rho}\right)^{(n-1)/2} I_{n-1}(\sqrt{2\rho x}) \quad (3.1-9a)$$

$$= \frac{1}{2} e^{-x/2} (x/2)^{n-1} / \Gamma(n), \rho = 0 \quad (3.1-9b)$$

where $I_m(x)$ is the modified Bessel function of the first kind and order m and $\Gamma(\cdot)$ is the gamma function.

3.1.2 Formulation of the Conditional Error Probability.

From Section 2.2, the conditional symbol error probability is

$$\begin{aligned}
 P_s(e|[\underline{v}]) &= P_s(e|\ell_1, \ell_2, \dots, \ell_M) \\
 &= 1 - \Pr\{z_2 < z_1, z_3 < z_1, \dots, z_M < z_1 | \underline{\ell}\} \\
 &= 1 - \Pr\{u_2 < u_1, u_3 < u_1, \dots, u_M < u_1 | \underline{\ell}\} \\
 &= 1 - \int_0^\infty d\alpha p_{u_1}(\alpha) \prod_{m=2}^M \int_0^\alpha d\beta_m p_{u_m}(\beta_m). \quad (3.1-10)
 \end{aligned}$$

From (3.1-8) and (3.1-9),

$$F_L(\alpha; \ell_m) \triangleq \int_0^\alpha d\beta_m p_{u_m}(\beta_m) = \begin{cases} 1 - \Gamma(L; \alpha/2)/\Gamma(L), & \ell_m = 0; & (3.1-11a) \\ 1 - \Gamma(L; \alpha/2K)/\Gamma(L), & \ell_m = L; & (3.1-11b) \\ 1 - \sum_{n=0}^{\infty} b_n \Gamma(L+n; \alpha/2)/\Gamma(L+n), & 0 < \ell_m < L; & (3.1-11c) \end{cases}$$

where $\Gamma(n;t)$ is the incomplete gamma function,

$$\Gamma(n;t) = \int_t^\infty dx e^{-x} x^{n-1} = \Gamma(n) e^{-t} \sum_{r=0}^{n-1} \frac{t^r}{r!}. \quad (3.1-12)$$

Formally, (3.1-10) can be written

$$P_s(e|\underline{\ell}) = 1 - \int_0^\infty d\alpha p_{u_1}(\alpha) [F_L(\alpha; 0)]^{n_0} [F_L(\alpha; 1)]^{n_1} \dots [F_L(\alpha; L)]^{n_L}, \quad (3.1-13a)$$

where n_i is the number of non-signal channels with $\ell_m = i$, and it is true that

$$n_0 + n_1 + \dots + n_L = M - 1. \quad (3.1-13b)$$

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3.1.3 Powers of Non-Signal Channel Probabilities.

We now show that the probabilities $F_L(\alpha; \lambda_m)$ in the general expression (3.1-13a) can be written in terms of power series.

For $\lambda_m = 0$, from (3.1-11) and (3.1-12)

$$\begin{aligned}
 [F_L(\alpha; 0)]^{n_0} &= \left[1 - e^{-\alpha/2} \sum_{r=0}^{L-1} \frac{(\alpha/2)^r}{r!} \right]^{n_0} \\
 &= \sum_{r_0=0}^{n_0} \binom{n_0}{r_0} (-1)^{r_0} e^{-r_0 \alpha/2} \left[\sum_{r=0}^{L-1} \frac{(\alpha/2)^r}{r!} \right]^{r_0} \\
 &= \sum_{r_0=0}^{n_0} \binom{n_0}{r_0} (-1)^{r_0} e^{-r_0 \alpha/2} \sum_{k_0=0}^{r_0(L-1)} \frac{C(k_0, r_0)}{k_0!} \left(\frac{\alpha}{2}\right)^{k_0}, \quad (3.1-14)
 \end{aligned}$$

where the coefficients $C(k_0, r_0)$ are [1, Appendix 4A] the functions

$$C(0, r) = 1 \quad (3.1-15a)$$

$$C(k, r) = \frac{1}{k} \sum_{n=1}^{\min(k, L-1)} \binom{k}{n} [(r+1)n - k] C(k-n, r), \quad (3.1-15b)$$

$$k > 0, \quad L \geq 2.$$

For example, when $L = 2$ the coefficients are simply

$$\begin{aligned}
 C(k, r) &= (r+1-k) C(k-1, r) \\
 &= r! / (r-k)! \quad (3.1-16a)
 \end{aligned}$$

so that the series raised to the r_0 power is

$$\sum_{k_0=0}^{r_0} \frac{r_0!}{k_0!(r_0-k_0)!} \left(\frac{\alpha}{2}\right)^{k_0} = (1+\alpha/2)^{r_0} \quad (3.1-16b)$$

as required.

Similarly, for $\ell_m = L$, we find that

$$[F_L(\alpha; L)]^{n_L} = \sum_{r_L=0}^{n_L} \binom{n_L}{r_L} (-1)^{r_L} e^{-r_L \alpha/2K} \sum_{k_L=0}^{r_L(L-1)} \frac{C(k_L, r_L)}{k_L!} \left(\frac{\alpha}{2K}\right)^{k_L}, \quad (3.1-17)$$

using the same $C(k, r)$ function as for $\ell_m=0$.

For $0 < \ell_m < L$, $L \geq 2$, the evaluation of the probability is more challenging, but does indeed reduce to a closed form. The development begins by recognizing that

$$\begin{aligned} F_L(\alpha; \ell_m) &= 1 - e^{-\alpha/2} \sum_{n=0}^{\infty} b_n \sum_{r=0}^{n+L-1} \frac{(\alpha/2)^r}{r!} \\ &= 1 - e^{-\alpha/2} \sum_{r=0}^{L-1} \frac{(\alpha/2)^r}{r!} - e^{-\alpha/2} \sum_{n=0}^{\infty} b_{n+1} \sum_{r=0}^n \frac{(\alpha/2)^{r+L}}{(r+L)!} \end{aligned} \quad (3.1-18)$$

where b_n is defined by (3.1-8d). The double summation in the last term can be manipulated in the following way:

$$\begin{aligned} \sum_{n=0}^{\infty} b_{n+1} \sum_{r=0}^n \frac{(\alpha/2)^{r+L}}{(r+L)!} &= \sum_{n=0}^{\infty} \sum_{r=0}^{\infty} b_{n+r+1} \frac{(\alpha/2)^{r+L}}{(r+L)!} \\ &= K^{-L} \sum_{r=0}^{\infty} \left(\frac{K-1}{K}\right)^{r+1} \frac{(\alpha/2)^{r+L}}{(r+L)!} \binom{r+L}{L-1} {}_2F_1\left(r+L+1, 1; r+2; \frac{K-1}{K}\right). \end{aligned} \quad (3.1-19)$$

The hypergeometric function can be transformed using [2, Eq. 15.3.5] to obtain

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on the summation were zero, the summation would be a Taylor's series:

$$\begin{aligned} & \sum_{n=0}^{\infty} \frac{a^n}{n!} \frac{n!}{(n+L-\ell)!} {}_1F_1(n+1; n+L-\ell+1; b) \\ &= \frac{1}{(L-\ell)!} \sum_{n=0}^{\infty} \frac{a^n}{n!} \left[\frac{d^n}{dx^n} {}_1F_1(1; L-\ell+1; x) \right] \Big|_{x=b} \end{aligned} \quad (3.1-23a)$$

$$= \left[e^{a+b} - \sum_{n=0}^{L-\ell-1} \frac{(a+b)^n}{n!} \right] \frac{1}{(a+b)^{L-\ell}}; \quad (3.1-23b)$$

therefore, (3.1-22) is seen to be

$$e^{\alpha/2} - \sum_{n=0}^{L-\ell-1} \frac{(\alpha/2)^n}{n!} - \sum_{n=0}^{\ell-1} \left(\frac{1}{K} \right)^n \frac{(\alpha/2)^{n+L-\ell}}{n!} \frac{d^n}{dx^n} x^{\ell-L} \left[e^x - \sum_{r=0}^{L-\ell-1} \frac{x^r}{r!} \right] \Big|_{x=(K-1)\alpha/2K} \quad (3.1-24)$$

Now, substituting (3.1-24) into (3.1-21) and the resulting expression into (3.1-18) gives

$$\begin{aligned} F_L(\alpha; \ell) &= 1 - e^{-\alpha/2} \left[\sum_{r=0}^{L-1} \frac{(\alpha/2)^r}{r!} + \sum_{r=0}^{\infty} \frac{(\alpha/2)^{r+L}}{(r+L)!} \right] \\ &\quad + e^{-\alpha/2} [(3.1-24)] \\ &= 0 + e^{-\alpha/2} [(3.1-24)] \\ &= 1 - e^{-\alpha/2} \left\{ \sum_{n=0}^{L-\ell-1} \frac{(\alpha/2)^n}{n!} \right. \\ &\quad \left. + \left(\frac{\alpha}{2} \right)^{L-\ell} \sum_{n=0}^{\ell-1} \frac{(\alpha/2K)^n}{n!} \frac{d^n}{dx^n} x^{\ell-L} \left[e^x - \sum_{r=0}^{L-\ell-1} \frac{x^r}{r!} \right] \Big|_{x=(K-1)\alpha/2K} \right\} \end{aligned}$$

$$\begin{aligned}
 &= 1 - e^{-\alpha/2} \left\{ \sum_{n=0}^{L-\ell-1} \frac{(\alpha/2)^n}{n!} \left[1 - \left(\frac{K}{K-1}\right)^{L-\ell-n} \sum_{r=0}^{\ell-1} \binom{L-\ell+r-n-1}{r} \frac{(-1)^r}{(K-1)^r} \right] \right\} \\
 &\quad - e^{-\alpha/2K} \left(\frac{K}{K-1}\right)^{L-\ell} \sum_{n=0}^{\ell-1} \frac{(\alpha/2K)^n}{n!} \sum_{r=0}^{\ell-1-n} \binom{L-\ell+r-1}{r} \left(\frac{-1}{K-1}\right)^r. \quad (3.1-25)
 \end{aligned}$$

For example, if $L = 2$ and $\ell = 1$,

$$\begin{aligned}
 F_2(\alpha; 1) &= 1 - e^{-\alpha/2} \left\{ 1 - \frac{K}{K-1} \right\} - e^{-\alpha/2K} \left(\frac{K}{K-1}\right) \\
 &= 1 - \frac{1}{K-1} \left\{ K e^{-\alpha/2K} - e^{-\alpha/2} \right\}. \quad (3.1-26)
 \end{aligned}$$

By direct algebraic calculation of (3.1-25), it can be shown that $F_L(\alpha, \ell)$ is of the form

$$F_L(\alpha; \ell) = 1 - \frac{1}{(K-1)^{L-1}} \left\{ e^{-\alpha/2K} f_1(\alpha; \ell, L) + e^{-\alpha/2} f_2(\alpha; \ell, L) \right\}, \quad (3.1-27)$$

where $f_1(\alpha; \ell, L)$ and $f_2(\alpha; \ell, L)$ are the polynomials given in Table 3.1-1.

Therefore

$$\begin{aligned}
 [F_L(\alpha; \ell)]^{n_\ell} &= \sum_{r_\ell=0}^{n_\ell} \binom{n_\ell}{r_\ell} \left[\frac{-1}{(K-1)^{L-1}} \right]^{r_\ell} \\
 &\quad \cdot \left[e^{-\alpha/2K} f_1(\alpha; \ell, L) + e^{-\alpha/2} f_2(\alpha; \ell, L) \right]^{r_\ell}
 \end{aligned}$$

TABLE 3.1-1 POLYNOMIALS FOR $F_L(\alpha; \ell)$

L	ℓ	$f_1(\alpha; \ell, L)$	$f_2(\alpha; \ell, L)$
1	0	0	1
	1	1	0
2	0	0	$(K-1)(1+\alpha/2)$
	1	K	-1
	2	$(K-1)(1+\alpha/2K)$	0
3	0	0	$(K-1)^2(1+\alpha/2+\alpha^2/8)$
	1	K^2	$-[2K-1+(K-1)\alpha/2]$
	2	$K(K-2)+(K-1)\alpha/2$	1
	3	$(K-1)^2[1+\alpha/2K+\alpha^2/8K^2]$	0
4	0	0	$(K-1)^3(1+\alpha/2+\alpha^2/8+\alpha^3/48)$
	1	K^3	$-[3K^2-3K+1+(2K^2-3K+1)\alpha/2+(K-1)^2\alpha^2/8]$
	2	$K^3-3K^2+K(K-1)\alpha/2$	$3K-1+(K-1)\alpha/2$
	3	K^3-3K^2+3K $+ (K^2-3K+2)\alpha/2$ $+ (K-1)^2\alpha^2/8K$	-1
	4	$(K-1)^3(1+\alpha/2K+\alpha^2/8K^2$ $+ \alpha^3/48K^3)$	0

$$\begin{aligned}
 &= \sum_{r_\ell=0}^{n_\ell} \binom{n_\ell}{r_\ell} \left[\frac{-1}{(K-1)^{L-1}} \right]^{r_\ell} \sum_{k_\ell=0}^{r_\ell} \binom{r_\ell}{k_\ell} e^{-(r_\ell-k_\ell)\alpha/2K-k_\ell\alpha/2} \\
 &\quad \cdot [f_1(\alpha; \ell, L)]^{r_\ell-k_\ell} [f_2(\alpha; \ell, L)]^{k_\ell}, \quad 0 < \ell < L, \quad (3.1-28)
 \end{aligned}$$

and the powers of the polynomials can be expressed as a higher order polynomial:

$$[f_1]^{r_\ell-k_\ell} [f_2]^{k_\ell} = \sum_{p_\ell=0}^P d(p_\ell) (\alpha/2)^{p_\ell}. \quad (3.1-29)$$

The coefficients $d(p)$ are given in Table 3.1-2 for L up to 4.

3.1.4 Expectation Over Signal Channel PDF.

Substitution of the powers of the non-signal channel probabilities into the conditional probability of symbol error equation yields the lengthy expression given in Table 3.1-3. The remaining analysis requires obtaining the expectation

$$\int_0^\infty d\alpha p_{u_1}(\alpha) e^{-a_0\alpha/2} (\alpha/2)^{b_0} = E_u \left\{ e^{-au_1/2} \left(\frac{u_1}{2} \right)^{b_0} \right\}, \quad (3.1-30)$$

where a_0 and b_0 are given in Table 3.1-3. From (3.1-7a), when the signal channel is not jammed ($\ell_j = 0$), the pdf $p_{u_1}(\alpha)$ is a straightforward noncentral chi-squared pdf, and the expectation is

$$\int_0^\infty d\alpha p_{\chi^2}(\alpha; 2L, 2L\rho_N) e^{-a_0\alpha/2} (\alpha/2)^{b_0}$$

TABLE 3.1-2 COEFFICIENTS FOR EQUATION (3.1-29)

L	ℓ	P = max p	d(p)
2	1	0	$K^{r_1-k_1} (-1)^{k_1}$
3	1	k_1	$(K^2)^{r_1-k_1} (-1)^{k_1} \binom{k_1}{p} (2K-1)^{k_1-p} (K-1)^p$
3	2	r_2-k_2	$\binom{r_2-k_2}{p} [K(K-2)]^{r_2-k_2-p} (K-1)^p$
4	1	$2k_1$	$(K^3)^{r_1-k_1} (-1)^{k_1} (3K^2-3K+1)^{k_1} \cdot g(p)$ where $g(0) = 1$, $g(1) = k_1 \cdot \frac{2K^2-3K+1}{3K^2-3K+1}$ $g(n) = \frac{1}{n} \left\{ (k_1+1-n) \cdot \frac{2K^2-3K+1}{3K^2-3K+1} \cdot g(n-1) \right.$ $\left. + [2(k_1+1)-n] \cdot \frac{(K-1)^2/2}{3K^2-3K+1} \cdot g(n-2) \right\}, n \geq 2$
4	2	r_2	$K^{2r_2-2k_2-p} (K-1)^p \sum_{q=\max(0, p-r_2+k_2)}^{\min(p, k_2)} \binom{r_2-k_2}{p-q} \binom{k_2}{q}$ $\cdot (K-3)^{r_2-k_2-p+q} (3K-1)^{k_2-q} K^q$
4	3	$2(r_3-k_3)$	$(-1)^{k_3} (K^3-3K^2+3K)^{r_3-k_3} g(p)$ where $g(0) = 1$, $g(1) = (r_3-k_3) \cdot \frac{K^2-3K+2}{K^3-3K^2+3K}$ $g(n) = \frac{1}{n} \left\{ (r_3-k_3+1-n) \cdot \frac{K^2-3K+2}{K^3-3K^2+3K} \cdot g(n-1) \right.$ $\left. + [2(r_3-k_3+1)-n] \cdot \frac{(K-1)^2/2K}{K^3-3K^2+3K} \cdot g(n-2) \right\}, n \geq 2$

TABLE 3.1-3 EXPRESSION FOR CONDITIONAL ERROR PROBABILITY

$$\begin{aligned}
 P_S(e|\underline{z}) &= 1 - \int_0^\infty d\alpha P_{u_1}(\alpha) [F_L(\alpha; 0)]^{n_0} [F_L(\alpha; 1)]^{n_1} \cdots [F_L(\alpha; L-1)]^{n_{L-1}} [F_L(\alpha; L)]^{n_L} \\
 &= 1 - \sum_{r_0=0}^{n_0} \binom{n_0}{r_0} (-1)^{r_0} \sum_{k_0=0}^{r_0(L-1)} \frac{C(k_0, r_0)}{k_0!} \sum_{r_L=0}^{n_L} \binom{n_L}{r_L} (-1)^{r_L} \sum_{k_L=0}^{r_L(L-1)} \frac{C(k_L, r_L)}{k_L!} \\
 &\quad \times \sum_{r_1=0}^{n_1} \binom{n_1}{r_1} \left[\frac{-1}{(K-1)^{L-1}} \right]^{r_1} \sum_{k_1=G}^{r_1} \binom{r_1}{k_1} \sum_{p_1=0}^{p_1} d(p_1) \cdots \\
 &\quad \times \sum_{r_{L-1}=0}^{n_{L-1}} \binom{n_{L-1}}{r_{L-1}} \left[\frac{-1}{(K-1)^{L-1}} \right]^{r_{L-1}} \sum_{k_{L-1}=0}^{r_{L-1}} \binom{r_{L-1}}{k_{L-1}} \sum_{p_{L-1}=0}^{p_{L-1}} d(p_{L-1}) \\
 &\quad \times \int_0^\infty d\alpha P_{u_1}(\alpha) e^{-a_0\alpha/2} (\alpha/2)^{b_0} \\
 a_0 &= r_0 + \frac{r_L}{K} + \sum_{\ell=1}^{L-1} \left[k_\ell + \frac{r_\ell - k_\ell}{K} \right] \\
 b_0 &= k_0 + k_L + \sum_{\ell=1}^{L-1} p_\ell
 \end{aligned}$$

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$$\begin{aligned}
 &= \int_0^{\infty} d\alpha e^{-L\rho_N - (1+a_0)\alpha/2} \left(\frac{\alpha}{2L\rho_N}\right)^{(L-1)/2} I_{L-1}\left(\sqrt{2L\rho_N\alpha}\right) \left(\frac{\alpha}{2}\right)^{b_0} \\
 &= \exp\left(\frac{-L\rho_N a_0}{1+a_0}\right) (1+a_0)^{-b_0-L} \\
 &\quad \cdot \int_0^{\infty} dx e^{-L\rho_N/(1+a_0) - x/2} \left[\frac{x}{2L\rho_N/(1+a_0)}\right]^{(L-1)/2} I_{L-1}\left(\sqrt{\frac{2L\rho_N x}{1+a_0}}\right) \left(\frac{x}{2}\right)^{b_0} \\
 &= \exp\left(\frac{-L\rho_N a_0}{1+a_0}\right) (1+a_0)^{-b_0-L} \cdot E\left[\left(\frac{x}{2}\right)^{b_0}\right], \tag{3.1-31}
 \end{aligned}$$

where x is distributed as a noncentral chi-squared random variable with $2L$ degrees of freedom and noncentral parameter $2L\rho_N/(1+a_0)$. The moments needed are

$$E\left[\left(\frac{x}{2}\right)^{b_0}\right] = b_0! \mathcal{L}_{b_0}^{L-1}\left(-\frac{L\rho_N}{1+a_0}\right) \tag{3.1-32a}$$

$$= \sum_{r=0}^{b_0} \binom{b_0+L-1}{b_0-r} \left(\frac{L\rho_N}{1+a_0}\right)^r \frac{1}{r!} b_0! \tag{3.1-32b}$$

where $\mathcal{L}_n^a(x)$ is the generalized Laguerre polynomial. Thus for the case of

$\ell_1 = 0$, the integral in (3.1-30) becomes the quantity

$$\exp\left(-\frac{L\rho_N a_0}{1+a_0}\right) \frac{b_0!}{(1+a_0)^{b_0+L}} \mathcal{L}_{b_0}^{L-1}\left(-\frac{L\rho_N}{1+a_0}\right). \quad (3.1-33)$$

In a similar way, when all the hops in the signal channel are jammed ($\ell_1 = L$), the integral in (3.1-30) becomes

$$\exp\left(-\frac{L\rho_T K a_0}{1+K a_0}\right) \frac{K^{b_0} b_0!}{(1+K a_0)^{b_0+L}} \mathcal{L}_{b_0}^{L-1}\left(-\frac{L\rho_T}{1+K a_0}\right). \quad (3.1-34)$$

Now, when the signal channel is jammed, but not on every hop ($0 < \ell_1 < L$), the channel pdf is a series of weighted noncentral chi-squared pdf's, as shown previously in (3.1-7c). This expectation (3.1-30) yields

$$\exp\left[-\frac{(L-\ell_1)\rho_N a_0}{1+a_0}\right] \sum_{n=0}^{\infty} c_n \frac{b_0!}{(1+a_0)^{b_0+L+n}} \mathcal{L}_{b_0}^{L+n-1}\left[-\frac{(L-\ell_1)\rho_N}{1+a_0}\right], \quad (3.1-35)$$

where the weights $\{c_n\}$ are

$$c_n = e^{-\ell_1 \rho_T} \left(\frac{K-1}{K}\right)^n K^{-\ell_1} \mathcal{L}_n^{\ell_1-1}\left[-\frac{\ell_1 \rho_T}{K-1}\right]. \quad (3.1-36)$$

In order to reduce (3.1-35) to a finite summation, it is necessary to seek an identity for

$$\sum_{n=0}^{\infty} A^n \mathcal{L}_n^a(x) \mathcal{L}_r^{n+k}(y). \quad (3.1-37)$$

To accomplish this objective, we note that [3, eq. 8.970.1]

$$\mathcal{L}_r^{n+k}(y) \triangleq \frac{1}{r!} e^y y^{-n-k} \frac{d^r}{dy^r} (e^{-y} y^{r+n+k}); \quad (3.1-38)$$

substituting in (3.1-37) yields the development

$$\begin{aligned} & \frac{1}{r!} e^y y^{-k} \sum_{n=0}^{\infty} \left(\frac{A}{y}\right)^n \mathcal{L}_n^a(x) \frac{d^r}{dy^r} (e^{-y} y^{r+n+k}) \\ &= \frac{e^y y^{-k}}{r!} \frac{\partial^r}{\partial z^r} \left[e^{-z} z^{r+k} \sum_{n=0}^{\infty} \left(\frac{Az}{y}\right)^n \mathcal{L}_n^a(x) \right] \Big|_{z=y} \\ &= \frac{e^y y^{-k}}{r!} \sum_{q=0}^r \binom{r}{q} \frac{\partial^{r-q}}{\partial z^{r-q}} (e^{-z} z^{r+k}) \frac{\partial^q}{\partial z^q} \left[\sum_{n=0}^{\infty} \left(\frac{Az}{y}\right)^n \mathcal{L}_n^a(x) \right] \Big|_{z=y} \\ &= e^y y^{-k} \sum_{q=0}^r \frac{1}{q!} e^{-z} z^{k+q} \mathcal{L}_{r-q}^{k+q}(z) \sum_{n=0}^{\infty} \left(\frac{A}{y}\right)^n \frac{n!}{(n-q)!} z^{n-q} \mathcal{L}_n^a(x) \Big|_{z=y}. \end{aligned} \quad (3.1-39)$$

The second summation, when manipulated, gives the result

$$\begin{aligned} & \left(\frac{A}{y}\right)^q \sum_{n=0}^{\infty} \frac{(n+q)!}{n!} \left(\frac{Az}{y}\right)^n \mathcal{L}_{n+q}^a(x) \\ &= \left(\frac{A}{y}\right)^q \sum_{n=0}^{\infty} \left(\frac{Az}{y}\right)^n \frac{(n+q+a)!}{n! a!} {}_1F_1(-n-q; a+1; x) \end{aligned}$$

$$\begin{aligned}
 &= \left(\frac{A}{y}\right)^q e^x \sum_{n=0}^{\infty} \left(\frac{Az}{y}\right)^n \frac{(n+q+a)!}{n! a!} {}_1F_1(n+q+a+1; a+1; -x) \\
 &= \left(\frac{A}{y}\right)^q e^x \sum_{n=0}^{\infty} \left(\frac{Az}{y}\right)^n \frac{1}{n!} \sum_{m=0}^{\infty} \frac{(-x)^m}{m!} \frac{(n+q+a+m)!}{(a+m)!} \\
 &= \left(\frac{A}{y}\right)^q e^x \sum_{m=0}^{\infty} \frac{(-x)^m}{m!} \frac{(q+a+m)!}{(a+m)!} (1 - Az/y)^{-q-a-m-1} \\
 &= \left(\frac{A}{z}\right)^q e^x \frac{(q+a)!}{a!} (1 - Az/y)^{-q-a-1} {}_1F_1(q+a+1; a+1; \frac{-x}{1 - Az/y}) \\
 &= \left(\frac{A}{y}\right)^q \frac{\exp\left(\frac{-Axz}{y-Az}\right)}{(1-Az/y)^{q+a+1}} q! \rho_q^a\left(\frac{xy}{y-Az}\right). \tag{3.1-40}
 \end{aligned}$$

After substitution in (3.1-39), we find that

$$\begin{aligned}
 &\sum_{n=0}^{\infty} A^n \rho_n^a(x) \rho_r^{n+k}(y) \\
 &= \frac{1}{(1-A)^{a+1}} \exp\left(\frac{-Ax}{1-A}\right) \sum_{q=0}^r \left(\frac{A}{1-A}\right)^q \rho_{r-q}^{k+q}(y) \rho_q^a\left(\frac{x}{1-A}\right). \tag{3.1-41}
 \end{aligned}$$

And, substituting appropriately, the desired expectation (3.1-35) becomes

$$\exp\left[-\frac{(L-\ell_1)\rho_N a_0}{1+a_0} - \frac{\ell_1 \rho_T K a_0}{1+K a_0}\right] \frac{b_0!}{(1+a_0)^{b_0+L-\ell_1} (1+K a_0)^{\ell_1}}$$

$$\cdot \sum_{q=0}^{b_0} \left(\frac{K-1}{1+K a_0}\right)^q \mathcal{L}_{b_0-q}^{q+L-1} \left[-\frac{(L-\ell_1)\rho_N}{1+a_0}\right] \mathcal{L}_q^{\ell_1-1} \left[-\frac{\ell_1 \rho_T}{K-1} \cdot \frac{K(1+a_0)}{1+K a_0}\right] \cdot$$

(3.1-42)

3.1.5 Special case: L=1 and M=2.

For L=1 and M=2 the FH/RMFSK total error probability is

$$P(e) = \pi_0 \cdot \frac{1}{2} e^{-\rho_N/2} + \pi_1 \cdot e^{-\rho_N/(K+1)} + \pi_2 \cdot \frac{1}{2} e^{-\rho_T/2} \quad (3.1-43a)$$

where $\rho_N = E_b/N_0$, $\rho_T = E_b/N_T$; (3.1-43b)

and $K = \sigma_T^2/\sigma_N^2 = \rho_N/\rho_T$. (3.1-43c)

3.2 NUMERICAL RESULTS

3.2.1 Soft-Decision Receiver.

Numerical results were obtained using two computational methods. In regions where the series converge rapidly enough, the form given in Table 3.1-3 and equation (3.1-42) was used for the computations. However, the presence of the term $(K-1)^L$ in the denominator of several terms causes difficulty when K is nearly equal to 1. In these cases, and on other occasions when round-off errors became excessive, the computations were performed by direct numerical integration of (3.1-10) using the densities (3.1-7) and (3.1-8) and the identity

$$1 - \int_0^{\infty} d\alpha p(\alpha) g(\alpha) \equiv \int_0^{\infty} d\alpha p(\alpha) [1-g(\alpha)] \quad (3.2-1)$$

which holds for all density functions $p(\alpha)$ for which $p(\alpha) \equiv 0$ if $\alpha < 0$, and hence by the properties of a p.d.f.

$$\int_0^{\infty} d\alpha p(\alpha) = 1. \quad (3.2-2)$$

Then (3.2-1) follows immediately from the fact that integration is a linear operation. The form on the right-hand side of (3.2-1) has the computational advantage that only the integrand need be computed to high accuracy, rather than the integral. For example, if $P_s(e) \approx 10^{-5}$ and we desire 4 significant digits in the answer, then the integral on the left-hand side must be computed numerically to 8-digit accuracy (e.g. 0.9999xxxx) in order to leave 4 non-zero

correct digits after subtracting from 1. But if we use the form on the right-hand side of (3.2-1) we could demand only 4-place relative accuracy from the numerical integration; the burden of many-place accuracy is placed only on the function $g(\cdot)$, which is usually much simpler and faster to compute than the overall integral. A listing of the computer program is given in Appendix D.

In Figures 3.2-1 through 3.2-4 we show the bit error probability as a function of bit energy-to-jamming density ratio with jamming fraction γ as a parameter for the case of $M=2$ (binary FSK) and $L=1,2,3$, and 4 hops per symbol (bit), respectively. In these figures the ratio of the bit energy to thermal noise density is set at 13.3525 dB, which corresponds to a bit error probability of 10^{-5} for one hop per bit in the absence of jamming. We note that for a given E_b/N_j ratio there is an optimum value of γ which maximizes the jammer's effectiveness. Furthermore, an incorrect choice of γ by the jammer can reduce the effectiveness (as measured by the communicator's bit error probability) by possibly as much as two orders of magnitude.

Figure 3.2-5 shows the envelopes of the curves in Figures 3.2-1 through 3.2-4, which represent the performance of the square-law combining receiver in worst-case partial-band noise jamming. We note that increasing L , the number of hops per bit, consistently degrades the performance. This implies that the noncoherent combining loss dominates over any diversity gain effects.

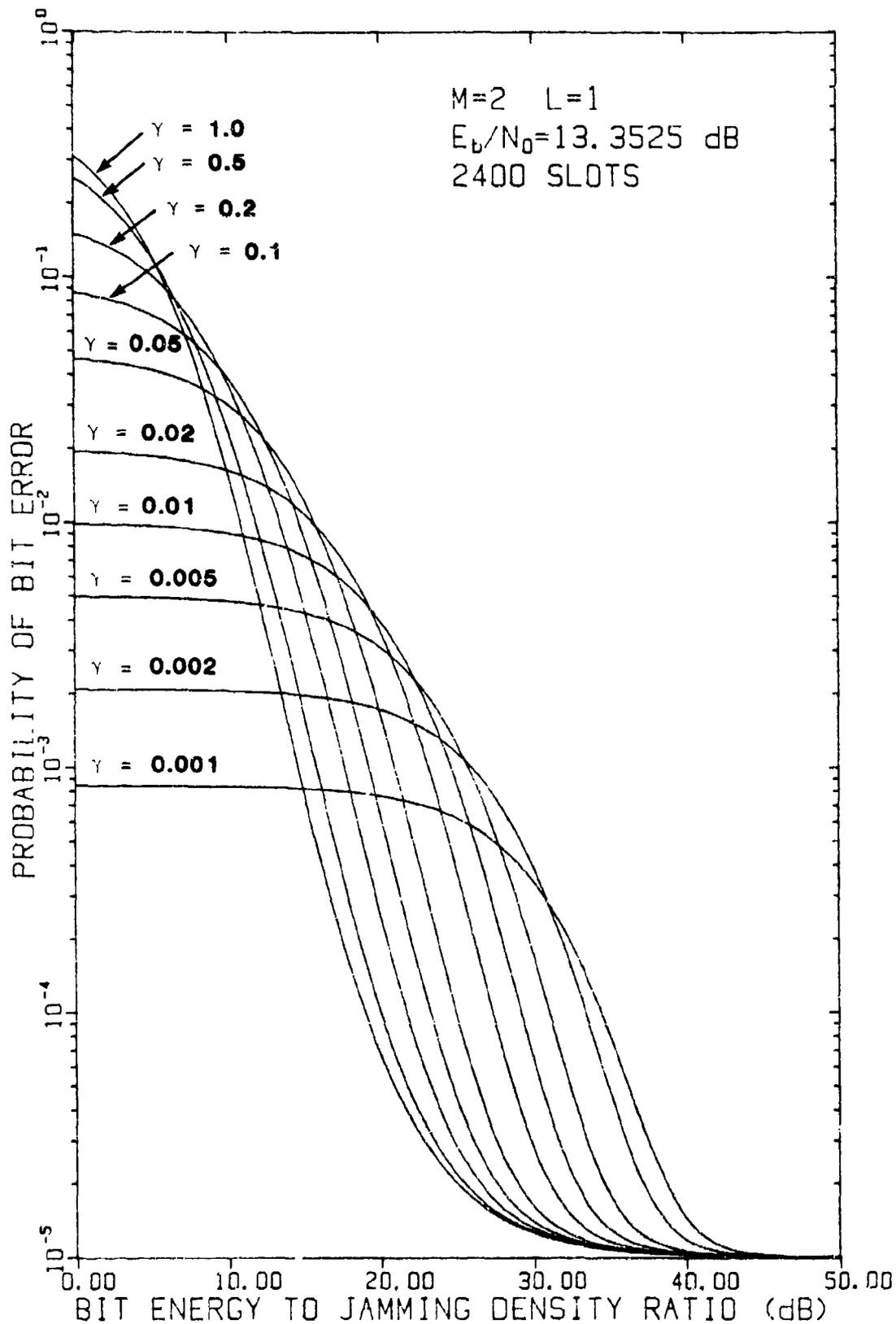


FIGURE 3.2-1 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH $M = 2$ AND $L = 1$ HOP/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.3525$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

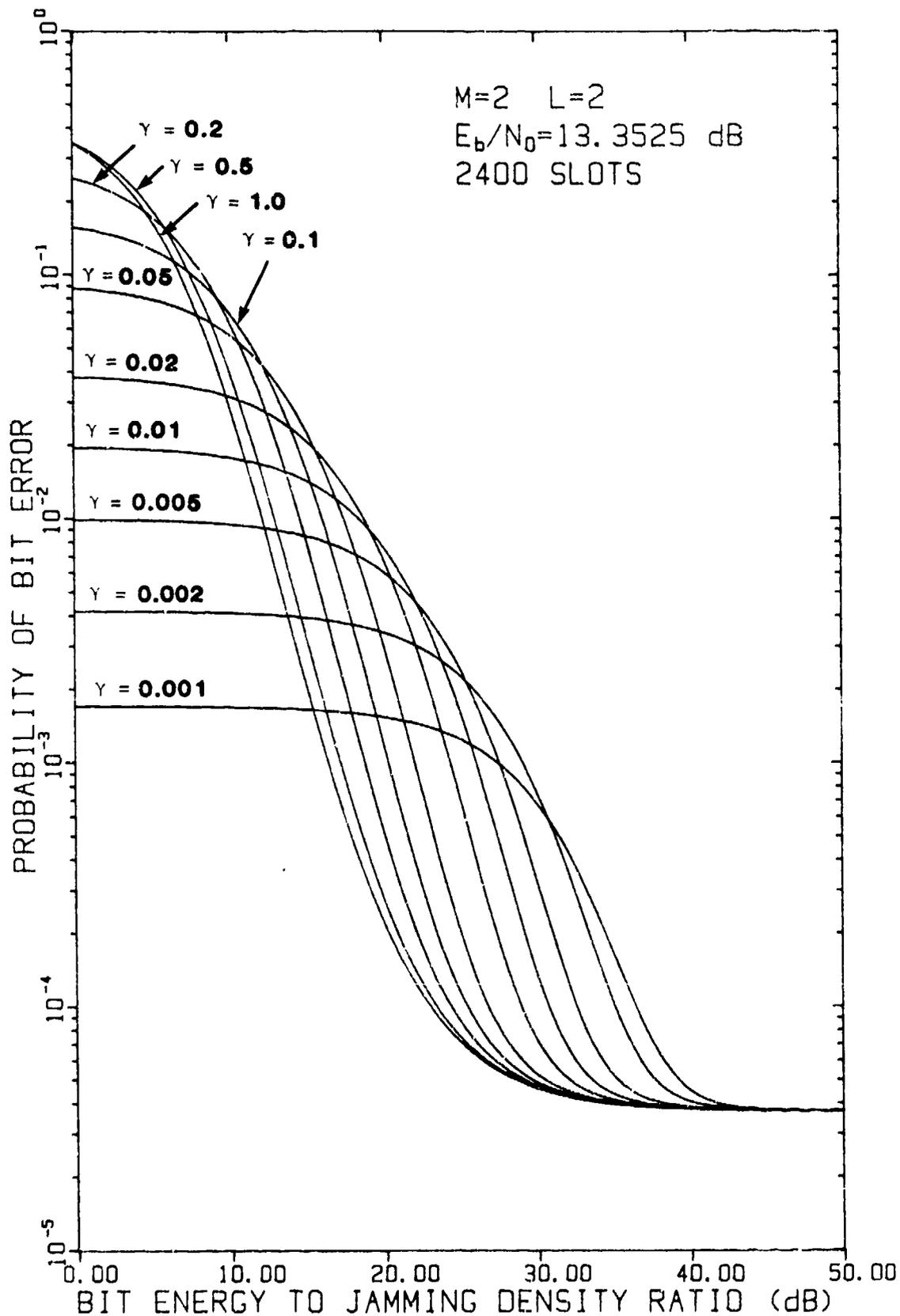


FIGURE 3.2-2 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH $M=2$ AND $L=2$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.3525$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$.) WITH JAMMING FRACTION γ AS A PARAMETER

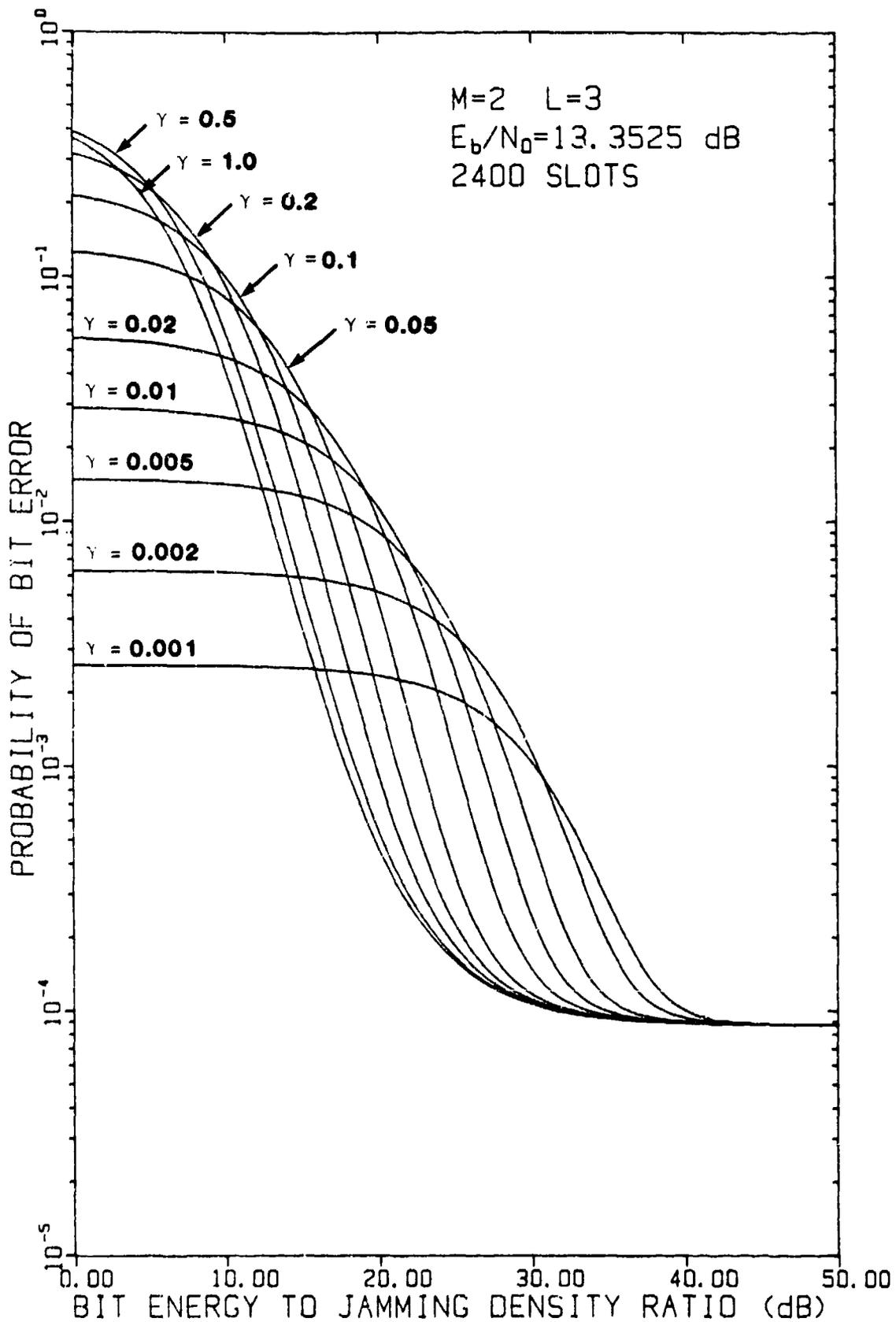


FIGURE 3.2-3 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH $M=2$ AND $L=3$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.3525$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ AS A PARAMETER

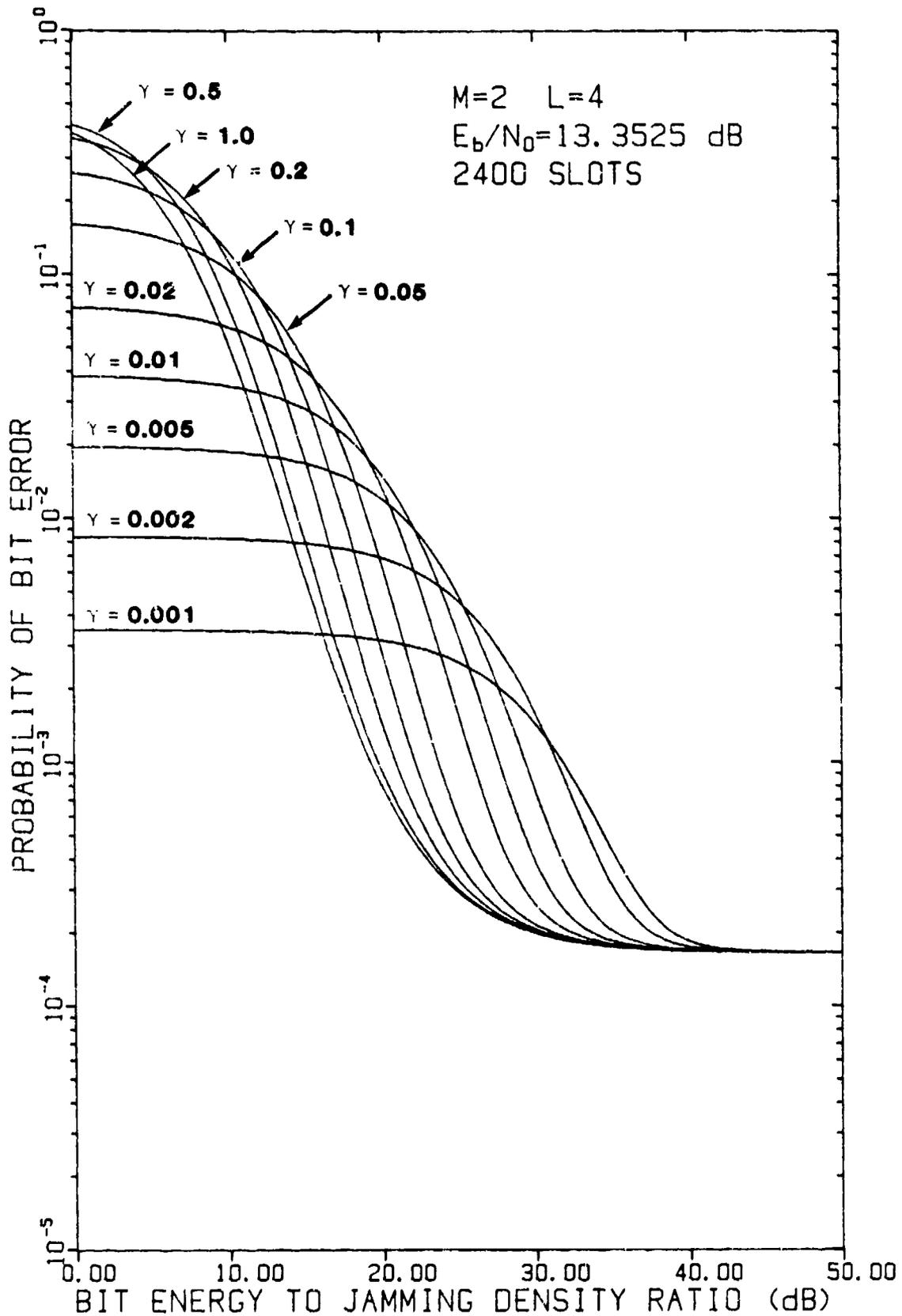


FIGURE 3.2-4 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH $M=2$ AND $L=4$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.3525 \text{ dB}$ (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ AS A PARAMETER

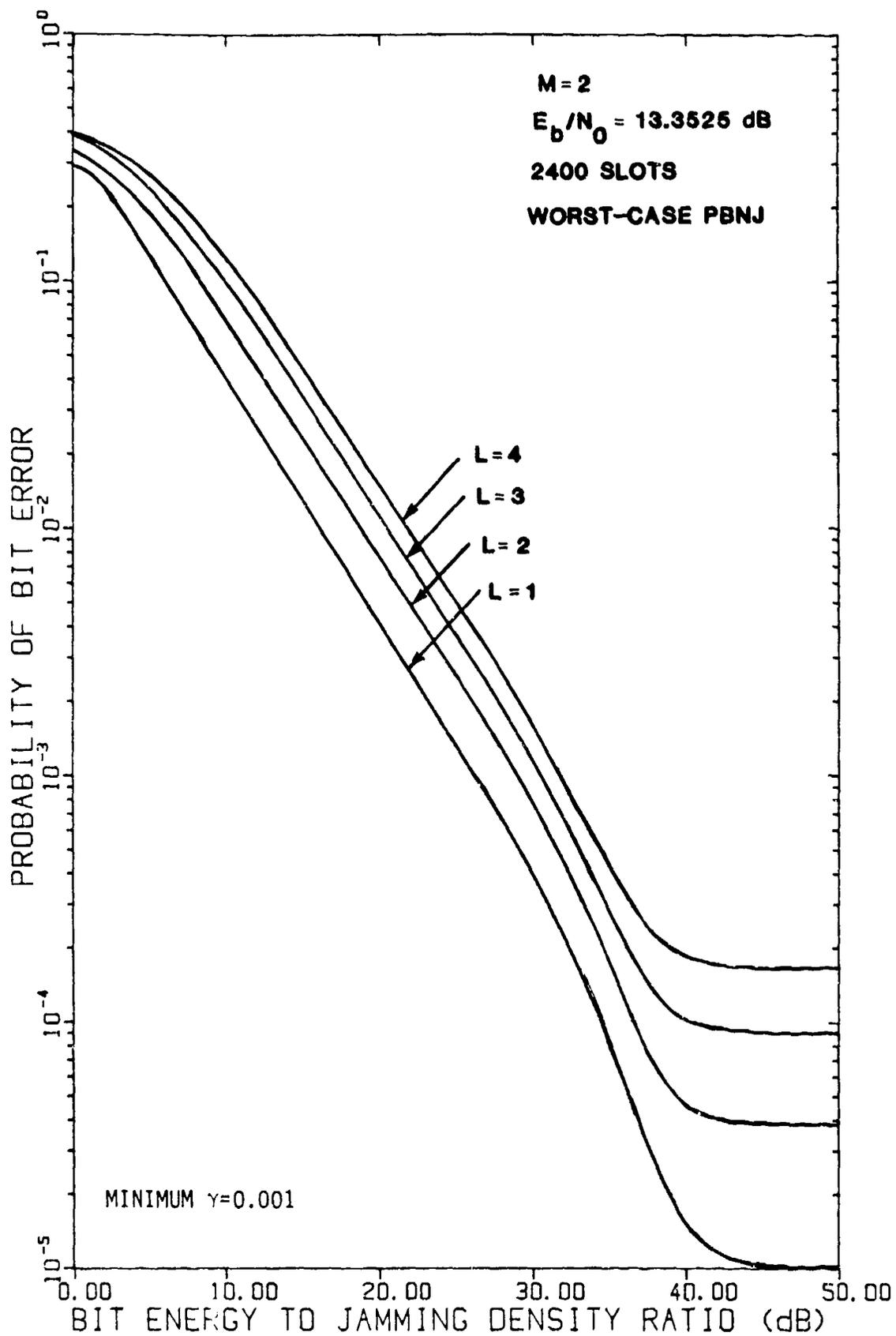


FIGURE 3.2-5 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER AND $M = 2$ WITH NUMBER OF HOPS/BIT AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING

Similar results for the case of $M=4$ are shown in Figures 3.2-6 through 3.2-9 for $L=1,2,3$, and 4 hops/symbol, respectively. Again we note that the jammer must carefully choose the proper partial-band fraction or risk reducing his effectiveness by more than an order of magnitude. We also observe that full-band jamming ($\gamma=1.0$) is not optimum until the jamming becomes very strong, i.e. $E_b/N_j < 0$ dB.

Figure 3.2-10 shows the envelope of the curves in Figures 3.2-6 through 3.2-9, which gives the performance in worst-case partial-band noise jamming. We note that increasing the number of hops per symbol consistently degrades the performance of the 4-ary system, just as it does for the binary system.

Finally, Figure 3.2-11 shows the worst-case partial-band noise jamming performance of the square-law receiver for $L=1$ hop/symbol and $M=2,4$, and 8. We observe that for strong jamming increasing M from 2 to 4 provides a very small performance improvement; but further increase to $M=8$ degrades the performance. This behavior is similar to that of a block-hopping system in tone jamming [1, Section 8].

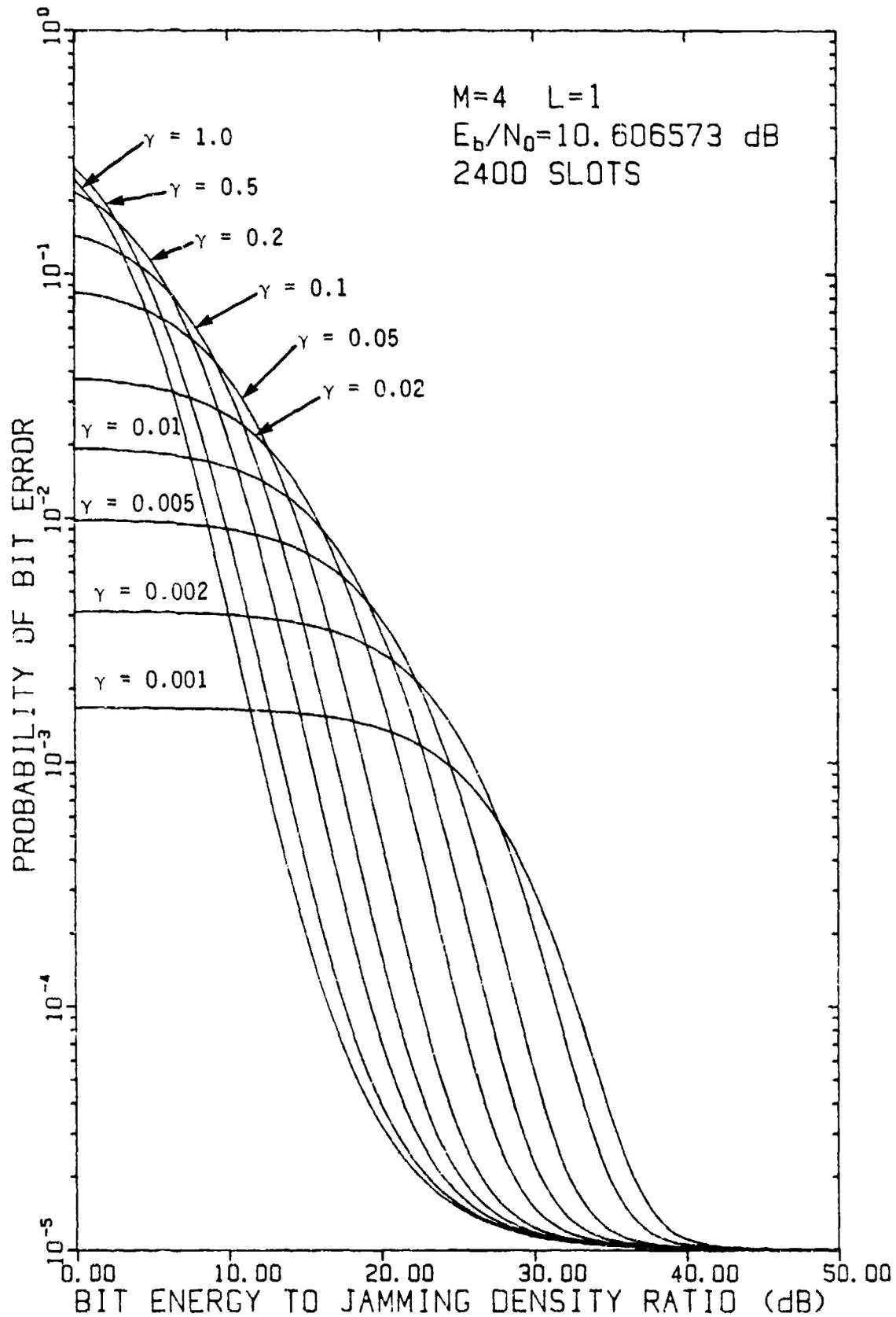


FIGURE 3.2-6 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH $M = 4$ AND $L = 1$ HOP/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

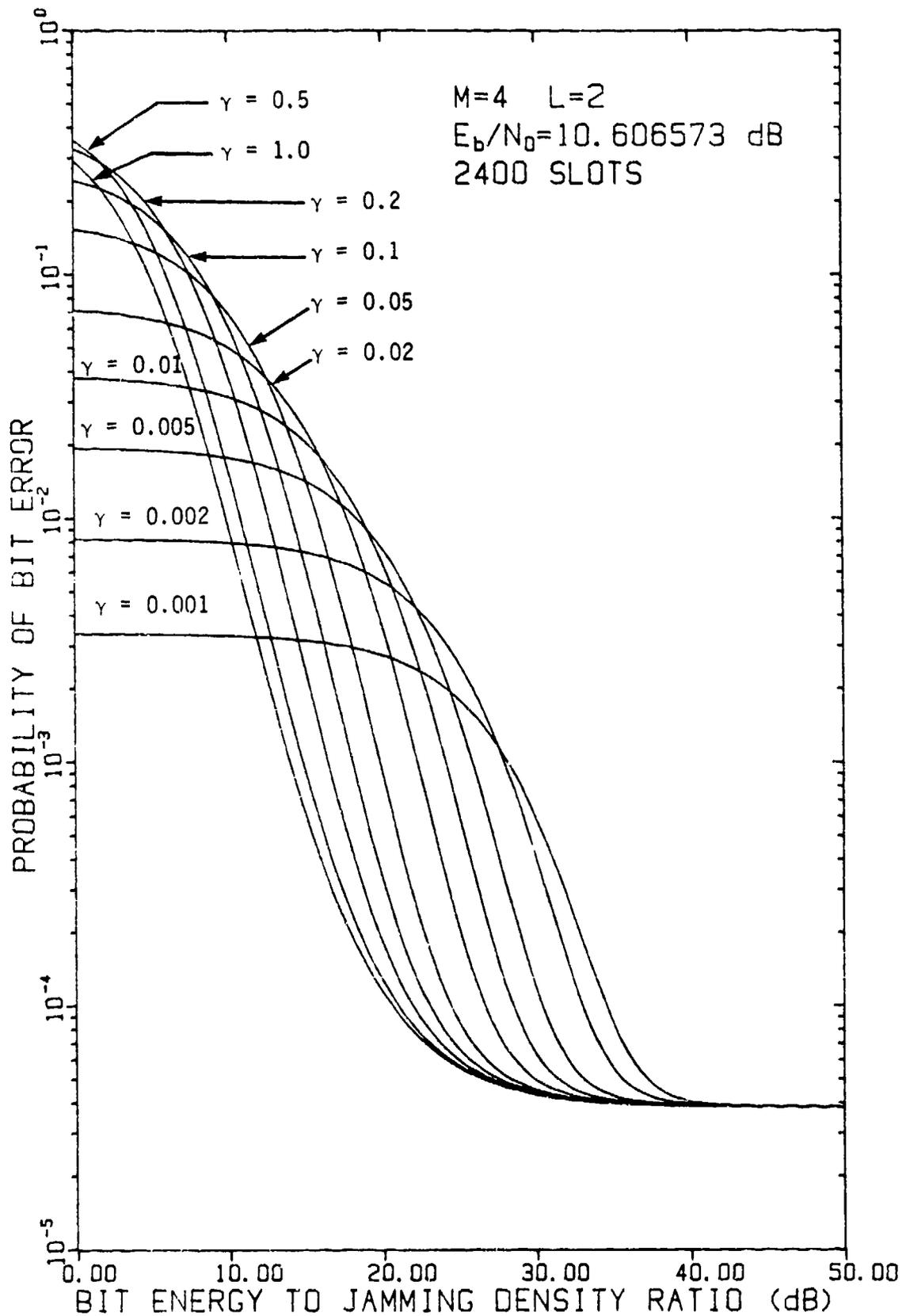


FIGURE 3.2-7 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH $M = 4$ AND $L = 2$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573 \text{ dB}$ (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

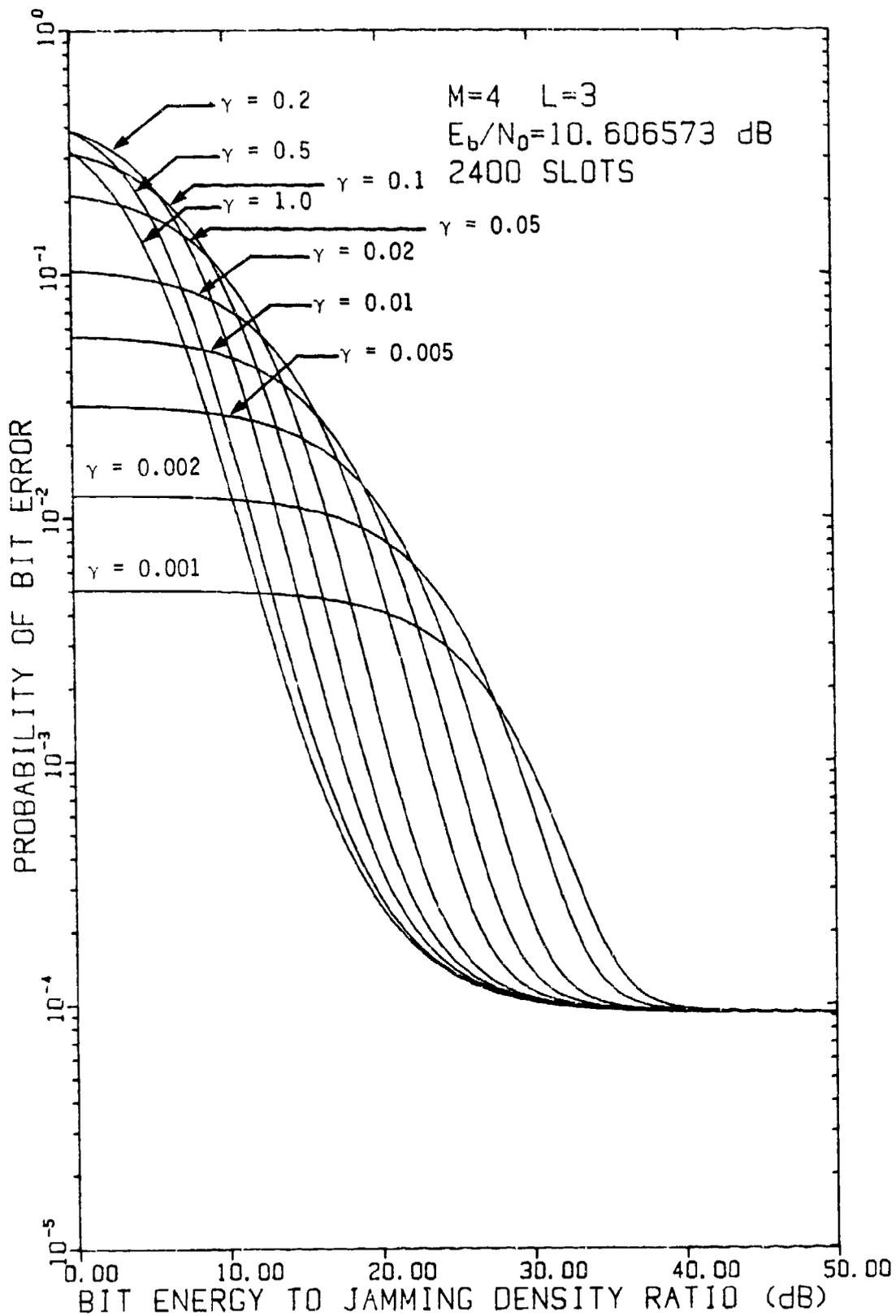


FIGURE 3.2-8 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH $M = 4$ AND $L = 3$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

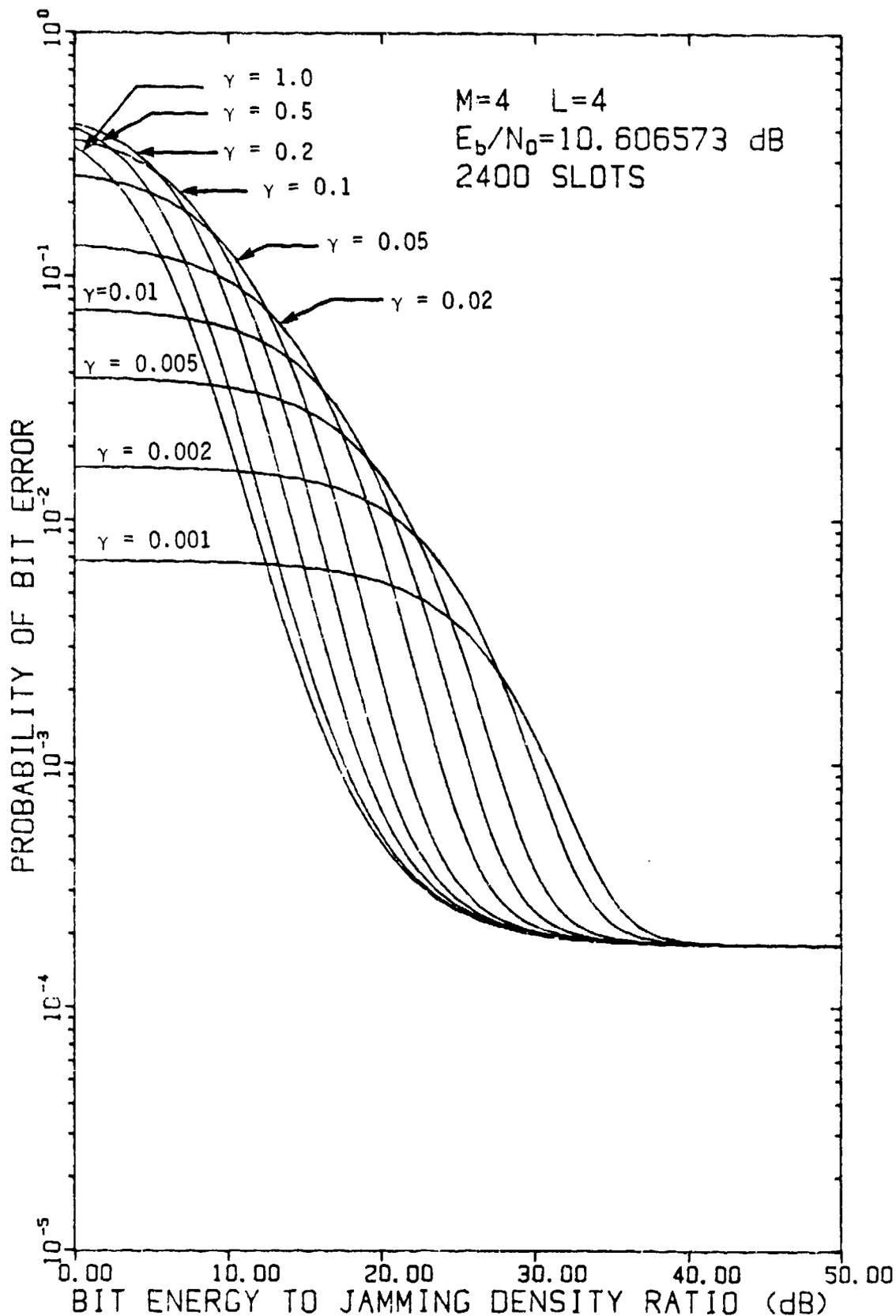


FIGURE 3.2-9 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH $M = 4$ AND $L = 4$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

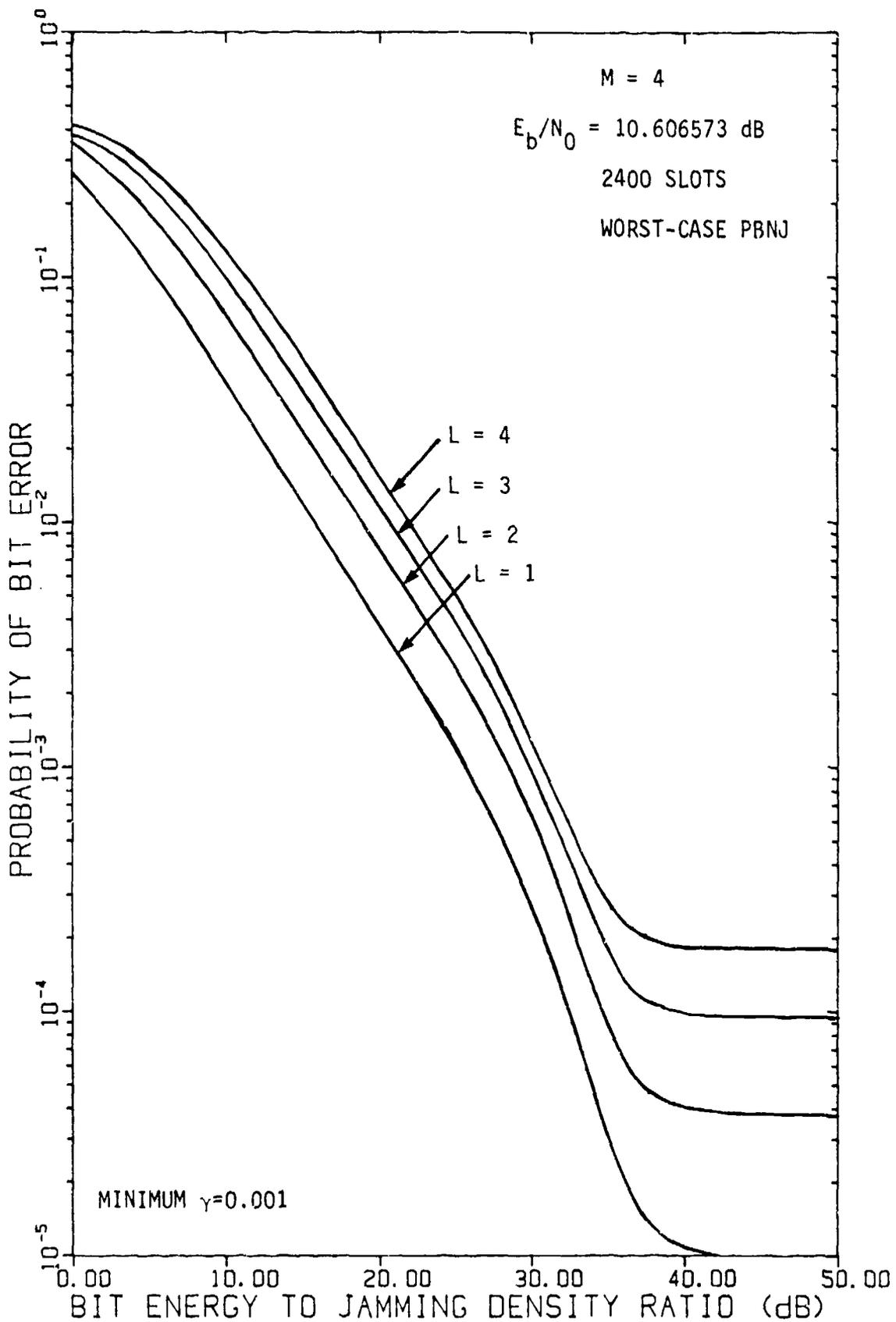


FIGURE 3.2-10 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER AND $M = 4$ WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING

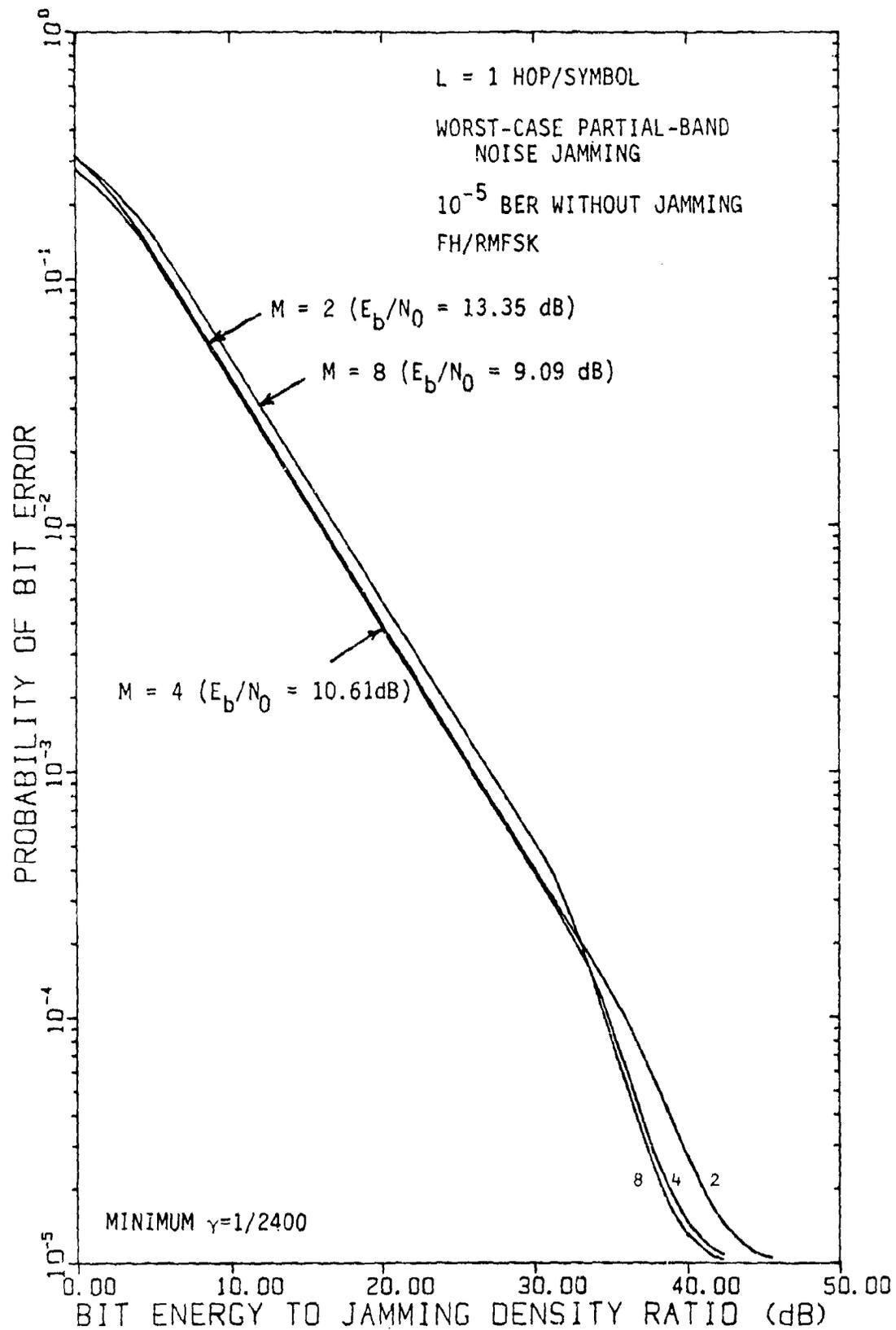


FIGURE 3.2-11 WORST-CASE PARTIAL-BAND NOISE JAMMING PERFORMANCE OF FH/RMFSK RECEIVERS FOR $L = 1$ HOP/SYMBOL AND $M = 2, 4, 8$ WHEN E_b/N_0 GIVES A 10^{-5} BER WITHOUT JAMMING

3.2.2 Hard-Decision Receiver

In this subsection we apply the explicit form of the error probability expression (2.4-14) to evaluate the symbol error probability, $P_s(e)$, for a square-law receiver with hard decisions. We consider M-ary cases of $M=2, 4$, and 8 with L values (hops per symbol) ranging from one through five. The worst-case or maximum probability of error is obtained by computing $P_s(e)$ upon varying the number of noise jammed hopping slots q , for an FH/RMFSK system comprised of 2400 hopping slots; i.e. a partial-band noise jamming (PBNJ) model.

We first present plots of numerical results for $P(e)$ versus the variable E_b/N_j with thermal noise (E_b/N_0) as a parameter. Practical values of E_b/N_0 were chosen for which the probability of error becomes 10^{-5} in the absence of jamming. These values are: 13.35247, 10.60657, and 9.09401 dB for M-ary signalling alphabets of $M=2, 4$, and 8 respectively. Corresponding performance plots are shown in Figures 3.2-12 to 3.2-14.

A comprehensive view of Figure 3.2-12 ($M=2$) reveals that all five of the L error curves could be grouped into three E_b/N_j regions of relative jamming strength: (1) below 11 dB (strong), (2) 11 dB to 28 dB (medium), and (3) beyond 28 dB (weak). Within the strong jamming region, we see a consistent $P(e)$ L -curve ordering of 4,5,2,3,1 and a 4,2,5,3,1 ranking in the weak region. The region defined as medium strength jamming exhibits "cross-overs" of the various L $P(e)$ curves. Similarly, for $M=4$ (Figure 3.2-13) and $M=8$ (Figure 3.2-14) this same three-region behavior exists as follows:

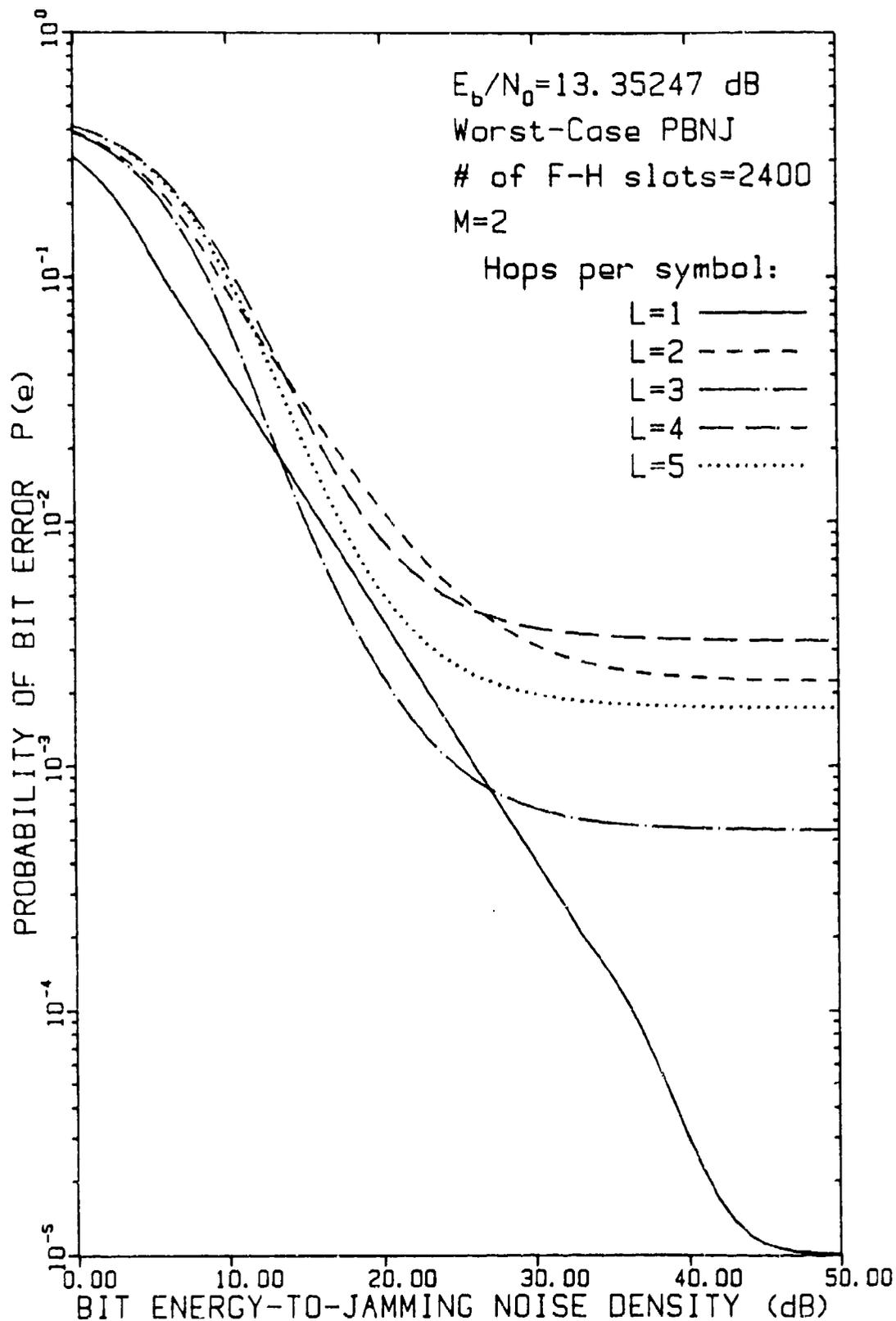


FIGURE 3.2-12 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO
 FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR $M=2$
 IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT
 $E_b/N_0 = 13.35247$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$)

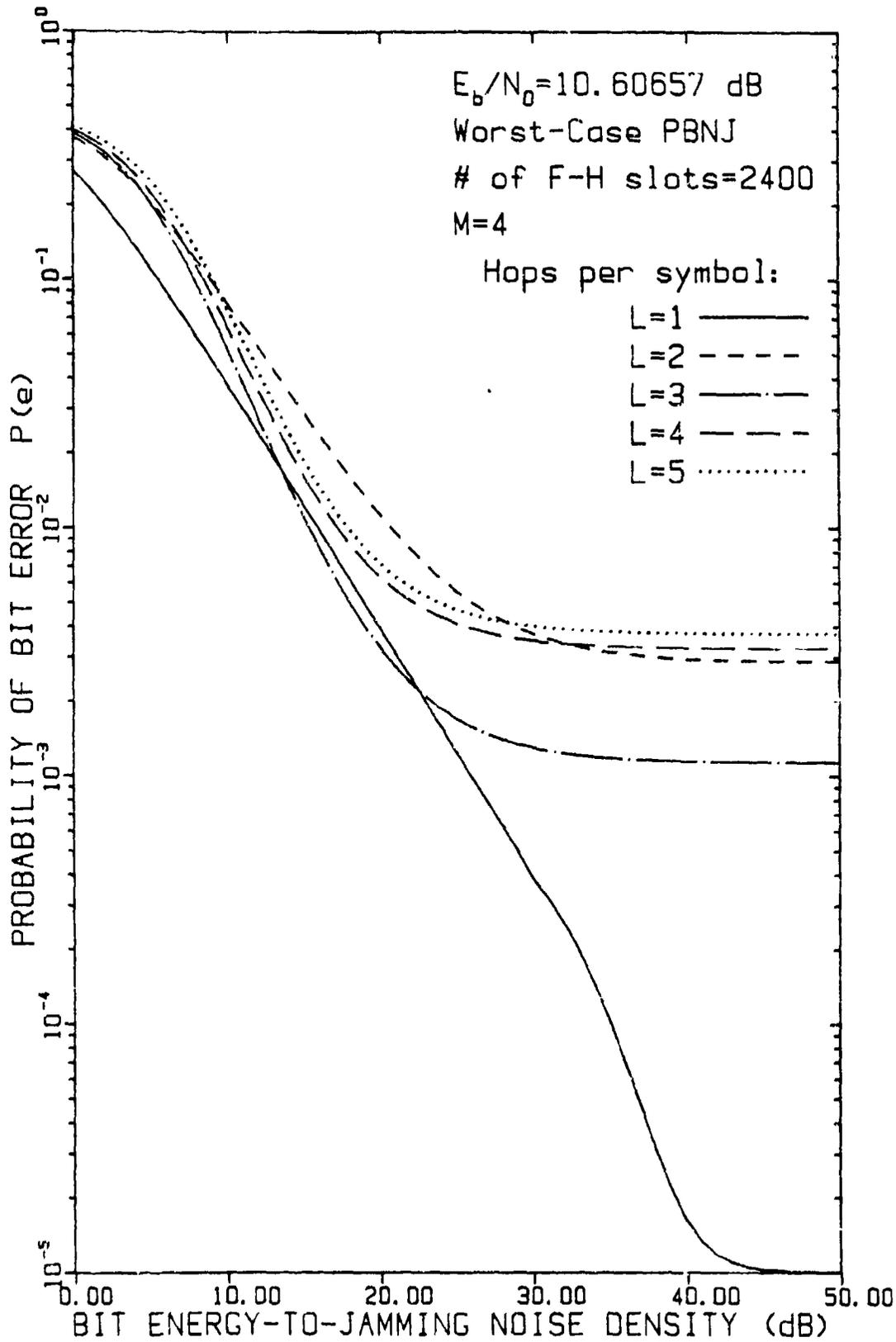


FIGURE 3.2-13 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR $M=4$ IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT $E_b/N_0 = 10.60657$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$)

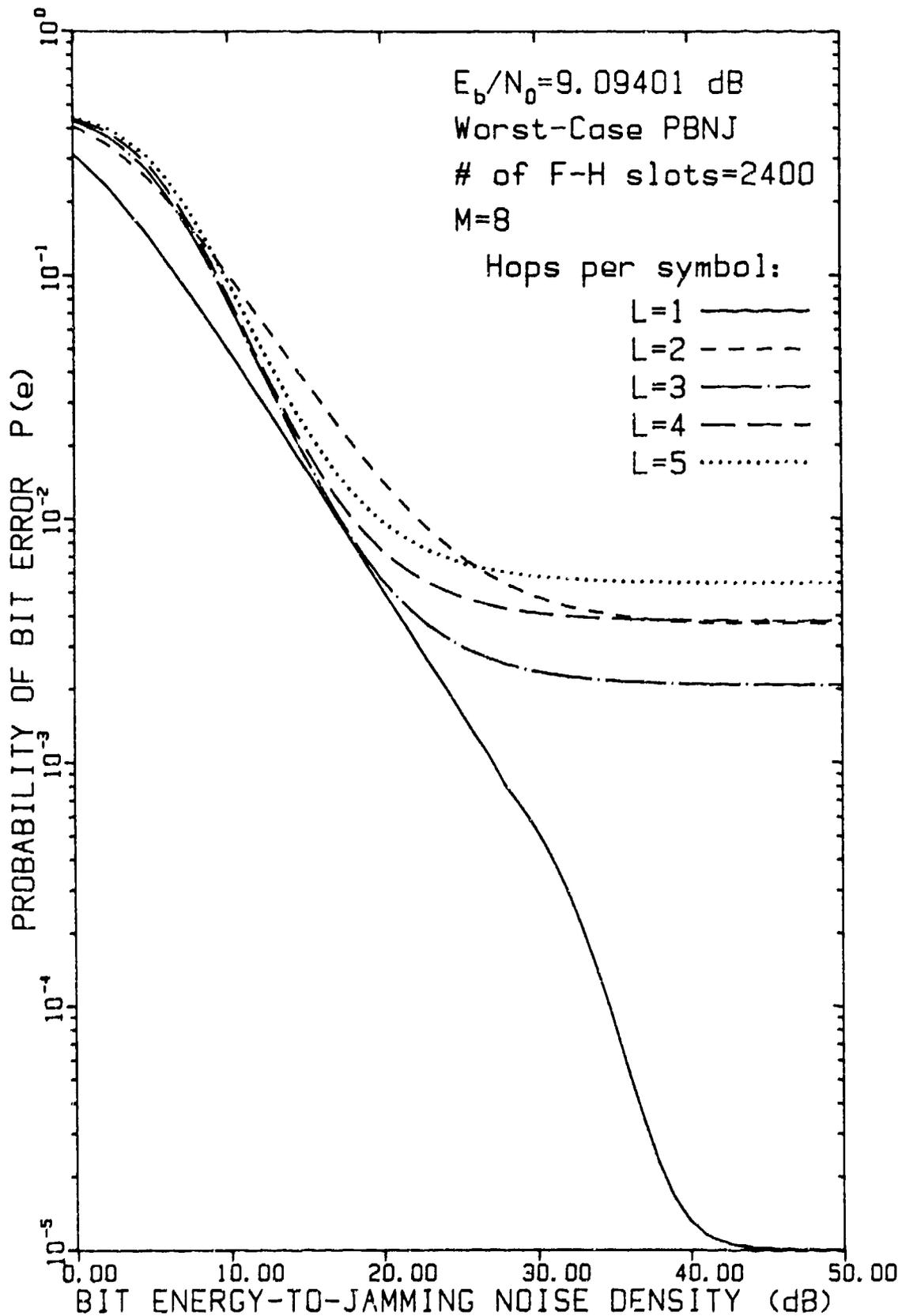


FIGURE 3.2-14 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR $M=8$ IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT $E_b/N_0 = 9.09401$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$)

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Relative Jamming Region of E_b/N_j (dB) Values

<u>M</u>	<u>Strong</u>	<u>Medium</u>	<u>Weak</u>
4	< 4	4 to 32	> 32
8	< 6	6 to 38	> 38 .

The ranking of the $L P(e)$ curves for both $M=4$ and $M=8$ in the strong jamming region is 5,4,3,2,1 while for weak jamming a 5,4,2,3,1 ordering is observed.

It is plain to see that for the most part no diversity improvement is realized by the communicator with the exception of a portion of the $L=3 P(e)$ curve in medium jamming for $M=2$ and $M=4$. This general behavior can be attributed to the dominance of the noncoherent combining loss existing for the stated thermal noise levels.

A somewhat different trend is noticed when the effect of thermal noise is minimized. Figures 3.2-15 through 3.2-17 show $P(e)$ results for $M=2, 4,$ and 8 respectively at E_b/N_0 levels of 20 dB. Clearly, the regions of strong and weak jamming are now quite discernable with a smaller crossover region (medium jamming) existing among the $L P(e)$ curves. However, we do notice a ranking in the weak jamming areas that differs from those of the 10^{-5} parameter E_b/N_0 curves previously presented. Here it is easily seen that the hard-decision receiver is uniformly 3 dB better for $L=1$ than for $L=2$ as described by (2.4-15) for any value of M . Also, a form of diversity improvement is realized as L becomes greater than 2 for all M -ary cases at E_b/N_j values of more than about 12 dB.

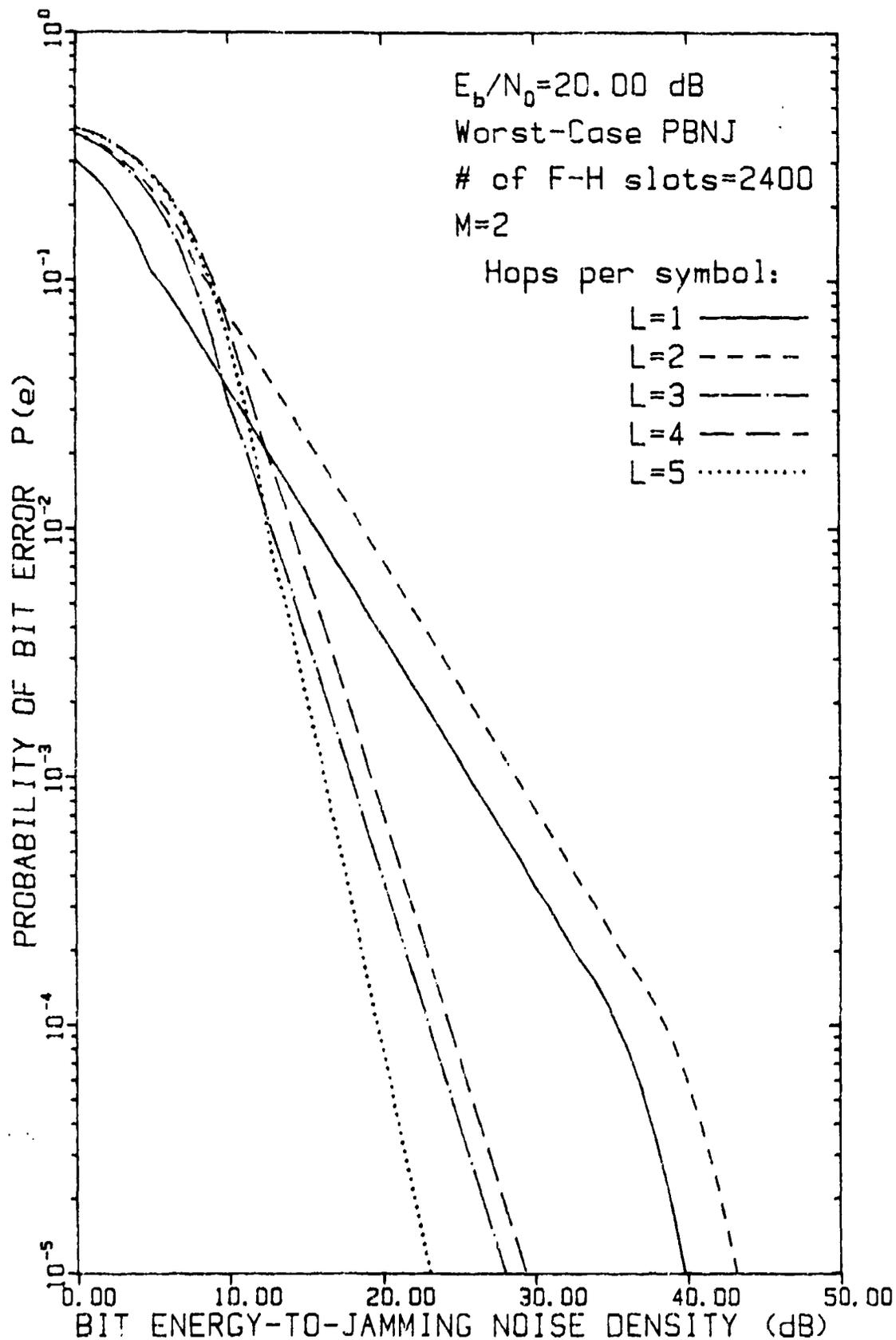


FIGURE 3.2-15 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR $M=2$ IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT $E_b/N_0=20$ dB (FOR MINIMIZATION OF THERMAL NOISE EFFECT)

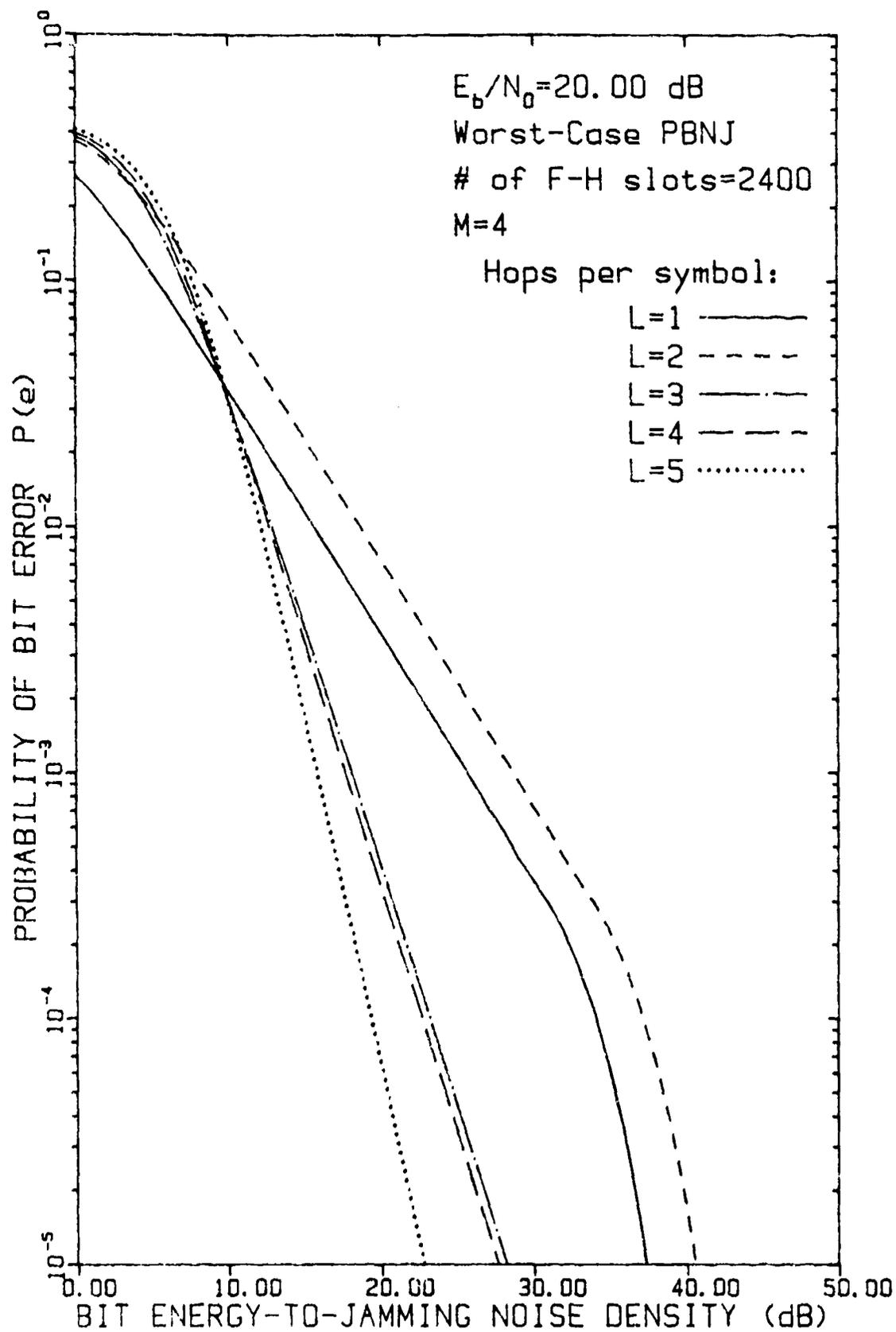


FIGURE 3.2-16 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR $M=4$ IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT $E_b/N_0 = 20$ dB (FOR MINIMIZATION OF THERMAL NOISE EFFECT)

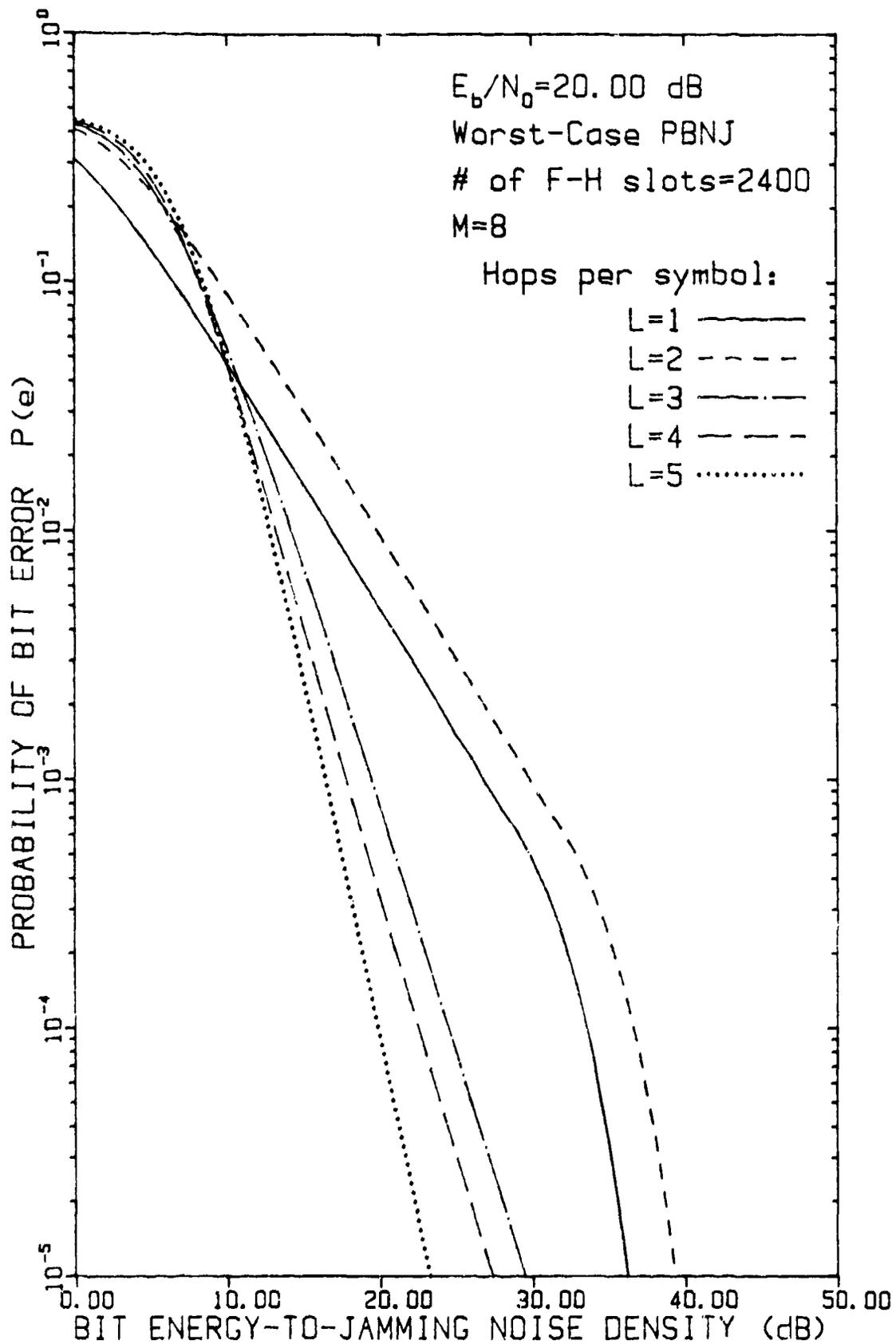


FIGURE 3.2-17 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR $M=8$ IN THE PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING AT $E_b/N_0=20$ dB (FOR MINIMIZATION OF THERMAL NOISE EFFECT)

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The case of $M=2$ is an exception to this trend for $L=3$ and $L=4$ as explained in subsection 2.4.2.3 where $L=4$ is shown to be 1.25 dB worse than that for $L=3$.

Thus we conclude that when thermal noise is minimized, a form of diversity improvement does exist in all M cases for E_b/N_j values greater than around 12 dB for $L > 2$ hops per symbol.

We now determine the optimum number of jammed slots (Q_{\max}) which yields the maximum probability of error for a given value of E_b/N_j . Figures 3.2-18 to 3.2-20 show such plots for case of $M=2, 4$, and 8 with L values ranging from one to five. It is seen that in all cases a definite ascending order of the L Q_{\max} curves exists for increasing E_b/N_j values as is to be expected for worst-case jamming calculations. For example, in Figure 3.2-18 ($M=2$) we see that at a 30 dB E_b/N_j value, the Q_{\max} value is 2 for $L=1$ and over 200 for $L=5$. Also in Figure 3.2-18 we note the "plateau" effect for all the curves at Q_{\max} equal to 2400. Now the breakpoint at which each individual L -curve falls off from the "plateau" represents that E_b/N_j value for which full-band jamming ($\gamma=1.0$) will not cause maximum probability of error. We can also characterize each L -curve behavior of Figure 3.2-18 as per three definite regions with respect to the "slope" of each curve. These regions, in terms of Q_{\max} values, are: (1) 2400 to about 900, (2) 900 down to approximately 20, and (3) below 20. Note that distinguishable breaks in the curves below 20 are due to the smaller quantized values of Q becoming more discernable for the lower values of the logarithmic Q_{\max} scale.

With regard to Figures 3.2-19 ($M=4$) and 3.3-20 ($M=8$), we see an asymptotic merging of the L -curves within the region of approximately $Q_{\max} = 800$ to 2400. Below a Q_{\max} of about 800 we have two more noticeable regions

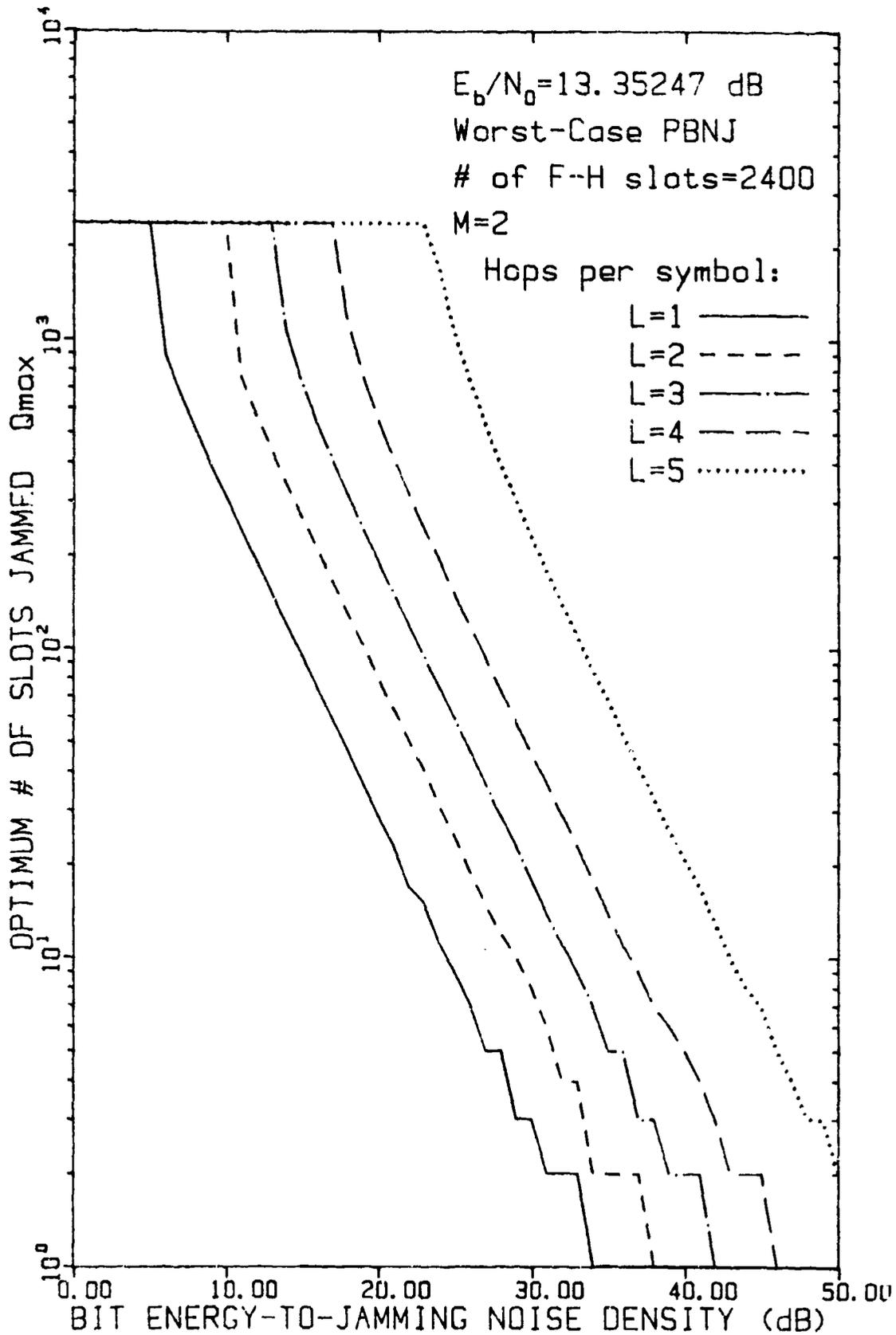


FIGURE 3.2-18 OPTIMUM NUMBER OF HOPPING SLOTS JAMMED (Q_{max}) THAT PRODUCES $P(e)_{max}$ VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR THE SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR $M=2$ AT $E_b/N_0 = 13.35246$ dB (FOR 10^{-5} BER WHEN $L=1$)

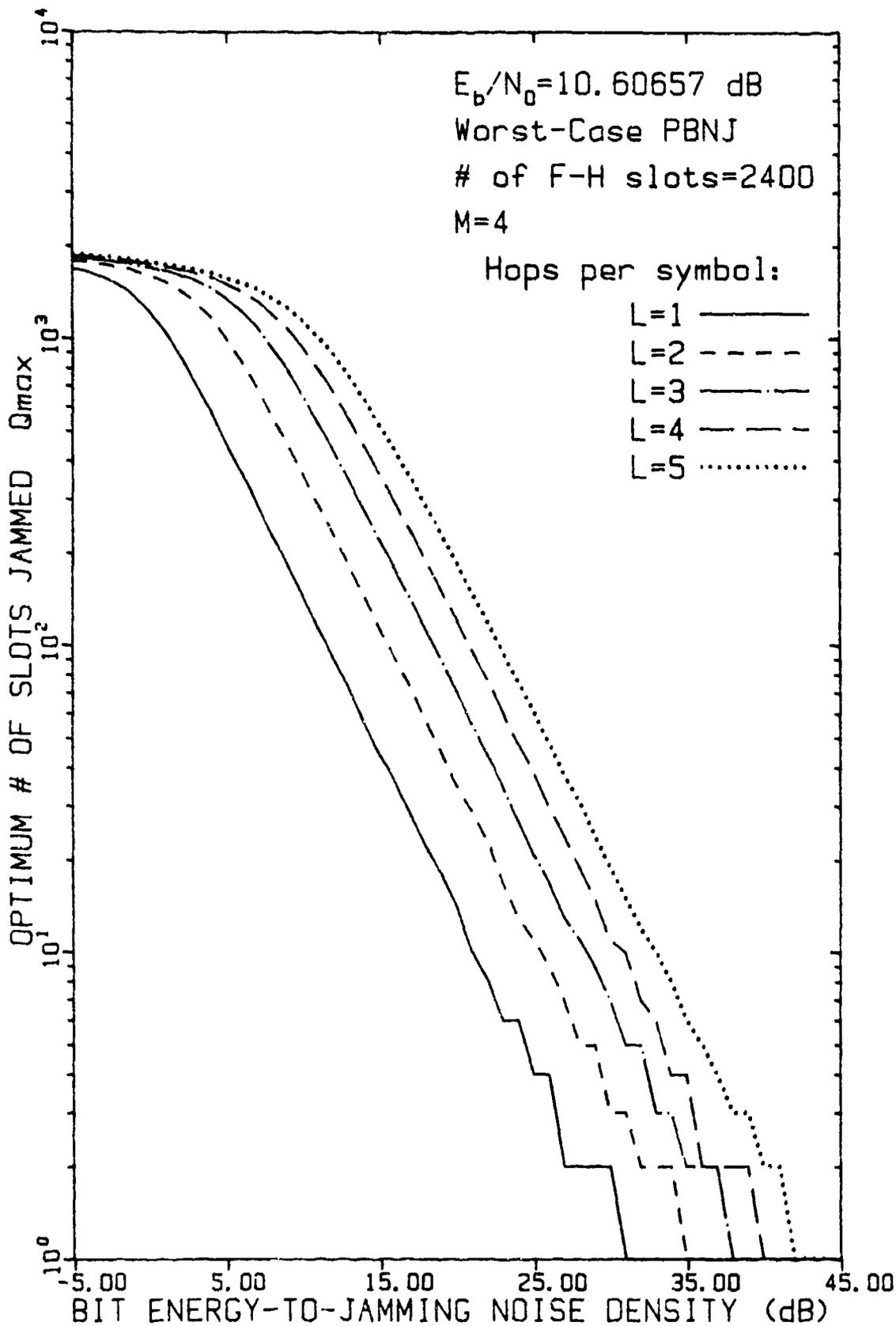


FIGURE 3.2-19 OPTIMUM NUMBER OF HOPPING SLOTS JAMMED (Q_{max} THAT PRODUCES $P(e)_{max}$) VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR THE SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR $M=4$ AT $E_b/N_0 = 10.60657$ dB (FOR 10^{-5} BER WHEN $L=1$)

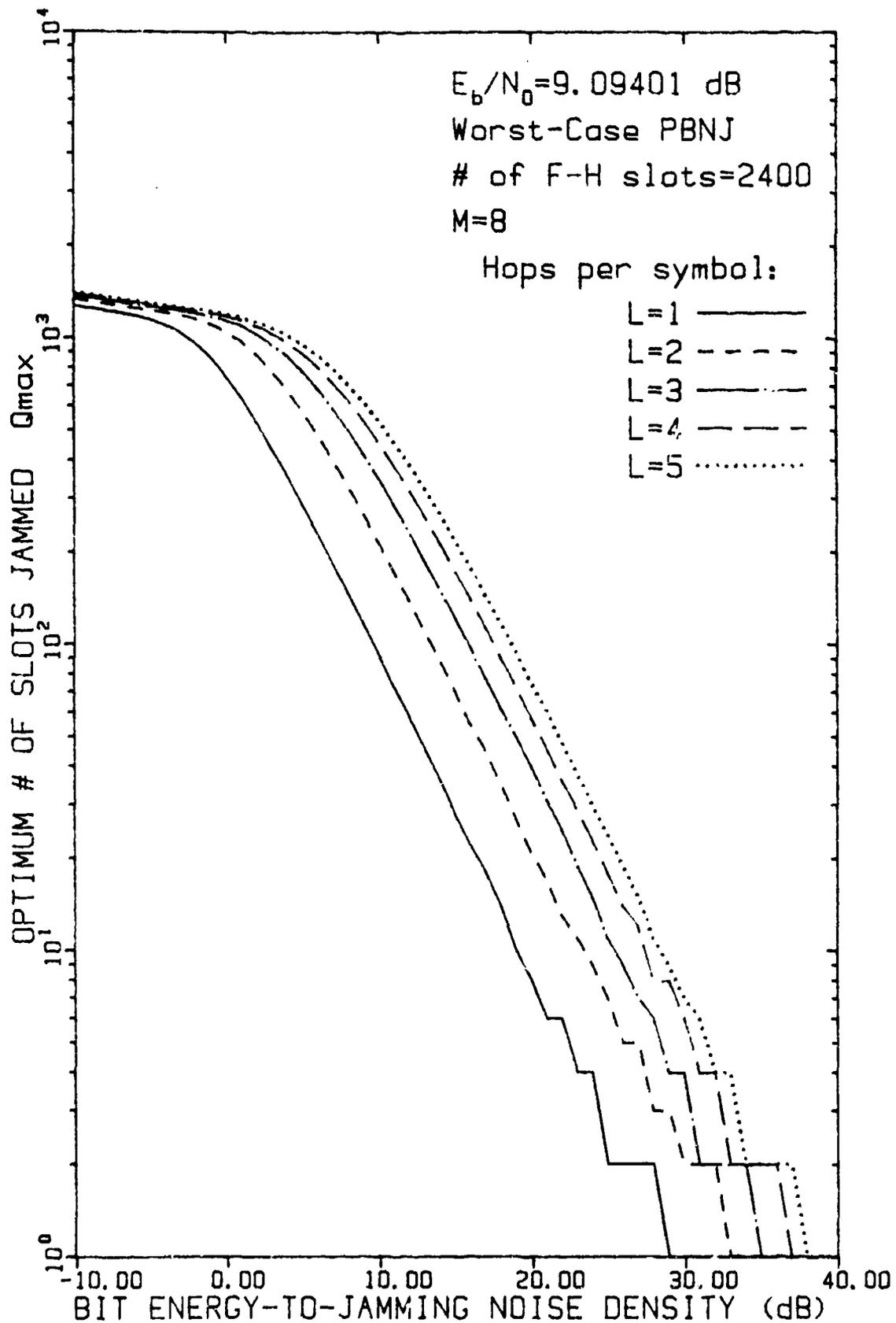


FIGURE 3.2-20 OPTIMUM NUMBER OF HOPPING SLOTS JAMMED (Q_{max} THAT PRODUCES $P(e)_{max}$) VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR THE SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR $M=8$ AT $E_b/N_0=9.09401$ dB (FOR 10^{-5} BER WHEN $L=1$)

similar to those in Figure 3.2-18. Additionally, we see that this asymptotic merging has yet to reach a full-band value of $Q_{\max}=2400$ for the minimum E_b/N_j value utilized in the calculations. Further computations (not shown) for lower values of E_b/N_j reveal the following points at which the $L=1$ curve breaks from the $Q_{\max}=2400$ ($\gamma=1.0$) value: $M=4$ at -149.0 dB, $M=8$ at -150.0 dB. Hence, in these cases, full-band jamming would only be optimum for a very large amount of available jamming power.

A final point of interest is shown in Figure 3.2-21 with respect to minimization of the thermal noise component. Here we see that for the binary ($M=2$) case, increasing the E_b/N_0 value over that used in Figure 3.2-18 causes the $L=5$ curve to move around 10 dB (E_b/N_j) lower while the $L=1$ curve decreases only about 1 dB.

The results indicate that the hard symbol decision receiver can be considered an ECCM receiver (for sufficiently high E_b/N_0), while the linear combining receiver cannot. Therefore, it is not diversity as such that yields an ECCM effect, but the combining technique. Hard decisions (in form of repetition coding) in effect limit the jamming effects on a given hop to that hop, whereas with linear combining a single, strongly jammed hop can dominate the soft decision.

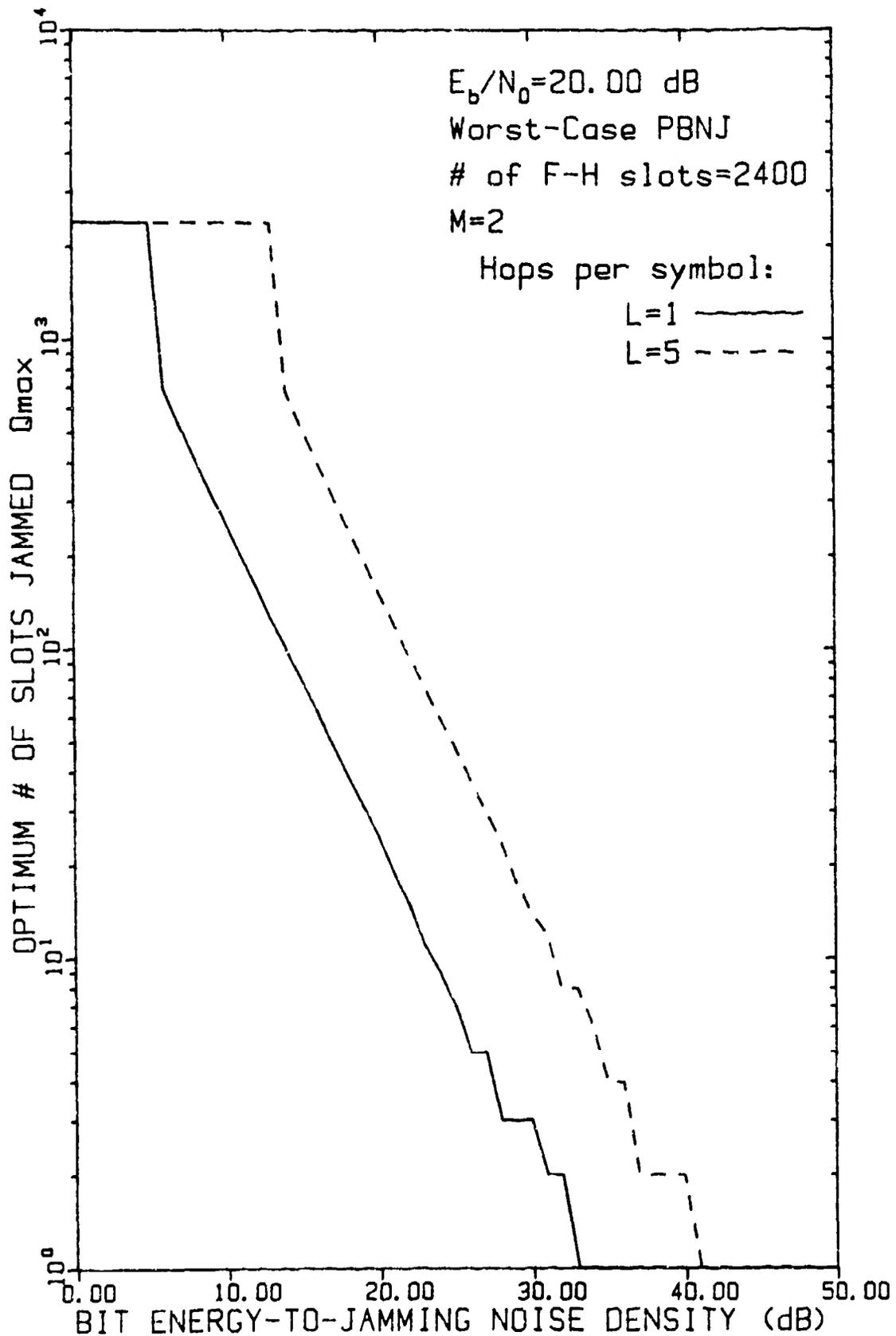


FIGURE 3.2-21 OPTIMUM NUMBER OF HOPPING SLOTS JAMMED (Q_{\max} THAT PRODUCES $P(e)_{\max}$) VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR THE SQUARE-LAW COMBINING RECEIVER WITH HARD DECISIONS FOR $M=2$ AND $L=1,5$ AT $E_b/N_0=20$ dB (FOR MINIMIZATION OF THERMAL NOISE)

4.0 FH/RMFSK PERFORMANCE USING SQUARE-LAW AGC RECEIVER

In this section we consider the case of the generic receiver shown in Figure 2.2-1 when the envelope samples are processed using the function

$$f(x_{mk}) = x_{mk}^2 \cdot w_{mk} = z_{mk} \quad (4.0-1)$$

That is, the decision statistics $\{z_m\}$ are weighted sums of samples of the squared envelope in each channel over multiple (L) hops.

For conventional FH/MFSK, where the symbol frequency slots are hopped together, it was assumed in [1] that all the slots are jammed or all the slots are not jammed on a given hop, and the weights were taken to be

$$w_{mk} = 1/\sigma_k^2 = \begin{cases} 1/\sigma_N^2, & \text{hop not jammed} \\ 1/\sigma_J^2, & \text{hop jammed.} \end{cases} \quad (4.0-2)$$

This weighting or normalization scheme was predicated on use of a separate channel, or perhaps a look-ahead scheme, to measure the noise power (perfectly) on each hop. The effect of the weighting is to de-emphasize the jammed hops in the summations

$$z_m = \sum_{k=1}^L z_{mk} \quad (4.0-3)$$

and therefore to mitigate the effect of the jamming on the symbol decision.

For FH/RMFSK, in general the different symbol frequency slots are independently jammed or not jammed when the system bandwidth contains power from a partial-band noise jammer. In Section 4.1, we discuss several normalization schemes of the AGC (adaptive gain control) type. The performances of two of these schemes are analyzed in Sections 4.2 and 4.3. We also consider the effect of hard-limiting the hop statistics $\{z_{mk}\}$ prior to combining.

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4.1 POSSIBLE AGC WEIGHTING SCHEMES FOR FH/RMFSK

The squared envelope samples in the M receiver channels on a given hop are weighted chi-squared random variables:

$$\begin{aligned}x_{1k}^2 &\sim \sigma_{1k}^2 \chi^2(2; \lambda_{1k} = 2S/\sigma_{1k}^2) \\ &= (\sigma_N^2 + v_{1k} \sigma_J^2) \chi^2[2; 2S/(\sigma_N^2 + v_{1k} \sigma_J^2)]\end{aligned}\quad (4.1-1a)$$

in the signal channel, and

$$x_{mk}^2 \sim \sigma_{mk}^2 \chi^2(2) = (\sigma_N^2 + v_{mk} \sigma_J^2) \chi^2(2), \quad m > 1, \quad (4.1-1b)$$

in the non-signal channels, where

$$v_{mk} = \begin{cases} 1, & \text{channel } m \text{ jammed on hop } k \\ 0, & \text{channel } m \text{ not jammed on hop } k. \end{cases} \quad (4.1-2)$$

We shall consider three approaches to AGC normalization, as illustrated in Figure 4.1-1:

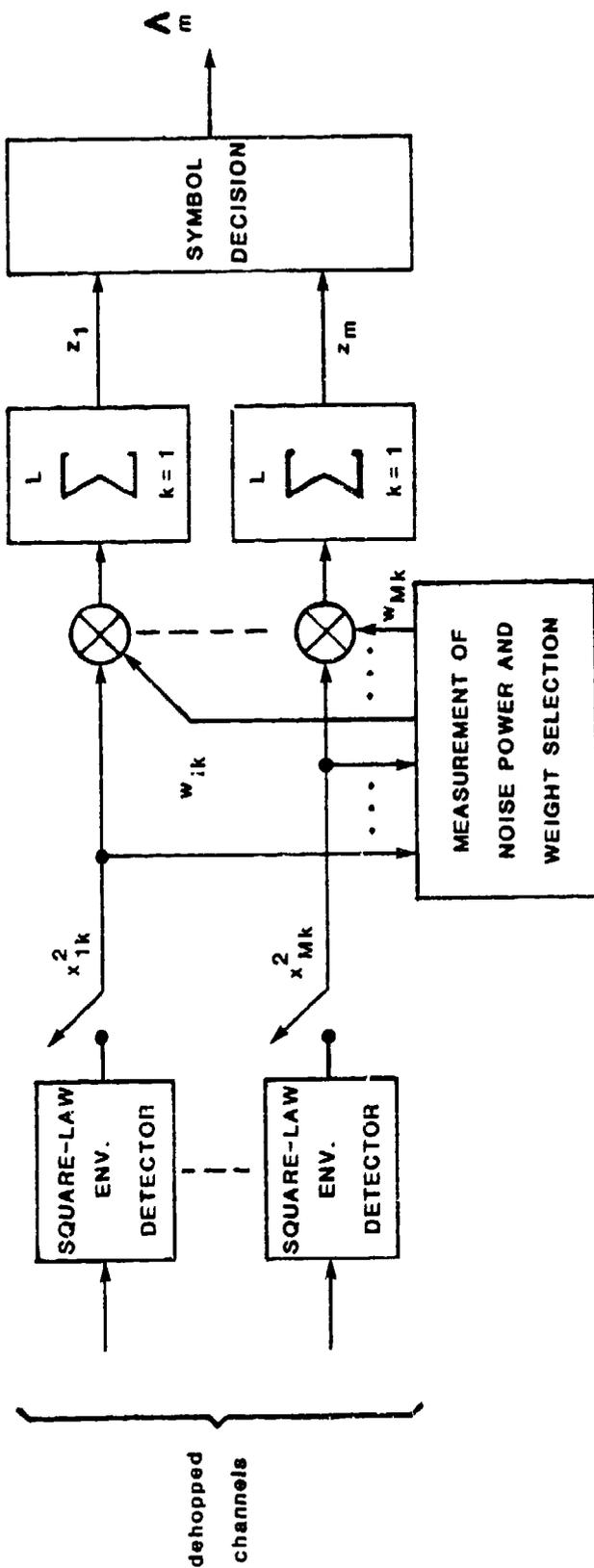
(a) Measurement of noise power in the M de-hopped channels on each hop and normalization (division) by the average received noise power (variance) in these channels.

(b) Individual measurement and normalization of each of the M channels on each hop.

(c) Normalization of the M channels by the same amount, depending on whether one or more of the channels are jammed, or none are jammed.

4.1.1 Average AGC Scheme.

An ideal measurement of average noise power in the M channels would yield the weights



Weighting schemes:

(a) Average AGC approach:

$$w_{mk} = \left[\frac{1}{2} E \left\{ \sum_{m=1}^M x_{Mk}^2 - S \right\}^{-1} \right]^{-1} = \left[\frac{1}{M} \sum_{m=1}^M \sigma_{mk}^2 \right]^{-1} = w_k$$

(b) Individual-channel AGC approach: $w_{mk} = (\sigma_{mk}^2)^{-1}$

(c) Any-channel-jammed AGC approach: $w_{mk} = w_k = \begin{cases} (\sigma_N^2)^{-1} & \text{no channels jammed} \\ (\sigma_T^2)^{-1} & \text{one or more channels jammed} \end{cases}$

FIGURE 4.1-1 POSSIBLE AGC NORMALIZATION SCHEMES

$$\begin{aligned}
 w_{mk} \equiv w_k &= \left(\frac{1}{M} \sum_{m=1}^M \sigma_{mk}^2 \right)^{-1} \\
 &= \left(\sigma_N^2 + \sigma_J^2 \sum_{m=1}^M v_{mk}/M \right)^{-1}.
 \end{aligned} \tag{4.1-3}$$

There are M possible values to these weights. After normalization, the $\{z_{mk}\}$ become

$$z_{1k} \sim w_k \sigma_{1k}^2 \chi^2(2; 2S/\sigma_{1k}^2) \tag{4.1-4a}$$

$$z_{mk} \sim w_k \sigma_{mk}^2 \chi^2(2), \quad m > 1. \tag{4.1-4b}$$

The effective weights $w_k \equiv w_k \sigma_{mk}^2$ on the chi-squared variables in a given channel can take $2(M-1)$ values. Therefore the decision statistics for this normalization scheme have the form

$$z_m \sim \sum_{k=1}^L w_k \chi^2(2; \lambda_{mk}). \tag{4.1-5}$$

The distribution of sums of non-equally weighted chi-squared random variables is extremely difficult to compute. For this reason, it is not feasible to consider calculation of the error performance using such a weighting scheme.

4.1.2 Individual Channel AGC Scheme.

Ideal measurements of noise power in each of the M channels would yield the weights

$$w_{mk} = (\sigma_{mk}^2)^{-1} \tag{4.1-6}$$

and the decision statistics

$$z_1 = \sum_{k=1}^L \chi^2(2; \lambda_{1k}) = \chi^2(2L; \lambda_1) \tag{4.1-7a}$$

in the signal channel, where

$$\lambda_1 = \sum_{k=1}^L \lambda_{1k} = (L - \ell_1) \cdot 2S / \sigma_N^2 + \ell_1 \cdot 2S / \sigma_T^2, \quad (4.1-7b)$$

and in the non-signal channels,

$$z_m = \sum_{k=1}^L \chi^2(2) = \chi^2(2L). \quad (4.1-7c)$$

Thus whatever else the merits of this normalization scheme may be, it yields decision statistics which are purely chi-squared random variables with $2L$ degrees of freedom. In fact, from (4.1-7) we observe that the discernable jamming events are characterized solely by the number of hops jammed in the signal channel, ℓ_1 .

4.1.3 Any-Channel-Jammed AGC Scheme.

This scheme takes the approach that if any of the channels on a given hop is jammed, then all the channels are normalized by $\sigma_T^2 = \sigma_N^2 + \sigma_J^2$; otherwise they are all normalized by σ_N^2 . Expressed mathematically, the weights are

$$w_{mk} \equiv w_k = \begin{cases} (\sigma_N^2)^{-1}, & \sum_{m=1}^M v_{mk} = 0 \\ (\sigma_T^2)^{-1}, & \text{otherwise.} \end{cases} \quad (4.1-8)$$

The result of this approach is that the hop statistics fall into three categories:

- (a) channel not jammed, normalized by σ_N^2
- (b) channel not jammed, normalized by σ_T^2
- (c) channel jammed, normalized by σ_T^2 .

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If we define ℓ_0 as the number of hops on which at least one of the channels is jammed, we find that in channel m ($m=1,2,\dots,M$) there would be

- * $(L-\ell_0)$ hops with noise power σ_N^2 normalized by σ_N^2
- * $(\ell_0-\ell_m)$ hops with noise power σ_N^2 normalized by σ_T^2
- * ℓ_m hops with noise power σ_T^2 normalized by σ_T^2 .

Therefore, for a given ℓ_0 and jamming event vector $\underline{\ell} = (\ell_1, \ell_2, \dots, \ell_M)$, the decision statistics would have the following distributions:

$$\begin{aligned} z_1 &\sim \chi^2[2(L-\ell_0); 2(L-\ell_0)S/\sigma_N^2] + (\sigma_N^2/\sigma_T^2)\chi^2[2(\ell_0-\ell_1); 2(\ell_0-\ell_1)S/\sigma_N^2] \\ &\quad + \chi^2[2\ell_1; 2\ell_1S/\sigma_T^2] \\ &= \chi^2[2(L+\ell_1-\ell_0); 2(L-\ell_0)S/\sigma_N^2 + 2\ell_1S/\sigma_T^2] \\ &\quad + K^{-1}\chi^2[2(\ell_0-\ell_1); 2(\ell_0-\ell_1)S/\sigma_N^2], \ell_1 \neq \ell_0; \end{aligned} \quad (4.1-9a)$$

and

$$\begin{aligned} z_m &\sim \chi^2[2(L-\ell_0)] + (\sigma_N^2/\sigma_T^2)\chi^2[2(\ell_0-\ell_m)] + \chi^2(2\ell_m) \\ &= \chi^2[2(L+\ell_m-\ell_0)] + K^{-1}\chi^2[2(\ell_0-\ell_m)], m > 1; \ell_m \neq \ell_0. \end{aligned} \quad (4.1-9b)$$

When $\ell_m = \ell_0$ the case of noise power σ_N^2 normalized by σ_T^2 does not occur and the distributions are:

$$z_1 \sim \chi^2[2L; 2(L-\ell_0)S/\sigma_N^2 + 2\ell_0S/\sigma_T^2], \ell_1 = \ell_0; \quad (4.1-9c)$$

and

$$z_m \sim \chi^2(2L), \ell_m = \ell_0. \quad (4.1-9d)$$

As in Section 3, we use $K \triangleq \sigma_T^2/\sigma_N^2 > 1$. We see from (4.1-9) that the decision statistics are in general sums of two unequally weighted chi-squared random variables. Analysis of this distribution is difficult but has been accomplished previously, in Section 3. There is the additional complexity, however, that the jamming events now must be described by an additional parameter: ℓ_0 , the number of hops with at least one channel jammed. This task can be achieved as shown below.

4.2 ANALYSIS OF FH/RMFSK PERFORMANCE USING INDIVIDUAL CHANNEL AGC SCHEME

Now we obtain the probability of bit error for the FH/RMFSK receiver using the individual channel AGC normalization scheme.

4.2.1 Conditional Probability Of Error.

The probability of a symbol error, given a jamming event described by $\underline{\ell} = (\ell_1, \ell_2, \dots, \ell_M)$, is

$$\begin{aligned} P_s(e|\underline{\ell}) &= 1 - P_s(C|\underline{\ell}) \\ &= 1 - \int_0^\infty d\alpha p_{z_1}(\alpha) \prod_{m=2}^M \int_0^\alpha d\beta_m p_{z_m}(\beta_m). \end{aligned} \quad (4.2-1)$$

From (4.1-7) we observe that the non-signal channel decision statistics $\{z_m, m>1\}$ are identically distributed as chi-squared random variables with $2L$ degrees of freedom. Thus

$$\begin{aligned} \int_0^\alpha d\beta_m p_{z_m}(\beta_m) &= \int_0^\alpha d\beta p_{z_2}(\beta), \quad m = 2, 3, \dots, M \\ &= 1 - \frac{\Gamma(L; \alpha/2)}{\Gamma(L)} \end{aligned} \quad (4.2-2a)$$

$$= 1 - e^{-\alpha/2} \sum_{r=0}^{L-1} \frac{(\alpha/2)^r}{r!}, \quad (4.2-2b)$$

and

$$\begin{aligned} P_s(C|\underline{\ell}, \alpha) &\triangleq \prod_{m=2}^M \int_0^\alpha d\beta_m p_{z_m}(\beta_m) = \left[1 - e^{-\alpha/2} \sum_{r=0}^{L-1} \frac{(\alpha/2)^r}{r!} \right]^{M-1} \\ &= \sum_{k=0}^{M-1} \binom{M-1}{k} (-1)^k e^{-k\alpha/2} \left[\sum_{r=0}^{L-1} \frac{(\alpha/2)^r}{r!} \right]^k. \end{aligned} \quad (4.2-3)$$

From Section 3, equation (3.1-14), we find that

$$P_S(C|\underline{\ell}, \alpha) = \sum_{k=0}^{M-1} \binom{M-1}{k} (-1)^k e^{-k\alpha/2} \sum_{r=0}^{k(L-1)} \frac{C(r,k)}{r!} \left(\frac{\alpha}{2}\right)^r, \quad (4.2-4a)$$

where

$$\begin{aligned} C(0,k) &= 1 \\ C(r,k) &= \frac{1}{r} \sum_{n=1}^{\min(r,L-1)} \binom{r}{n} [(k+1)n - r] C(r-n,k), \end{aligned} \quad (4.2-4b)$$

$r > 0, L \geq 2.$

Substitution in (4.2-1) yields

$$P_S(e|\underline{\ell}) = 1 - \sum_{k=0}^{M-1} \binom{M-1}{k} (-1)^k \sum_{r=0}^{k(L-1)} \frac{C(r,k)}{r!} \int_0^{\infty} d\alpha p_{z_1}(\alpha) e^{-k\alpha/2} \left(\frac{\alpha}{2}\right)^r. \quad (4.2-5)$$

From (4.1-7) and (3.1-9) the pdf for z_1 is

$$p_{z_1}(\alpha) = \frac{1}{2} e^{-(\alpha+\lambda_1)/2} \left(\frac{\alpha}{\lambda_1}\right)^{(L-1)/2} I_{L-1}\left(\sqrt{\alpha\lambda_1}\right), \quad (4.2-6)$$

with $\lambda_1 = 2(L-\ell_1)\rho_N + 2\ell_1\rho_T$. Thus the required integral in (4.2-5) is

$$\begin{aligned} &\left(\frac{1}{1+k}\right)^{r+L} \int_0^{\infty} dx \left(\frac{x}{2}\right)^r \exp\left(-\frac{k\lambda_1/2}{k+1}\right) p_{\chi^2}\left(x; 2L, \lambda = \frac{\lambda_1}{1+k}\right) \\ &= \left(\frac{1}{1+k}\right)^{r+L} \exp\left(-\frac{k\lambda_1/2}{k+1}\right) r! \mathcal{L}_r^{L-1}\left(\frac{\lambda_1/2}{k+1}\right), \end{aligned} \quad (4.2-7)$$

giving the error probability

$$P_s(e|\underline{\ell}) = \sum_{k=1}^{M-1} \binom{M-1}{k} \frac{(-1)^{k+1}}{(1+k)^L} \sum_{r=0}^{k(L-1)} \frac{C(r,k)}{(1+k)^r} \exp\left(-\frac{k\lambda_1/2}{k+1}\right) \rho_r^{L-1} \left(\frac{-\lambda_1/2}{1+k}\right) \quad (4.2-8)$$

$$\equiv P_s(e|\underline{\ell}_1) .$$

4.2.2 Total Error Probability.

Since the symbol error probability depends only on whether the signal channel is jammed, the total bit error probability is

$$P_b(e) = \sum_{\ell_1=0}^L \binom{L}{\ell_1} \gamma^{\ell_1} (1-\gamma)^{L-\ell_1} \frac{M/2}{M-1} P_s(e|\underline{\ell}_1), \quad (4.2-9a)$$

where

$$\gamma = \text{Pr}\{\text{channel 1 jammed on hop } k\} . \quad (4.2-9b)$$

Noting that (4.2-8) and (4.2-9) are mathematically identical to equations (4-26) in [1], we observe that MFSK and RMFSK give equivalent PBNJ performances for individual channel AGC normalization.

4.3 ANALYSIS OF FH/RMFSK PERFORMANCE USING ANY-CHANNEL-JAMMED AGC SCHEME

In what follows we find the probability of bit error for the FH/RMFSK receiver using the any-channel-jammed AGC normalization scheme (ACJ).

4.3.1 Conditional Probability Of Error.

The probability of a symbol error, given the jamming event $(\ell_0, \underline{\ell})$, is

$$P_s(e|\ell_0, \underline{\ell}) = 1 - P_s(C|\ell_0, \underline{\ell})$$

$$= 1 - \int_0^\infty d\alpha \, p_{Z_1}(\alpha; \ell_0, \underline{\ell}_1) \prod_{m=2}^M \int_0^\alpha d\beta_m \, p_{Z_m}(\beta_m; \ell_0, \underline{\ell}_m). \quad (4.3-1)$$

$$\begin{aligned} \text{Since } P_s(C|\underline{\ell}) &= \Pr\{z_2 < z_1, z_3 < z_1, \dots, z_M < z_1\} \\ &= \Pr\{Kz_2 < Kz_1, Kz_3 < Kz_1, \dots, Kz_M < Kz_1\}, \end{aligned} \quad (4.3-2)$$

we may analyze the error probability using the statistics $\{u_m\}$ instead of $\{z_m\}$

$$\begin{aligned} \text{where } u_1 \equiv Kz_1 &= \chi^2[2(\ell_0 - \ell_1); 2(\ell_0 - \ell_1)\rho_N] \\ &\quad + K\chi^2[2(L + \ell_1 - \ell_0); 2(L - \ell_0)\rho_N + 2\ell_1\rho_T] \end{aligned} \quad (4.3-3a)$$

and

$$u_m \equiv Kz_m = \chi^2[2(\ell_0 - \ell_m)] + K\chi^2[2(L + \ell_m - \ell_0)] \quad m > 1. \quad (4.3-3b)$$

From Appendix A, the pdf's of these random variables are as follows: for the signal channel,

$$p_{u_1}(\alpha) = \begin{cases} p_{\chi^2}(\alpha; 2L, 2L\rho_N), & \ell_0 - \ell_1 = L \end{cases} \quad (4.3-4a)$$

$$p_{u_1}(\alpha) = \begin{cases} \frac{1}{K} p_{\chi^2}(\alpha/K; 2L, 2(L - \ell_0)\rho_N + 2\ell_1\rho_T), & \ell_0 - \ell_1 = 0; \end{cases} \quad (4.3-4b)$$

$$p_{u_1}(\alpha) = \begin{cases} \sum_{n=0}^{\infty} c_n p_{\chi^2}(\alpha; 2L + 2n, 2(\ell_0 - \ell_1)\rho_N), & 0 < \ell_0 - \ell_1 < L; \end{cases} \quad (4.3-4c)$$

where

$$c_n = e^{-(L - \ell_0)\rho_N - \ell_1\rho_T} \left(\frac{K-1}{K}\right)^n \left(\frac{1}{K}\right)^{L - \ell_0 + \ell_1} \frac{\rho_n^{L - \ell_0 + \ell_1 - 1}}{\Gamma(L - \ell_0 + \ell_1)} \left[\frac{-(L - \ell_0)\rho_N \ell_1 \rho_T}{K-1} \right]. \quad (4.3-4d)$$

For the nonsignal channels ($m > 1$),

$$p_{u_m}(\alpha) = \begin{cases} p_{\chi^2}(\alpha; 2L), & \ell_0 - \ell_m = L \\ \frac{1}{K} p_{\chi^2}(\alpha/K; 2L), & \ell_0 - \ell_m = 0 \\ \sum_{n=0}^{\infty} b_n p_{\chi^2}(\alpha; 2L+2n), & 0 < \ell_0 - \ell_m < L \end{cases} \quad \begin{array}{l} (4.3-5a) \\ (4.3-5b) \\ (4.3-5c) \end{array}$$

using the coefficients

$$b_n = \left(\frac{K-1}{K}\right)^n \left(\frac{1}{K}\right)^{L-\ell_0+\ell_m} \binom{n+L-\ell_0+\ell_m-1}{n} \quad (4.3-5d)$$

The symbol error probability expression for the $\{u_m\}$ statistics is (4.3-1) with the subscripts $\{u_m\}$ instead of $\{z_m\}$, $m=1,2,\dots,M$.

4.3.1.1 Formulations of nonsignal channel probabilities

Since the nonsignal channel pdf's are identical except for the parameters $\{\ell_m\}$, the number of hops jammed in the individual channels, we may express the product

$$\prod_{m=2}^M \int_0^{\alpha} d\beta_m p_{u_m}(\beta_m; \ell_0, \ell_m) \prod_{m=2}^M \Pr\{u_m < \alpha\} \quad (4.3-6)$$

in terms of the numbers of channels with certain combinations of ℓ_0 and the $\{\ell_m\}$. The probabilities needed are

$$\Pr\{u_m < \alpha\} = \begin{cases} 1 - \Gamma(L; \alpha/2)/\Gamma(L), & \ell_0 - \ell_m = L \\ 1 - \Gamma(L; \alpha/2K)/\Gamma(L), & \ell_0 - \ell_m = 0 \\ 1 - \sum_{n=0}^{\infty} b_n \Gamma(L+n; \alpha/2)/\Gamma(L+n), & 0 < \ell_0 - \ell_m < L. \end{cases} \quad \begin{array}{l} (4.3-7a) \\ (4.3-7b) \\ (4.3-7c) \end{array}$$

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Upon comparing (4.3-7) with (3.1-11), we observe that

$$\Pr\{u_m < \alpha\} \equiv F_L(\alpha; L - \ell_0 + \ell_m). \quad (4.3-8)$$

Thus we can write the product in (4.3-6) as

$$\prod_{m=2}^M \Pr\{u_m < \alpha\} = [F_L(\alpha; 0)]^{n_0} [F_L(\alpha; 1)]^{n_1} \dots [F_L(\alpha; L)]^{n_L} \quad (4.3-9)$$

where

$$n_i \triangleq \# \text{ (channels with } L - \ell_0 + \ell_m = i), \quad (4.3-10)$$

and $[F_L(\alpha; p)]^{n_p}$ is given by (3.1-14), (3.1-17), and (3.1-28).

4.3.1.2 Formulation of symbol error probability in terms of previous results (Section 3).

If we denote the conditional probability of symbol error for the square-law linear combining FH/RMFSK receiver studied in Section 3 by

$$P_s(e; \rho_N, \rho_T | \underline{\ell})_{LC}, \quad (4.3-11)$$

we can by analogy express the conditional probability of symbol error for the any-channel-jammed AGC receiver as

$$P_s(e; \rho_N, \rho_T | \ell_0, \underline{\ell})_{ACJ} = P_s(e; \rho_N, \rho'_T | \underline{v})_{LC} \quad (4.3-12a)$$

where

$$\rho'_T = \begin{cases} \frac{\ell_1 \rho_T + (L - \ell_0) \rho_N}{L - \ell_0 + \ell_1}, & \ell_0 - \ell_1 \neq L \\ \rho_T, & \ell_0 - \ell_1 = L \end{cases} \quad (4.3-12b)$$

and

$$\underline{v} \triangleq (L - \ell_0 + \ell_1, L - \ell_0 + \ell_2, \dots, L - \ell_0 + \ell_M). \quad (4.3-13)$$

4.3.2 Enumeration and Probabilities for Jamming Events.

The enumeration of jamming events, and their probabilities, has already been accomplished in Section 2.2 for the situation in which the jamming event is sufficiently described by the vector $\underline{\ell} = (\ell_1, \ell_2, \dots, \ell_M)$. Now our task is to develop for use in (4.3-12) the conditional probabilities $\Pr\{\ell_0 | \underline{\ell}\}$ for the parameter ℓ_0 , the number of hops on which at least one channel is jammed, given the vector $\underline{\ell}$.

The enumeration technique treated in Section 2.2 recognized the arbitrariness of the channel numbers m for $m > 1$ (nonsignal channels) by assuming that the calculations will generate the partially ordered $\underline{\ell}$ vector

$$\underline{\ell}' = \{(\ell_1, \ell_2, \dots, \ell_M) : \ell_2 \leq \ell_3 \leq \dots \leq \ell_M\}. \quad (4.3-14)$$

Thus, with this ordering the range of ℓ_0 is

$$\ell_x \triangleq \max(\ell_1, \dots, \ell_M) \leq \ell_0 \leq \min(L, \ell_1 + \ell_2 + \dots + \ell_M). \quad (4.3-15)$$

The number of elementary or $[v]$ -matrix jamming events characterized by a given $\underline{\ell}$ or $\underline{\ell}'$ vector is

$$\#([v] \rightarrow \underline{\ell}) = \binom{L}{\ell_1} \binom{L}{\ell_2} \dots \binom{L}{\ell_M} = \prod_{m=1}^M \binom{L}{\ell_m}, \quad (4.3-16)$$

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and each $\underline{\ell}'$ vector represents

$$\binom{M-1}{n_0, n_1, \dots, n_L}, \quad n_k = \#(\ell_m = k), \quad m > 1 \quad (4.3-17)$$

$\underline{\ell}$ vectors. Thus

$$\sum_{\underline{\ell}} \binom{M-1}{n_0, n_1, \dots, n_L} \Pr\{\underline{\ell}'; M, L\} = 1. \quad (4.3-18)$$

Now for jamming events specified by ℓ_0 as well as $\underline{\ell}$, the number of elementary jamming events thus specified can be shown to be

$$\#([v] \rightarrow \ell_0, \underline{\ell}) \binom{L}{\ell_0} \sum_{r=0}^{\ell_0 - \ell_x} \binom{\ell_0}{r} (-1)^r \prod_{m=1}^M \binom{\ell_0 - r}{\ell_m}. \quad (4.3-19)$$

For example, if $\ell_0 = \ell_x = \max(\ell_m)$, there are

$$\binom{L}{\ell_0} \prod_{m=1}^M \binom{\ell_0}{\ell_m} \quad (4.3-20)$$

$[v]$ events. Summation of (4.3-19) over the values of ℓ_0 given by (4.3-15) can be shown numerically to give (4.3-16). (See Appendix B.4.)

What is needed for evaluation of the ACJ total probability of error are the probabilities of the jamming events $(\ell_0, \underline{\ell}')$ and the number of $(\ell_0, \underline{\ell}')$ events represented by the ordered version. Since $(\ell_0, \underline{\ell}')$ is a subset of $\underline{\ell}$ for any $\underline{\ell}$, and permutations of the nonsignal channel elements of $\underline{\ell}$ do not affect ℓ_0 , it is reasonable that (4.3-17) gives the required number. This fact is confirmed

by the consideration that

$$\Pr\{\underline{\ell}'; M, L\} = \sum_{\ell_0} \Pr\{\underline{\ell}', \ell_0; M, L\} \quad (4.3-21)$$

which can be substituted in (4.3-18) to show that $\Pr\{\ell_0, \underline{\ell}\}$ must be multiplied by (4.3-17).

The probability of the event $(\ell_0, \underline{\ell})$ is derived in the following manner:

$$\begin{aligned} \Pr\{\underline{\ell}, \ell_0\} &= \binom{L}{\ell_0} \Pr\{\underline{v}_1 + \underline{v}_2 + \dots + \underline{v}_{\ell_0} = \underline{\ell}, \underline{v}_1 \neq \underline{0}, \underline{v}_2 \neq \underline{0}, \dots, \\ &\quad \underline{v}_{\ell_0} \neq \underline{0}, \underline{v}_{\ell_0+1} = \underline{0}, \dots, \underline{v}_L = \underline{0}\} \\ &= \binom{L}{\ell_0} \pi_0^{L-\ell_0} \Pr\{\underline{v}_1 + \underline{v}_2 + \dots + \underline{v}_{\ell_0} = \underline{\ell}, \underline{v}_1 \neq \underline{0}, \dots, \underline{v}_{\ell_0} \neq \underline{0}\} \end{aligned} \quad (4.3-22)$$

The probability required in (4.3-22) can be computed using the convolutional method described in Section 2.2.5, modified to give

$$\begin{aligned} &\Pr\{\underline{v}_1 + \underline{v}_2 + \dots + \underline{v}_{\ell_0} = \underline{\ell}, \underline{v}_1 \neq \underline{0}, \underline{v}_2 \neq \underline{0}, \dots, \underline{v}_{\ell_0} \neq \underline{0}\} \\ &= \sum_{\underline{v}_1 > \underline{0}} \sum_{\underline{v}_2 > \underline{0}} \dots \sum_{\substack{\underline{v}_{\ell_0} > \underline{0} \\ \underline{v}_{\ell_0} = \underline{\ell} - \sum_{r=1}^{\ell_0} \underline{v}_r}} \Pr\{\underline{v}_1\} \Pr\{\underline{v}_2\} \dots \Pr\{\underline{v}_{\ell_0}\} \delta \left[\underline{\ell} - \sum_{r=1}^{\ell_0} \underline{v}_r \right]. \end{aligned} \quad (4.3-23)$$

This method is useful for M tending to be large; for M=2, it is simpler to recognize that (2.2-24), repeated here as

$$\begin{aligned} \Pr\{\underline{\ell}; 2, L\} &= \sum_{n=0}^L \binom{L}{n, L-\ell_2-n, L-\ell_1-n, \ell_1+\ell_2+n-L} \\ &\quad \times \pi_0^n \pi_1^{2L-\ell_1-\ell_2-2n} \pi_2^{\ell_1+\ell_2-L+n}, \end{aligned} \quad (4.3-24)$$

is the sum of $\Pr\{\underline{l}, l_0; 2, L\}$ over l_0 , with $n \equiv L - l_0$. Therefore,

$$\Pr\{\underline{l}, l_0; 2, L\} = \binom{L}{L-l_0, l_0-l_2, l_0-l_1, l_1+l_2-l_0} \pi_1^{2l_0-l_1-l_2} \pi_2^{l_1+l_2-l_0} \quad (4.3-25)$$

For $M=4$, we recognize the same principle in the equation given in Table 2.2-4 for $\Pr\{\underline{l}; 4, L\}$ to give the $\Pr\{l_0, \underline{l}; 4, L\}$ equation shown in Table 4.3-1.

TABLE 4.3-1 PROBABILITIES OF $(\ell_0, \underline{\ell})$ JAMMING EVENTS FOR M = 4

$$\Pr \{ \ell_0, \underline{\ell}; 4, L \} = \sum_{n_1=0}^L \sum_{n_2=0}^L \cdots \sum_{n_{15}=0}^L \frac{L!}{(L-\ell)! n_1! \cdots n_{15}!} \pi_0^{L-\ell} \pi_1^{n_1+n_2+n_4+n_8} \pi_2^{n_3+n_5+n_6+n_9+n_{10}+n_{12}} \pi_3^{n_7+n_{11}+n_{13}+n_{14}} \pi_4^{n_{15}} \pi_5^{n_4}$$

CONSTRAINTS

CONSTRAINTS: $n_1 + n_3 + n_5 + n_7 + n_9 + n_{11} + n_{13} + n_{15} = \ell_1$
 $n_2 + n_3 + n_6 + n_7 + n_{10} + n_{11} + n_{14} + n_{15} = \ell_2$
 $n_4 + n_5 + n_6 + n_7 + n_{12} + n_{13} + n_{14} + n_{15} = \ell_3$
 $n_8 + n_9 + n_{10} + n_{11} + n_{12} + n_{13} + n_{14} + n_{15} = \ell_4$

$$\sum_{i=1}^{15} n_i = \ell_0$$

4.4 NUMERICAL RESULTS

In this section we present numerical results for the performance of the individual-channel AGC receiver and the any channel-jammed (ACJ) AGC receiver.

4.4.1 Numerical Results for Individual-Channel AGC Receiver

The numerical computations for the performance of the individual-channel AGC receiver were performed using (4.2-8) and (4.2-9). In this particular case, no unusual computational difficulties are encountered in the computations. A listing of the computer program is given in Appendix E.

Figures 4.4-1 through 4.4-4 show the performance of binary ($M=2$) RMFSK/FH with $L=1,2,3$, and 4 hops/bit, respectively with the jamming fraction $\gamma=q/N$ as a parameter. We observe that the choice of jamming fraction is critical to the effective operation of the jammer. This is similar to the behavior of the square-law combining receiver. However, unlike the square-law combining receiver, the optimum jamming fraction against the individual-channel AGC receiver is $\gamma=1.0$ over a wider range of E_b/N_j , especially for higher values of L , the number of hops/bit.

Figure 4.4-5 compares the worst-case jamming performance of binary RMFSK/FH as L varies. We observe that over the range of about $E_b/N_j=8$ dB to $E_b/N_j=39$ dB, the optimum choice for the communicator is $L=2$ or 3 hops/bit. However, outside this range $L=1$ is optimum. In no case does increasing L beyond 3 hops/bit improve the performance.

Figures 4.4-6 through 4.4-9 show the performance of RMFSK/FH when $M=4$ and $L=1,2,3$, and 4, respectively. Again, the importance to the jammer of

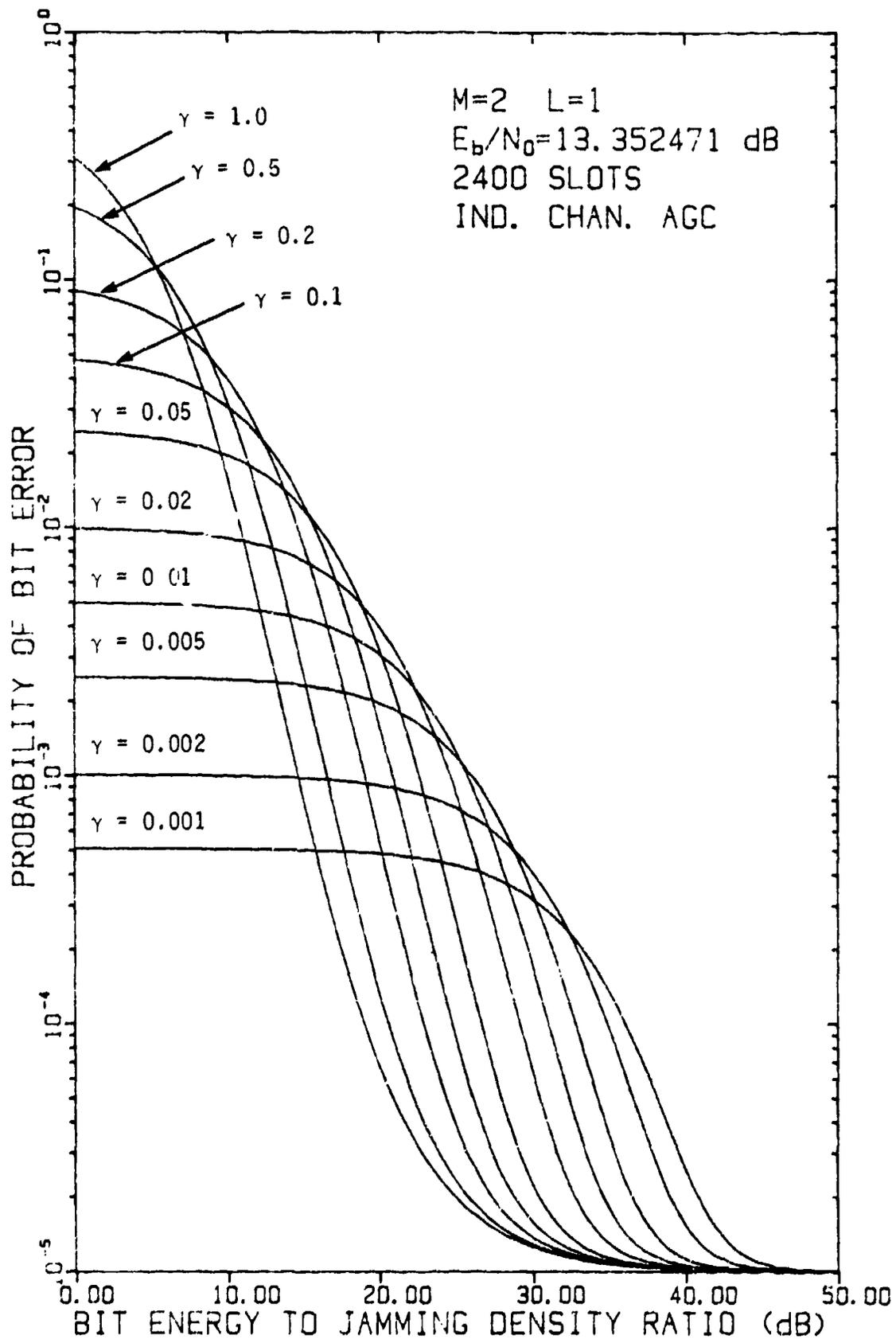


FIGURE 4.4-1 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 2$ AND $L = 1$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

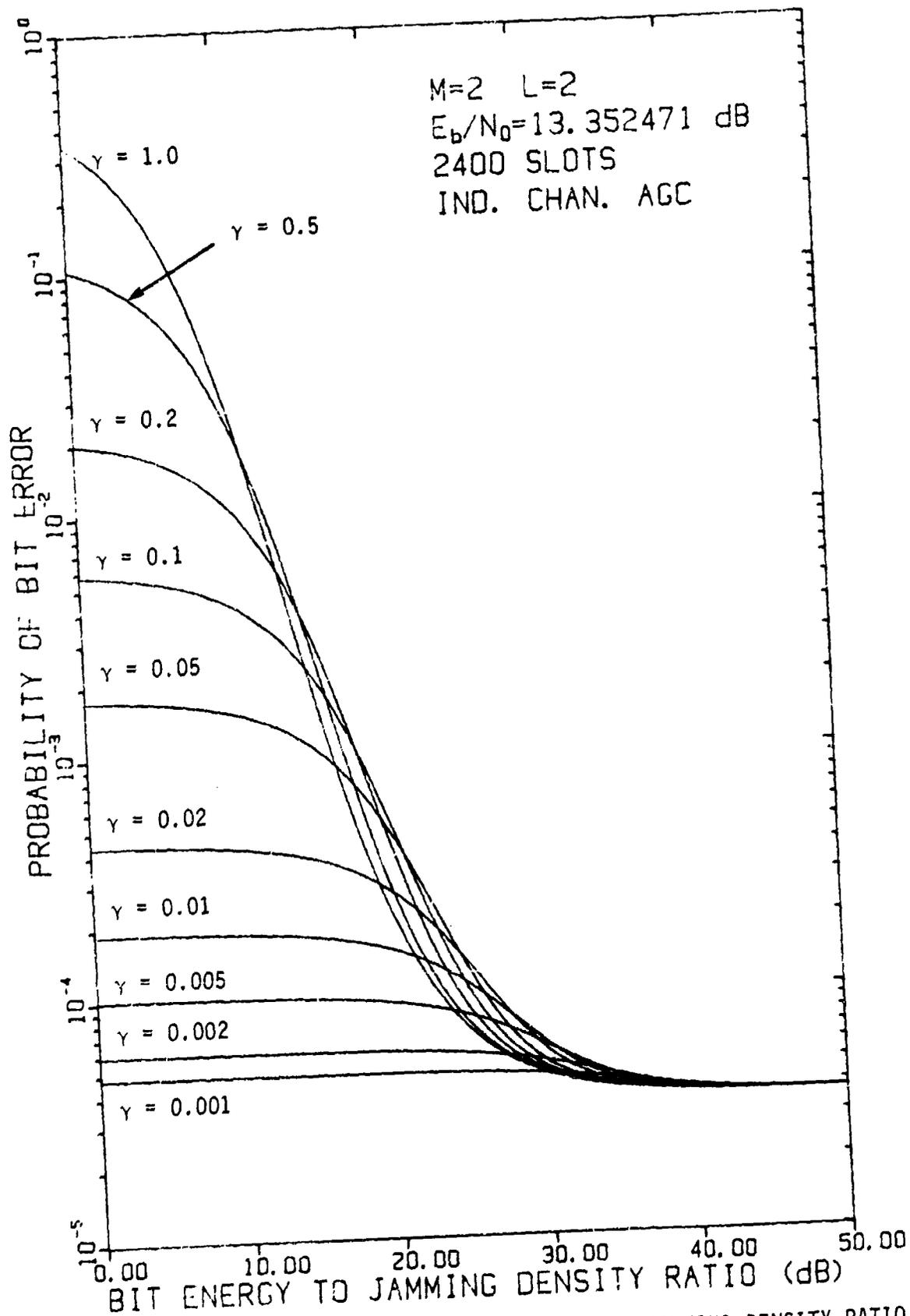


FIGURE 4.4-2 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 2$ AND $L = 2$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471 \text{ dB}$ (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

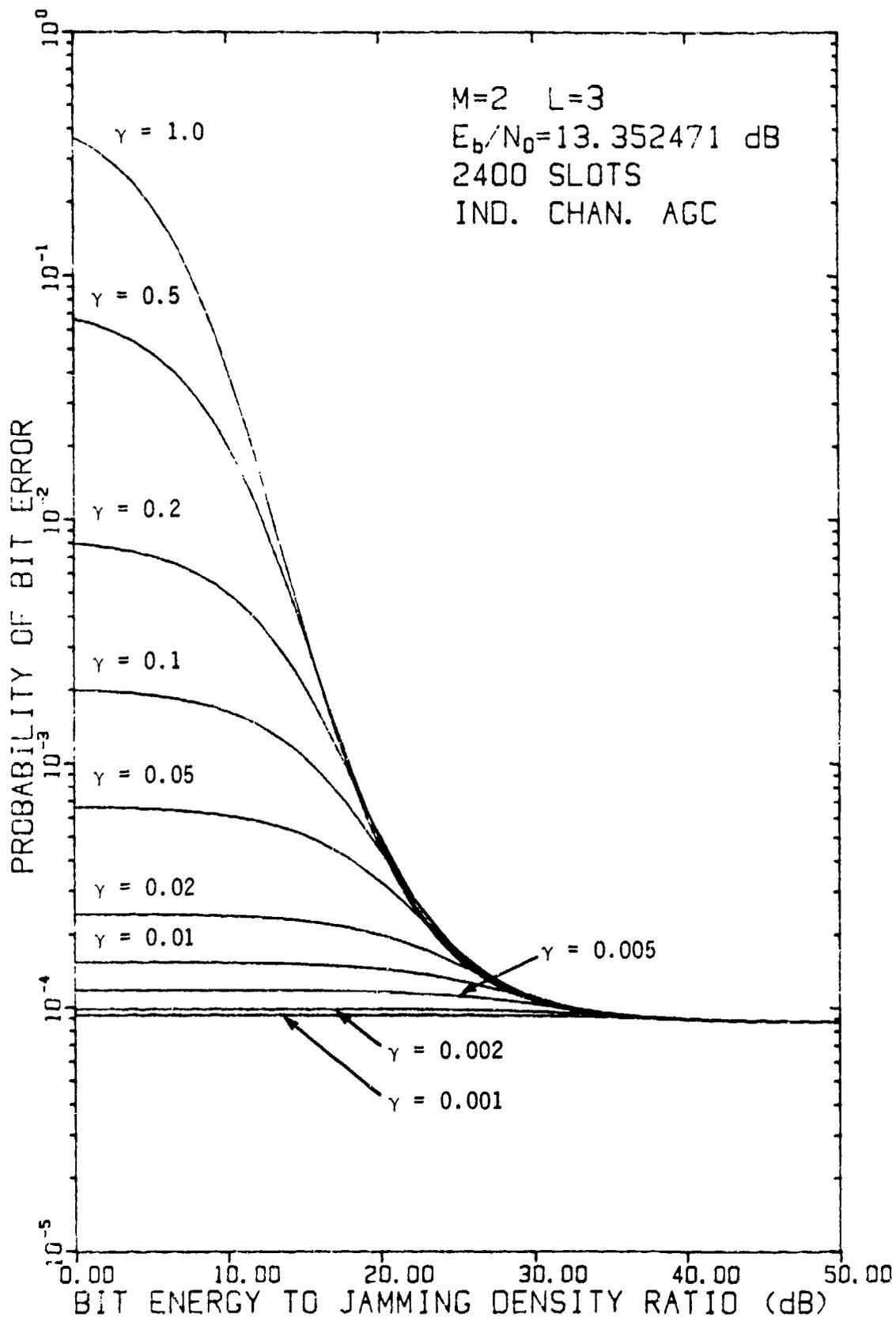


FIGURE 4.4-3 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 2$ AND $L = 3$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

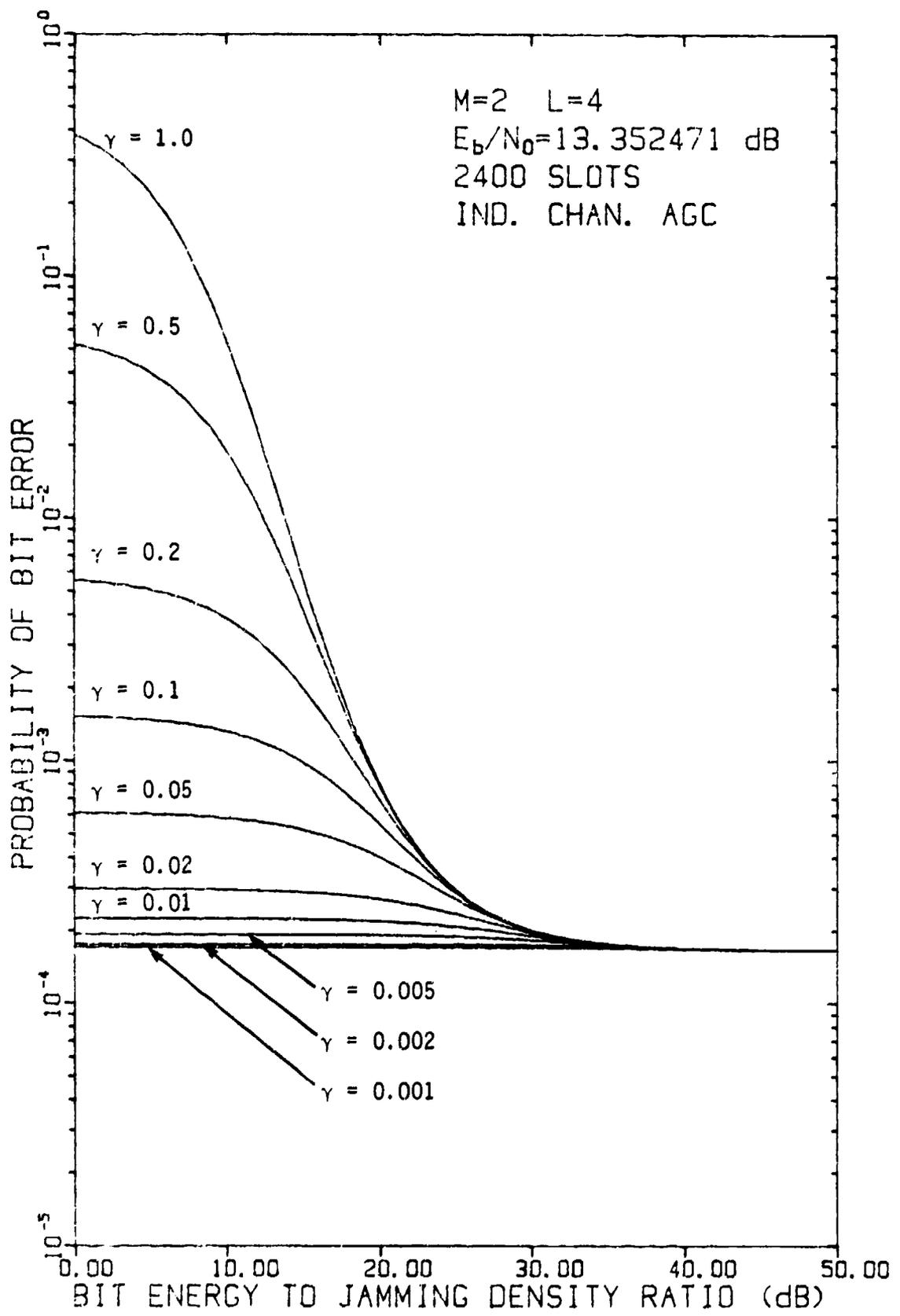


FIGURE 4.4-4 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 2$ AND $L = 4$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471 \text{ dB}$ (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

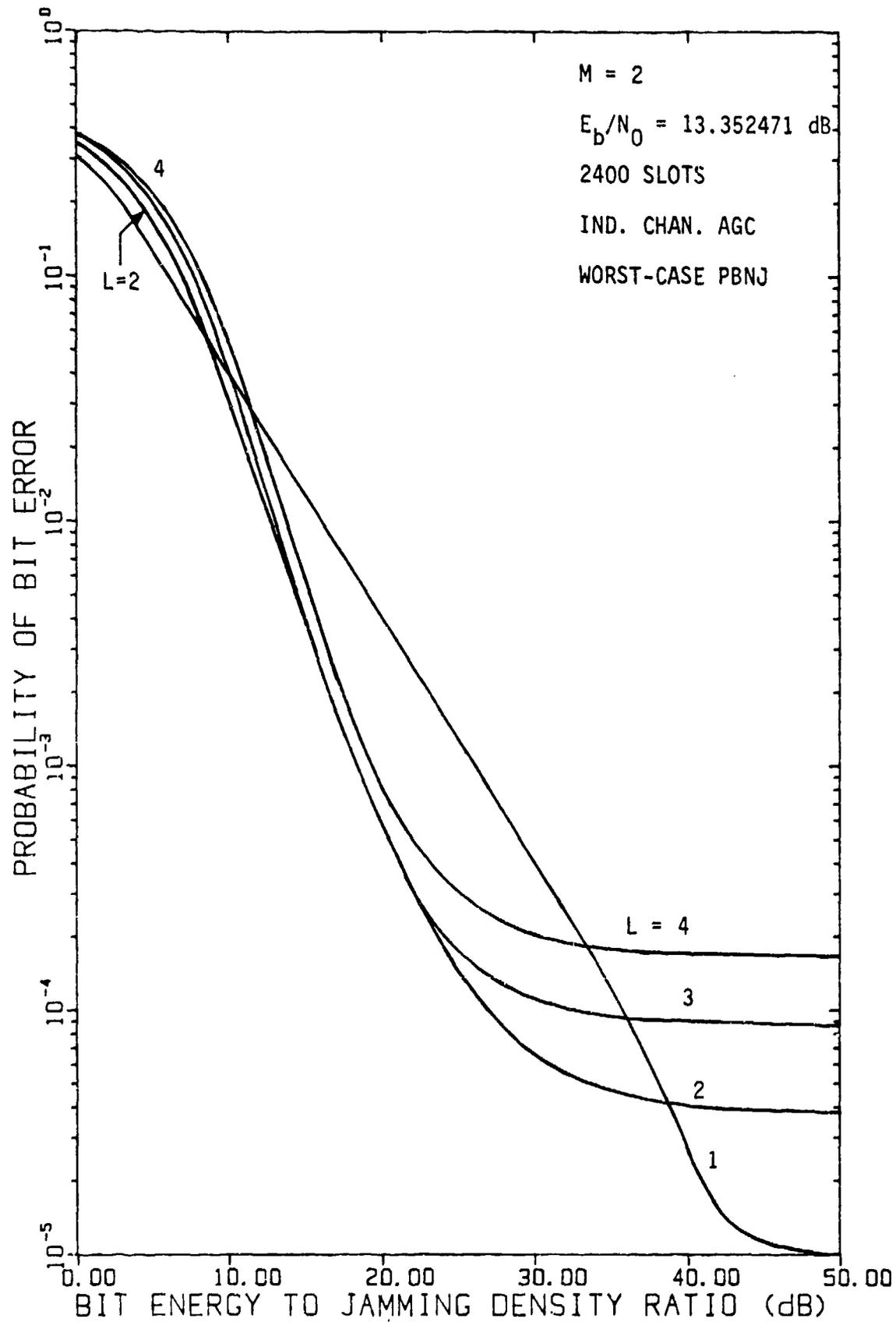


FIGURE 4.4-5 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER AND $M = 2$ WITH NUMBER OF HOPS/SYMBOL L AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING

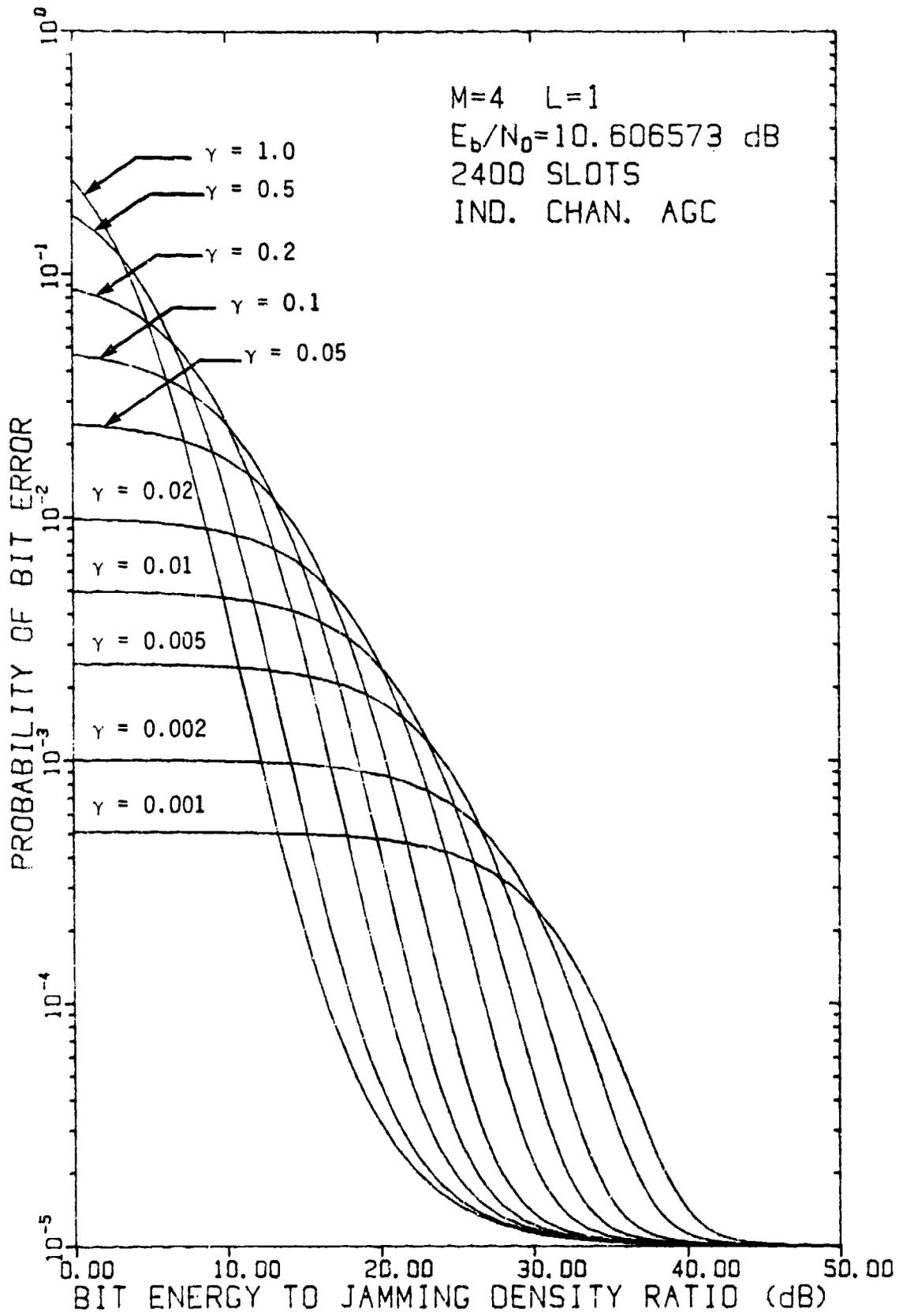


FIGURE 4.4-6 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 4$ AND $L = 2$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

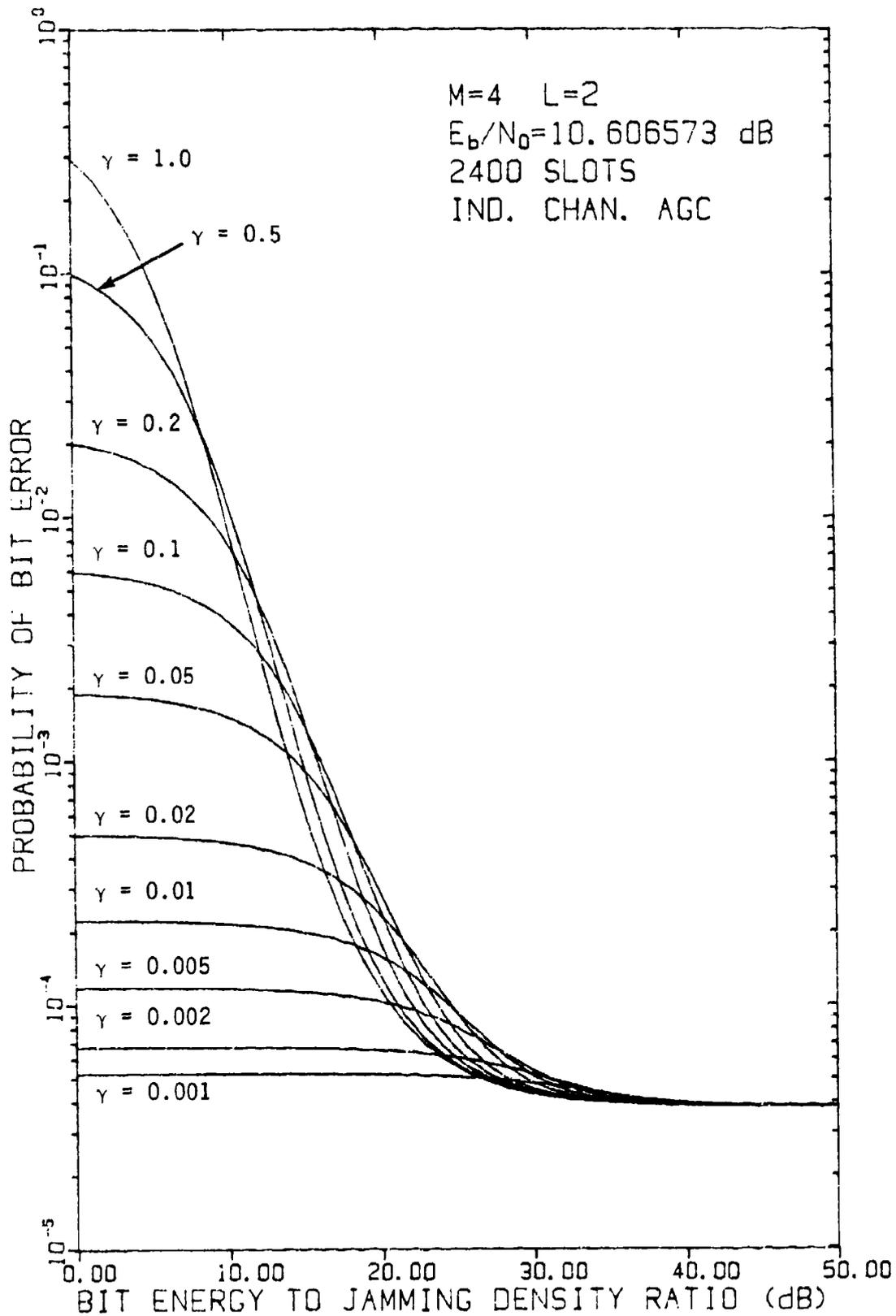


FIGURE 4.4-7 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 4$ AND $L = 2$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

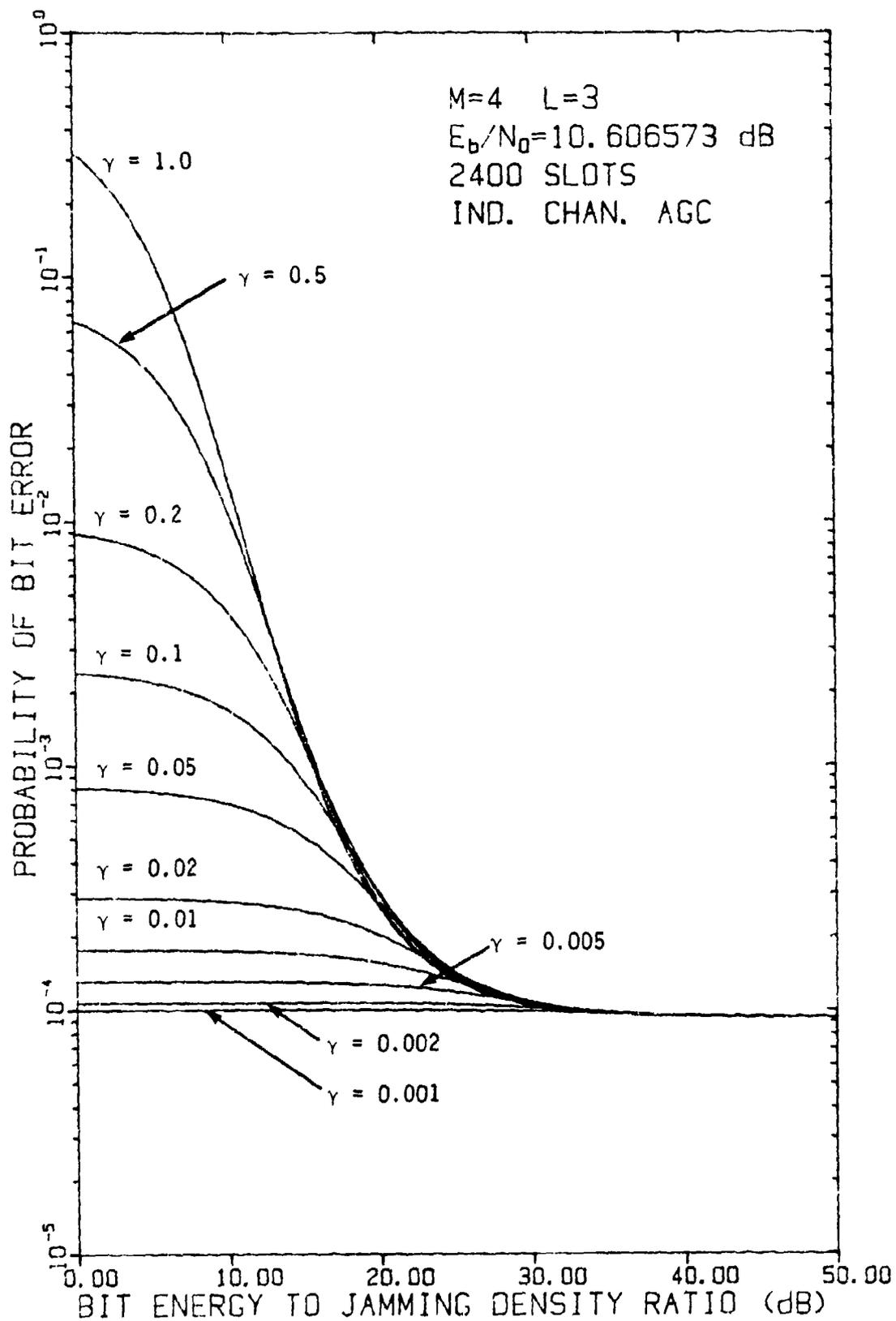


FIGURE 4.4-8 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 4$ AND $L = 3$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

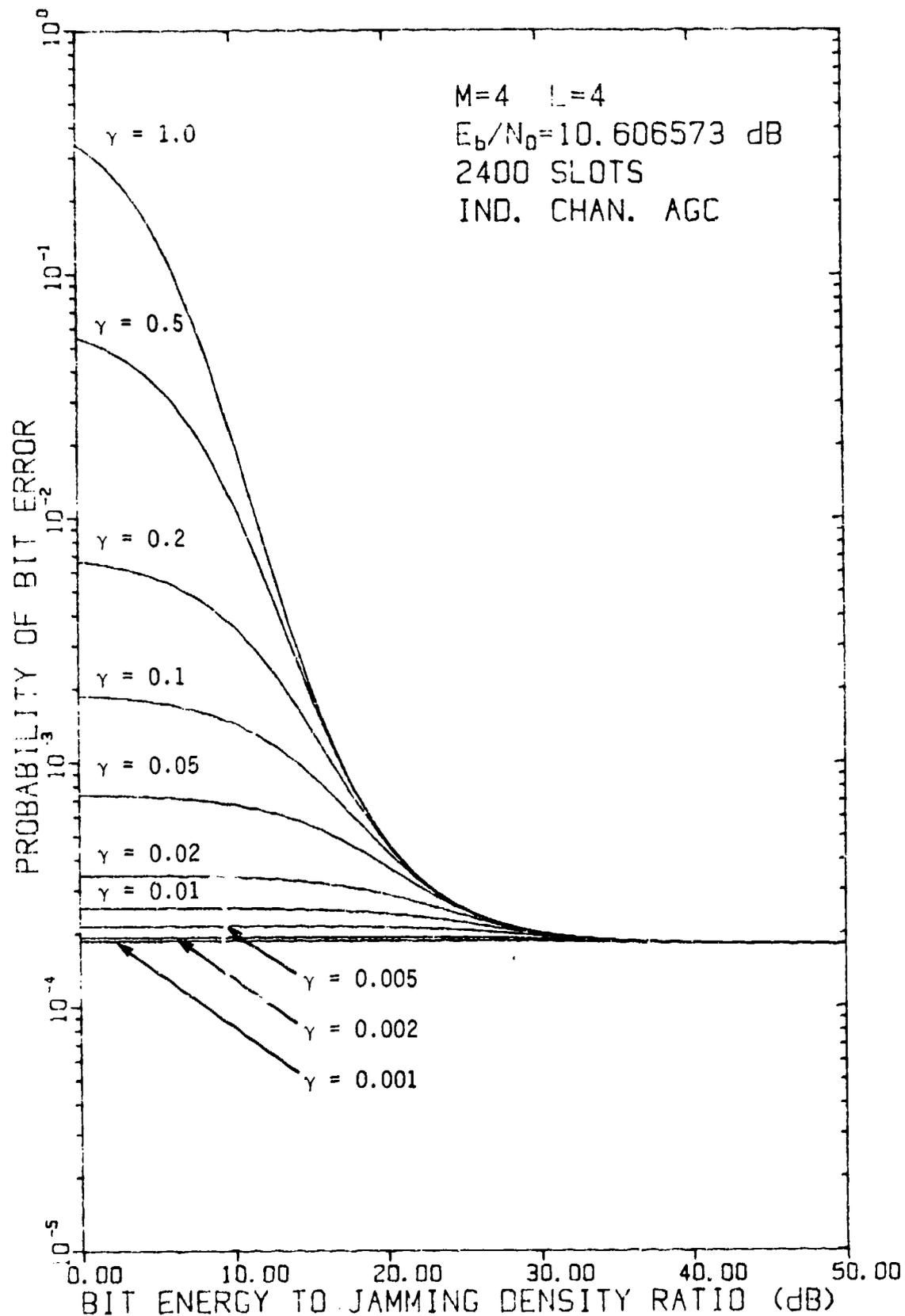


FIGURE 4.4-9 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 4$ AND $L = 4$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573 \text{ dB}$ (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

the correct selection of γ stands out clearly. Figure 4.4-10 compares the performance of 4-ary RMFSK/FH in worst-case partial-band noise jamming as L varies. We see that for E_b/N_J in the range of about 7 to 36 dB, $L=2$ or 3 is optimum; elsewhere, $L=1$ is optimum.

Figures 4.4-11 through 4.4-14 show the curves for $M=8$ with γ as a parameter and $L=1,2,3$, and 4, respectively. Figure 4.4-15 shows performance for $M=8$ with L as a parameter in worst-case partial-band noise jamming. Again, from about $E_b/N_J=5$ dB to 35 dB the optimum L is 2 or 3, but elsewhere $L=1$ is optimum.

Two important conclusions can be drawn from these curves:

- The correct choice of fraction γ is critical for the jammer;
- The communicator can obtain only a small benefit by using multiple hops/symbol, and then only over a limited range of jamming conditions.

4.4.2 Numerical Results for Any-Channel-Jammed AGC (ACJ-AGC) Receiver

The numerical computations for the performance of the ACJ-AGC receiver required the use of two alternative forms. We used (4.3-12) for the computations, in conjunction with the computational techniques discussed in Section 3.3. The switch-over criteria for choosing between series and numerical integration were determined empirically. A listing of the computer program for numerical computations is given in Appendix F and a listing of the program which produced the plots is given in Appendix G. This latter program is typical of the plotting programs for all receivers; for sake of brevity only this one version of the plot program is included in our report.

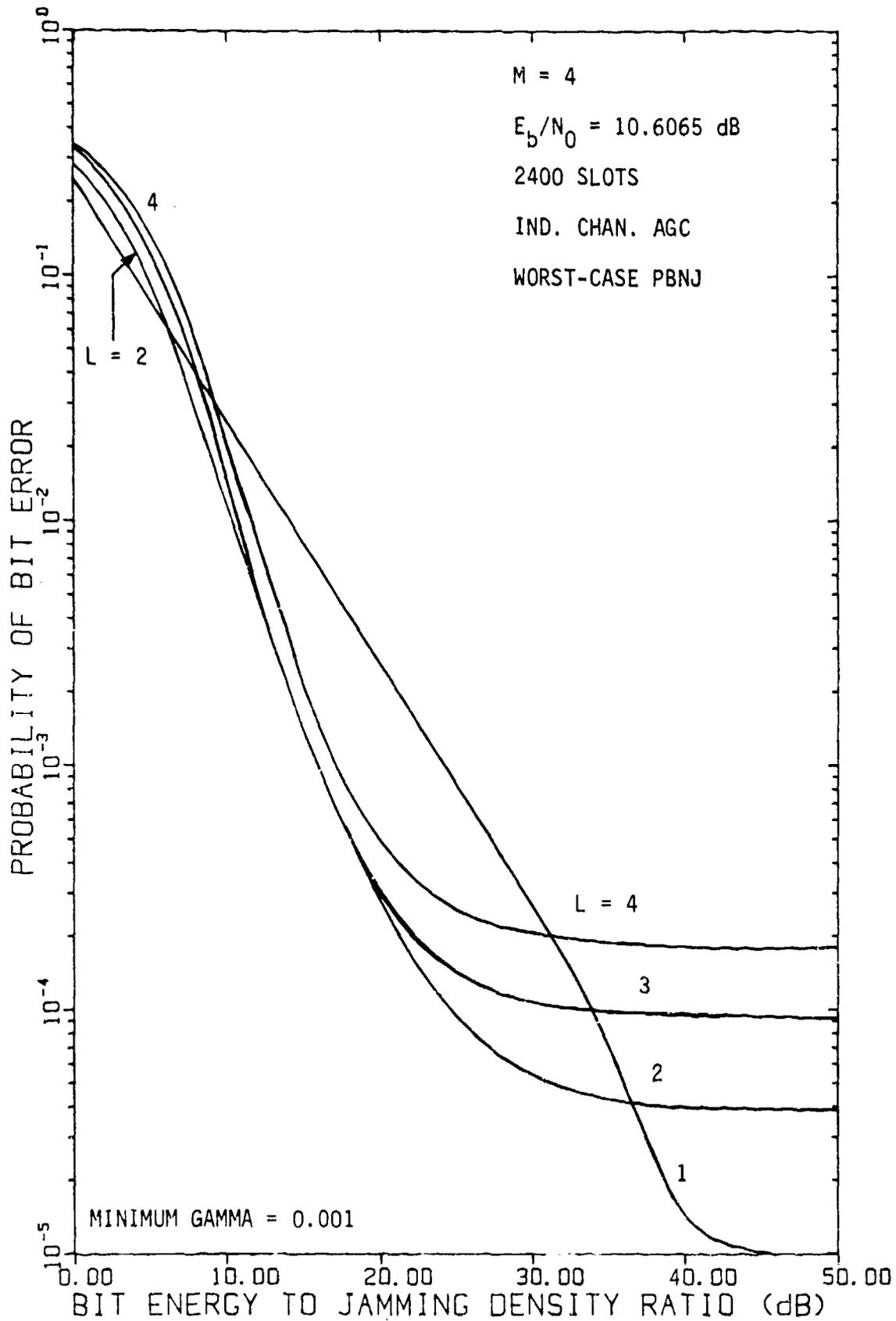


FIGURE 4.4-10 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER AND $M = 4$ WITH NUMBER OF HOPS/SYMBOL L AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING

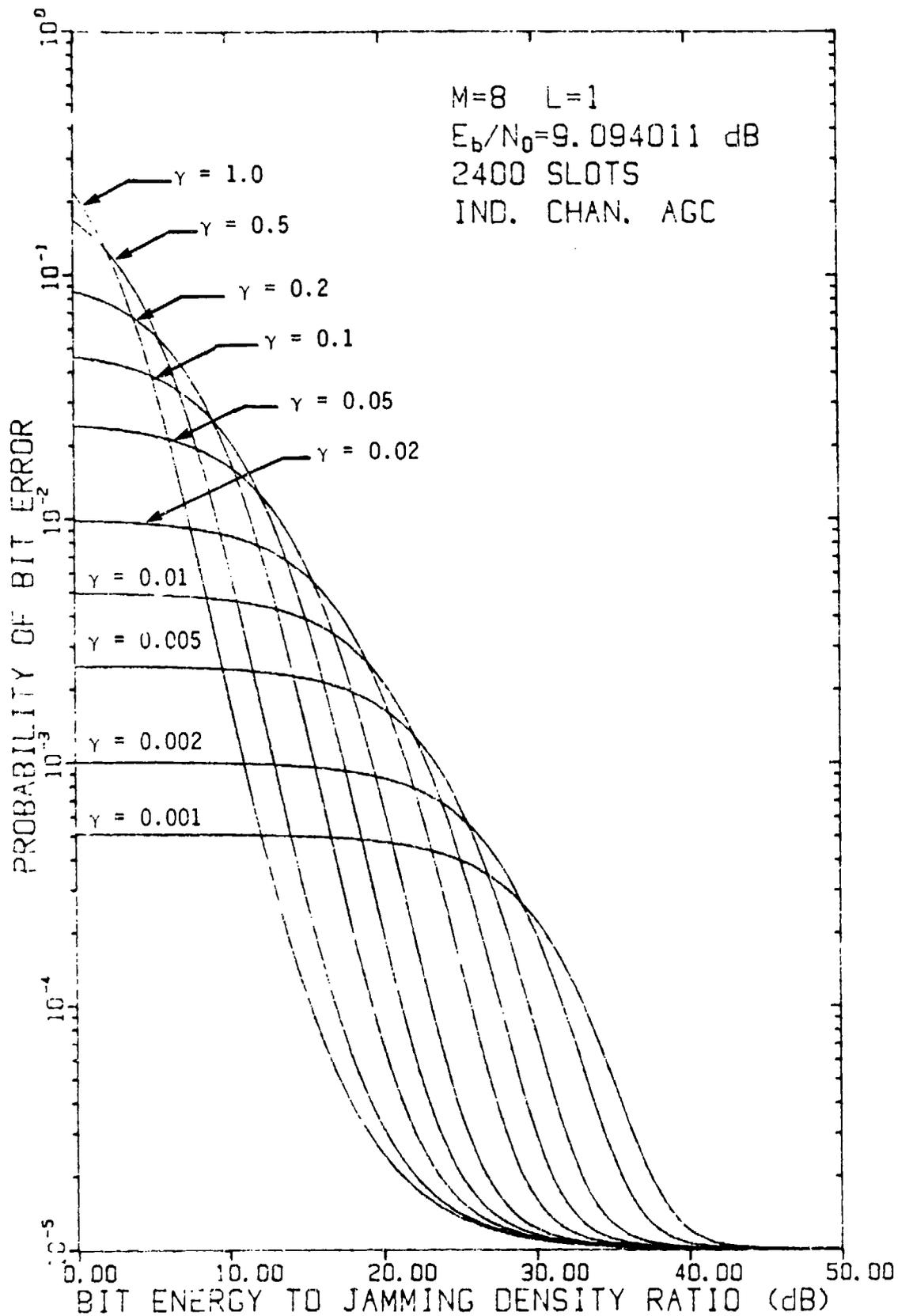


FIGURE 4.4-11 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 8$ AND $L = 1$ HOP/SYMBOL, 2400 HOPPING SLOTS, AND $E_t/N_0 = 9.094011$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

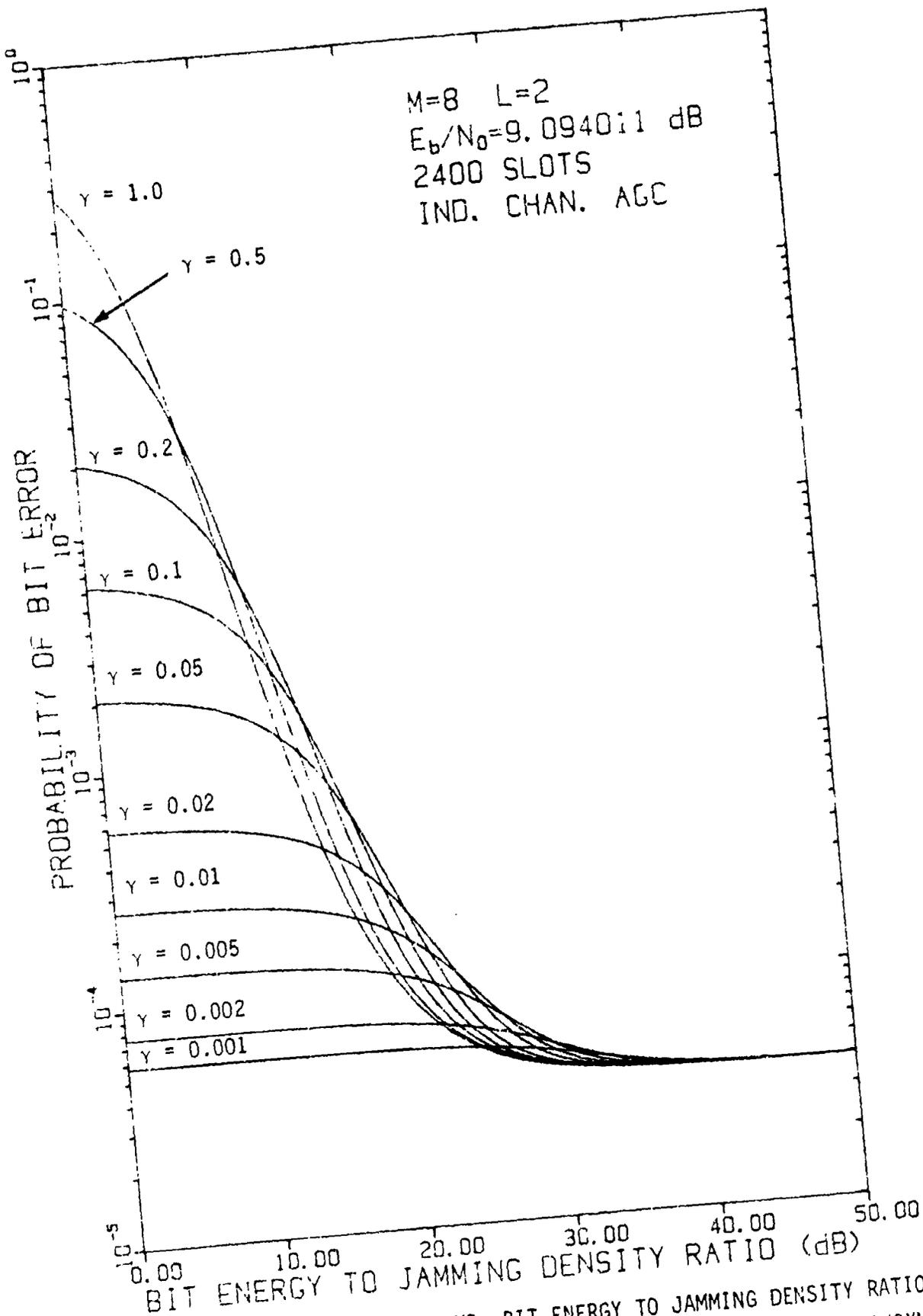


FIGURE 4.4-12 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 8$ AND $L = 2$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 9.094011 \text{ dB}$ (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

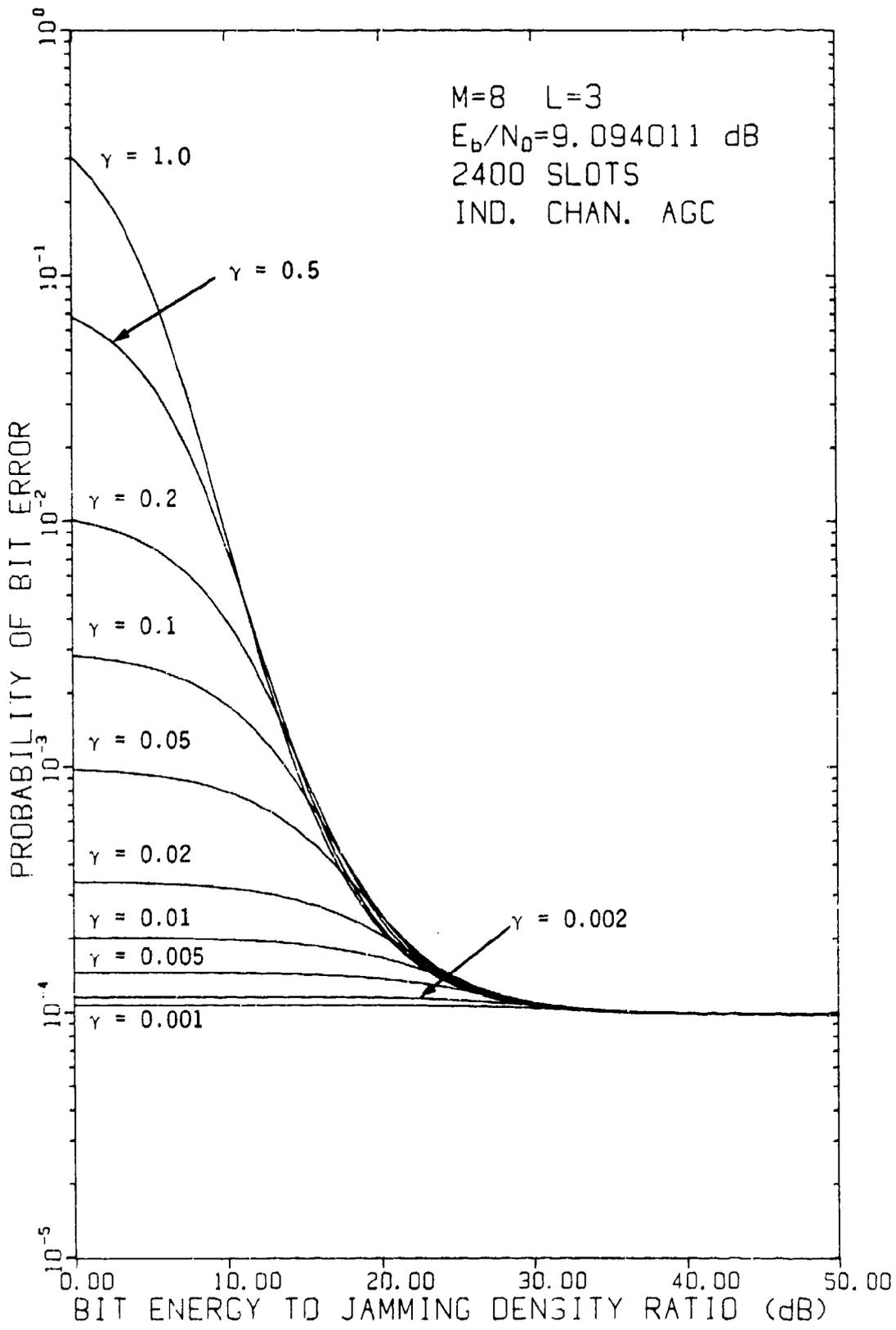


FIGURE 4.4-13 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 8$ AND $L = 3$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 9.094011$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

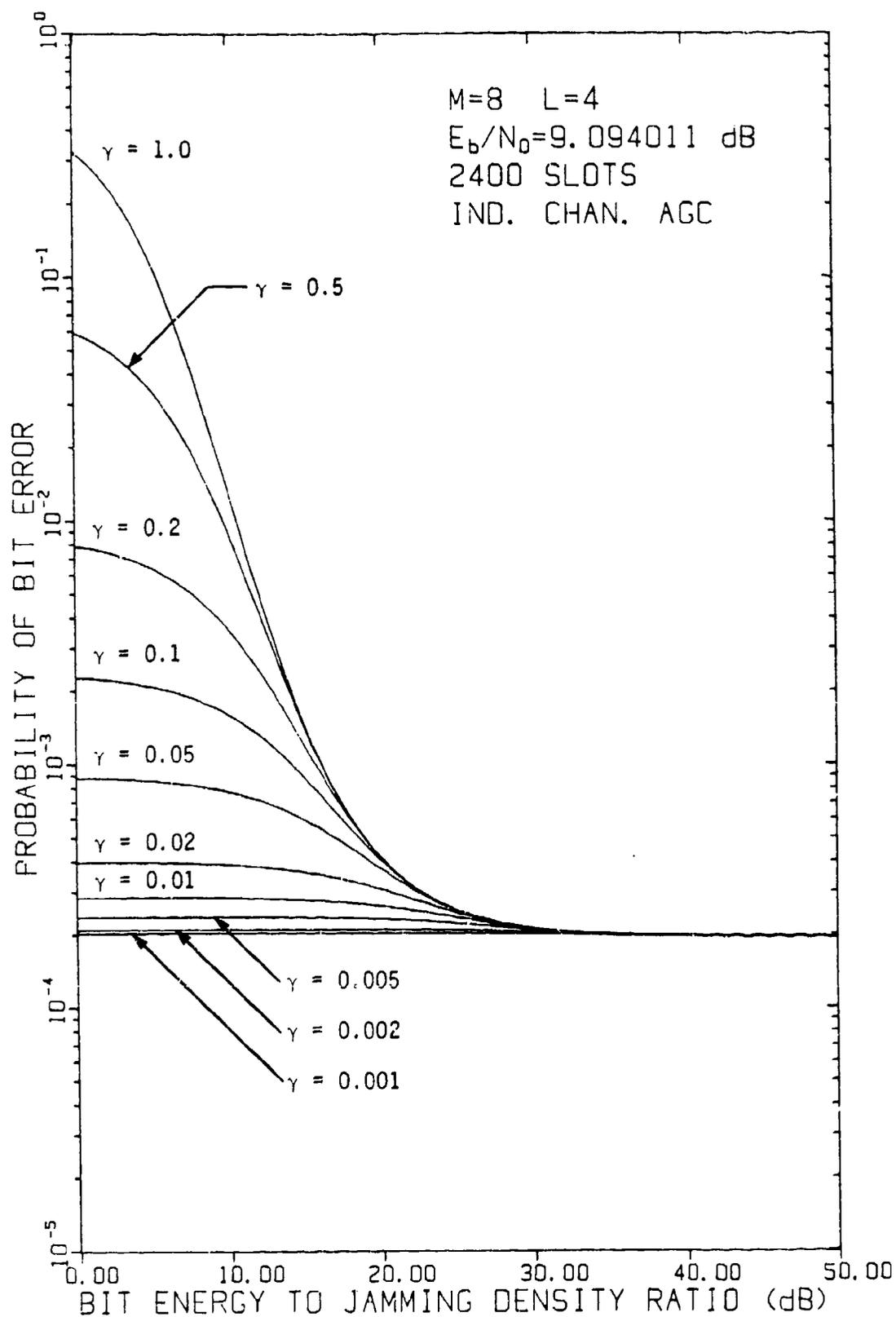


FIGURE 4.4-14 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER WITH $M = 8$ AND $L = 4$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 9.094011$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L = 1$) WITH JAMMING FRACTION γ AS A PARAMETER

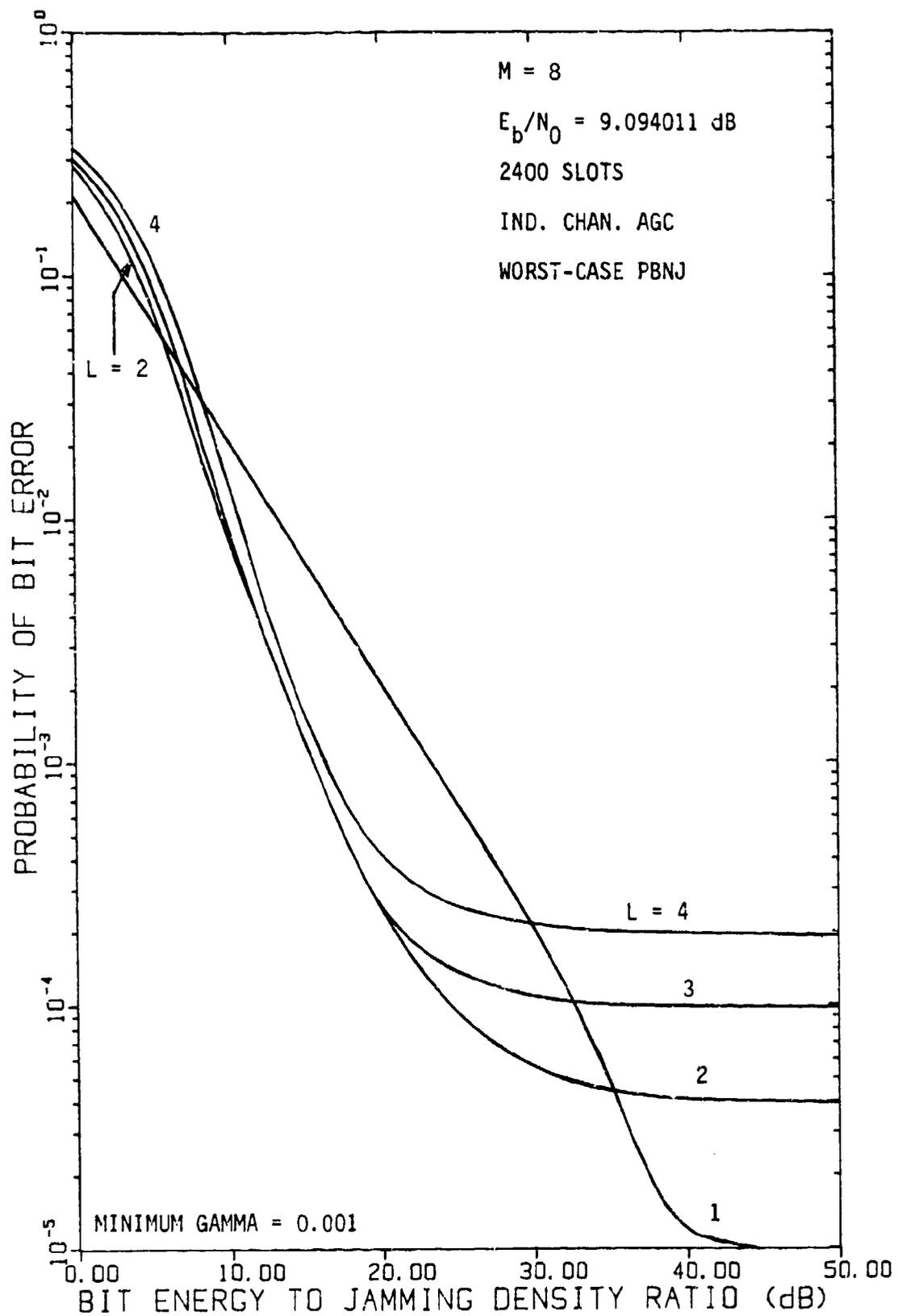


FIGURE 4.4-15 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER AND $M=8$ WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING

Figures 4.4-16 through 4.4-19 show the performance of the ACJ-AGC receiver with $M=2$ for $L=1,2,3$, and 4 hops/bit, respectively. Figure 4.4-20 summarizes the performance in worst-case partial-band noise jamming with L as a parameter. Again, we find a limited range, roughly $10 \text{ dB} \leq E_b/N_j \leq 35 \text{ dB}$, where the optimum diversity is $L=2$ or 3 hops per bit; elsewhere $L=1$ is optimum.

Figures 4.4-21 through 4.4-24 show the performance of the ACJ-AGC receiver with $M=4$ for $L=1,2,3$, and 4 hops/symbol, respectively. These curves show the same general behavior as those for the other receivers. Figure 4.4-25 shows the performance in worst-case partial-band noise jamming with L as a parameter. Again, we find a limited range where $L=2$ or 3 is optimum, but elsewhere $L=1$ is optimum.

Finally, Figures 4.4-26 and 4.4-27 show the performance for $M=8$ and $L=1$ and 2, respectively. Because of the large computer time required, $L>2$ was not considered for $M=8$. Figure 4.4-28 summarizes performance in worst-case partial-band noise jamming. Again, there is a region where diversity ($L=2$) offers some advantage.

In summary, a small amount of diversity ($L=2$ or 3) is of meaningful benefit to RMFSK/FH over a limited range of signal-to-jamming ratios. However, outside this range no diversity ($L=1$) gives better performance. In some cases, e.g. Figure 4.4-20 for $M=2$, the penalty for using $L=3$ in the absence of jamming is nearly the same as the benefit of using $L=3$ when $E_b/N_j=25 \text{ dB}$.

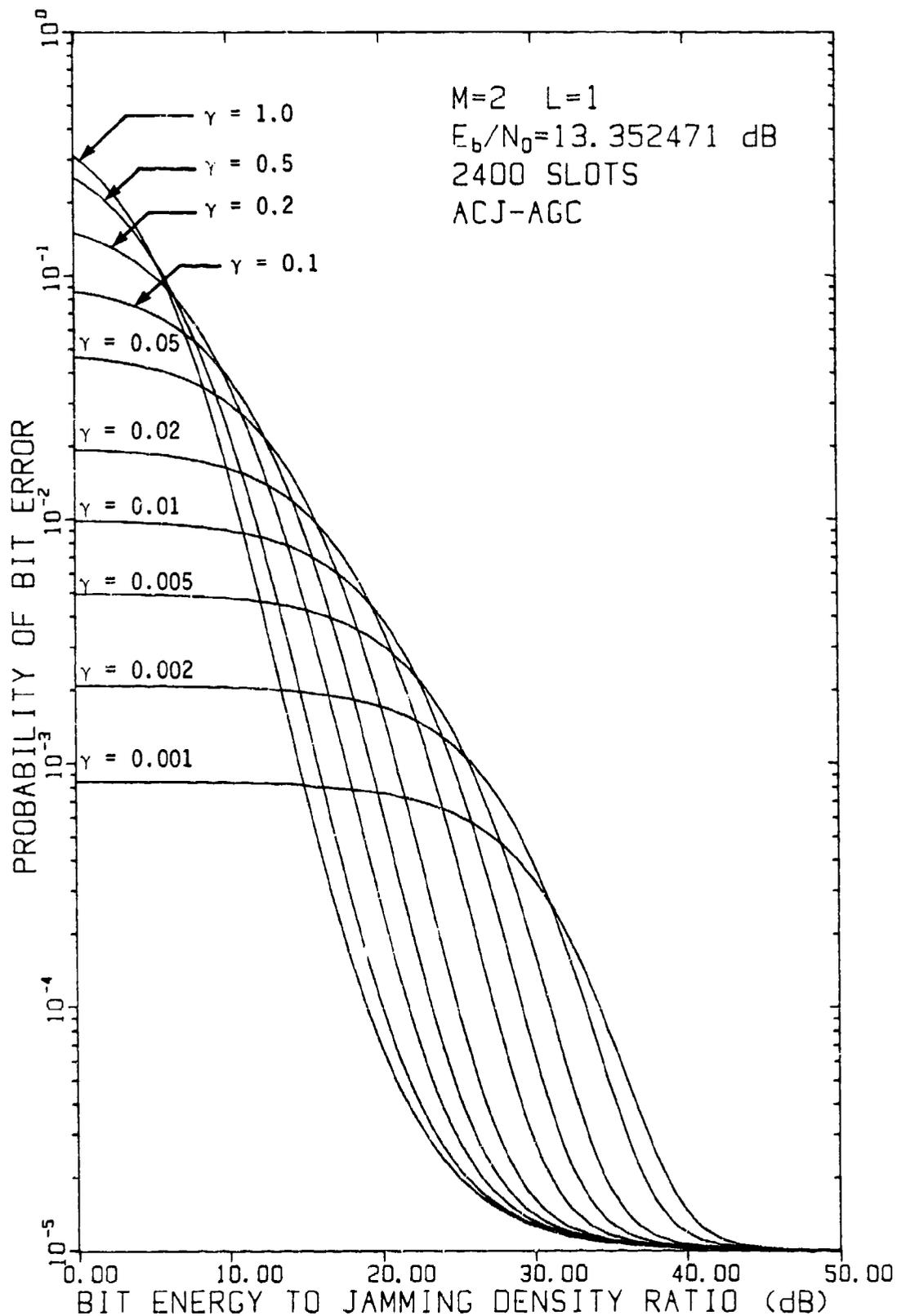


FIGURE 4.4-16 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH $M=2$ AND $L=1$ HOP/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471 \text{ dB}$ (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ AS A PARAMETER

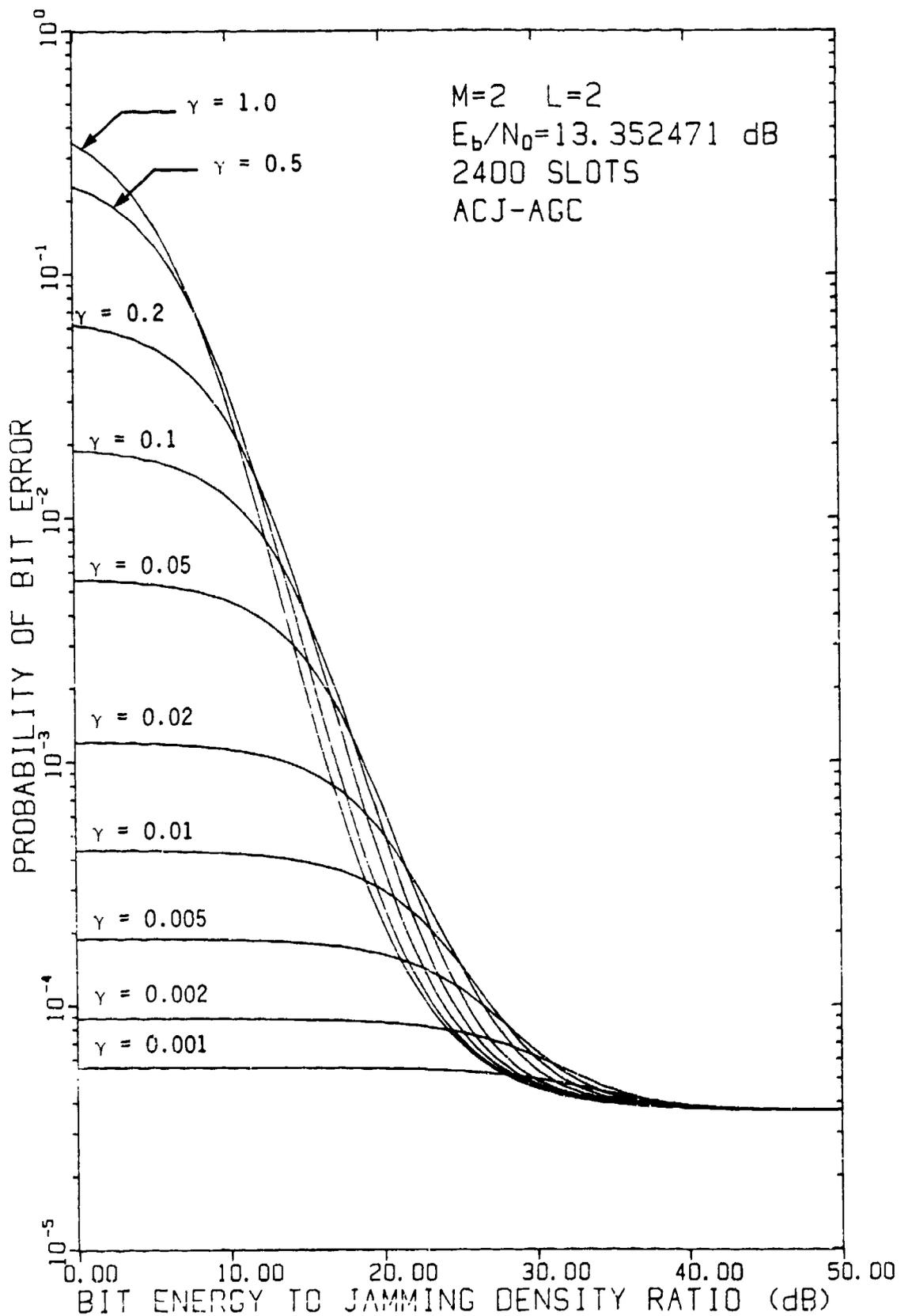


FIGURE 4.4-17 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH $M=2$ AND $L=2$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471 \text{ dB}$ (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ AS A PARAMETER

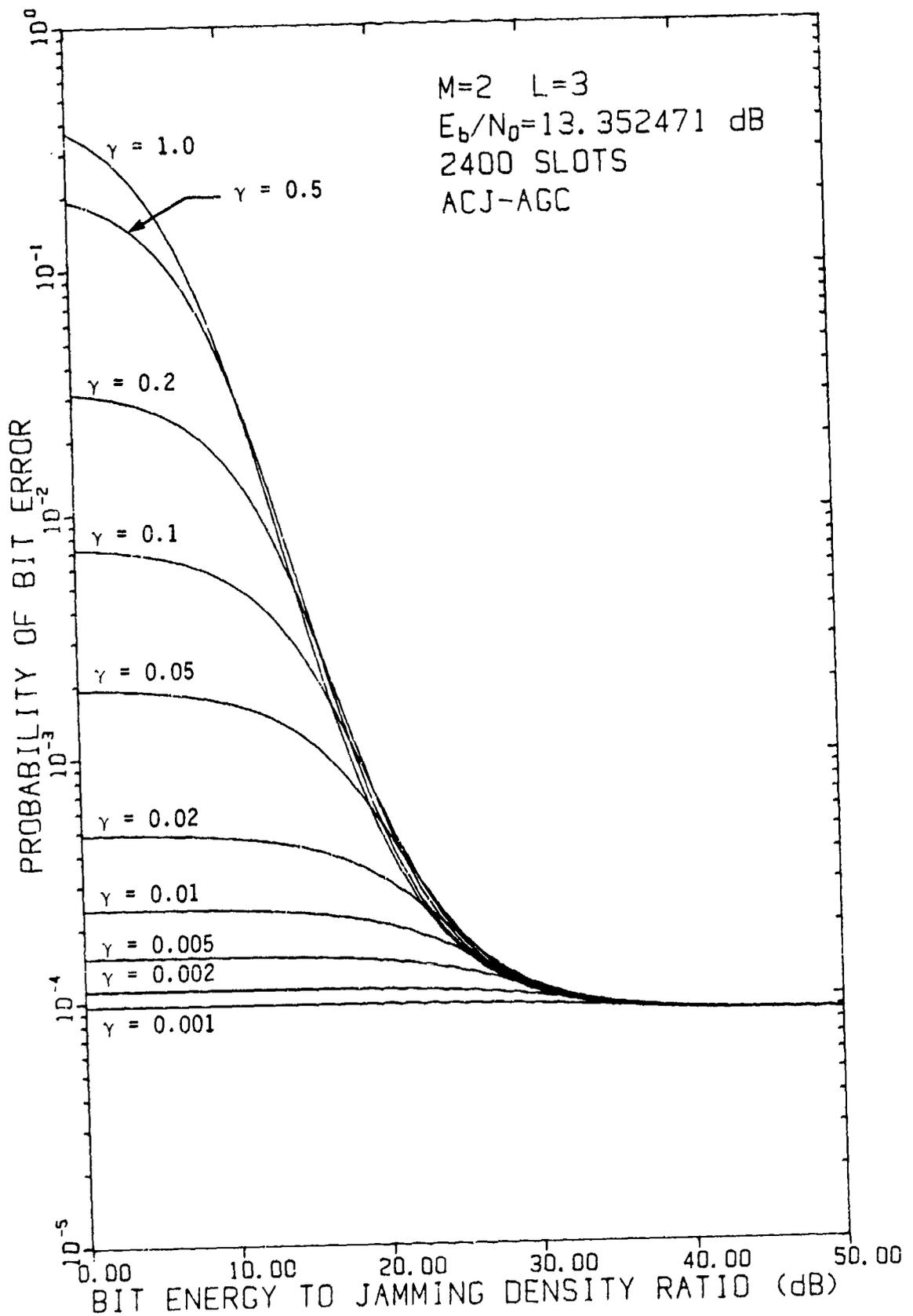


FIGURE 4.4-18 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL JAMMED AGC RECEIVER WITH $M=2$ AND $L=3$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ AS A PARAMETER

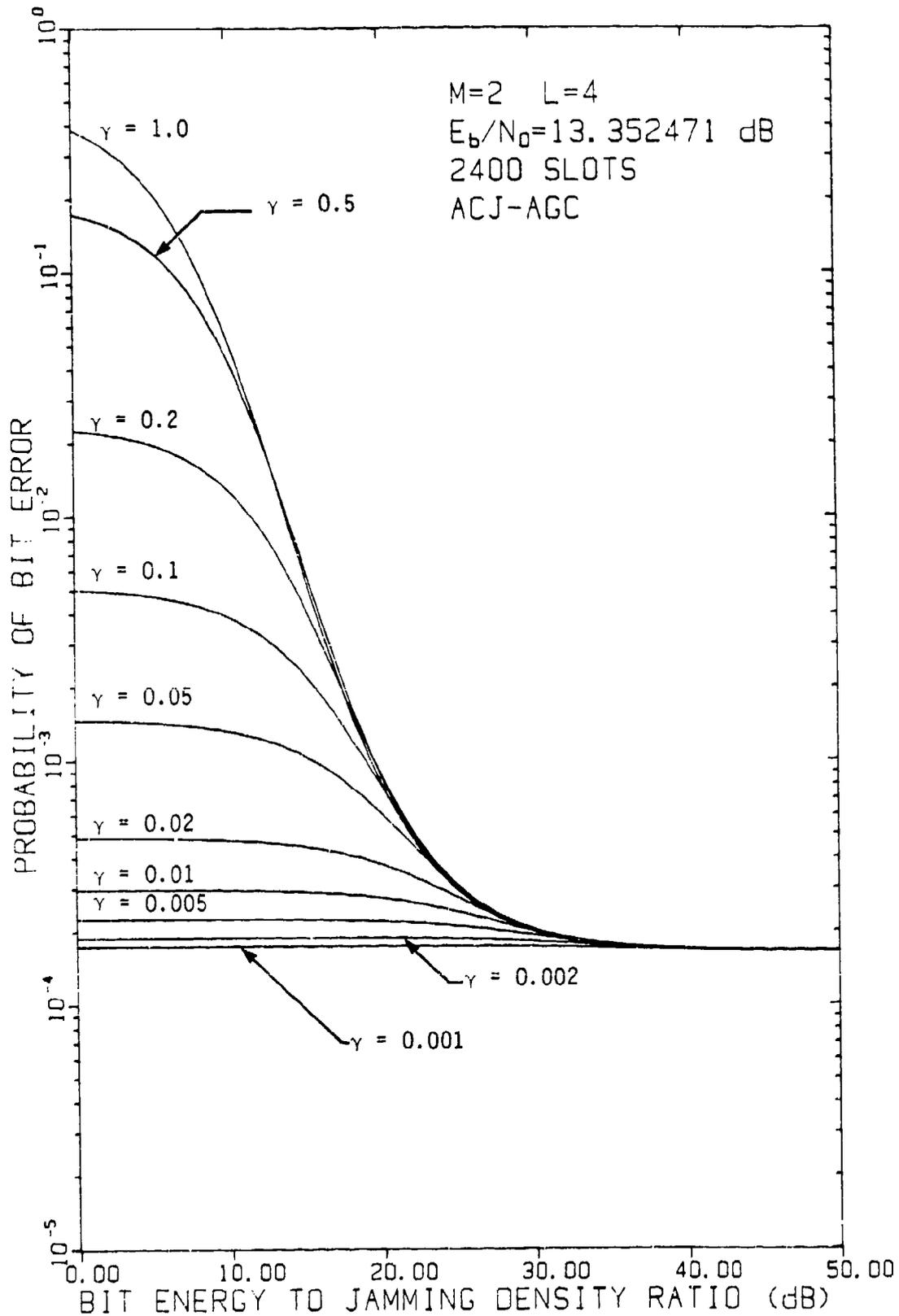


FIGURE 4.4-19 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH $M=2$ AND $L=4$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.352471$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ AS A PARAMETER

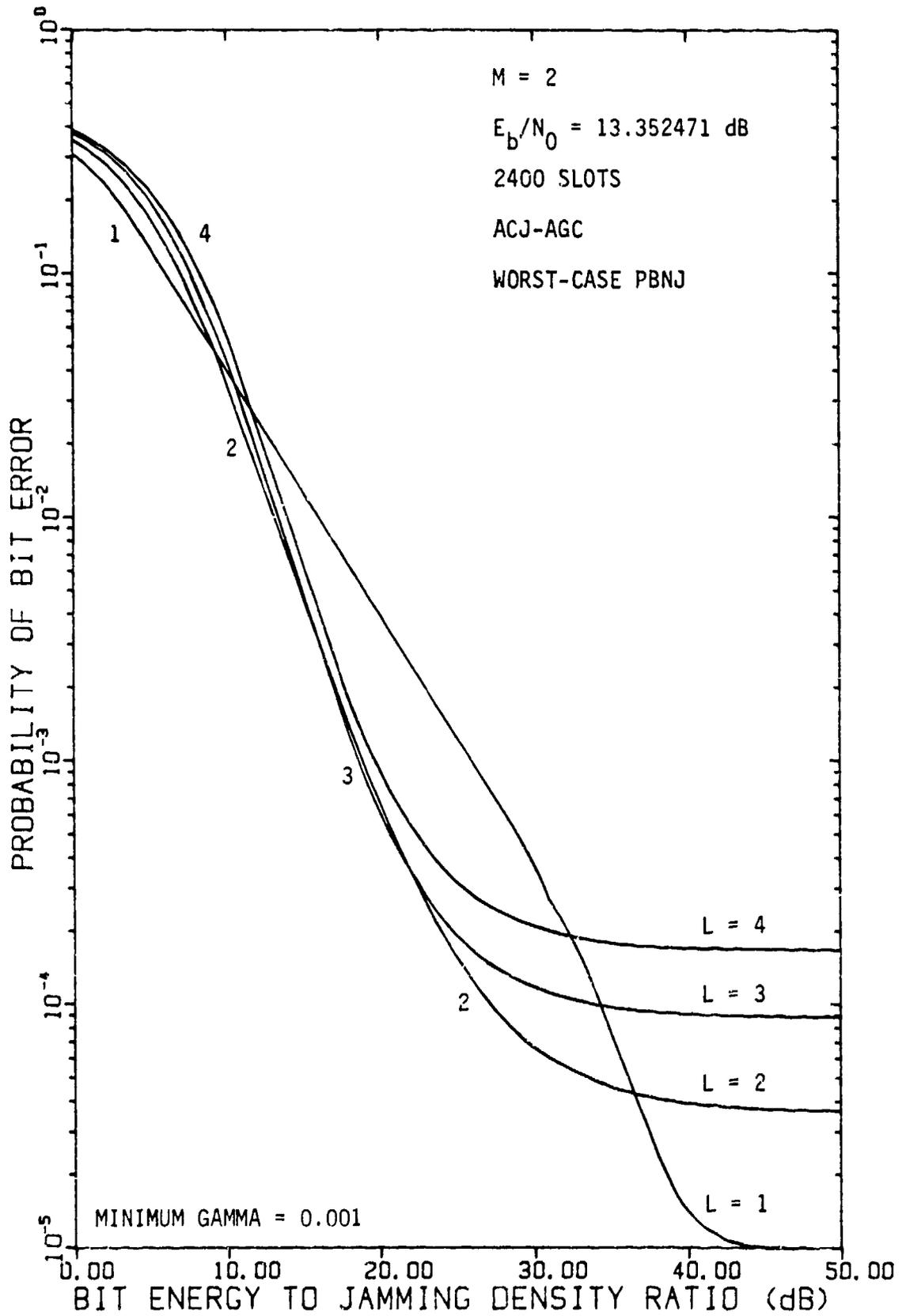


FIGURE 4.4-20 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER AND $M=2$ WITH NUMBER OF HOPS/BIT AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING

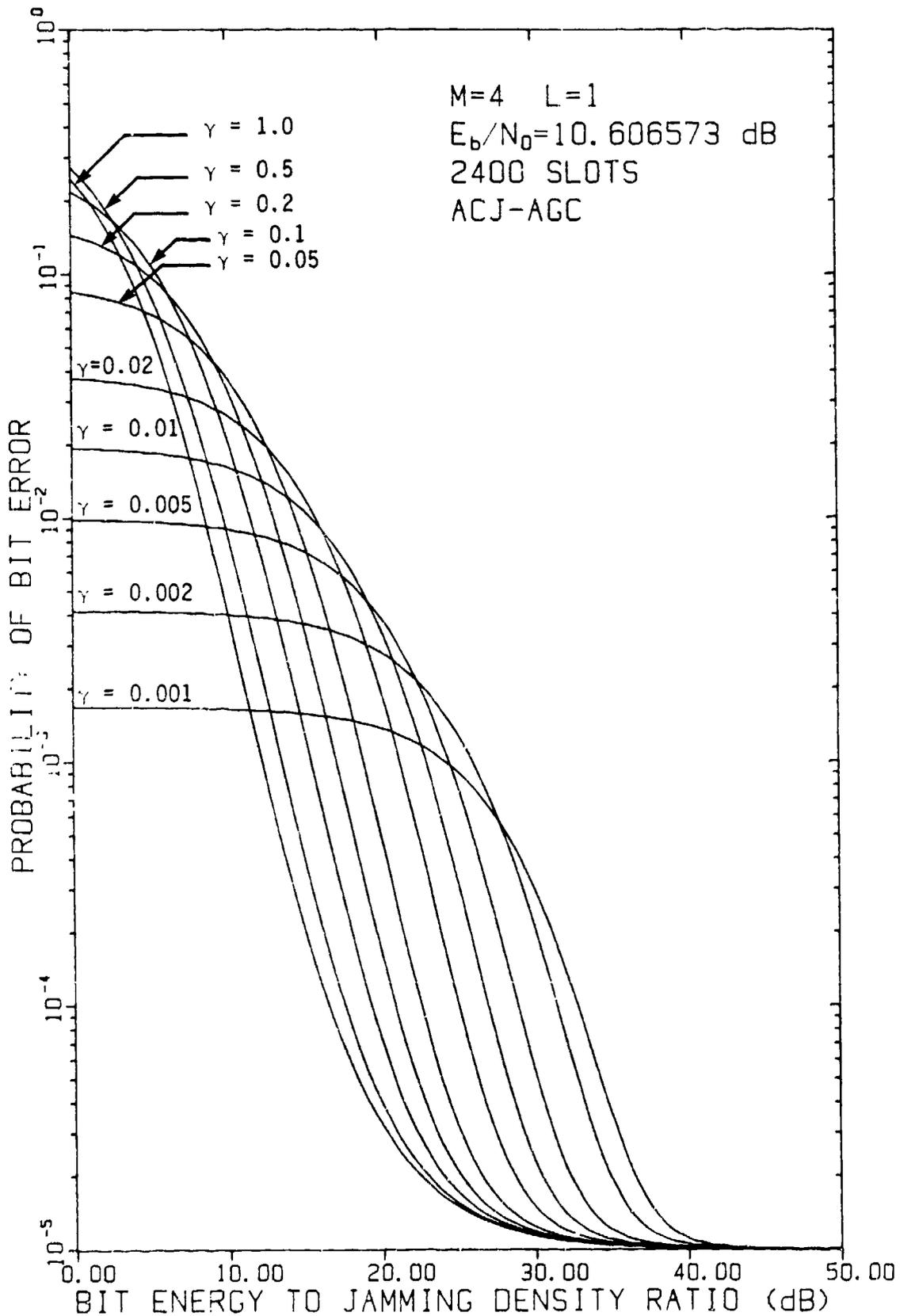


FIGURE 4.4-21 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH $M=4$ AND $L=1$ HOP/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ AS A PARAMETER

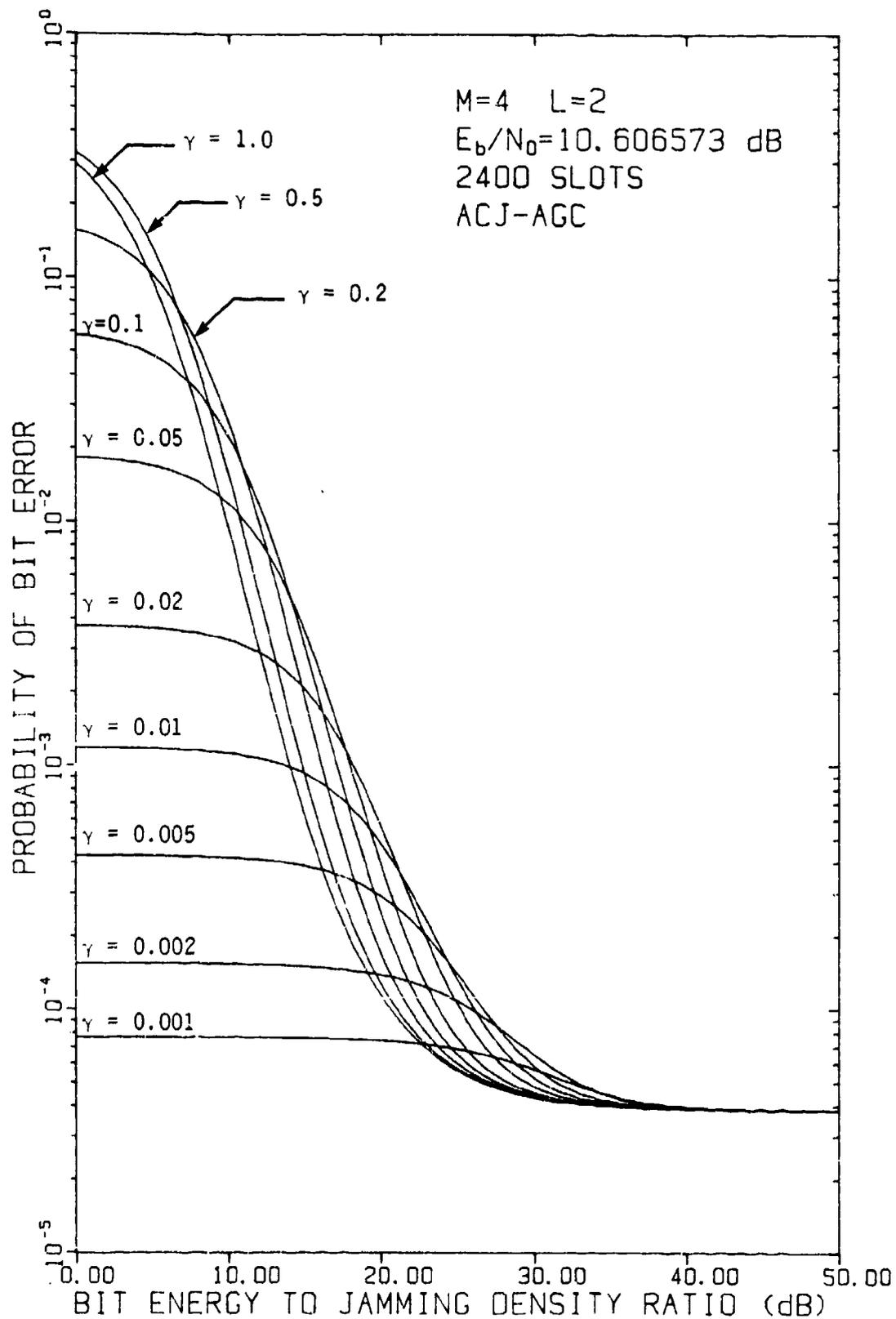


FIGURE 4.4-22 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH $M=4$ AND $L=2$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ AS A PARAMETER

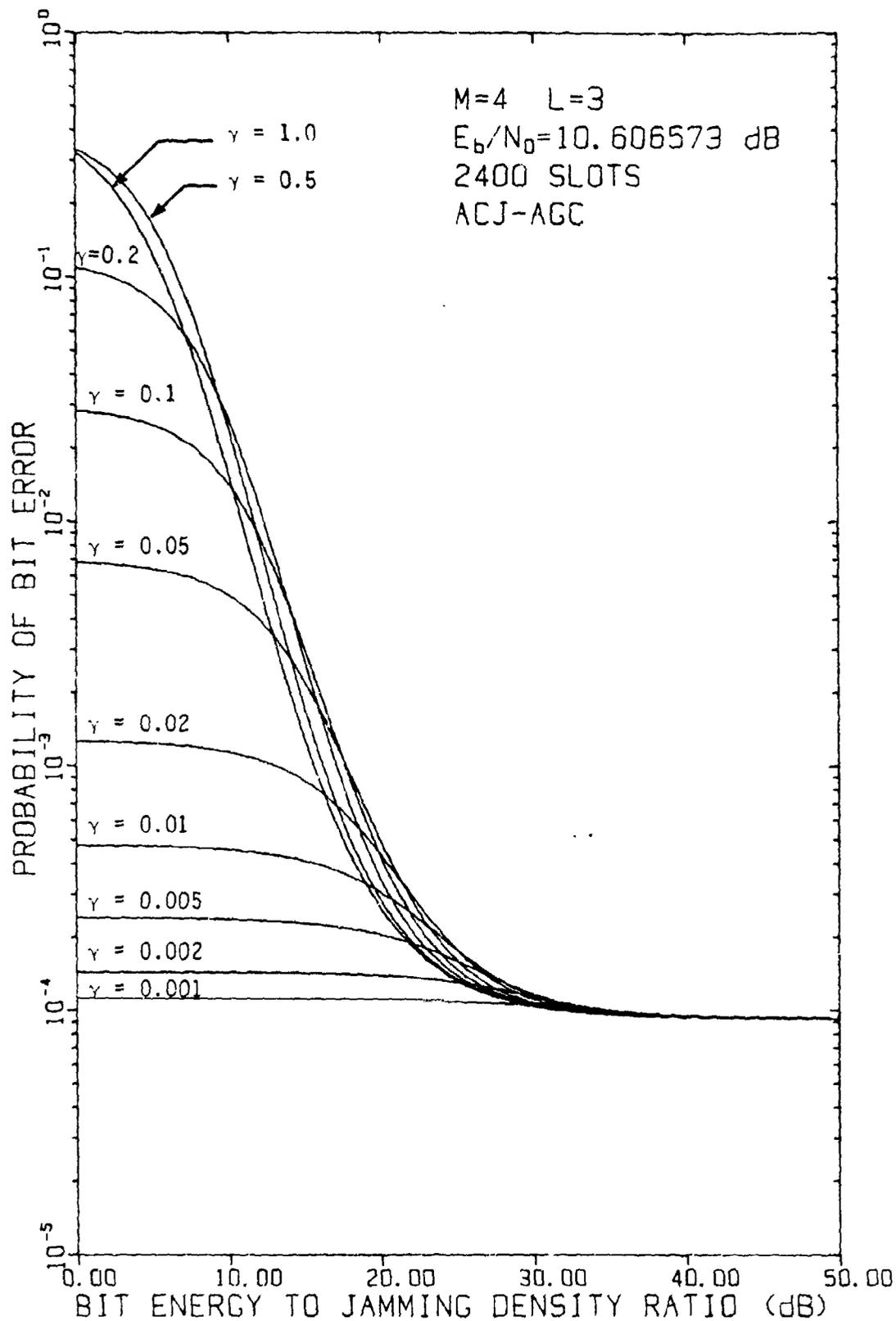


FIGURE 4.4-23 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH $M=4$ AND $L=3$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ AS A PARAMETER

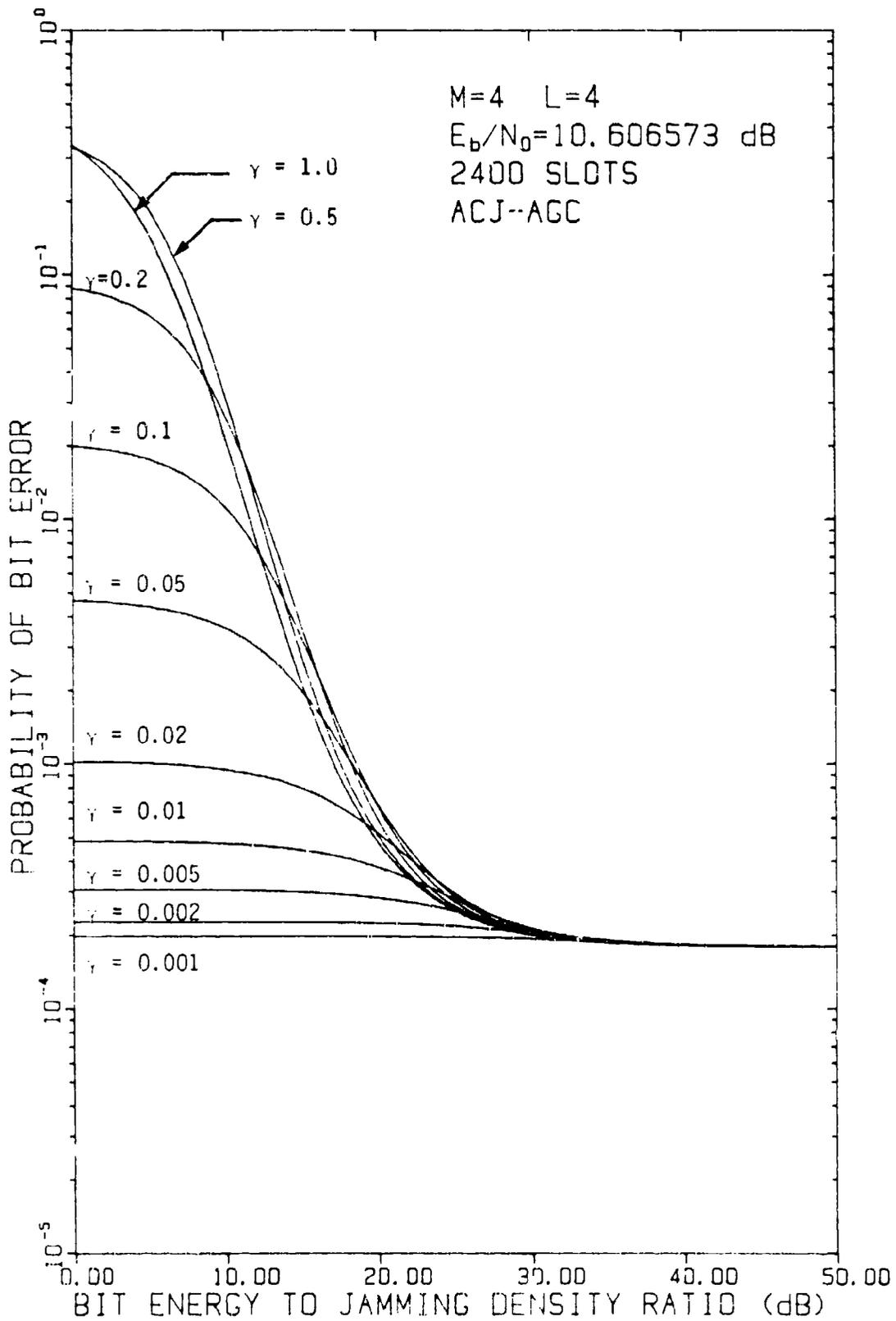


FIGURE 4.4-24 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH $M=4$ AND $L=4$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 10.606573 \text{ dB}$ (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ , AS A PARAMETER

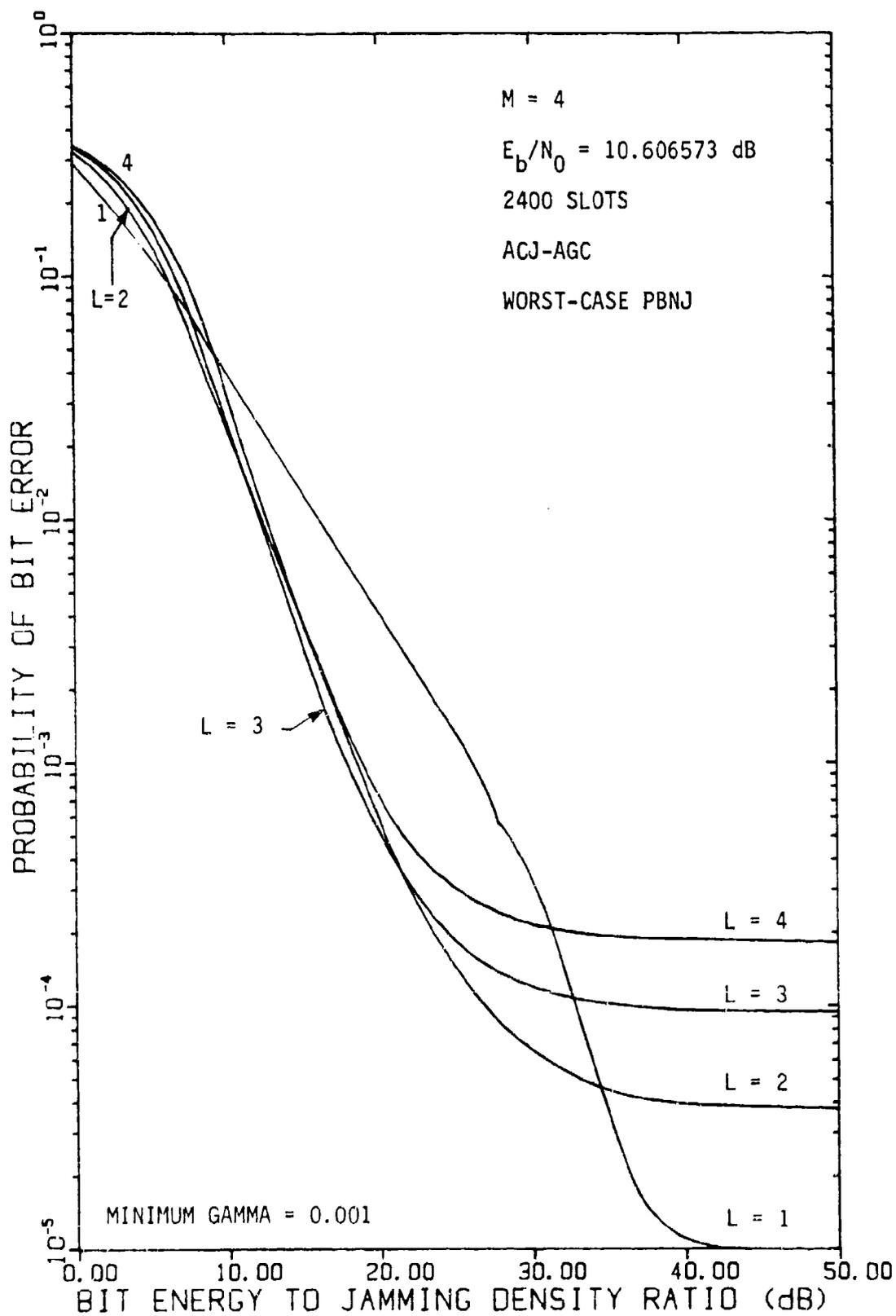


FIGURE 4.4-25 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CANNEL-JAMMED AGC RECEIVER AND $M=4$ WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING

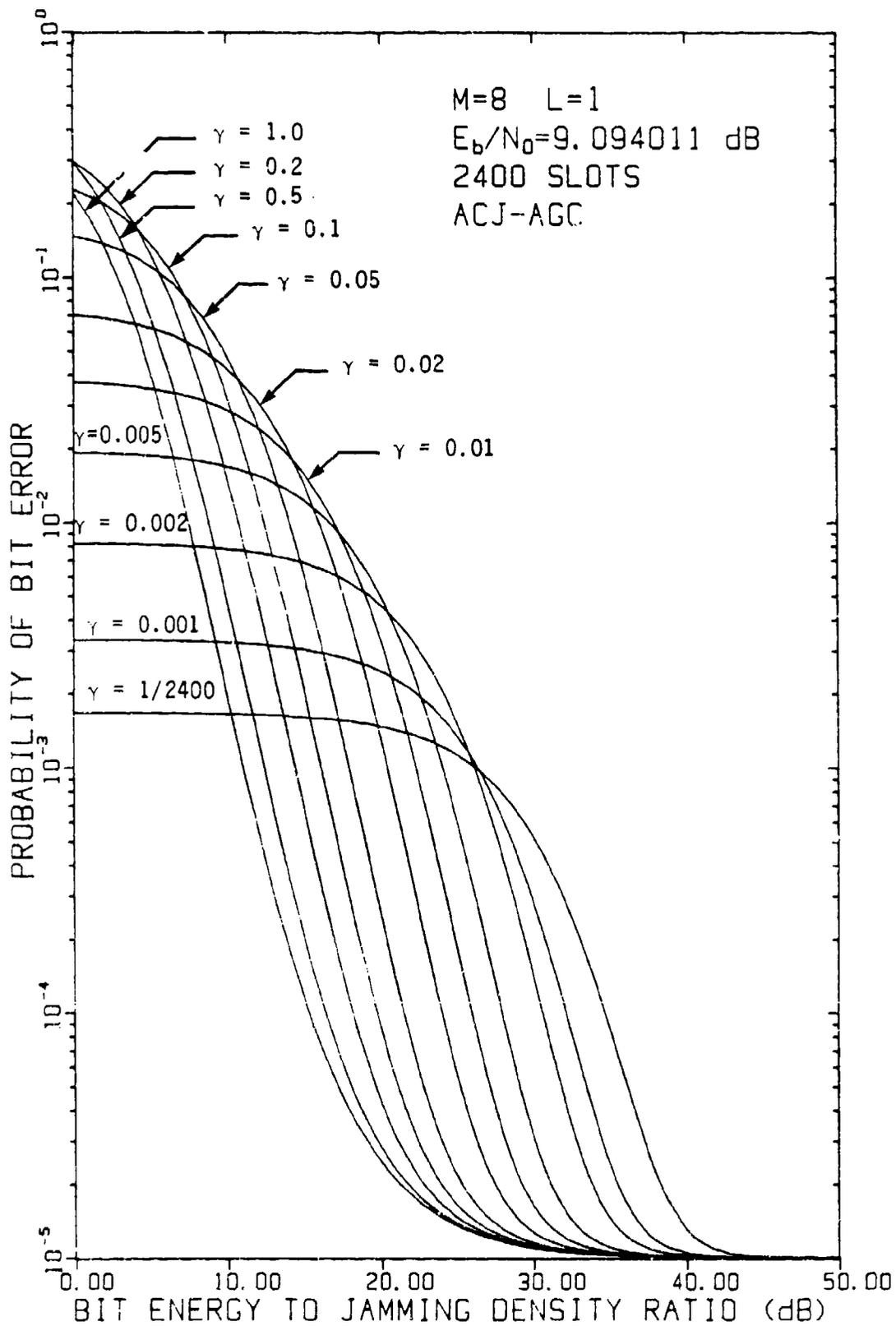


FIGURE 4.4-26 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH $M=8$ AND $L=1$ HOP/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 9.094011 \text{ dB}$ (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ AS A PARAMETER

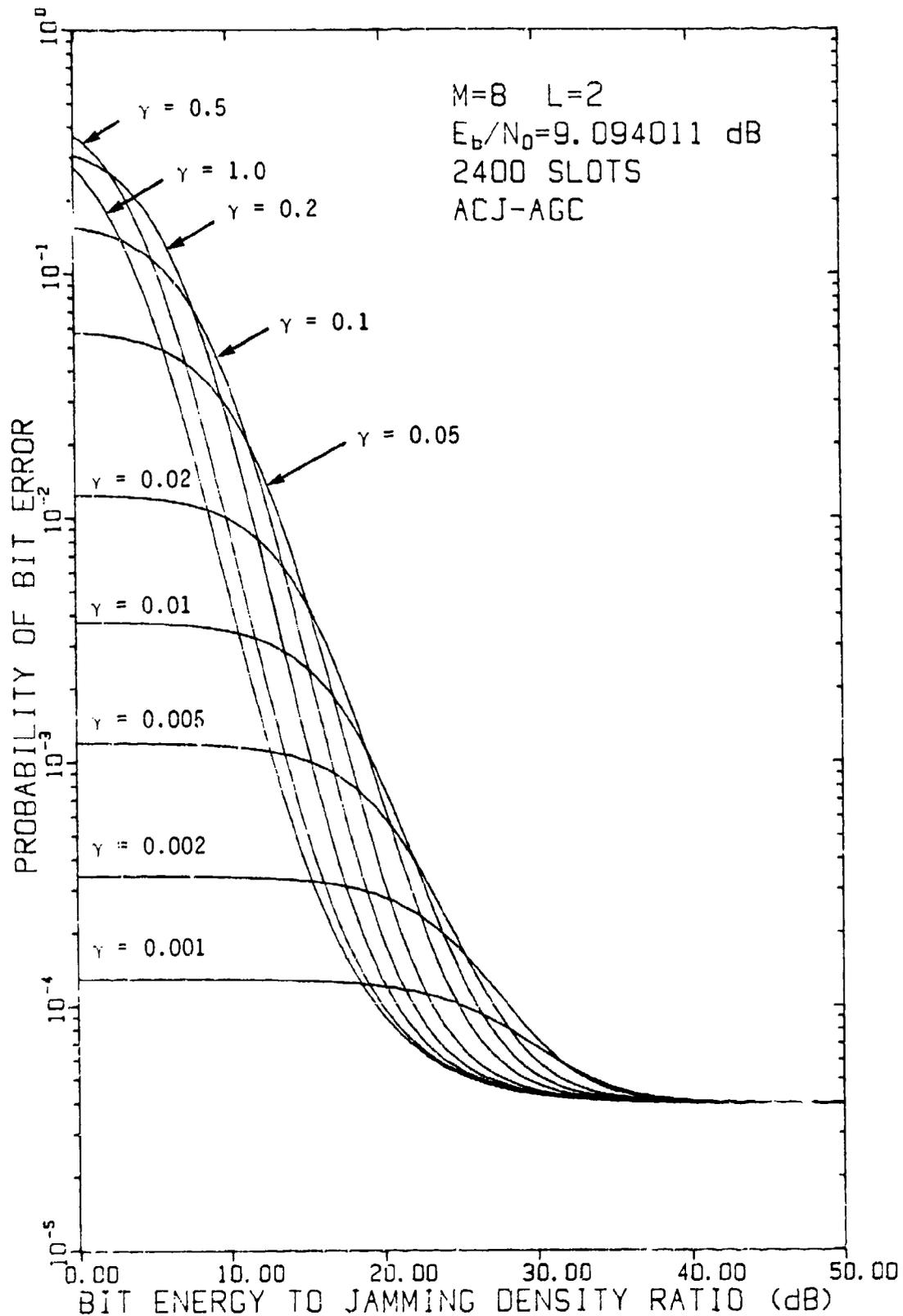


FIGURE 4.4-27 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER WITH $M=8$ AND $L=2$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 9.094011$ dB (FOR 10^{-5} BER WITHOUT JAMMING WHEN $L=1$) WITH JAMMING FRACTION γ AS A PARAMETER

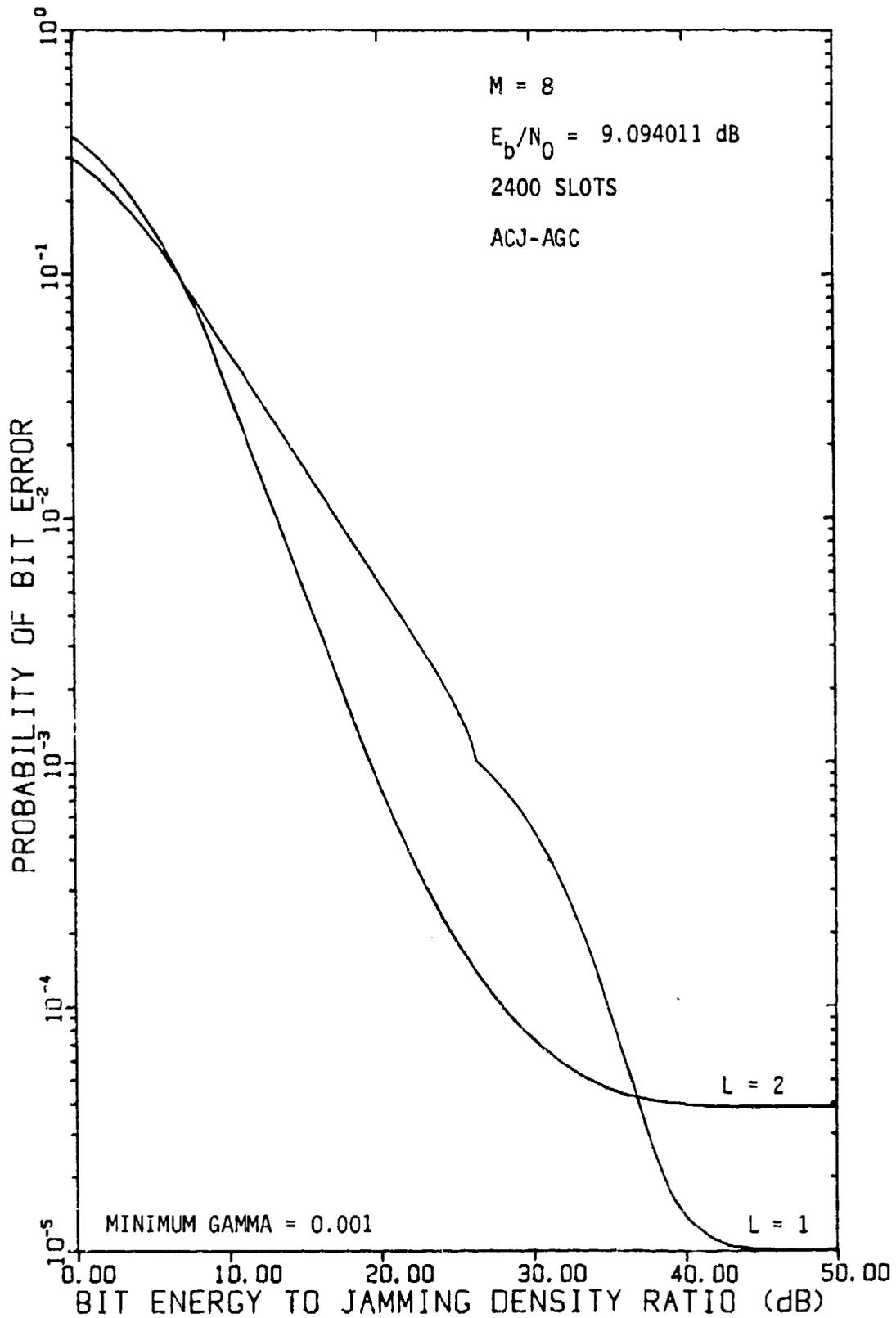


FIGURE 4.4-28 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER AND $M=8$ WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING

5.0 FH/RMFSK PERFORMANCE USING CLIPPER RECEIVER

We now undertake analysis of a third type of ECCM receiver for FH/RMFSK, in which the effect of jamming on the symbol decision is reduced by soft-limiting or clipping the per-hop symbol decision variables $\{z_{mk}; m=1,2,\dots,M; k=1,2,\dots,L\}$. The receiver structure is diagrammed in Figure 5.0-1. In each of the M dehopped symbol channels, the square-law envelope detector samples are clipped at some level η prior to summing to perform the symbol decision. Because the contribution of a jammed hop to the decision variables is at most η , no matter how strong the jammer noise power, it is expected that an improved performance will result. The clipping threshold η is to be chosen to minimize the error probability when there is no jamming.

In previous analyses of the clipper receiver (for conventional FH/MFSK) we had employed a numerical convolution technique to obtain the distributions of the decision variables. Here we shall obtain the needed probability density functions (pdf's) directly, through analysis.

5.1 DISTRIBUTIONS OF THE DECISION VARIABLES

We first discuss the general form for the pdf of the sum of clipped square-law envelope detector samples, then apply this form to non-signal and signal channels.

5.1.1 General Form of the pdf.

If the input to a clipper with clipping level η has the pdf $f_1(x), x \geq 0$, then the output has the pdf

$$p_1(x) = \begin{cases} f_1(x) + q \delta(x-\eta) & 0 \leq x < \eta \\ 0, & \text{otherwise;} \end{cases} \quad (5.1-1a)$$

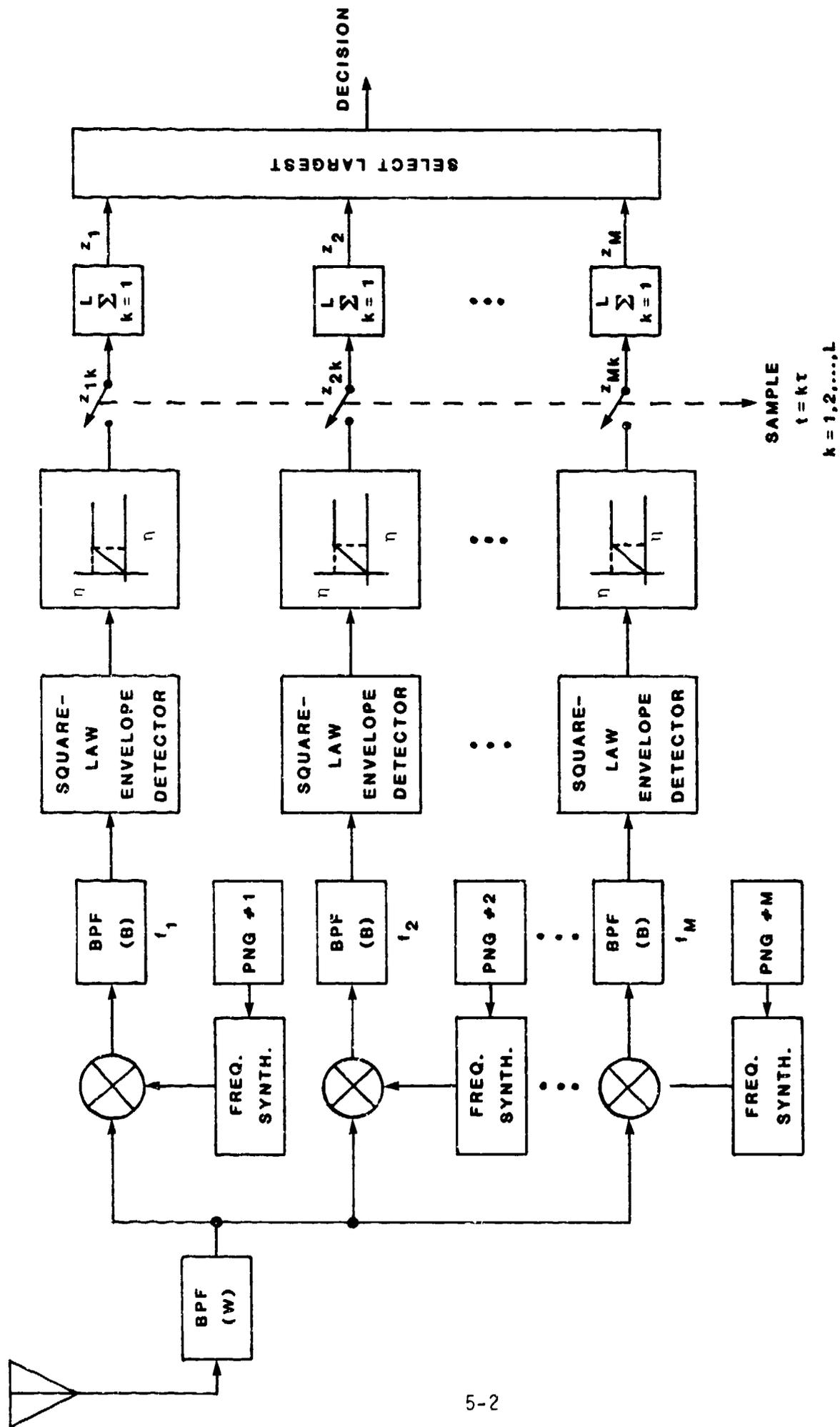


FIGURE 5.0-1 SOFT-DECISION FH/RMFSK RECEIVER WITH CLIPPLERS

where

$$q = \Pr\{\text{input} > \eta\} = \int_{\eta}^{\infty} d\alpha f_1(\alpha) \quad (5.1-1b)$$

This fact is illustrated by figure 5.1-1.

Now, since individual hops are jammed independently and in any combination, we introduce the notations

$$f_L(x; \ell) \equiv \text{non-delta function part of the pdf for the sum of } L \text{ clipped samples when } \ell \text{ hops in that channel are jammed.} \quad (5.1-2a)$$

$$q_0 = \Pr\{\text{one sample} > \eta \mid \text{not jammed}\} \quad (5.1-2b)$$

$$q_1 = \Pr\{\text{one sample} > \eta \mid \text{jammed}\}. \quad (5.1-2c)$$

Note that it is sufficient to specify only ℓ , the number of hops jammed; the order in which the jamming occurs does not affect the sum. Using this notation, the pdf for a single clipped envelope sample is

$$p_1(x; 0) = f_1(x; 0) + q_0 \delta(x - \eta), \text{ hop not jammed;} \quad (5.1-3a)$$

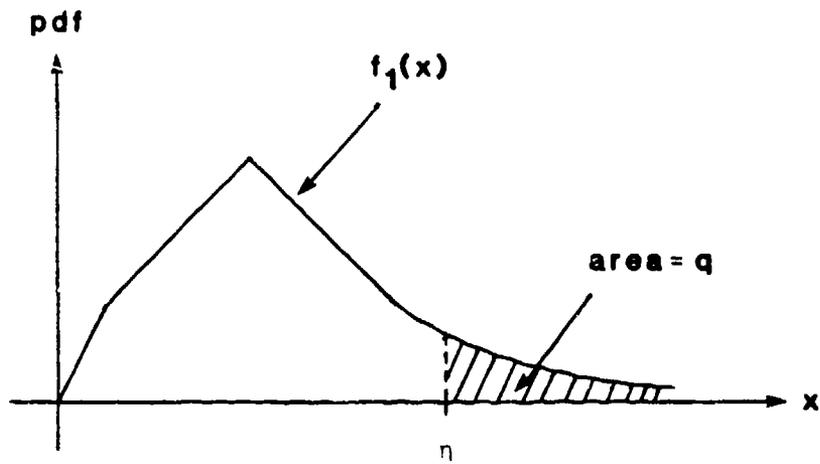
$$p_1(x; 1) = f_1(x; 1) + q_1 \delta(x - \eta), \text{ hop jammed;} \quad (5.1-3b)$$

and it is understood that the pdf is zero outside the interval $0 \leq x \leq \eta$.

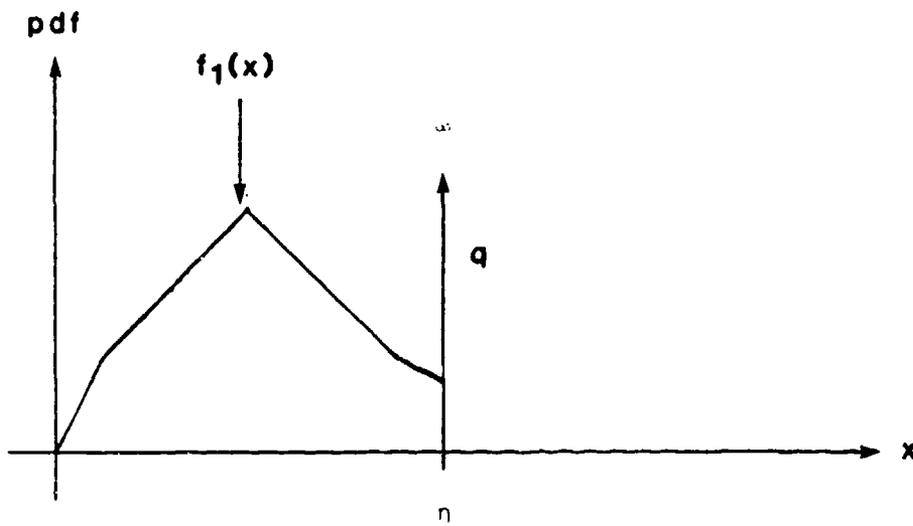
For ℓ hops jammed, the pdf of the sum of L clipped samples can be expressed as the convolution

$$\underbrace{p_1(x; 0) * p_1(x; 0) * \dots * p_1(x; 0)}_{L - \ell \text{ pdf's}} * \underbrace{p_1(x; 1) * \dots * p_1(x; 1)}_{\ell \text{ pdf's}}. \quad (5.1-4)$$

Thus we have the following general expressions for the sum's pdf for $L=2$ to 4:



(a) pdf before clipping



(b) pdf after clipping

FIGURE 5.1-1 EFFECTS OF CLIPPING ON pdf

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$$p_2(x;0) = f_2(x;0) + q_0^2 \delta(x-2\eta) , \quad 0 \leq x \leq 2\eta; \quad (5.1-5a)$$

where

$$f_2(x;0) = \begin{cases} \int_0^x dw f_1(x-w;0) f_1(w;0) , & 0 \leq x < \eta; \\ \int_{x-\eta}^{\eta} dw f_1(x-w;0) f_1(w;0) \\ \quad + 2q_0 f_1(x-\eta;0) , & \eta \leq x \leq 2\eta. \end{cases} \quad (5.1-5b)$$

$$p_2(x;1) = f_2(x;1) + q_0 q_1 \delta(x-2\eta) , \quad 0 \leq x \leq 2\eta; \quad (5.1-6a)$$

where

$$f_2(x;1) = \begin{cases} \int_0^x dw f_1(x-w;1) f_1(w;0) , & 0 \leq x < \eta; \\ \int_{x-\eta}^{\eta} dw f_1(x-w;1) f_1(w;0) \\ \quad + q_0 f_1(x-\eta;1) + q_1 f_1(x-\eta;0) , & \eta \leq x \leq 2\eta. \end{cases} \quad (5.1-6b)$$

$$p_2(x;2) = f_2(x;2) + q_1^2 \delta(x-2\eta) , \quad 0 \leq x \leq 2\eta; \quad (5.1-7a)$$

where

$$f_2(x;2) = \begin{cases} \int_0^x dw f_1(x-w;1) f_1(w;1) , & 0 \leq x < \eta; \\ \int_{x-\eta}^{\eta} dw f_1(x-w;1) f_1(w;1) \\ \quad + 2q_1 f_1(x-\eta;1) , & \eta \leq x \leq 2\eta. \end{cases} \quad (5.1-7b)$$

$$p_3(x;0) = f_3(x;0) + q_0^3 \delta(x-3\eta) , \quad 0 \leq x \leq 3\eta; \quad (5.1-8a)$$

where

$$f_3(x;0) = \begin{cases} \int_0^x dw f_1(x-w;0) f_2(w;0) , & 0 \leq x < \eta; \\ \int_{x-\eta}^{\eta} dw f_1(x-w;0) f_2(w;0) \\ \quad + q_0 f_2(x-\eta;0) , & \eta \leq x < 2\eta; \end{cases} \quad (5.1-8b)$$

and

$$f_3(x;0) = \begin{cases} \int_{x-n}^{2n} dw f_1(x-w;0) f_2(w;0) \\ + q_0 f_2(x-n;0) \\ + q_0^2 f_1(x-2n;0) , \quad 2n \leq x \leq 3n . \end{cases} \quad (5.1-8c)$$

$$p_3(x;1) = f_3(x;1) + q_0^2 q_1 \delta(x-3n) , \quad 0 \leq x \leq 3n; \quad (5.1-9a)$$

where

$$f_3(x;1) = \begin{cases} \int_0^x dw f_1(x-w;1) f_2(w;0) , \quad 0 \leq x < n \\ \int_{x-n}^x dw f_1(x-w;1) f_2(w;0) \\ + q_1 f_2(x-n;0) , \quad n \leq x < 2n \\ \int_{x-n}^{2n} dw f_1(x-w;1) f_2(w;0) \\ + q_1 f_2(x-n;0) \\ + q_0^2 f_1(x-2n;1) , \quad 2n \leq x \leq 3n . \end{cases} \quad (5.1-9b)$$

$$p_3(x;2) = f_3(x;2) + q_0 q_1^2 \delta(x-3n) , \quad 0 \leq x \leq 3n; \quad (5.1-10a)$$

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where

$$f_3(x;2) = \left\{ \begin{array}{l} \int_0^x dw f_1(x-w;0)f_2(w;2) , \quad 0 \leq x < n; \\ \int_{x-n}^x dw f_1(x-w;0)f_2(w;2) \\ \quad + q_0 f_2(x-n;2) , \quad n \leq x < 2n; \\ \int_{x-n}^{2n} dw f_1(x-w;0)f_2(w;2) \\ \quad + q_0 f_2(x-n;2) \\ \quad + q_1^2 f_1(x-2n;0) , \quad 2n \leq x \leq 3n. \end{array} \right. \quad (5.1-10b)$$

$$p_3(x;3) = f_3(x;3) + q_1^3 \delta(x-3n) , \quad 0 < x < 3n ; \quad (5.1-11a)$$

where

$$f_3(x;3) = \left\{ \begin{array}{l} \int_0^x dw f_1(x-w;1)f_2(w;2) , \quad 0 \leq x < n; \\ \int_{x-n}^x dw f_1(x-w;1)f_2(w;2) \\ \quad + q_1 f_2(x-n;2) , \quad n \leq x < 2n; \\ \int_{x-n}^{2n} dw f_1(x-w;1)f_2(w;2) \\ \quad + q_1 f_2(x-n;2) \\ \quad + q_1^2 f_1(x-2n;1) , \quad 2n \leq x \leq 3n. \end{array} \right. \quad (5.1-11b)$$

$$p_4(x;0) = f_4(x;0) + q_0^4 \delta(x-4n), \quad 0 \leq x \leq 4n; \quad (5.1-12a)$$

where

$$f_4(x;0) = \left\{ \begin{array}{l} \int_0^x dw f_1(x-w;0) f_3(w;0), \quad 0 \leq x < n; \\ \int_{x-n}^x dw f_1(x-w;0) f_3(w;0) \\ \quad + q_0 f_3(x-n;0), \quad n \leq x < 3n \\ \int_{x-n}^{3n} dw f_1(x-w;0) f_3(w;0) \\ \quad + q_0 f_3(x-n;0) \\ \quad + q_0^3 f_1(x-3n;0), \quad 3n \leq x \leq 4n. \end{array} \right. \quad (5.1-12b)$$

$$p_4(x;1) = f_4(x;1) + q_0^3 q_1 \delta(x-4n), \quad 0 \leq x \leq 4n; \quad (5.1-13a)$$

where

$$f_4(x;1) = \left\{ \begin{array}{l} \int_0^x dw f_1(x-w;1) f_3(w;0), \quad 0 \leq x < n; \\ \int_{x-n}^x dw f_1(x-w;1) f_3(w;0) \\ \quad + q_1 f_3(x-n;0), \quad n \leq x < 3n; \\ \int_{x-n}^{3n} dw f_1(x-w;1) f_3(w;0) \\ \quad + q_1 f_3(x-n;0) \\ \quad + q_0^3 f_1(x-3n;1), \quad n < x < 4n. \end{array} \right. \quad (5.1-13b)$$

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$$p_4(x;2) = f_4(x;2) + q_0^2 q_1^2 \delta(x-4n), \quad 0 \leq x \leq 4n; \quad (5.1-14a)$$

where

$$f_4(x;2) = \begin{cases} \int_0^x dw f_2(x-w;2) f_2(w;0), & 0 \leq x < 2n; \\ \int_{x-2n}^{2n} dw f_2(x-w;2) f_2(w;0) \\ \quad + q_0^2 f_2(x-2n;2) \\ \quad + q_1^2 f_2(x-2n;0), & 2n \leq x \leq 4n. \end{cases} \quad (5.1-14b)$$

$$p_4(x;3) = f_4(x;3) + q_0 q_1^3 \delta(x-4n), \quad 0 \leq x \leq 4n; \quad (5.1-15a)$$

where

$$f_4(x;3) = \begin{cases} \int_0^x dw f_1(x-w;0) f_3(w;3), & 0 \leq x < n; \\ \int_{x-n}^x dw f_1(x-w;0) f_3(w;3) \\ \quad + q_0 f_3(x-n;3), & n \leq x < 3n; \\ \int_{x-n}^{3n} dw f_1(x-w;0) f_3(w;3) \\ \quad + q_0 f_3(x-n;3) \\ \quad + q_1^3 f_1(x-3n;0), & 3n \leq x \leq 4n. \end{cases} \quad (5.1-15b)$$

$$p_4(x;4) = f_4(x;4) + q_1^4 \delta(x-4n), \quad 0 \leq x \leq 4n; \quad (5.1-16a)$$

$$f_4(x;4) = \begin{cases} \int_0^x dw f_1(x-w;1)f_3(w;3) , & 0 \leq x < \eta; \\ \int_{x-\eta}^x dw f_1(x-w;1)f_3(w;3) \\ \quad + q_1 f_3(x-\eta;3) , & \eta \leq x < 3\eta ; \\ \int_{x-\eta}^{3\eta} dw f_1(x-w;1)f_3(w;3) \\ \quad + q_1 f_3(x-\eta;3) \\ \quad + q_1^3 f_1(x-3\eta;1) , & 3\eta \leq x \leq 4\eta. \end{cases} \quad (5.1-16b)$$

5.1.2 Non-Signal Channel pdf.

Assuming without loss of generality that the received signal power S is present in the first ($m=1$) of M dehopped symbol frequency channels, the remaining channels ($m=2,3,\dots,M$) contain only background noise and possibly jamming noise. The samples of the square-law envelope detectors in these channels are independent chi-squared random variables with two degrees of freedom, multiplied by σ_{mk}^2 , where

$$\sigma_{mk}^2 = \begin{cases} \sigma_N^2 = N_0 B, & \text{hop not jammed} \\ \sigma_T^2 = (N_0 + N_J/\gamma) B, & \text{hop jammed.} \end{cases} \quad (5.1-17)$$

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Consequently, the pdf of the unclipped samples is

$$f_1(x) = \frac{1}{2\sigma_{mk}^2} e^{-x/2\sigma_{mk}^2}, \quad x \geq 0;$$

$$= \begin{cases} a e^{-ax} & , x \geq 0, \text{ hop not jammed;} \\ b e^{-bx} & , x \geq 0, \text{ hop jammed;} \end{cases} \quad (5.1-18a)$$

using

$$a \equiv 1/2\sigma_N^2, \quad b \equiv 1/2\sigma_T^2. \quad (5.1-18b)$$

Also, we have from (5.1-2b and c)

$$q_0 = e^{-a\eta}, \quad q_1 = e^{-b\eta}. \quad (5.1-19)$$

In order to distinguish the non-signal channel pdf's from that of the signal channel, we adopt the notation

$$g_L(x; \ell) = f_L(x; \ell, S=0) \quad (5.1-20)$$

for the non-delta function part of the pdf of the sum of L clipped samples when ℓ hops in that channel are jammed. Thus we have for channels $\{m:m>2\}$, the sum pdf

$$p_{Z_m}(x) = \begin{cases} g_L(x; \ell_m) + (q_0)^{L-\ell_m} (q_1)^{\ell_m} \delta(x-L\eta), & 0 \leq x \leq L\eta; \\ 0, & \text{otherwise.} \end{cases} \quad (5.1-21)$$

Substituting (5.1-18) in the general convolutional formulas in Section 5.1.1 yields the pdf's listed in Table 5.1-1 for $L=1$ to 3.

TABLE 5.1-1 NON-SIGNAL CHANNEL PROBABILITY DENSITY FUNCTIONS, L = 1 TO 3

L	ℓ	$p_{zm}(x; \ell)$
1	0	$ae^{-ax} + e^{-an} \delta(x-n), 0 \leq x \leq n; 0, \text{ otherwise.} \quad a \equiv 1/2\sigma_T^2$
	1	$be^{-ax} + e^{-bn} \delta(x-n), 0 \leq x \leq n; 0, \text{ otherwise.} \quad b \equiv 1/2\sigma_T^2$
2	0	$a^2 x e^{-ax}, 0 \leq x < n; [2a + a^2(2n-x)]e^{-ax} + e^{-2an} \delta(x-2n), n \leq x \leq 2n; 0, \text{ otherwise}$
	1	$\frac{ab}{a-b} (e^{-bx} - e^{-ax}), 0 < x < n; \frac{a^2}{a-b} e^{-bn-a(x-n)} - \frac{b^2}{a-b} e^{-an-b(x-n)} + e^{-(a+b)n} \delta(x-2n), n \leq x < 2n; 0, \text{ otherwise}$
3	2	same as for $\ell = 0$, but with a replaced by b
	0	$\frac{1}{2} a^3 x^2 e^{-ax}, 0 \leq x < n; 0, x > 3n, x < 0$ $a^2 [\frac{1}{2} a n^2 + (3+an)(x-n) - a(x-n)^2] e^{-ax}, n \leq x < 2n$ $a [3+3an + \frac{1}{2} a^2 n^2 - a(3+an)(x-2n) + \frac{1}{2} a^2 (x-2n)^2] e^{-ax} + e^{-3an} \delta(x-3n), 2n \leq x \leq 3n.$
1	1	$\frac{a^2 b}{(a-b)^2} \left\{ e^{-bx} - e^{-ax} [1+(a-b)x] \right\}, 0 \leq x < n; 0, x > 3n, x < 0;$ $\frac{a}{a-b} e^{-a(x-n)} \left\{ a \left[\frac{b}{a-b} + a(x-n) \right] e^{-bn} + b \left[\frac{a}{a-b} - (2+an) + a(x-n) \right] e^{-an} \right\} - \frac{2ab^2}{(a-b)^2} e^{-an-b(x-n)}, n \leq x < 2n$ $\frac{a^2}{a-b} e^{-(a+b)n-a(x-2n)} \left[\frac{2a-3b}{a-b} + an-a(x-2n) \right] + \frac{b^3}{(a-b)^2} e^{-2an-b(x-2n)} + e^{-(2a+b)n} \delta(x-3n), 2n \leq x < 3n.$
	2	same as for $\ell = 1$, but with a and b exchanged
3	3	same as for $\ell = 0$, but with a replaced by b

In the conditional probability of error expression,

$$P_s(e|\ell, L) = 1 - E_{z_1} \left\{ \prod_{m=2}^M \Pr\{z_m < z_1 | \ell, L\} \right\}, \quad (5.1-22)$$

the cumulative distribution function for the non-signal channels is needed, written

$$G_L(x; \ell) \triangleq \Pr\{z_m \leq x | \ell, L\}. \quad (5.1-23)$$

This function is given in Table 5.1-2 for L=1 to 3.

5.1.3 Signal Channel pdf.

The samples of the square-law envelope detector in the signal channel are independent noncentral chi-squared random variables with two degrees of freedom, multiplied by σ_{1k}^2 , and with noncentrality parameters

$$\lambda_k = 2S/\sigma_{1k}^2 = \begin{cases} 2S/\sigma_N^2 = 2\rho_N, & \text{hop not jammed} \\ 2S/\sigma_T^2 = 2\rho_T, & \text{hop jammed.} \end{cases} \quad (5.1-24)$$

Consequently the pdf of the unclipped samples is

$$f_1(x) = \frac{1}{2\sigma_{1k}^2} e^{-(x+2S)/2\sigma_{1k}^2} I_0(\sqrt{2Sx}/\sigma_{1k}^2) \\ = \begin{cases} a e^{-a(x+2S)} I_0(2a\sqrt{2Sx}), & x \geq 0, \text{ hop not jammed;} \\ b e^{-b(x+2S)} I_0(2b\sqrt{2Sx}), & x \geq 0, \text{ hop jammed;} \end{cases} \quad (5.1-25)$$

where a and b are given by (5.1-18b). To distinguish the signal case from the non-signal case, the q_0 and q_1 defined by (5.1-2) will be written in the upper case; the values are

TABLE 5.1-2 NON-SIGNAL CHANNEL CUMULATIVE DISTRIBUTION FUNCTIONS, L=1 TO 3

L ℓ $G_L(x; \ell) = \Pr\{Z_m \leq x | \ell, L\}, m \geq 2$

1	0	$1 - e^{-ax}, 0 \leq x < n; 1, x \geq n$	$a \equiv 1/2\sigma_N^2$
	1	$1 - e^{-bx}, 0 \leq x < n; 1, x \geq n$	$b \equiv 1/2\sigma_I^2$
2	0	$1 - e^{-ax}(1+ax), 0 \leq x < n; 1 - e^{-ax}[1+a(2n-x)], n \leq x < 2n; 1, x \geq 2n$	
	1	$1 - \frac{1}{a-b}[ae^{-bx} - be^{-ax}], 0 \leq x < n; 1 - \frac{1}{a-b}[ae^{-bn-a(x-n)} - be^{-an-b(x-n)}], n \leq x < 2n; 1, x \geq 2n$	
3	2	$1 - e^{-bx}(1+bx), 0 \leq x < n; 1 - e^{-bx}[1+b(2n-x)], n \leq x < 2n; 1, x \geq 2n$	
	0	$1 - e^{-ax}(1+ax+\frac{1}{2}a^2x^2), 0 \leq x < n; 1 - e^{-ax}[1+an+\frac{1}{2}a^2n^2+a(1+an)(x-n)-a^2(x-n)^2], n \leq x < 2n;$ $1 - e^{-ax}[1+2an+\frac{1}{2}a^2n^2-a(2+an)(x-2n)+\frac{1}{2}a^2(x-2n)^2], 2n \leq x < 3n; 1, x \geq 3n$	
1	1	$1 - \frac{a^2}{(a-b)}e^{-bx} + \frac{b}{a-b}e^{-ax} \left[\frac{2a-b}{a-b} + ax \right], 0 \leq x < n;$ $1 - e^{-a(x-n)} \left\{ e^{-bn} \left[\frac{a^2}{(a-b)^2} + \frac{a^2}{a-b}(x-n) \right] + e^{-an} \left[\frac{b^2}{(a-b)^2} + \frac{ab}{a-b}(x-2n) \right] \right\} + \frac{2ab}{(a-b)^2} e^{-an-b(x-n)}, n \leq x < 2n;$ $1 - e^{-(a+b)n-a(x-2n)} \left[\frac{a(1+an)}{a-b} - \frac{ab}{(a-b)^2} - \frac{a^2}{a-b}(x-2n) \right] - \frac{b^2}{(a-b)^2} e^{-2an-b(x-2n)}, 2n \leq x < 3n; 1, x \geq 3n$	
	2	same as for $\ell = 1$, but with a and b exchanged	
3	3	same as for $\ell = 0$, but with a replaced by b	

$$\begin{aligned}
 Q_0 &= \Pr\{z_{1k} > \eta \mid \text{not jammed}\} \\
 &= Q(2\sqrt{aS}, \sqrt{2a\eta})
 \end{aligned}
 \tag{5.1-26a}$$

and

$$\begin{aligned}
 Q_1 &= \Pr\{z_{1k} > \eta \mid \text{jammed}\} \\
 &= Q(2\sqrt{bS}, \sqrt{2b\eta}),
 \end{aligned}
 \tag{5.1-26b}$$

where $Q(x,y)$ is Marcum's Q-function.

For the sum of clipped samples, the pdf is

$$p_{z_1}(x) = \begin{cases} f_L(x; \ell_1) + (Q_0)^{L-\ell} (Q_1)^{\ell} \delta(x-L\eta), & 0 \leq x \leq L\eta; \\ 0, & \text{otherwise.} \end{cases}
 \tag{5.1-27}$$

Substituting (5.1-25) in the general convolutional formulas in Section 5.1.1 yields the pdf's listed in Table 5.1-3 for $L=1, 2$ and Table 5.1-4 for $L=3$.

5.2 ERROR PROBABILITY FORMULATION

Having the pdf's for the FH/RMFSK clip-and-sum decision variables $\{z_m\}$, we can formulate the probability of error.

5.2.1 Conditional Probability of Error.

The probability of symbol error, conditioned on the jamming event $\underline{\ell} = (\ell_1, \ell_2, \dots, \ell_M)$ where ℓ_m is the number of hops jammed (out of L hops) in channel m , can be expressed as parametric in η , the clipping threshold, by

TABLE 5.1-3 SIGNAL CHANNEL PROBABILITY DENSITY FUNCTIONS, L = 1,2

$$P_{Z_1}(x|x, L)$$

L	λ	$P_{Z_1}(x x, L)$
1	0	$ae^{-a(x+2S)} I_0(2a\sqrt{2xS}) + Q_0\delta(x-n), 0 \leq x \leq n. \quad a \equiv 1/2\sigma_N^2, Q_0 = Q(2\sqrt{aS}, \sqrt{2an})$
	1	$be^{-b(x+2S)} I_0(2b\sqrt{2xS}) + Q_1\delta(x-n), 0 \leq x \leq n. \quad b \equiv 1/2\sigma_N^2, Q_1 = Q(2\sqrt{bS}, \sqrt{2bn})$
2	0	$\frac{1}{2}a\sqrt{x/S} e^{-a(x+4S)} I_1(4a\sqrt{Sx}), 0 \leq x < n; \quad 0, x > 2n, x < 0;$ $2aQ_0 e^{-a(x-n+2S)} I_0(2a\sqrt{2S(x-n)}) + a^2 e^{-a(x+4S)} \int_{x-n}^n dw I_0(2a\sqrt{2Sw}) I_0(2a\sqrt{2S(x-w)})$ $+ Q_0^2 \delta(x-2n), n \leq x \leq 2n.$
	1	$abe^{-bx-2(a+b)S} \int_0^x dw e^{-(a-b)w} I_0(2a\sqrt{2Sw}) I_0(2b\sqrt{2S(x-w)}), 0 \leq x < n; \quad 0, x > 2n, x < 0;$ $bQ_0 e^{-b(x-n+2S)} I_0(2b\sqrt{2S(x-n)}) + aQ_1 e^{-a(x-n+2S)} I_0(2a\sqrt{2S(x-n)})$ $+ abe^{-bx-2(a+b)S} \int_{x-n}^n dw e^{-(a-b)w} I_0(2a\sqrt{2Sw}) I_0(2b\sqrt{2S(x-w)}) + Q_0 Q_1 \delta(x-2n), n \leq x \leq 2n.$
2	2	same as for $\lambda = 0$, but with a replaced by b and Q_0 replaced by Q_1 .

TABLE 5.1-4 SIGNAL CHANNEL PROBABILITY DENSITY FUNCTIONS, L = 3

L	i	$P_{z1}(x z, L)$
3	0	$(ax/6S) e^{-a(x+6S)} I_2(2a\sqrt{6Sx}), 0 \leq x < n; 0, x > 3n, x < 0;$ $\frac{3}{2} aQ_0 \sqrt{(x-n)/S} e^{-a(x-n+4S)} I_1(4a\sqrt{S(x-n)}) + \frac{1}{2} a^2 e^{-a(x+6S)} \int_{x-n}^n dw \sqrt{w/S} I_1(4a\sqrt{Sw}) I_0(2a\sqrt{2S(x-w)})$ $+ a^3 e^{-a(x+6S)} \int_n^x dw I_0(2a\sqrt{2S(x-w)}) \int_{w-n}^n dz I_0(2a\sqrt{2Sz}) I_0(2a\sqrt{2S(w-z)}), n \leq x < 2n;$ $3aQ_0 e^{-a(x-2n+2S)} I_0(2a\sqrt{2S(x-2n)})$ $+ 3a^2 Q_0 e^{-a(x-n+4S)} \int_{x-n}^{2n} dw I_0(2a\sqrt{2S(x-w)}) I_0(2a\sqrt{2S(w-n)}) + Q_0^3 \delta(x-3n)$ $+ a^3 e^{-a(x+6S)} \int_{x-n}^{2n} dw I_0(2a\sqrt{2S(x-w)}) \int_{w-n}^n dz I_0(2a\sqrt{2Sz}) I_0(2a\sqrt{2S(w-z)}), 2n \leq x \leq 3n.$
	1	$\frac{1}{2} ab e^{-bx-2(2a+b)S} \int_0^x dw e^{-(a-b)w} \sqrt{w/S} I_0(2b\sqrt{2S(x-w)}) I_1(4a\sqrt{Sw}), 0 \leq x < n; 0, x > 3n, x < 0;$ $\frac{1}{2} aQ_1 \sqrt{(x-n)/S} e^{-a(x-n+4S)} I_1(4a\sqrt{S(x-n)}) + \frac{1}{2} abe^{-bx-(2a+b)2S} \int_{x-n}^n dw \sqrt{w/S} I_1(2b\sqrt{2S(x-w)}) e^{-(a-b)w}$ $+ 2abQ_0 e^{-b(x-n)-(a+b)2S} \int_0^{x-n} dw I_0(2a\sqrt{2Sw}) I_0(2b\sqrt{2S(x-n-w)}) e^{-(a-b)w}$ $+ a^2 b e^{-bx-(2a+b)2S} \int_n^x dw e^{-(a-b)w} I_0(2b\sqrt{2S(x-w)}) \int_{w-n}^n dz I_0(2a\sqrt{2Sz}) I_0(2a\sqrt{2S(w-z)}), n \leq x \leq 2n;$ $bQ_0^2 e^{-b(x-2n+2S)} I_0(2b\sqrt{2S(x-2n)}) + 2aQ_0Q_1 e^{-a(x-2n+2S)} I_0(2a\sqrt{2S(x-2n)})$ $+ a^2 Q_1 e^{-a(x-n+4S)} \int_{x-n}^{2n} dw I_0(2a\sqrt{2S(x-w)}) I_0(2a\sqrt{2S(w-n)})$ $+ 2abQ_0 e^{-bx+a\eta-(a+b)2S} \int_{x-n}^{2n} dw I_0(2b\sqrt{2S(x-w)}) I_0(2a\sqrt{2S(w-n)}) e^{-(a-b)w}$ $+ a^2 b e^{-bx-(2a+b)2S} \int_{x-n}^{2n} dw e^{-(a-b)w} I_0(2b\sqrt{2S(x-w)}) \int_{w-n}^n dz I_0(2a\sqrt{2Sz}) I_0(2a\sqrt{2S(w-z)}),$ <p style="text-align: right;">$2n \leq x \leq 3n.$</p>
	2	<p>same as for $i = 1$, but with a and b exchanged and Q_0 and Q_1 exchanged</p>
	3	<p>same as for $i = 0$, but with a replaced by b and Q_0 replaced by Q_1</p>

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$$P_s(e; n|L, \underline{\ell}) = 1 - \Pr\{C \equiv \text{correct decision}; n|L, \underline{\ell}\}. \quad (5.2-1)$$

Since there is clipping, there is a finite probability that one or more of the $\{z_m\}$ are equal to L . Thus an appropriate formulation, assuming a randomized decision rule, is

$$\begin{aligned} \Pr\{C; n|L, \underline{\ell}\} &= \sum_{p=0}^M \Pr\{C \text{ and } (p \text{ channels} = L_n); n|L, \underline{\ell}\} \\ &= \sum_{p=0}^M \Pr\{C \text{ and } (p \text{ channels} = L_n, \text{ including signal channel}); n|L, \underline{\ell}\} \\ &= \Pr\{C \text{ and no channels} = L_n; n|L, \underline{\ell}\} \\ &\quad + \sum_{p=0}^{M-1} \Pr\{C \text{ and } (\text{signal channel} = L_n) \text{ and } (p \text{ non-signal channels} = L_n); \\ &\quad \quad \quad n|L, \underline{\ell}\}. \end{aligned} \quad (5.2-2)$$

The first term in (5.2-2) is

$$\int_0^{L_n} dx f_L(x; \ell_1) \prod_{m=2}^M G_L(x; \ell_m), \quad (5.2-3)$$

where $f_L(x; \ell_1)$ is the non-delta function part of the signal channel's pdf, and $G_L(x; \ell_m)$, $m \geq 2$, is the cumulative distribution function for the non-signal channels. (We assume without loss of generality that the signal channel is the first one, i.e., $m=1$.) The sum in (5.2-2) can be expanded as

$$\begin{aligned} &\sum_{p=0}^{M-1} \Pr\{C; n| (z_1 = L_n) \text{ and } (p \text{ non-signal channels} = L_n); L, \underline{\ell}\} \\ &\quad \cdot \Pr\{p \text{ non-signal channels} = L_n | L, \ell_2, \dots, \ell_m\}. \\ &\quad \cdot \Pr\{z_1 = L_n | L, \ell_1\} \\ &= \sum_{p=0}^{M-1} \frac{1}{p+1} \Pr\{p \text{ non-signal channels} = L_n | L, \ell_2, \dots, \ell_M\} \cdot P_{1L}(\ell_1), \end{aligned} \quad (5.2.4a)$$

and we use $P_{1L}(\ell_1) \triangleq Q_0^{L-\ell_1} Q_1^{\ell_1}$. (5.2-4b)

Using a similar notation, the probability of a non-signal channel's being equal to L_n , i.e., $z_m = L_n$ for $m \geq 2$, is

$$P_{2L}(\ell_m) \triangleq q_0^{L-\ell_m} q_1^{\ell_m} = e^{-(L-\ell_m)a_n - \ell_m b_n} \quad (5.2-5)$$

Now, there are $\binom{M-1}{p}$ ways for p of the $M-1$ non-signal channels to be selected as either $z_m = L_n$ or $z_m < L_n$. However, it is necessary to account for the fact that these channels may have different numbers of hops jammed, ℓ_m . Let

$$v_m = \begin{cases} 1 & \text{if } z_m = L_n \\ 0 & \text{if } z_m < L_n \end{cases} \quad (5.2-6)$$

using this indicator variable, and the vector

$$\underline{v} = (v_2, v_3, \dots, v_M), \quad (5.2-7)$$

we can write

$$\begin{aligned} & \Pr\{p \text{ non-signal channels} = L_n | L, \ell_2, \dots, \ell_M\} \\ &= \sum_{\underline{v}} \prod_{m=2}^M \left\{ v_m P_{2L}(\ell_m) + (1-v_m) [1 - P_{2L}(\ell_m)] \right\} \cdot \delta\left(\sum_m v_m, p\right) \end{aligned} \quad (5.2-8a)$$

where

$$\delta(n, p) \triangleq \begin{cases} 1, & p = n \\ 0, & p \neq n. \end{cases} \quad (5.2-8b)$$

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For example, if all the non-signal channels have the same number of jammed hops, $\ell_m = \ell$, then (5.2-8a) is evaluated as

$$\begin{aligned} & [P_{2L}(\ell)]^p [1 - P_{2L}(\ell)]^{m-1-p} \sum_{\underline{v}} \delta(\sum_m v_m, p) \\ &= \binom{M-1}{p} [P_{2L}(\ell)]^p [1 - P_{2L}(\ell)]^{-1-p}. \end{aligned} \quad (5.2-9)$$

Substituting (5.2-8) and (5.2-3) into the error expression results in

$$\begin{aligned} P_s(e; n | L, \underline{\ell}) &= 1 - \int_0^{Ln} dx f_L(x; \ell_1) \prod_{m=2}^M G_L(x; \ell_m) \\ &- P_{1L}(\ell_1) \sum_{p=0}^{M-1} \frac{1}{1+p} \sum_{\underline{v}} \prod_{m=2}^M \{v_m P_{2L}(\ell_m) + (1-v_m) [1 - P_{2L}(\ell_m)]\} \delta(\sum_m v_m, p). \end{aligned} \quad (5.2-10)$$

5.2.1.1 Special case: $L=1$ (one hop/symbol).

For $L=1$, we have

$$f_1(x; \ell_1) = c_1 e^{-c_1(x+2s)} I_0(2c_1 \sqrt{2Sx}); \quad (5.2-11)$$

using

$$c_m = \begin{cases} a & \text{if } \ell_m = 0 \\ b & \text{if } \ell_m = 1; \end{cases} \quad (5.2-12)$$

$$P_{21}(\ell_m) = e^{-c_m n}; \quad (5.2-13)$$

and

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$$\begin{aligned}
 \prod_{m=2}^M G_L(x; \xi_m) &= \prod_{m=2}^M (1 - e^{-c_m x}) \\
 &= (1 - e^{-ax})^{n_0} (1 - e^{-bx})^{M-1-n_0} \\
 &= \sum_{k=0}^{n_0} \sum_{r=0}^{M-1-n_0} \binom{n_0}{k} \binom{M-1-n_0}{r} (-1)^{k+r} e^{-(ka+rb)x}
 \end{aligned} \tag{5.2-14a}$$

where

$$n_0 \triangleq \#(\xi_m = 0, m \geq 2), \tag{5.2-14b}$$

that is, n_0 is the number of unjammed, non-signal channels. Substituting in the error expression (5.2-10) results in

$$\begin{aligned}
 P_S(e; n | 1, \underline{\xi}) &= \sum_{k=0}^{n_0} \sum_{\substack{r=0 \\ k+r > 0}}^{M-1-n_0} \binom{n_0}{k} \binom{M-1-n_0}{r} (-1)^{k+r+1} \\
 &\quad \cdot \int_0^n dx c_1 e^{-2c_1 S - (c_1 + ka + rb)x} I_0(2c_1 \sqrt{Sx}) + Q_{\xi_1} \\
 &\quad - Q_{\xi_1} \sum_{p=0}^{M-1} \frac{1}{1+p} \sum_{\underline{v}} \prod_{m=2}^M \left\{ (2v_m - 1) e^{-c_m v_m} + 1 - v_m \right\} \delta(\sum_m v_m, p).
 \end{aligned} \tag{5.2-15}$$

The integral equals

$$\frac{c_1}{c_1 + ka + rb} \exp \left\{ \frac{-(ka + rb) 2c_1 S}{c_1 + ka + rb} \right\} \left[1 - Q \left(\sqrt{\frac{4c_1^2 S}{c_1 + ka + rb}}, \sqrt{2(c_1 + ka + rb) \eta} \right) \right]. \tag{5.2-16}$$

Also, since $\xi_m = 0$ or 1 when $L = 1$, the last term can be written

$$\begin{aligned}
 & - Q(2\sqrt{c_1 S}, \sqrt{2c_1 n}) \sum_{p=0}^{M-1} \frac{1}{1+p} \sum_{p_0=p_{\min}}^{p_{\max}} \binom{n_0}{p_0} \binom{M-1-n_0}{p-p_0} e^{-p_0(a-b)n-pbn} \\
 & \times (1-e^{-an})^{n_0-p_0} (1-e^{-bn})^{M-1-n_0-(p-p_0)}, \quad (5.2-17)
 \end{aligned}$$

where

$$p_{\min} = \max[0, p-(M-1-n_0)], \quad p_{\max} = \min(p, n_0). \quad (5.2-17)$$

For example, if $M=2$, (5.2-15) becomes, since $n_0 = 0$ or 1 ($n_0 \equiv 1-\ell_2$),

$$\begin{aligned}
 P_b(e; n | 1, \underline{\ell}) &= \ell_2 \cdot \frac{c_1}{c_1+b} \exp\left\{\frac{-2bc_1 S}{c_1+b}\right\} \left[1 - Q\left(\sqrt{\frac{4c_1^2 S}{c_1+b}}, \sqrt{2(c_1+b)n}\right) \right] \\
 &+ (1-\ell_2) \cdot \frac{c_1}{c_1+a} \exp\left\{\frac{-2ac_1 S}{c_1+a}\right\} \left[1 - Q\left(\sqrt{\frac{4c_1^2 S}{c_1+a}}, \sqrt{2(c_1+a)n}\right) \right] \\
 &- Q(2\sqrt{c_1 S}, \sqrt{2c_1 n}) \left\{ \ell_2 \cdot \left[1 - e^{-bn} + \frac{1}{2} e^{-bn} \right] \right. \\
 &\quad \left. + (1-\ell_2) \cdot \left[1 - e^{-an} + \frac{1}{2} e^{-an} \right] \right\} \\
 &+ Q(2\sqrt{c_1 S}, \sqrt{2c_1 n}) \quad (5.2-18)
 \end{aligned}$$

5.2.1.2 Special case: $\underline{\ell} = \underline{0}$ (no jamming).

For this case, we substitute $\underline{\ell} = \underline{0}$ in (5.2-10) to get

$$P_s(e; n | L, \underline{Q}) = 1 - \int_0^{Ln} dx f_L(x; 0) [G_L(x; 0)]^{M-1} - Q_0^L \sum_{p=0}^{M-1} \frac{1}{1+p} \binom{M-1}{p} e^{-Lpan} (1 - e^{-Lan})^{M-1-p} \quad (5.2-19a)$$

$$= \sum_{k=1}^{M-1} \binom{M-1}{k} (-1)^{k+1} \int_0^{Ln} dx f_L(x; 0) [1 - G_L(x; 0)]^k + Q_0^L - Q_0^L \cdot \frac{e^{Lan}}{M} \left[1 - (1 - e^{-Lan})^M \right]. \quad (5.2-19b)$$

Now, $1 - G_L(x; 0)$ in the integral has the form (see Tables 5.1-3 and 5.1-4)

$$1 - G_L(x; 0) = \begin{cases} 1, & x < 0; \\ e^{-ax} h_r(x - rn), & rn \leq x < (r+1)n, \quad r=0, 1, 2, \dots, L-1; \\ 0, & x > Ln; \end{cases} \quad (5.2-20)$$

where $h_r(x)$ is an $(L-1)$ degree polynomial. Using this form, the integral in (5.2-19) becomes

$$\sum_{r=0}^{L-1} \int_{rn}^{(r+1)n} dx f_L(x; 0) [h_r(x - rn)]^k e^{-kax} = \sum_{r=0}^{L-1} \int_0^n dx f_L(x + rn; 0) [h_r(x)]^k e^{-ka(x+rn)}. \quad (5.2-21)$$

Noting also from Tables 5.1-5 and 5.1-6 that the signal channel pdf can be written

$$f_L(x;0) = e^{-ax} v_r(x-m), r=0,1,\dots,L-1, \quad (5.2-22)$$

we further manipulate (5.2-21) to obtain

$$\sum_{r=0}^{L-1} e^{-(k+1)ra} \int_0^n dx e^{-(k+1)ax} v_r(x) [h_r(x)]^k. \quad (5.2-23)$$

For example, if $L=1$, then $h_0(x) \equiv 1$ and $v_0(x) = a e^{-2aS} I_0(2a\sqrt{2Sx})$, giving for (5.2-23) the value

$$\frac{1}{1+k} \exp\left\{-\frac{2kaS}{k+1}\right\} \left[1 - Q\left(\sqrt{\frac{4aS}{1+k}}, \sqrt{2a(k+1)n}\right)\right] \quad (5.2-24)$$

and

$$P_S(e;n|1,0) = \sum_{k=1}^{M-1} \binom{M-1}{k} \frac{(-1)^{k+1}}{k+1} \exp\left\{\frac{-2kaS}{1+k}\right\} \left[1 - Q\left(\sqrt{\frac{4aS}{1+k}}, \sqrt{2a(k+1)n}\right)\right] + Q_0 - Q_0 \frac{e^{an}}{M} \left[1 - (1 - e^{-an})^M\right]. \quad (5.2-25)$$

5.2.2 Total Probability of Error.

For a given number of hops/symbol, L , the total symbol probability of error is

$$P_S(e;n,L) = \sum_{\underline{\ell}} \Pr\{\underline{\ell}\} P_S(e;n|L,\underline{\ell}); \quad (5.2-26)$$

the bit error probability is

$$P_b(e;n,L) = \frac{M/2}{M-1} P_s(e;n,L). \quad (5.2-27)$$

5.2.2.1 Choice of clipping threshold.

The procedure we have adopted for choosing η , the clipping threshold, is the following: choose η to minimize the error probability when there is no jamming. That is,

$$\eta^* : \min_{\eta} P_s(e;\eta|L,0). \quad (5.2-28)$$

Differentiation of the error expression (5.2-19a) gives an equation for the optimum η thus defined. This equation may be written

$$\begin{aligned} & - f_L(L\eta;0) (1 - e^{-La\eta})^{M-1} \\ & + \sum_{r=1}^{L-1} \int_0^{\eta} dx \frac{\partial}{\partial \eta} \left\{ f_L(x+r\eta;0) [G_L(x+r\eta;0)]^{M-1} \right\} \\ & + L Q_0^{L-1} f_1(\eta;0) \frac{e^{La\eta}}{M} [1 - (1 - e^{-La\eta})^M] \\ & - Q_0^L \frac{La e^{La\eta}}{M} \left\{ 1 - (1 - e^{-La\eta})^M - Me^{-La\eta} (1 - e^{-La\eta})^{M-1} \right\} \end{aligned} \quad (5.2-29)$$

For $L=1$, the second term is zero and the equation can be put in the form

$$\frac{e^{a\eta}}{M} [Q_0 a - f_1(\eta;0)] \left\{ [1 + (M-1)e^{-a\eta}] (1 - e^{-a\eta})^{M-1} - 1 \right\}; \quad (5.2-30)$$

this partial derivative with respect to η is negative, indicating the error decreases as η increases, indefinitely*. Thus for $L=1$ the optimum threshold is infinite (no clipping):

$$\eta^*(L=1) \rightarrow \infty. \quad (5.2-31)$$

For $L>1$, it is not feasible to find the optimum threshold by differentiation; it must be done numerically.

5.2.2.2 Total error for $L=1$.

Since the optimum threshold for $L=1$ is $\eta^* \rightarrow \infty$, we may express the total error probability by using (5.2-15) to obtain

$$P_b(e; L=1) = \frac{M/2}{M-1} \sum_{\underline{\lambda}} \Pr\{\underline{\lambda}\} \sum_{k=0}^{n_0} \sum_{r=0}^{M-1-n_0} \binom{n_0}{k} \binom{M-1-n_0}{r} (-1)^{k+r+1} \cdot \frac{c_1}{c_1+ka+rb} \exp\left\{\frac{-(ka+rb)2c_1S}{c_1+ka+rb}\right\}, \quad (5.2-32a)$$

$$\text{where } c_1 = (1-\lambda_1)a + \lambda_1 b \quad (5.2-32b)$$

$$\text{and } n_0 = M-1 - \sum_{m=2}^M \lambda_m. \quad (5.2-32c)$$

This, of course, gives exactly the same performance as the other receiver processing schemes for $L=1$.

* The last factor in (5.2-30) can be recognized as the quantity

$$- \sum_{m=2}^M \binom{M}{m} (1-e^{-a\eta})^{n-m} e^{-m a \eta} < 0. \quad \text{The second factor is always positive since } aQ_0 = f_1(\eta; 0) + \exp\{-a(2S+\eta)\} \sum_{k>1} a(2S/\eta)^k I_k(2a\sqrt{S\eta}).$$

5.3 NUMERICAL RESULTS

In this section we present some numerical results for the clipper receiver's performance. These results are less voluminous than those obtained for other receivers because of the extremely long computer run times for the clipper equations.*

Two stages of computation are required for the clipper receiver. First, the optimum clipping level in the absence of jamming, η_0/σ_N^2 , must be found by a numerical search. Then this value must be used in computing the jammed performance. Whenever L, M, or E_b/N_0 changes the optimum clipping level must be recomputed.

The many numerical integrations required to evaluate (5.2-10) using the forms for $f_L(x;\ell_1)$ from Tables 5.1-3 and 5.1-4 and for $G_L(x;\ell)$ from Table 5.1-2 result in very lengthy computations. Consider, for example, the case of $M=4$, $L=2$, in which the numerical integrations which are required have the structure

$$1 - \left[\int f_1 g + \int (f_2 + \int f_3) g \right], \quad \ell_1 = 0 \text{ or } \ell_1 = 2 \quad (5.3-1a)$$

$$1 - \left[\int f_1 \int f_2 g + \int (f_3 + \int f_4) g \right], \quad \ell_1 = 1 \quad (5.3-1b)$$

where g is a function of ℓ_2 . Each conditional error probability involves one or two double integrations which must be evaluated numerically to sufficient accuracy as to leave several significant digits after subtracting from 1. This subtractive cancellation problem is especially severe for high E_b/N_J when P(e) is small.

For $L=3$ the situation is even worse, for the numerical integrations take the forms

*It has been noted that for $L=1$ and any M value, the clipper receiver with optimum threshold is merely a conventional receiver, since that threshold is infinite for $L>1$. Thus the results computed previously for $L=1$ apply to this Section as well.

$$1 - \left\{ \int f_1 g + \int [f_1 + \int f_2 + \int (f_3 \int f_4)] g + \int [f_5 + \int f_6 + \int (f_7 \int f_8)] g \right\},$$

$$\ell_1 = 0 \text{ or } \ell_1 = 3 \quad (5.3-2a)$$

$$1 - \left\{ \int (\int f_1) g + \int [f_2 + \int f_3 + \int f_4 + \int f_5 (\int f_6)] g \right. \\ \left. + \int [f_7 + \int f_8 + \int f_9 + \int f_{10} (\int f_{11})] g \right\}, \ell_1 = 1 \text{ or } \ell_1 = 2 \quad (5.3-2b)$$

which results in a worst-case of 2 one-dimensional integrations, 5 two-dimensional integrations, and 2 three-dimensional integrations to be performed numerically. The inner-most integrals must be evaluated to very high precision in order to evaluate the outer integrals to sufficient precision so as to reduce subtractive cancellation to acceptable levels. The result is a very slow computer program.

Under these conditions, the available computational facilities (a PDP-11/44 minicomputer) restricted the number of performance curves we were able to generate.

5.3.1 The Optimum Threshold Setting.

The optimum clipping threshold η_0 is defined as the level η which minimizes the bit error probability in the absence of jamming. This is accomplished by the first part of the computer programs for calculating the performance in partial-band noise jamming. The thresholds are normalized by the thermal noise density; thus we actually find $\eta_0/2\sigma_N^2$. The optimum thresholds found by the computer programs given in appendices H (for M=2, L=2), I (for M=4, L=2), and J (for M=2 or 4, L=3) are given in Table 5.3-1.

TABLE 5.3-1

OPTIMUM NORMALIZED CLIPPING THRESHOLD $\eta_0/2\sigma_N^2$

L	M		
	2 ($E_b/N_0 = 13.35247$ dB)	4 ($E_b/N_0 = 10.60657$ dB)	8 ($E_b/N_0 = 9.09401$ dB)
2	10.20	10.55	10.89
3	7.91	8.15	---

We note that in terms of signal power $\eta_0 = (1.89S, 1.83S, 1.79S)$ for $L=2$ and $M=(2,4,8)$; $\eta_0 = (2.19S, 2.13S)$ for $L=3$ and $M=(2,4)$. The threshold is almost a function only of S , L , and M .

5.3.2 Probability of Bit Error .

For $M=2$ and $L=2$, the computations using the program given in Appendix H were sufficiently rapid to permit obtaining a full set of curves for jamming fractions from $\gamma = 0.001$ through $\gamma = 1.0$, as shown in Figure 5.3-1.

For $M=4$ and $L=2$, the computations were much slower, due to the increased number of jamming events and the need to compute products of the function $G_L(x)$. Therefore, the program in Appendix I was used to search for the optimum value of γ for each value of E_b/N_j . To aid the speed of the search, we used the a priori knowledge that $\gamma_{opt} = 1/N$ where N is the number of hopping slots when E_b/N_j is very high, and that γ_{opt} increases as E_b/N_j decreases. Thus the computations started at $E_b/N_j = 50$ dB and decreased (in rather large steps to conserve computer time) to 0 dB. The result is the curve of $P_b(e)$ vs. E_b/N_j in worst-case partial-band noise jamming as shown for $M=4$ in Figure 5.3-2. For comparison, the envelope of the curves from Figure 5.3-1 is shown in Figure 5.3-2. We see that $M=4$ FH/RMFSK is about 2 dB better than $M=2$ FH/RMFSK in strong jamming.

Selected runs for $M=8$ and $L=2$ with $E_b/N_0=9.09$ dB were made in order to examine the dependence of the worst-case jamming performance upon M . These runs yielded the threshold shown in Table 5.3-1 and the following points for $\gamma=0.01$: $[E_b/N_j, P(e)] = [15, 8.963(-4)], [20, 4.680(-4)], [22.5, 2.604(-4)], [25, 1.286(-4)], [\infty, 3.91(-5)]$. From these points, a curve for $\gamma=0.01$ was constructed, and the inflection point was taken to be a point on the worst-case jamming curve. This point, estimated as the 22.5 dB point given above, is shown on Figure 5.3-2.

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Although the program given in Appendix J contains code to compute $P_b(e)$ as well as $n_0/2\sigma_N^2$, excessive run time (nearly 8 hours to obtain just $n_0/2\sigma_N^2$), prevented us from allowing it to run to completion to obtain performance curves for the case $L=3$ hops per symbol.

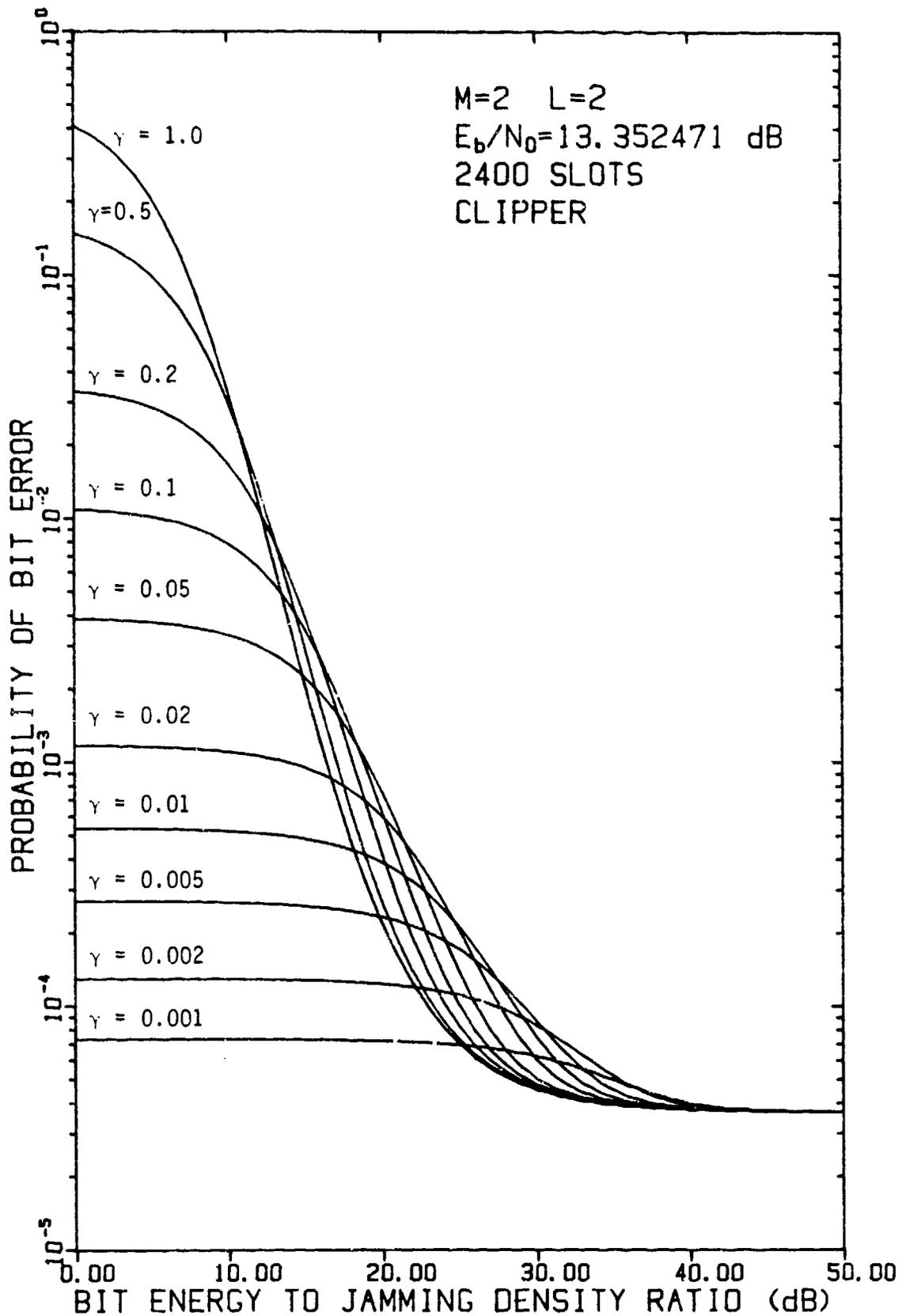


FIGURE 5.3-1 PERFORMANCE OF CLIPPER RECEIVER FOR FH/RMFSK WITH $M=2$, $L=2$ HOPS/SYMBOL, AND $E_b/N_0 = 13.35247 \text{ dB}$ (FOR $P_b(e) = 10^{-5}$ IN THE ABSENCE OF JAMMING WHEN $L=1$)

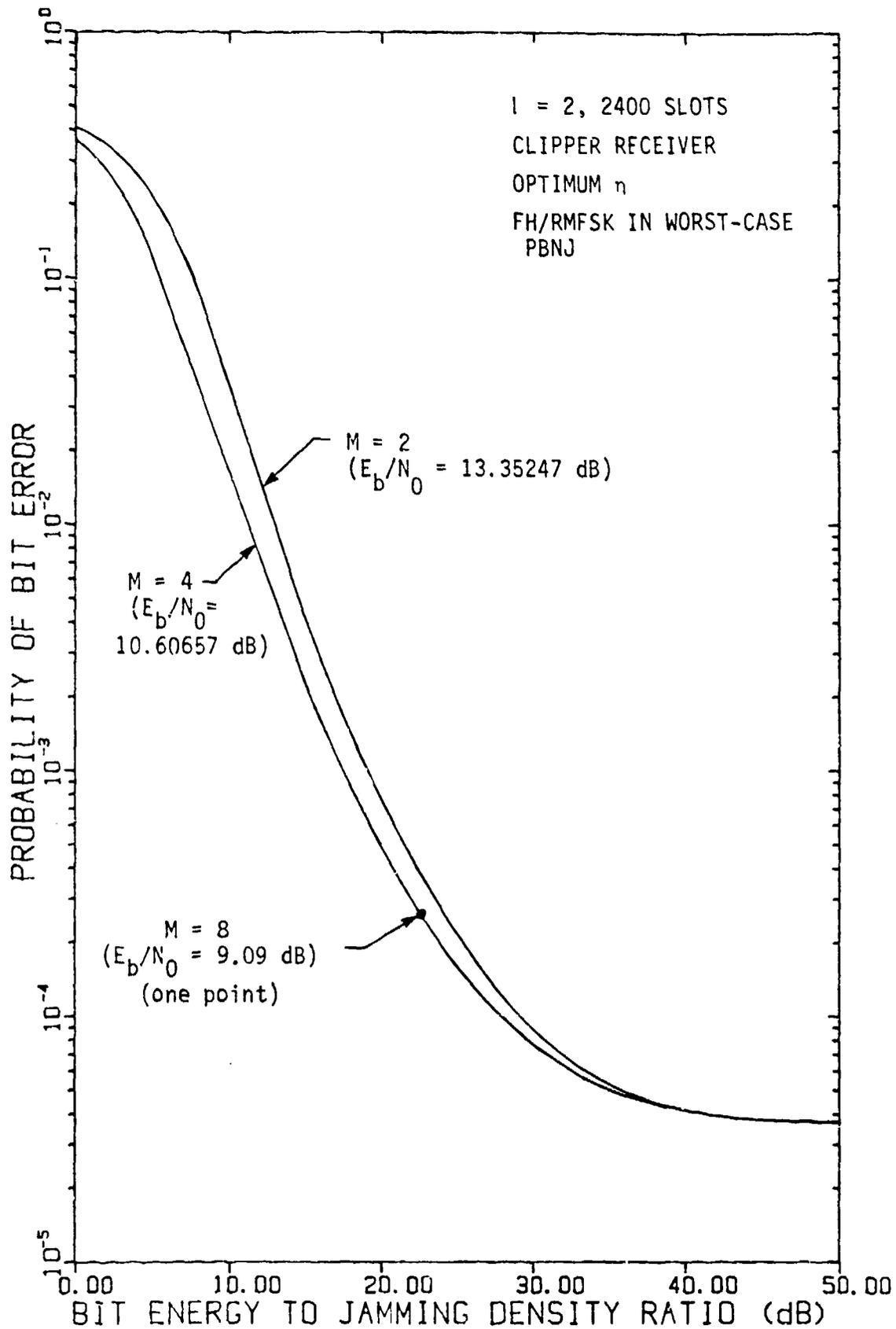


FIGURE 5.3-2 PERFORMANCE OF CLIPPER RECEIVER FOR FH/RMFSK WHEN $L=2$ HOPS/
 SYMBOL WITH M AS A PARAMETER AND E_b/N_0 CORRESPONDING TO
 $P_b(e) = 10^{-5}$ IN THE ABSENCE OF JAMMING (WHEN $L=1$ HOP/SYMBOL)

6.0 FH/RMFSK PERFORMANCE USING SQUARE-LAW SELF-NORMALIZING RECEIVER

The ECCM weighting schemes which make the (ideal) AGC and clipper soft-decision receivers for FH/MFSK and FH/RMFSK work depend upon a priori knowledge of system parameters, or else real-time measurements. (The feasibility of these measurements is discussed in a later section.) It is evident that the clipper strategy, which requires setting an SNR-dependent threshold, would be easier to implement than the AGC receiver, which requires detection of which hops are jammed and knowledge or measurement of thermal noise and jamming noise levels. Meanwhile we have seen that the hard-decision receiver accomplishes a form of ECCM protection, much in the manner of the clipper receiver - the jammed hops are prevented from dominating the decision. If there is sufficient SNR, one might well choose then to employ the hard-decision scheme, since it does not require any a priori knowledge or measurements.

In this section we consider soft-decision weighting schemes which are not predicated on using signal or noise parameters. In particular, we find the FH/RMFSK performance of a "self-normalizing" receiver in partial-band noise jamming.

6.1 THE SELF-NORMALIZATION SCHEME

The general FH/RMFSK soft-decision receiver shown in Figure 2.2-1 is rendered what we call the "self-normalizing" (SNORM) receiver by use of the weighting function

$$z_{mk} = f(x_{mk}) = \frac{x_{mk}^2}{x_{1k}^2 + x_{2k}^2 + \dots + x_{Mk}^2} \quad (6.1-1)$$

That is, on each hop (indexed by k , $k=1,2,\dots,L$), the squared envelope samples in each channel ($m=1,2,\dots,M$) are normalized (divided) by their sum. In this manner, hops which are jammed in one or more MFSK slots can be expected to be weighted less than unjammed hops.

Thus the decision variables for the SNORM receiver are

$$\begin{aligned} z_m &= \sum_{k=1}^L z_{mk} \\ &= \sum_{k=1}^L w_k x_{mk}^2 \end{aligned} \quad (6.1-2a)$$

where

$$w_k = \left(\sum_{m=1}^M x_{mk}^2 \right)^{-1} \quad (6.1-2b)$$

6.1.1 Single-hop Distribution of Decision Variables.

With the FH/RMFSK hopping scheme, any, none, or all of the channels can be jammed on a particular hop. Using $u_{mk} \equiv x_{mk}^2$ for the square-law envelope samples and

$$\begin{aligned} a &\equiv 1/2\sigma_N^2 \\ b &\equiv 1/2\sigma_J^2, \end{aligned} \quad (6.1-3)$$

for a general one-hop jamming event, we can write (assuming the signal is in channel 1)

$$p_{u_1}(\alpha; c_1) = c_1 e^{-c_1 \alpha - \rho_1} I_0(2\sqrt{\rho_1 c_1 \alpha}), \quad \alpha > 0; \quad (6.1-4a)$$

and

$$p_{u_m}(\alpha; c_m) = c_m e^{-c_m \alpha}, \quad m=2, \dots, M; \alpha > 0; \quad (6.1-4b)$$

where

$$c_m = \begin{cases} a, & \text{channel not jammed} \\ b, & \text{channel jammed, } m=1, 2, \dots, M. \end{cases} \quad (6.1-4c)$$

Thus the joint pdf for the square-law envelope detector samples is, conditioned on the jamming,

$$p_{\underline{u}}(\alpha_1, \dots, \alpha_M | c_1, \dots, c_M) = c_1 c_2 \dots c_M \exp \left\{ -\rho_1 - \sum_{m=1}^M c_m \alpha_m \right\} I_0(2\sqrt{\rho_1 c_1 \alpha_1}), \quad \alpha_m \geq 0. \quad (6.1-5)$$

By a change of variables,

$$\begin{aligned} u_1 &= \xi z_1 \\ u_1 + u_2 &= \xi(z_1 + z_2) \\ \sum_{i=1}^k u_i &= \xi \sum_{i=1}^k z_i \end{aligned} \quad (6.1-6)$$

$$u_1 + u_2 + \dots + u_M = \xi,$$

we can express the joint pdf of $\{z_1, z_2, \dots, z_{M-1}\}$ by, using $\underline{c} \equiv (c_1, c_2, \dots, c_M)$,

$$p_{\underline{z}}(\alpha_1, \alpha_2, \dots, \alpha_{M-1} | \underline{c}) = \int_0^\infty d\xi \xi^{M-1} p_{\underline{u}} \left[\xi \alpha_1, \xi \alpha_2, \dots, \xi \alpha_{M-1}, \xi - \sum_{m=1}^{M-1} \xi \alpha_m \right]$$

$$\begin{aligned}
 &= c_1 c_2 \dots c_M e^{-\rho_1} \int_0^\infty d\xi \xi^{M-1} \exp \left\{ -\xi \left[c_M + \sum_{m=1}^{M-1} (c_m - c_M) \alpha_m \right] \right\} \\
 &\quad \times I_0(2\sqrt{\rho_1 c_1 \xi \alpha_1}) \\
 &= \frac{(M-1)! e^{-\rho_1} \prod_m c_m}{\left[c_M + \sum_{m=1}^{M-1} (c_m - c_M) \alpha_m \right]^M} {}_1F_1 \left[M; 1; \frac{c_1 \rho_1 \alpha_1}{c_M + \sum_{m=1}^{M-1} (c_m - c_M) \alpha_m} \right]. \quad (6.1-7)
 \end{aligned}$$

In this development we used equations 6.643.2 and 9.220.2 from [3]; ${}_1F_1(a; b, x)$ is the confluent hypergeometric function.

Note that α_M does not appear in (6.1-7). This occurrence is due to the fact that there are now only $M-1$ dependent random variables; the value of the M th channel variable is completely determined by the others. This fact can be made explicit by writing (using all variables)

$$p_{\underline{z}}(\alpha_1, \alpha_2, \dots, \alpha_M | \underline{c}) = p_{\underline{z}}(\alpha_1, \alpha_2, \dots, \alpha_{M-1} | \underline{c}) \delta \left(\sum_{m=1}^M \alpha_m - 1 \right). \quad (6.1-8)$$

Also, we note that the domains of these variables are interdependent:

$$\begin{aligned}
 0 &\leq z_m \leq 1 \\
 0 &\leq z_i + z_j \leq 1 \quad (\text{all pairs}) \\
 0 &\leq z_i + z_j + z_k \leq 1 \quad (\text{all triples}) \\
 &\vdots \\
 0 &\leq z_1 + z_2 + \dots + z_{M-1} \leq 1 \\
 \sum_{m=1}^M z_m &\equiv 1.
 \end{aligned} \quad (6.1-9)$$

This interdependence of finite domains makes analysis and computation difficult, as will be seen below.

6.1.2 Alternate Forms.

By using the identity

$$\begin{aligned} {}_1F_1(M;1;x) &= e^x {}_1F_1(1-M;1;-x) \\ &= e^x \mathcal{L}_{M-1}(-x), \end{aligned} \quad (6.1-10)$$

where $\mathcal{L}_n(x)$ is the Laguerre polynomial, we realize that the joint pdf given in (6.1-7) has the form of an exponential times an (M-1)-degree polynomial in $x(\underline{\alpha})$, divided by an M-degree polynomial in $y(\underline{\alpha})$:

$$p_{\underline{z}}(\underline{\alpha}|\underline{c}) = \frac{\text{const} \cdot e^{x(\underline{\alpha})} \mathcal{L}_{M-1}[-x(\underline{\alpha})]}{[y(\underline{\alpha})]^M} \delta\left(\sum_m \alpha_m - 1\right) \quad (6.1-11a)$$

where

$$x(\underline{\alpha}) = c_1 \rho_1 \alpha_1 / y(\underline{\alpha}) \quad (6.1-11b)$$

$$y(\underline{\alpha}) = c_M + \sum_{m=1}^{M-1} (c_m - c_M) \alpha_m \quad (6.1-11c)$$

and

$$\text{const} = (M-1)! e^{-\rho_1} \prod_{m=1}^M c_m. \quad (6.1-11d)$$

A somewhat simpler form results from recognizing that [3, equation 8.970.1]

$$\mathcal{L}_{M-1}(x) = \frac{1}{(M-1)!} e^x \frac{d^{M-1}}{dx^{M-1}} \left[e^{-x} x^{M-1} \right]. \quad (6.1-12)$$

Applying this relation results in

$$p_{\underline{z}}(\underline{\alpha}|\underline{c}) = \frac{\prod_m c_m}{\left[\sum_m c_m \alpha_m \right]^M} e^{-\rho_1} \frac{\partial^{M-1}}{\partial \rho_1^{M-1}} \left\{ \rho_1^{M-1} e^{\rho_1 X(\underline{\alpha})} \right\} \delta\left(\sum_m \alpha_m - 1\right). \quad (6.1-13)$$

6.1.2.1 Special case: M=2 (binary).

The various general expressions for the joint pdf reduce to the following ones for M=2:

$$p_{\underline{z}}(\alpha_1, \alpha_2 | c_1, c_2) = \frac{c_1 c_2 e^{-\rho_1}}{(c_1 \alpha_1 + c_2 \alpha_2)^2} {}_1F_1(2; 1; \frac{c_1 \rho_1 \alpha_1}{c_1 \alpha_1 + c_2 \alpha_2}) \delta(\alpha_1 + \alpha_2 - 1) \quad (6.1-14a)$$

$$= \frac{c_1 c_2 e^{-\rho_1}}{[c_2 + (c_1 - c_2) \alpha_1]^2} \exp\left\{ \frac{c_1 \rho_1 \alpha_1}{c_2 + (c_1 - c_2) \alpha_1} \right\} \left[1 + \frac{c_1 \rho_1 \alpha_1}{c_2 + (c_1 - c_2) \alpha_1} \right] \delta(\alpha_1 + \alpha_2 - 1) \quad (6.1-14b)$$

$$= \frac{c_1 c_2 e^{-\rho_1}}{[c_2 + (c_1 - c_2) \alpha_1]^2} \frac{\partial}{\partial \rho_1} \left[\rho_1 \exp\left\{ \frac{c_1 \rho_1 \alpha_1}{c_2 + (c_1 - c_2) \alpha_1} \right\} \right] \delta(\alpha_1 + \alpha_2 - 1) \quad (6.1-14c)$$

with

$$0 \leq \alpha_1 \leq 1, \quad 0 \leq \alpha_2 = 1 - \alpha_1 \leq 1. \quad (6.1-14d)$$

6.1.2.2 Special case: no jamming.

For no jamming, $c_1 = c_2 = \dots = c_M$ (also true for all channels jammed), the general pdf reduces to

$$p_{\underline{z}}(\underline{\alpha} | c_1 = c_2 = \dots = c_M) = (M-1)! e^{-\rho_1} {}_1F_1(M; 1; \rho_1 \alpha_1) \delta\left(\sum_m \alpha_m - 1\right) \quad (6.1-15a)$$

$$= (M-1)! e^{-\rho_1 + \rho_1 \alpha_1} \mathcal{L}_{M-1}(-\rho_1 \alpha_1) \delta\left(\sum_m \alpha_m - 1\right) \quad (6.1-15b)$$

$$= e^{-\rho_1} \frac{\partial^{M-1}}{\partial \rho_1^{M-1}} \left\{ \rho_1^{M-1} e^{\rho_1 \alpha_1} \right\} \delta\left(\sum_m \alpha_m - 1\right). \quad (6.1-15c)$$

6.1.3 Conditional pdf's for M=2.

We now show the explicit expressions for the decision variable for the binary case; there is only one decision variable $z \equiv z_1$ since $z_2 \equiv 1 - z_1$.

6.1.3.1 Single hop/bit case (L=1).

As far as computation of error probabilities is concerned, the L=1 case of the SNORM receiver pdf's is not needed since the normalization does not affect the outcome of the decision; we know in advance that the result will be the same as if no normalization were employed. However, to go on to the L=2 case, we need the L=1 pdf's.

Using $K \triangleq \sigma_T^2 / \sigma_N^2$ as in previous analyses, the pdf's conditioned on the possible jamming events $\underline{v} = (v_1, v_2)$ are as follows:

$$p_1[z | \underline{v} = (0,0)] = e^{-\rho_N} {}_1F_1(2;1; \rho_N z) \quad (6.1-16a)$$

$$= e^{-\rho_N + \rho_N z} (1 + \rho_N z) . \quad (6.1-16b)$$

$$p_1[z | \underline{v} = (0,1)] = \frac{K e^{-\rho_N}}{[1 + (K-1)z]^2} {}_1F_1\left(2;1; \frac{K \rho_N z}{1 + (K-1)z}\right) \quad (6.1-17a)$$

$$= \frac{K e^{-\rho_N}}{[1 + (K-1)z]^2} \exp\left[\frac{K \rho_N z}{1 + (K-1)z}\right] \left(1 + \frac{K \rho_N z}{1 + (K-1)z}\right) \quad (6.1-17b)$$

$$p_1[z | \underline{v} = (1,0)] = \frac{K e^{-\rho_T}}{[K - (K-1)z]^2} {}_1F_1\left(2;1; \frac{\rho_T z}{K - (K-1)z}\right) \quad (6.1-18a)$$

$$= \frac{K e^{-\rho_T}}{[K - (K-1)z]^2} \exp\left[\frac{\rho_T z}{K - (K-1)z}\right] \left(1 + \frac{\rho_T z}{K - (K-1)z}\right) \quad (6.1-18b)$$

$$p_1[z | \underline{v} = (1,1)] = e^{-\rho_T} {}_1F_1(2;1; \rho_T z) \quad (6.1-19a)$$

$$= e^{-\rho_T + \rho_T z} (1 + \rho_T z) . \quad (6.1-19b)$$

For all of these expressions, the domain of z is $0 \leq z \leq 1$. Note that when there is no signal ($\rho_N = \rho_T = 0$), the variable z is uniformly distributed when both channels have the same noise power.

6.1.3.2 Two hops/bit case ($L=2$).

To obtain the pdf for $L=2$, it is necessary to convolve the expressions (6.1-16) to (6.1-19) with each other for the jamming events $\underline{z} = \underline{v}_1 + \underline{v}_2$. The general form of the convolution is

$$p_2(z|\underline{z} = \underline{v}_1 + \underline{v}_2) = \int_{\max(0, z-1)}^{\min(1, z)} dv p_1(z-v|\underline{v}_1) p_1(v|\underline{v}_2), 0 \leq z \leq 2. \quad (6.1-20)$$

There are ten distinguishable jamming events, two for $\underline{z}=(1,1)$ and one each for other \underline{z} . For three of these ten cases, the convolution shown in (6.1-20) can be performed analytically without too much difficulty; the cases are the ones in which both channels are jammed or not jammed on a given hop: $\underline{z}=(0,0)$, $(1,1)^*$, $(2,2)$. For these cases, the pdf for one hop can be written

$$p_1(z) = e^{-\rho_1} \frac{\partial}{\partial \rho_1} \rho_1 e^{\rho_1 z}, 0 \leq z \leq 1 \quad (6.1-21)$$

The convolution then takes the form

$$\begin{aligned} p_2(z|\rho_1, \rho_2) &= e^{-\rho_1 - \rho_2} \frac{\partial^2}{\partial \rho_1 \partial \rho_2} \rho_1 \rho_2 \int_{\max(0, z-1)}^{\min(1, z)} dv e^{\rho_1(z-v) + \rho_2 v} \\ &= e^{-\rho_1 - \rho_2} \frac{\partial^2}{\partial \rho_1 \partial \rho_2} \frac{\rho_2 \rho_1}{\rho_2 \rho_1} \cdot \begin{cases} e^{\rho_2 z} - e^{\rho_1 z}, & 0 \leq z \leq 1 \\ e^{\rho_2 + \rho_1(z-1)} - e^{\rho_1 + \rho_2(z-1)}, & 1 \leq z \leq 2. \end{cases} \end{aligned} \quad (6.1-22)$$

*i.e., the case of $\underline{z} = (1,1)$ where $\underline{v}_1 = (1,1)$ and $\underline{v}_2 = (0,0)$ or vice versa.

Carrying out the partial differentiations results in

$$p_2(z|\rho_1, \rho_2) = \frac{e^{-\rho_1 - \rho_2}}{(\rho_1 - \rho_2)^3} \left\{ e^{\rho_1 z} \left[-2\rho_1\rho_2 + \rho_1^2(\rho_1 - \rho_2)z \right] + e^{\rho_2 z} \left[2\rho_1\rho_2 + \rho_2^2(\rho_1 - \rho_2)z \right] \right\}, \rho_1 \neq \rho_2, 0 \leq z \leq 1; \quad (6.1-23a)$$

$$= e^{-2\rho + \rho z} z(1 + \rho z + \rho^2 z^2/6), \rho_1 = \rho_2 = \rho, 0 < z < 1; \quad (6.1-23b)$$

$$= \frac{1}{(\rho_1 - \rho_2)^3} \left\{ e^{\rho_2(z-2)} \left[-2\rho_1\rho_2 + \rho_1^2(\rho_1 - \rho_2) + \rho_2(\rho_1 - \rho_2)(\rho_1^2 - \rho_1\rho_2 - \rho_2)(z-1) \right] + e^{\rho_1(z-2)} \left[2\rho_1\rho_2 + \rho_2^2(\rho_1 - \rho_2) + \rho_1(\rho_1 - \rho_2)(\rho_2^2 - \rho_1\rho_2 - \rho_1)(z-1) \right] \right\} \quad 1 < z \leq 2, \rho_1 \neq \rho_2 \quad (6.1-23c)$$

$$= e^{-2\rho + \rho z} \left[1 + \rho + \rho^2/6 - (1 - \rho^2/2)(z-1) - \rho(1 + \rho/2)(z-1)^2 - (\rho^2/6)(z-1)^3 \right], \rho_1 = \rho_2 = \rho, 1 < z \leq 2. \quad (6.1-23d)$$

This expression is applied to the pertinent jamming events using the following table:

k_1	ρ_1	ρ_2
0	ρ_N	ρ_N
1	ρ_N	ρ_T
2	ρ_T	ρ_T

(6.1-24)

The other seven cases must be handled by numerical convolution. (We have found an analytical expression, but it is no easier to compute than the convolutions.)

6.1.4 Conditional pdf's for M=4.

The system analysis for M>2 becomes very difficult, as we now demonstrate for M=4.

6.1.4.1 Single hop/symbol case (L=1).

For M=4 and L=1 there are sixteen possible jamming events, described by the vector $\underline{v} = (v_1, v_2, v_3, v_4)$, where $v_m = 1$ if the mth symbol frequency slot is jammed, and $v_m = 0$ if not. These events give rise to the conditional pdf

$$p(\underline{z}|\underline{v}) = \frac{6\mu_1}{[y(\underline{z})]^4} e^{-\rho_1 + \mu_2 \rho_1 z_1 / y(\underline{z})} \mathcal{L}_3[-\mu_2 \rho_1 z_1 / y(\underline{z})], \quad (6.1-25a)$$

where

$$\mathcal{L}_3(-u) = 1 + 3u + \frac{3}{2} u^2 + \frac{1}{6} u^3 \quad (6.1-25b)$$

and the parameters μ_1, μ_2 , and ρ_1 and the polynomials $y(\underline{z})$ are listed in Table 6.1-1. Since $z_4 \equiv 1 - z_1 - z_2 - z_3$, it does not appear in the pdf. We note that z_1 always appears in the conditional pdf, while z_2 and z_3 may or may not appear.

It is understood that the domain of values for the variables is $(z_1, z_2, z_3) \in \Omega_{4,1}$, where $\Omega_{4,1}$ is the volume

$$\Omega_{4,1}: \begin{cases} 0 \leq z_i \leq 1, \quad i = 1, 2, 3; \\ 0 \leq z_i + z_j \leq 1, \quad \text{all pairs}; \\ 0 \leq z_1 + z_2 + z_3 \leq 1. \end{cases} \quad (6.1-26)$$

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TABLE 6.1-1

L=1 PROBABILITY DENSITIES FOR M=4

	$v_1 v_2 v_3 v_4$	ρ_1	μ_1	μ_2	$y(z)$
A	0 0 0 0	ρ_N	1	1	1
B	0 0 0 1		K^3		$1 + (K-1)(z_1+z_2+z_3)$
C	0 0 1 0		K^3		$K - (K-1)z_3$
D	0 0 1 1		K^2		$1 + (K-1)(z_1+z_2)$
E	0 1 0 0		K^3	K	$K - (K-1)z_2$
F	0 1 0 1		K^2		$1 + (K-1)(z_1+z_3)$
G	0 1 1 0		K^2		$K - (K-1)(z_2+z_3)$
H	0 1 1 1		K		$1 + (K-1)z_1$
I	1 0 0 0	ρ_T	K^3		$K - (K-1)z_1$
J	1 0 0 1		K^2		$1 + (K-1)(z_2+z_3)$
K	1 0 1 0		K^2		$K - (K-1)(z_1+z_3)$
L	1 0 1 1		K		$1 + (K-1)z_2$
M	1 1 0 0		K^2	1	$K - (K-1)(z_1+z_2)$
N	1 1 0 1		K		$1 + (K-1)z_3$
O	1 1 1 0		K		$K - (K-1)(z_1+z_2+z_3)$
P	1 1 1 1		1		1

Form:

$$\rho_1(\underline{z}|\underline{v}) = \frac{6\mu_1}{[y(\underline{z})]^4} e^{-\rho_1 + \mu_2 \rho_1 z_1 / y(\underline{z})} \mathcal{L}_3(-\mu_2 \rho_1 z_1 / y(\underline{z}))$$

$$\mathcal{L}_3(-u) = 1 + 3u + \frac{3}{2}u^2 + \frac{1}{6}u^3$$

$$\underline{z} = (z_1, z_2, z_3) \in \Omega_{4,1}$$

$$z_4 = 1 - z_1 - z_2 - z_3$$

The domain $\Omega_{4,1}$ may also be described as the volume included by the planes $z_1=0$, $z_2=0$, $z_3=0$, and $z_1+z_2+z_3=1$, as illustrated by Figure 6.1-1. It is obvious from the mutual constraints among the variables that they are statistically dependent.

6.1.4.2 Two hop/symbol case (L=2) .

Since the four SNORM variables are dependent, we cannot analyze the M=4, L=2 case by finding the two-sample distributions of the separate channels as we did for other receivers. The convolution must be done in three dimensions (M-1 dimensions for the general case). The concept for doing this is unusual, but can be visualized. Figure 6.1-2 illustrates the fact that multi-dimensional convolution of two pdf's involves integration over the volume which is the intersection of the domains of the pdf's. In Figure 6.1-2(b), the simple case when the point (z_1, z_2, z_3) lies inside the domain of $p_1(z)$ is shown; this yields a rectangular-sided volume. If (z_1, z_2, z_3) lies outside the domain of $p_1(z)$, the intersection is much more complicated.

By careful study we have determined that the pdf for the SNORM receiver's decision variables for M=4 and L=2 has the general form

$$p_2(\underline{z}|\underline{\xi} = \underline{v}_1 + \underline{v}_2) = \int_{A_1}^{B_1} dv_1 \int_{A_2}^{B_2} dv_2 \int_{A_3}^{B_3} dv_3 p_1(\underline{v}|\underline{v}_1) p_1(\underline{z}-\underline{v}|\underline{v}_2), \quad (6.1-27a)$$

where

$$\begin{aligned} A_1 &= \max(0, z_1 - 1) \\ B_1 &= \min(1, z_1) \\ A_2 &= \max(0, z_1 + z_2 - v_1 - 1) \\ B_2 &= \min(1 - v_1, z_2) \\ A_3 &= \max(0, z_1 + z_2 + z_3 - v_1 - v_2 - 1) \\ B_3 &= \min(1 - v_1 - v_2, z_3). \end{aligned} \quad (6.1-27b)$$

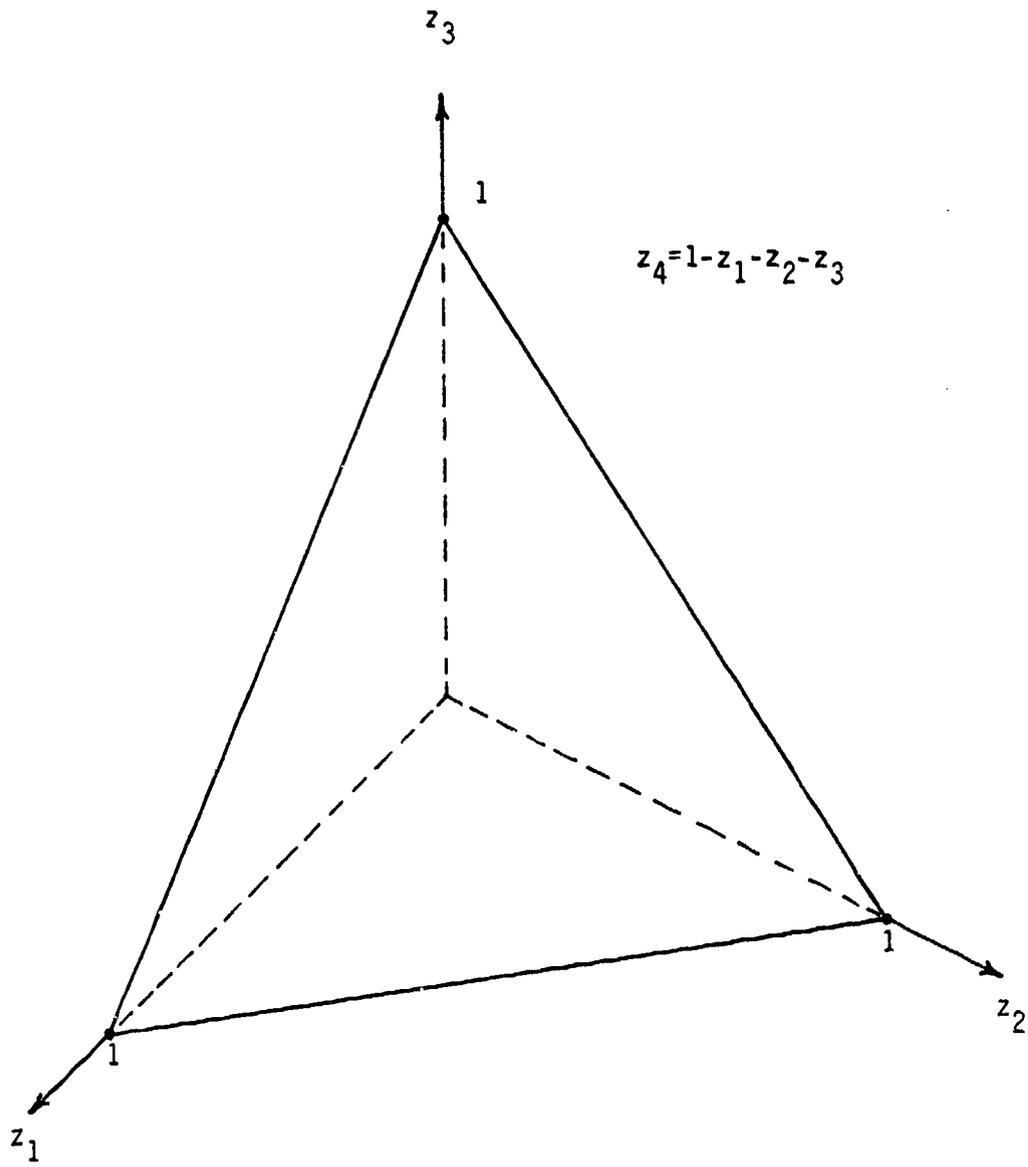
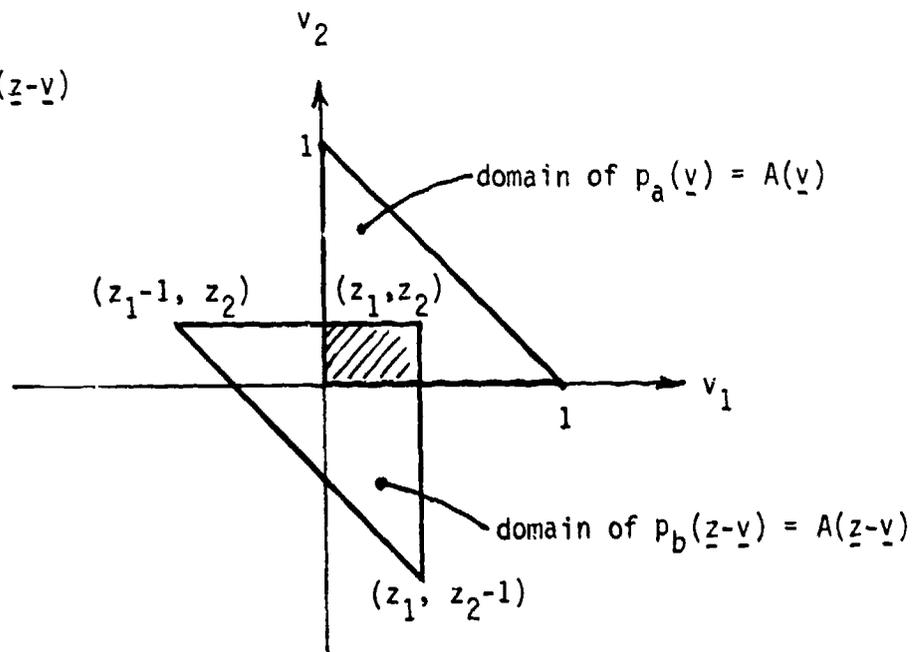


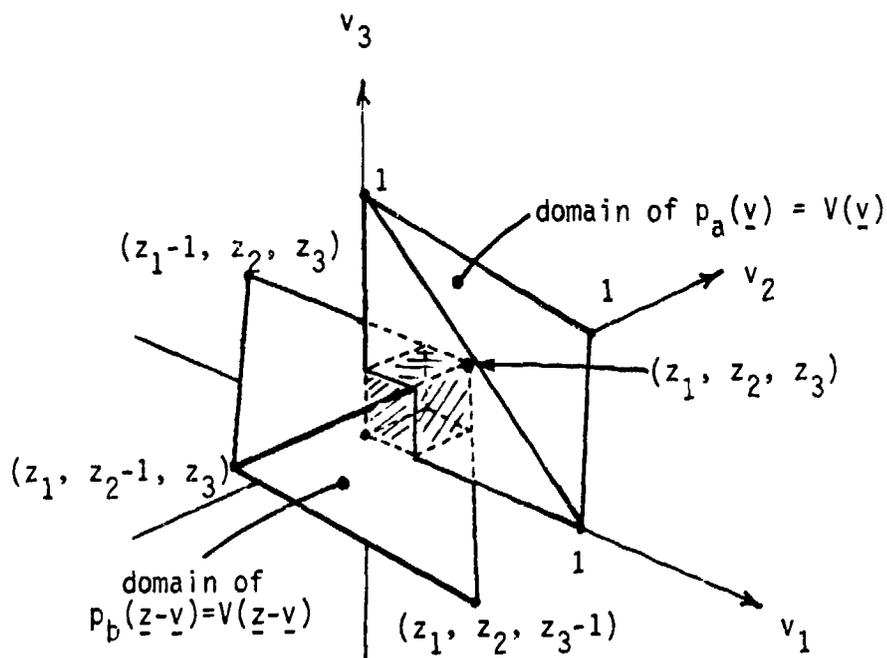
FIGURE 6.1-1 REGION CONTAINING DECISION VARIABLES

$$\iint d\underline{v} p_a(\underline{v}) p_b(\underline{z}-\underline{v})$$

$$A(\underline{v}) \cap A(\underline{z}-\underline{v})$$



(a) two-dimensional convolution



$$\iiint d\underline{v} p_a(\underline{v}) p_b(\underline{z}-\underline{v})$$

$$V(\underline{v}) \cap V(\underline{z}-\underline{v})$$

(b) three-dimensional convolution

FIGURE 6.1-2 TWO- AND THREE-DIMENSIONAL CONVOLUTIONS

Now, since there are sixteen cases of $p_1(\underline{z}|\underline{v})$ for $L=1$, there are $(16)^2 = 256$ cases for $L=2$. However since the numbering of symbol channels is arbitrary, there can be considered to be fewer, distinguishable jamming events. These are fully enumerated in Section 6.2. What we wish to note here is that if neither of the densities in (6.1-27) contains v_3 , the integral can be simplified to

$$p_2(\underline{z}|\underline{x}) = \int_{A_1}^{B_1} dv_1 \int_{A_2}^{B_2} dv_2 p_1(\underline{v}|\underline{v}_1) p_1(\underline{z}-\underline{v}|\underline{v}_2) (B_3-A_3) \quad (6.1-28a)$$

where

$$\begin{aligned} B_3 - A_3 &= \min(1-v_1-v_2, v_3) - \max(0, z_1+z_2+z_3-v_1-v_2-1) \\ &= \frac{1}{2} \left\{ 2-z_1-z_2 - |1-v_1-v_2-z_3| - |z_1+z_2+z_3-v_1-v_2-1| \right\}. \end{aligned} \quad (6.1-28b)$$

If neither pdf in (6.1-27) contains v_2 or v_3 , the integral can be further simplified to

$$p_2(\underline{z}|\underline{x}) = \int_{A_1}^{B_1} dv_1 p_1(v_1|\underline{v}_1) p_1(z_1-v_1|\underline{v}_2) \int_{A_2}^{B_2} dv_2 (B_3-A_3), \quad (6.1-29)$$

where

$$\begin{aligned} \int_{A_2}^{B_2} dv_2 (B_3-A_3) &= \frac{1}{2} (2-z_1-z_2) (B_2-A_2) \\ &\quad - \frac{1}{4} \left\{ (B_2+v_1+z_3-1) |B_2+v_1+z_3-1| \right. \\ &\quad \left. - (A_2+v_1+z_3-1) |A_2+v_1+z_3-1| \right. \\ &\quad \left. + (B_2+v_1+1-z_1-z_2-z_3) |B_2+v_1+1-z_1-z_2-z_3| \right. \\ &\quad \left. - (A_2+v_1+1-z_1-z_2-z_3) |A_2+v_1+1-z_1-z_2-z_3| \right\}, \end{aligned} \quad (6.1-30)$$

since

$$\int_A^B dx |x-a| = \int_A^B dx |a-x| = \frac{1}{2} (B-a)|B-a| - \frac{1}{2} (A-a)|A-a|. \quad (6.1-3)$$

Now, if v_3 is in the integrand of (6.1-27) but v_2 is not, we simply "switch labels" on v_2 and v_3 to get (6.1-28) with v_3 replacing v_2 .

6.2 JAMMING EVENTS AND ERROR PROBABILITY FOR L=2

We now extend the conditional distribution analysis in the last section to obtain the BER for the FH/RMFSK SNORM receiver under partial-band noise jamming. Since the jammed error for L=1 is the same for other receivers, we proceed to the case of L=2.

6.2.1 Jamming Events and Probabilities for M=2.

For L=2 and M=2 there are $2^{ML}=16$ elementary jamming events. As mentioned previously, for the SNORM receiver, only ten of these events are distinguishable in terms of jamming effects. These are listed in Table 6.2-1, along with the single-hop events which produce them and the probabilities of the L=2 events.

The error event, assuming the signal is in channel 1, is $z_1 < z_2 \equiv 2-z_1$. Thus

$$P_b(e; \gamma) = \Pr\{z_1 < 1\} = \sum_{\underline{\ell}} \Pr\{\underline{\ell}\} \Pr\{z_1 < 1 | \underline{\ell}\}. \quad (6.2-1)$$

The conditional error probabilities in (6.2-1) are calculated by integrating the pdf's shown previously in equation (6.1-23) or the convolution

TABLE 6.2-1

JAMMING EVENTS AND PROBABILITIES FOR M=2, L=2

\underline{l}	\underline{v}_1	\underline{v}_2	# events	Probability
0,0	0,0	0,0	1	π_0^2
0,1	0,1	0,0	2	$2\pi_0\pi_1$
0,2	0,1	0,1	1	π_1^2
1,0	1,0	0,0	2	$2\pi_0\pi_1$
1,1	0,0	1,1	2	$2\pi_0\pi_2$
	0,1	1,0	2	$2\pi_1^2$
1,2	0,1	1,1	2	$2\pi_1\pi_2$
2,0	1,0	1,0	1	π_1^2
2,1	1,0	1,1	2	$2\pi_1\pi_2$
2,2	1,1	1,1	1	π_2^2
Totals:			16	1

$$\pi_0 = \frac{\binom{N-2}{q}}{\binom{N}{q}} = \frac{(N-q)(N-q-1)}{N(N-1)}, \quad \pi_1 = \frac{\binom{N-2}{q-1}}{\binom{N}{q}} = \frac{q(N-q)}{N(N-1)}, \quad \pi_2 = \frac{\binom{N-2}{q-2}}{\binom{N}{q}} = \frac{q(q-1)}{N(N-1)}$$

$$p_2(z | \underline{l} = \underline{v}_1 + \underline{v}_2) = \int_{\max(0, z-1)}^{\min(1, z)} dv p_1(v | \underline{v}_1) p_1(z-v | \underline{v}_2)$$

of the pdf's in equations (6.1-16) to (6.1-19). The result is

$$\begin{aligned}
 P_b(e;\gamma) = & \pi_0^2 \cdot \frac{1}{2} e^{-\rho_N} (1+\rho_N/3) \\
 & + 2\pi_0\pi_1 \int_0^1 dv (1-v)e^{-\rho_N v} p_1(v|0,1) \\
 & + \pi_1^2 \int_0^1 dv p_1(v|0,1) \cdot \frac{K(1-v)}{v+K(1-v)} \exp \left\{ \frac{-\rho_N v}{v+K(1-v)} \right\} \\
 & + 2\pi_0\pi_1 \int_0^1 dv (1-v)e^{-\rho_N v} p_1(v|1,0) \\
 & + 2\pi_0\pi_2 \frac{1}{(\rho_N - \rho_T)^3} \left\{ e^{-\rho_T} \left[\rho_N(\rho_N - \rho_T) - (\rho_N + \rho_T) \right] \right. \\
 & \quad \left. + e^{-\rho_N} \left[\rho_T(\rho_N - \rho_T) + (\rho_N + \rho_T) \right] \right\} \\
 & + 2\pi_1^2 \int_0^1 dv p_1(v|0,1) \cdot \frac{1-v}{Kv+1-v} \exp \left\{ \frac{-K\rho_T v}{Kv+1-v} \right\} \\
 & + 2\pi_1\pi_2 \int_0^1 dv (1-v)e^{-\rho_T v} p_1(v|0,1) \\
 & + \pi_1^2 \int_0^1 dv p_1(v|1,0) \cdot \frac{1-v}{Kv+1-v} \exp \left\{ \frac{-K\rho_T v}{Kv+1-v} \right\} \\
 & + 2\pi_1\pi_2 \int_0^1 dv (1-v)e^{-\rho_T v} p_1(v|1,0) \\
 & + \pi_2^2 \cdot \frac{1}{2} e^{-\rho_T} (1+\rho_T/3).
 \end{aligned} \tag{6.2-2}$$

In this expression we have used the fact that

$$\begin{aligned}
 & \int_0^1 dz \int_{\max(0, z-1)}^{\min(1, z)} dv \, p_1(v|\underline{v}_1) \, p_1(z-v|\underline{v}_2) \\
 &= \int_0^1 dz \int_0^z dv \, p_1(v|\underline{v}_1) \, p_1(z-v|\underline{v}_2) \\
 &= \int_0^1 dv \, p_1(v|\underline{v}_1) \int_0^{1-v} dz \, p_1(z|\underline{v}_2) .
 \end{aligned} \tag{6.2-3}$$

Also, the parameters K , ρ_N , and ρ_T are

$$K = \frac{\sigma_T^2}{\sigma_N^2} , \quad \rho_N = \frac{1}{2} \cdot \frac{E_b}{N_0} , \quad \rho_T = \frac{1}{2} \cdot \frac{E_b}{N_T} . \tag{6.2-4}$$

6.2.2 Jamming Events and Probabilities for M=4.

For $L=2$ and $M=4$, there are 256 elementary jamming events, which can be represented by 47 distinguishable events. These are listed in Table 6.2-2, along with their probabilities of occurrence. The joint pdf of the decision variables, given the representative jamming event shown in the table, is the convolution (6.1-27) with the single-hop pdf's selected from Table 6.1-1 as indicated.

The error event for $M=4$ and $L=2$ is the complement of the condition for a correct symbol decision, so that the conditional error probability is

$$\begin{aligned}
 P_s(e|\underline{\ell}) &= 1 - \Pr\{z_1 > z_2, z_1 > z_3, z_1 > z_4 = 2 - z_1 - z_2 - z_3 | \underline{\ell}\} \\
 &= 1 - \Pr\{z_1 > z_2, z_1 > z_3, 2z_1 > 2 - z_2 - z_3 | \underline{\ell}\}
 \end{aligned} \tag{6.2-5a}$$

$$\equiv 1 - P_s(C|\underline{\ell}) . \tag{6.2-5b}$$

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TABLE 6.2-2
JAMMING EVENTS AND PROBABILITIES FOR M=4, L=2

\underline{i}	cases*	# events	prob.	\underline{i}	cases*	# events	prob.
0 0 0 0	A+A	1	π_0^2	1 2 2 0	F+N	6	$6\pi_2\pi_3$
0 1 0 0	A+E	6	$6\pi_0\pi_1$	1 2 2 1	H+N	6	$6\pi_3^2$
0 1 1 0	A+G	6	$6\pi_0\pi_2$		G+P	6	$6\pi_2\pi_4$
	C+E	6	$6\pi_1^2$	1 2 2 2	H+P	2	$2\pi_3\pi_4$
0 1 1 1	A+H	2	$2\pi_0\pi_3$				
	C+F	6	$6\pi_1\pi_2$	2 0 0 0	I+I	1	π_1^2
0 2 0 0	E+E	3	$3\pi_1^2$	2 1 0 0	I+M	6	$6\pi_1\pi_2$
0 2 1 0	E+G	12	$12\pi_1\pi_2$	2 1 1 0	I+N	6	$6\pi_1\pi_3$
0 2 1 1	E+H	6	$6\pi_1\pi_3$		K+M	6	$6\pi_2^2$
	F+G	6	$6\pi_2^2$	2 1 1 1	I+P	2	$2\pi_1\pi_4$
0 2 2 0	G+G	3	$3\pi_2^2$		K+N	6	$6\pi_2\pi_3$
0 2 2 1	G+H	6	$6\pi_2\pi_3$	2 2 0 0	M+M	3	$3\pi_2^2$
0 2 2 0	H+H	1	π_3^2	2 2 1 0	M+N	12	$12\pi_2\pi_3$
				2 2 1 1	M+P	6	$6\pi_2\pi_4$
1 0 0 0	A+I	2	$2\pi_0\pi_1$		L+N	6	$6\pi_3^2$
1 1 0 0	A+M	6	$6\pi_0\pi_2$	2 2 2 0	N+N	3	$3\pi_3^2$
	E+I	6	$6\pi_1^2$	2 2 2 1	N+P	6	$6\pi_3\pi_4$
1 1 1 0	A+N	6	$6\pi_0\pi_3$	2 2 2 2	P+P	1	π_4^2
	G+I	6	$6\pi_1\pi_2$		Totals:	256	1
	C+M	12	$12\pi_1\pi_2$				
1 1 1 1	A+P	2	$2\pi_0\pi_4$				
	H+I	2	$2\pi_1\pi_3$				
	C+N	6	$6\pi_1\pi_3$				
	F+K	6	$6\pi_2^2$				
1 2 0 0	E+M	6	$6\pi_1\pi_2$				
1 2 1 0	E+N	12	$12\pi_1\pi_3$				
	G+M	12	$12\pi_2^2$				
1 2 1 1	G+N	12	$12\pi_1\pi_3$				
	E+P	6	$6\pi_1\pi_4$				
	H+M	6	$6\pi_2\pi_3$				

*cases (A-P): See Table 6.1-1

$$\pi_0 = \frac{(N-q)(N-q-1)(N-q-2)(N-q-3)}{N(N-1)(N-2)(N-3)}$$

$$\pi_1 = \frac{q(N-q)(N-q-1)(N-q-2)}{N(N-1)(N-2)(N-3)}$$

$$\pi_2 = \frac{q(q-1)(N-q)(N-q-1)}{N(N-1)(N-2)(N-3)}$$

$$\pi_3 = \frac{q(q-1)(q-2)(N-q)}{N(N-1)(N-2)(N-3)}$$

$$\pi_4 = \frac{q(q-1)(q-2)(q-3)}{N(N-1)(N-2)(N-3)}$$

From this expression we observe that the probability of a correct symbol decision is obtained from the joint pdf of (z_1, z_2, z_3) by integrating it over the volume Ω_c implied in (6.2-5):

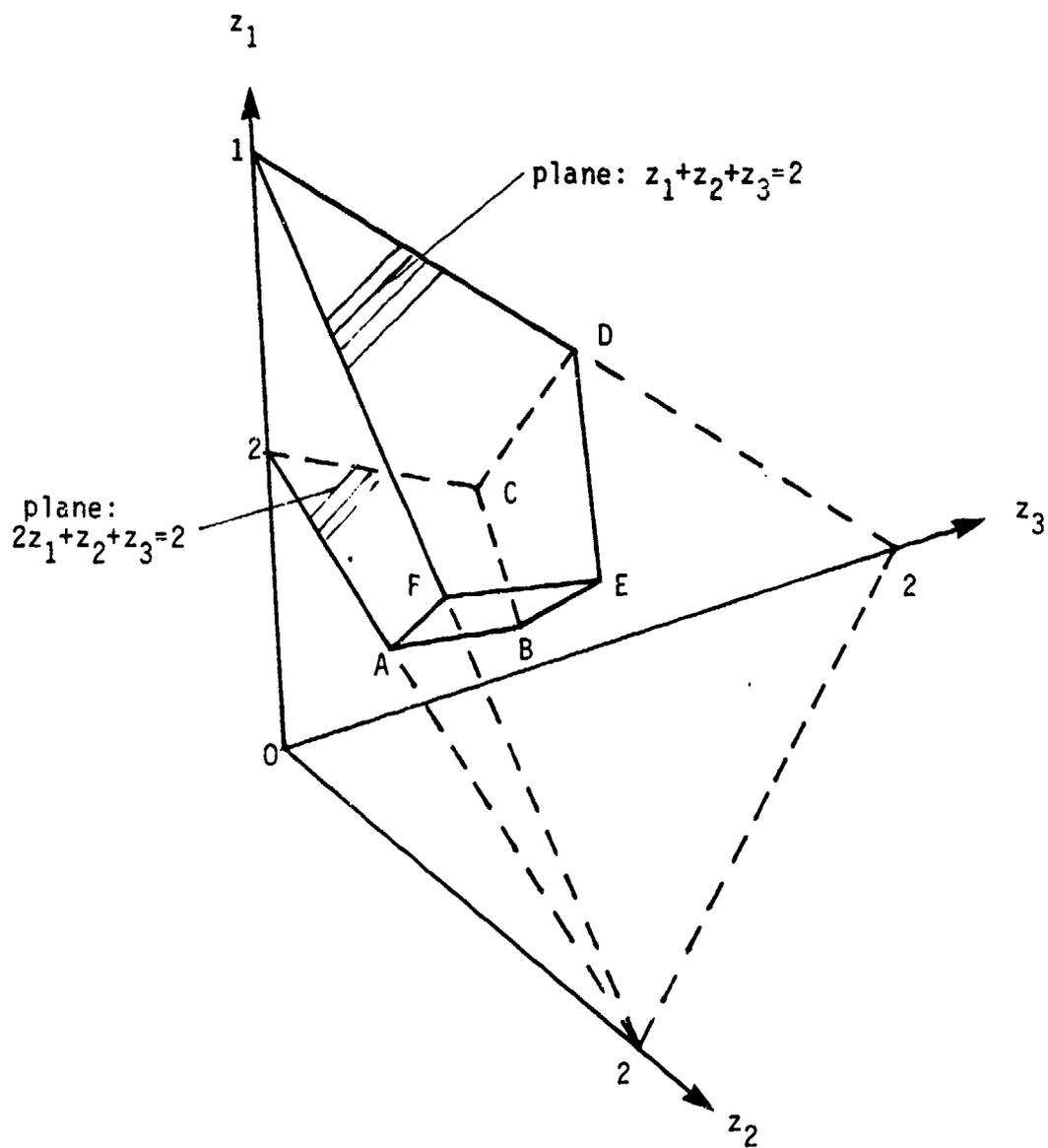
$$P_s(C|\underline{l}) = \iiint_{\Omega_c} dz_1 dz_2 dz_3 p_2(\underline{z} | \underline{l}). \quad (6.2-6)$$

As illustrated in Figure 6.2-1, the volume Ω_c may be described as that enclosed by the planes $z_2=0$, $z_3=0$, $2z_1+z_2+z_3=2$, $z_1+z_2+z_3=2$, $z_1=z_2$, and $z_1=z_3$. Thus

$$P_s(C|\underline{l}) = \int_{A_4}^{B_4} dz_1 \int_{A_5}^{B_5} dz_2 \int_{A_6}^{B_6} dz_3 p_2(\underline{z} | \underline{l}), \quad (6.2-7a)$$

where

$$\begin{aligned} A_4 &= 1/2, \quad B_4 = 2 \\ A_5 &= \max(0, 2-3z_1) \\ B_5 &= \min(z_1, 2-z_1) \\ A_6 &= \max(0, 2-2z_1-z_2) \\ B_6 &= \min(z_1, 2-z_1-z_2). \end{aligned} \quad (6.2-7b)$$



points: (z_1, z_2, z_3)

- A: $(2/3, 2/3, 0)$ D: $(1, 0, 1)$
 B: $(1/2, 1/2, 1/2)$ E: $(2/3, 2/3, 2/3)$
 C: $(2/3, 0, 2/3)$ F: $(1, 1, 0)$

FIGURE 6.2-1 VOLUME OF INTEGRATION FOR CORRECT SYMBOL DECISION, $M=4, L=2$

6.3 AN ALTERNATE APPROACH FOR M=2 AND L=3 HOPS/SYMBOL

In order to obtain a more computationally tractable form for the performance of the self-normalizing receiver in partial-band noise jamming, we may proceed as follows. The probability of a symbol error is

$$\begin{aligned} \Pr(e) &= \Pr\{z < 3/2\} \\ &= E_{\underline{v}}\{\Pr\{z < 3/2 | \underline{v}\}\} \\ &= E_{\underline{v}}\left\{\int_0^{3/2} p_3(\zeta | \underline{v}) d\zeta\right\} \end{aligned} \quad (6.3-1)$$

where $p_3(\zeta | \underline{v})$ is the probability density function conditioned on jamming event \underline{v} .

If we interchange the order of integration with respect to ζ and expectation with respect to \underline{v} in (6.3-1), we obtain

$$\Pr(e) = \int_0^{3/2} E_{\underline{v}}\{p_3(\zeta | \underline{v})\} d\zeta. \quad (6.3-2)$$

The expectation in (6.3-2) may be written as

$$\begin{aligned} p_3(\zeta) &\triangleq E_{\underline{v}}\{p_3(\zeta | \underline{v})\} \\ &= E_{\underline{v}}\{p_1(\zeta | \underline{v}) * p_2(\zeta | \underline{v})\} \\ &= p_1(\zeta) * p_2(\zeta) \end{aligned} \quad (6.3-3)$$

where the operator $*$ denotes convolution and

$$p_1(\zeta) \triangleq E_{\underline{v}_1}\{p_1(\zeta | \underline{v}_1)\} \quad (6.3-4)$$

with \underline{v}_1 being the first column of the event matrix \underline{v} and

$$p_2(\zeta) \triangleq E_{\underline{v}_2} \{p_2(\zeta | \underline{v}_2)\} \quad (6.3-5)$$

with \underline{v}_2 being the last two columns of the matrix \underline{v} . The expectation in (6.3-4) is given by

$$p_1(\zeta) = \pi_0 p_{00}(\zeta) + \pi_1 p_{01}(\zeta) + \pi_1 p_{10}(\zeta) + \pi_2 p_{11}(\zeta) \quad (6.3-6)$$

where the $p_{ij}(\zeta) \triangleq p_1[\zeta | \underline{v} = (i,j)]$ are given by (6.1-16)-(6.1-19) and the event probabilities π_0 , π_1 , and π_2 are given by Table 6.2-1.

The density $p_2(\zeta)$ from (6.3-5) contains 10 terms, as discussed in Section 6.1.3.2. Analytical results for three of these cases are given by (6.1-23) and (6.1-24).

By performing the convolution (6.3-3) using (6.3-6) and (6.1-23), we obtain a form containing the sum of seven numerical convolutions. However, each convolution involves a reasonably well-behaved integrand. Overall, the computational effort is also lightened by the reduction in total terms due to splitting up the 3-hop jamming events into 1-hop events in (6.3-6) and 2-hop events in (6.3-5) for which, at least in part, analytical results are available. This is the method implemented by the computer program given in Appendix L.

6.4 NUMERICAL RESULTS

The numerical computations for the self-normalizing receiver suffer difficulties similar to those encountered for the clipper receiver, namely a multitude of multiple-dimensional numerical integrations. For the case of $M=2$ and $L=2$, from (6.2-2) and (6.1-16), we have the simplest case to compute, consisting of 7 one-dimensional integrals. For the case of $M=2$ and $L=3$, we must do 5 two-dimensional numerical integrations and 2 three-dimensional numerical integrations to obtain a value of $P_b(e)$ for given E_b/N_0 , E_b/N_J , and γ . For the case of $M=4$ and $L=2$, we are faced with numerical integrations in five or six dimensions over non-standard regions (see, for example, Figure 6.2-1 for the region of integration of the outermost 3 dimensions).

At the computational throughput rate of the PDP-11/44 computer available for the computations, the CPU time to obtain results for even $M=4$ and $L=2$ are estimated to run to many months, or even years. Hence, we restrict our numerical computations to $M=2$ and $L=2$ and 3. For $L=2$, the speed of computation was sufficiently high to permit the full set of curves for $\gamma = 0.001$ to $\gamma = 1.0$, to be computed using the program in Appendix K. Figure 6.4-1 shows the performance as a function of E_b/N_J when $E_b/N_0 = 13.35247$ dB, which corresponds to $P_b(e) = 10^{-5}$ for ideal MFSK with $M=2$. We note in Figure 6.4-1 that there is a clustering of the cross-over around $E_b/N_J = 16$ dB.

If E_b/N_0 is increased to 20 dB, the performance curves shown in Figure 6.4-2 are obtained. In this figure we observe that many of the curves exhibit a breakpoint at which the direction of curvature changes. Clearly, the self-normalizing receiver departs considerably from the ideal receiver under certain ranges of operating conditions. The exact mechanisms which come into play to explain this behavior are not totally clear, but it appears to be the

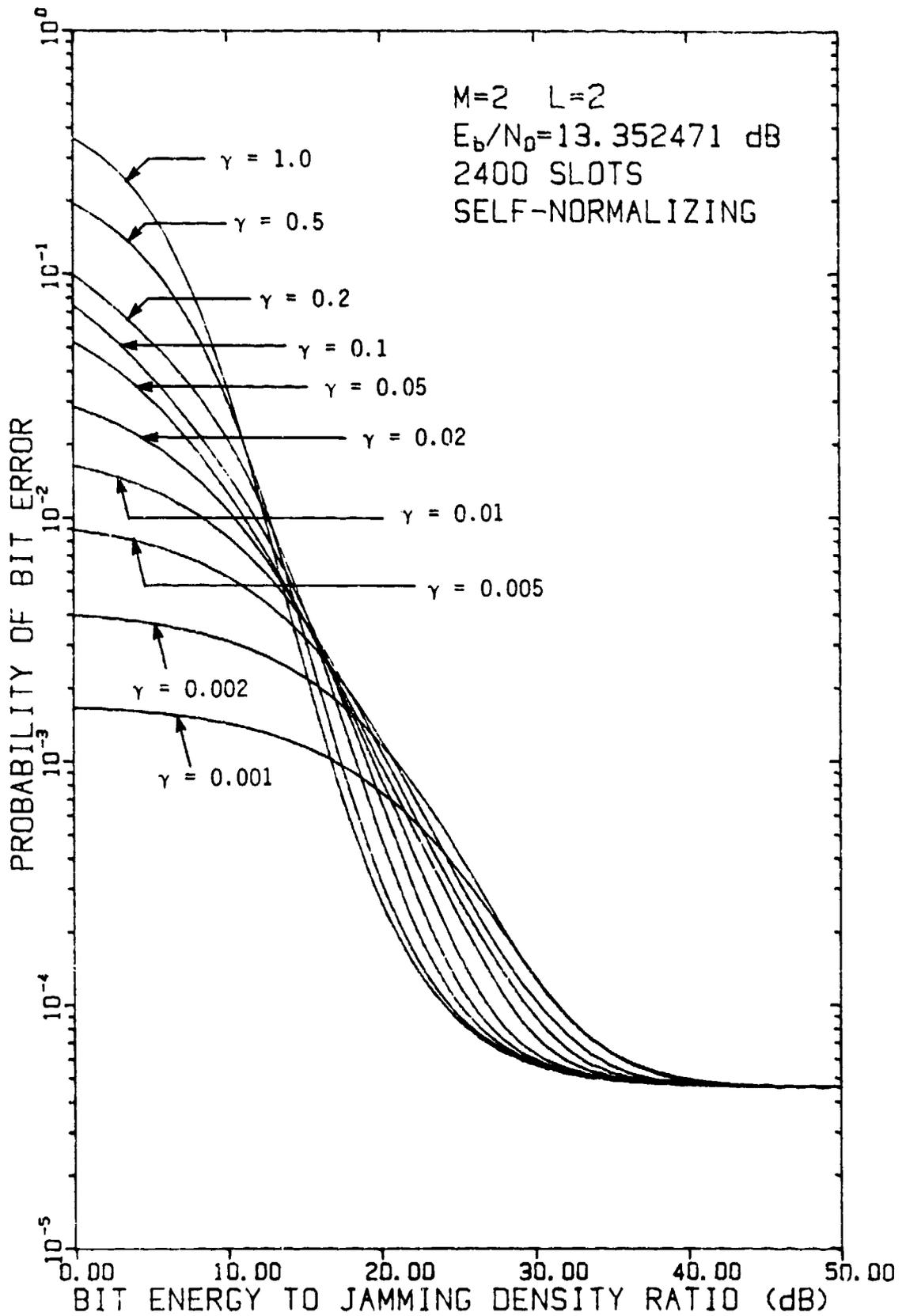


FIGURE 6.4-1 PERFORMANCE OF SELF-NORMALIZING RECEIVER FOR FH/RMFSK WITH $M=2$, $L=2$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.35247 \text{ dB}$ (FOR $P_b(e)=10^{-5}$ IN THE ABSENCE OF JAMMING WHEN $L=1$)

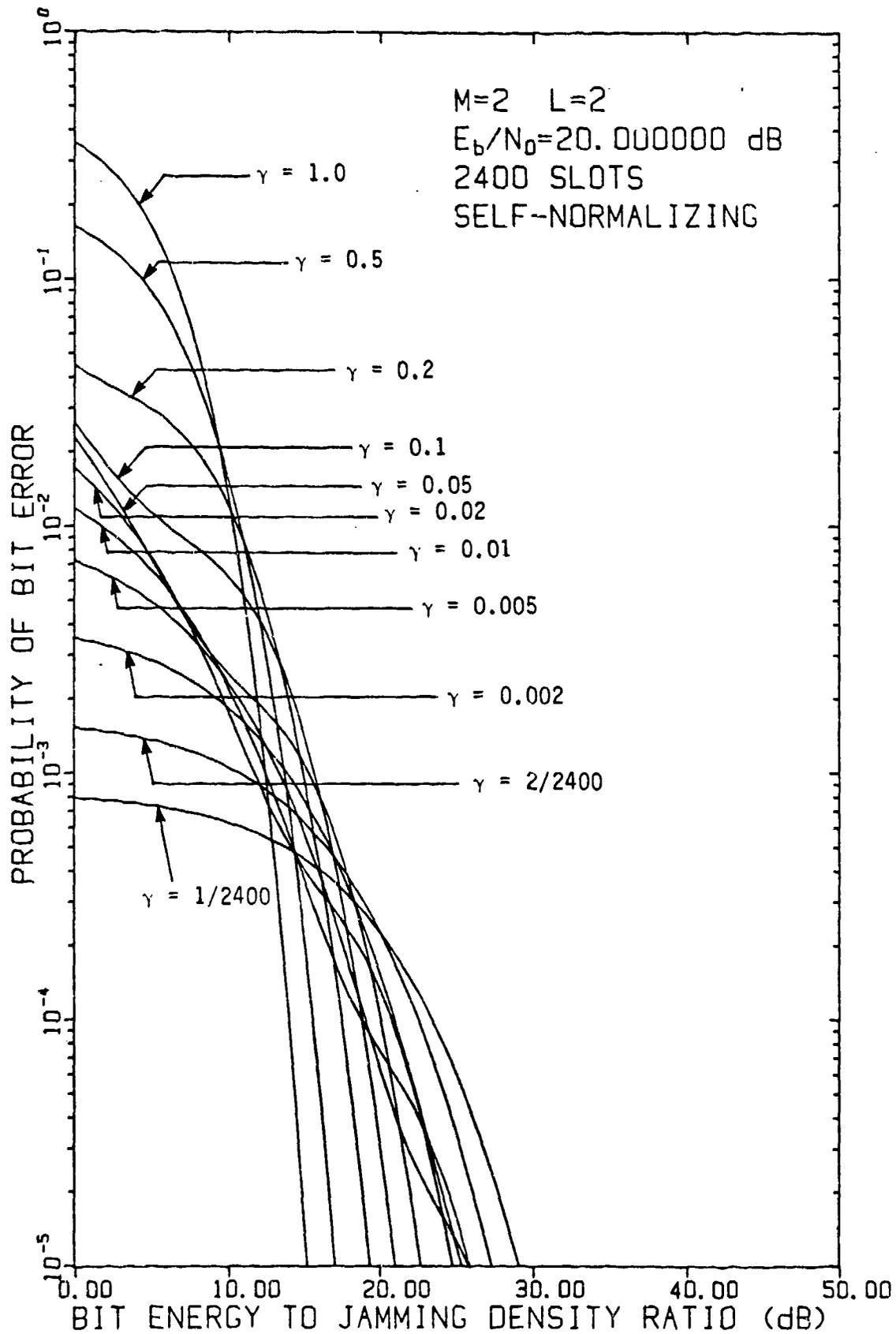


FIGURE 6.4-2 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR SELF-NORMALIZING RECEIVER IN WORST-CASE PARTIAL-BAND NOISE JAMMING FOR $M=2$, $L=2$ AT $E_b/N_0 = 20 \text{ dB}$

interaction of several different effects, with the switch-over from thermal-noise-limited operation to partial-band-jamming-limited operation playing a significant role. The importance of this switch-over is supported by the lack of apparent breakpoints in the curves for very small γ ($\gamma = 0.001, 0.002$) and very large γ ($\gamma = 0.5, 1.0$); these are the cases in which the one-slot-jammed jamming event predominates and the shape of the curves reflects essentially the performance conditioned on the dominant event.*

For the case of $L=3$ hops, the reduced speed of computation dictated that we search for the optimum jamming fraction at each E_b/N_J rather than run full curves for the various values of γ . Again, we started at $\gamma = 1/2400$ for $E_b/N_J = 50$ dB to speed the search, and then stepped to a lower value of E_b/N_J . The computer program in Appendix L was used to obtain the results presented in Figure 6.4-3 for $M=2, L=3$.

Finally, Figure 6.4-4 compares the performances of the self-normalizing receiver as L , the number of hops per symbol, varies. As we have observed with the other receivers, there is a limited range of E_b/N_J over which a "diversity" effect is achieved. For example, $L=3$ outperforms $L=2$ for $17 \text{ dB} < E_b/N_J < 29 \text{ dB}$. However, in the thermal-noise-limited region and in the strong-jamming region (where $\gamma = 1.0$ is the worst-case jamming), the noncoherent combining loss dominates and $L=1$ is optimum.

* For further discussion, see Section 7.3.3.5.

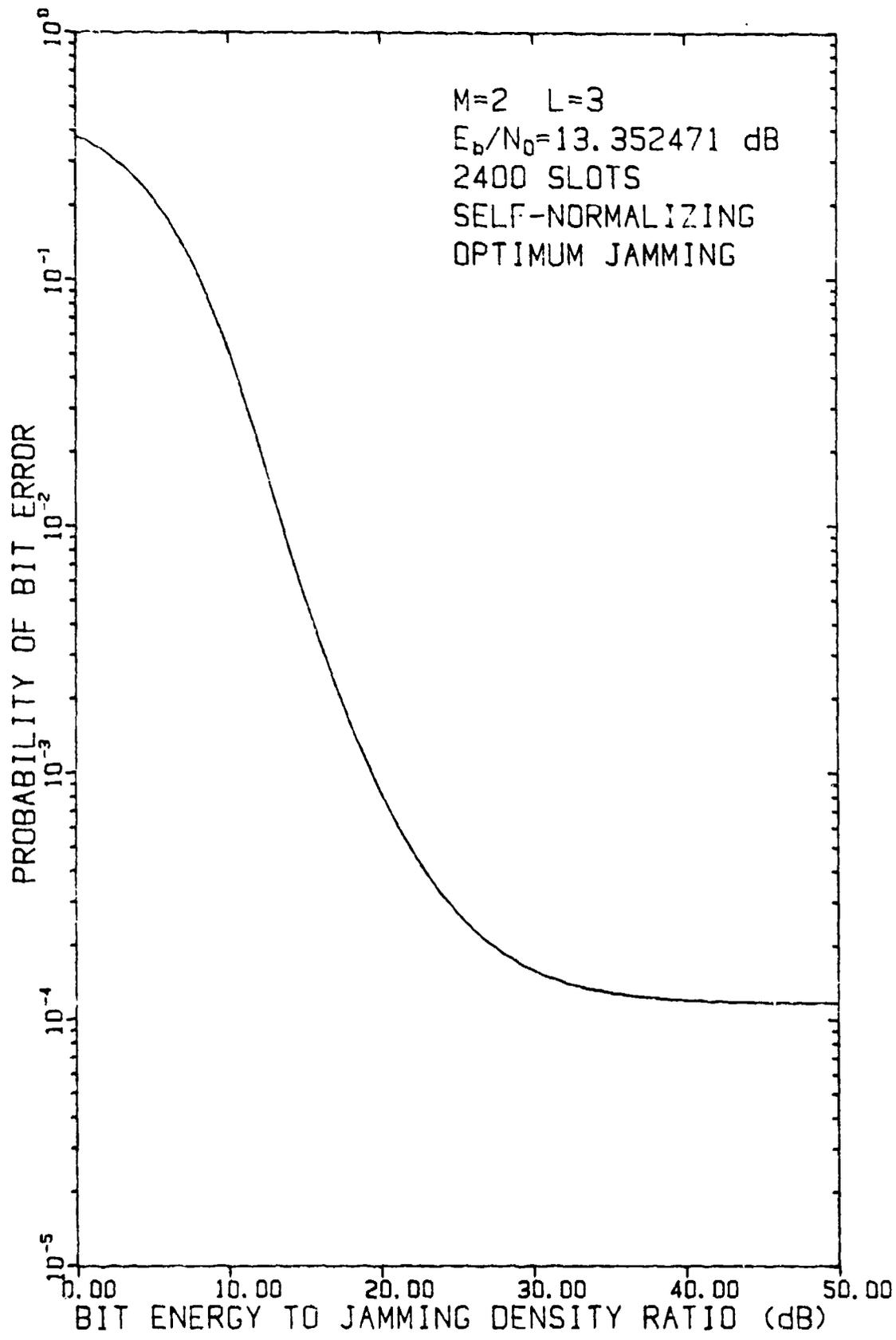


FIGURE 6.4-3 PERFORMANCE OF SELF-NORMALIZING RECEIVER FOR FH/RMFSK WITH $M=2$, $L=3$ HOPS/SYMBOL, 2400 HOPPING SLOTS, AND $E_b/N_0 = 13.35247 \text{ dB}$ (FOR $P_b(e) = 10^{-5}$ IN THE ABSENCE OF JAMMING WHEN $L=1$)

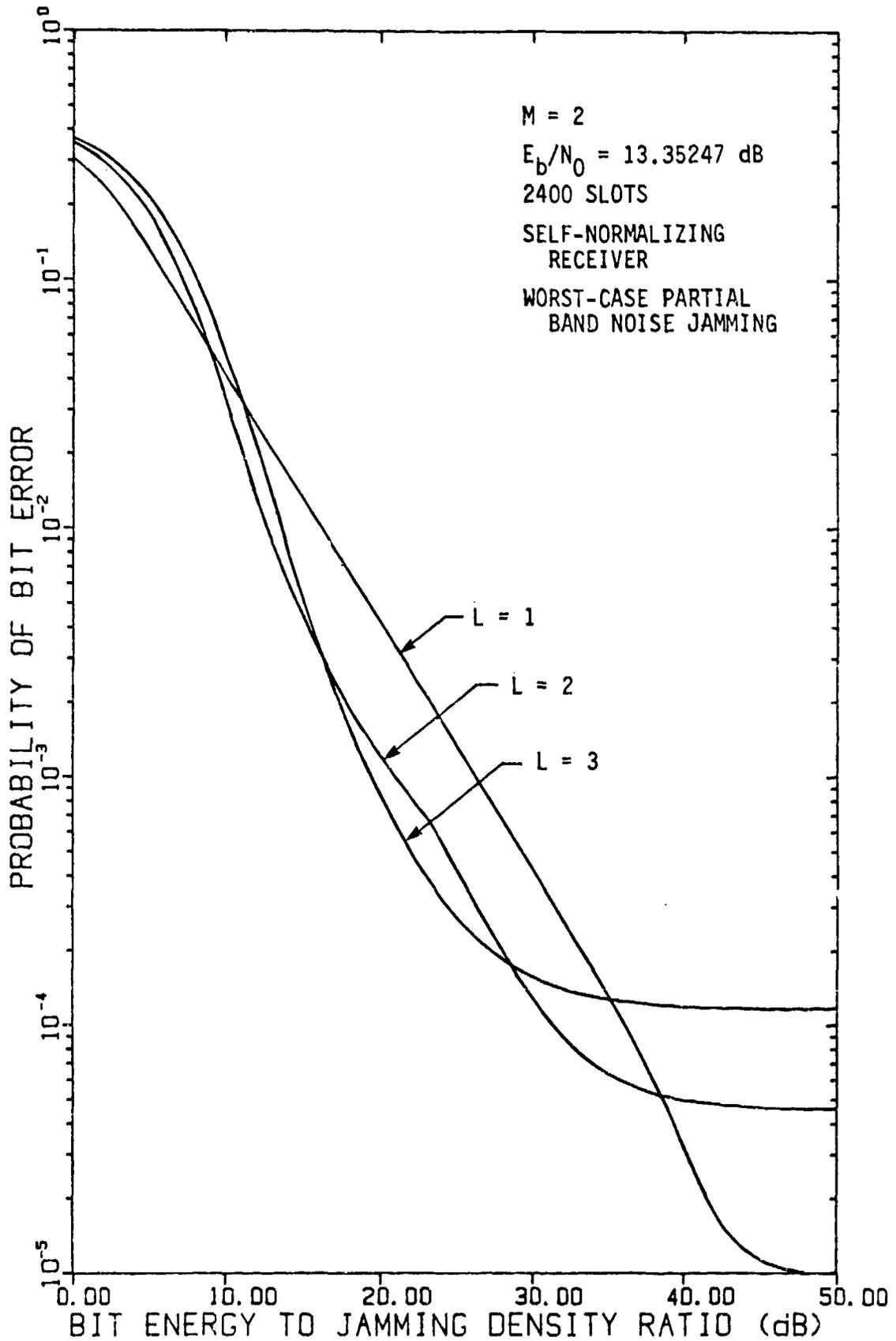


FIGURE 6.4-4 PERFORMANCE OF SELF-NORMALIZING RECEIVER FOR FH/RMFSK WITH $M=2$, AND $E_b/N_0 = 13.35247 \text{ dB}$ WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER

7.0 COMPARISONS OF RECEIVER PERFORMANCES

The information we have generated separately on the performances of the various FH/RMFSK receivers in Sections 3-6 can now be compared to learn which ECCM processing scheme is most effective in worst-case partial-band noise jamming. However, since the random MFSK waveform is specifically designed to counter follow-on jamming, we first develop the performances of these receiver processing schemes in follow-on jamming, both for conventional hopping and for random hopping.

7.1 RECEIVER PERFORMANCES IN FOLLOW-ON NOISE JAMMING

7.1.1 Formulation of Follow-on Jamming Analysis: Simple Jammer.

Under follow-on noise jamming (FNJ), it is assumed that on each hop, the jammer places his available power, J watts, in a relatively narrow band centered on the signal's hop frequency. If this band is at least $2(M-1) + 1$ slots wide, then the jammer is guaranteed to jam all M slots of conventional FH/MFSK on every hop*. The hop SNR for FNJ therefore is

$$\rho_T = \frac{S}{\sigma_N^2 + \sigma_J^2} = \frac{E_h}{N_0 + N_J/\gamma_r} = \frac{\log_2 M}{L} \cdot \frac{E_b}{N_0 + N_J/\gamma_r} \quad (7.1-1a)$$

where the jammer spectral density N_J is defined over the entire hopping system bandwidth, $N_J = J/W$, and therefore the effective jamming fraction is

$$\gamma_r = \frac{W_J}{W} \geq \frac{2M-1}{N} \quad (7.1-1b)$$

In terms of the jamming events indexed by the vector $\underline{\ell} = (\ell_1, \ell_2, \dots, \ell_M)$, where ℓ_m is the number of hops jammed in symbol frequency

*In Section 7.1.2, we consider an "advanced" FNJ which excludes jamming from the signal slot.

channel m , for repeat jamming we have

$$\ell_1 = L. \quad (7.1-2)$$

The values of the other $\{\ell_m\}$ depend on the hopping scheme.

7.1.2 Formulation of FNJ Analysis: Advanced Jammer.

The FNJ can avoid helping the communicator by not putting any jammer power in the signal slot. Assuming that this measure is taken, the hop SNR is ρ_N in the signal slot. For maximum effectiveness against FH/MFSK, we also assume that the jammer places half its power in each of the two slots on either side of the intercepted signal, as illustrated in Figure 7.1-1. Thus $\ell_1 = 0$, and for a single hop, for FH/MFSK,

$$\left. \begin{aligned} \Pr\{1 \text{ nonsignal slot jammed}\} &= \frac{2}{M} \\ \Pr\{2 \text{ nonsignal slots jammed}\} &= 1 - \frac{2}{M} \end{aligned} \right\} \quad (7.1-3a)$$

For random hopping and advanced FNJ,

$$\left. \begin{aligned} \Pr\{0 \text{ nonsignal slots jammed}\} &= \frac{(N-M)(N-M-1)}{(N-1)(N-2)} \\ \Pr\{1 \text{ nonsignal slot jammed}\} &= 2 \frac{(N-M)(M-1)}{(N-1)(N-2)} \\ \Pr\{2 \text{ nonsignal slots jammed}\} &= \frac{(M-1)(M-2)}{(N-1)(N-2)} \end{aligned} \right\} \quad (7.1-3b)$$

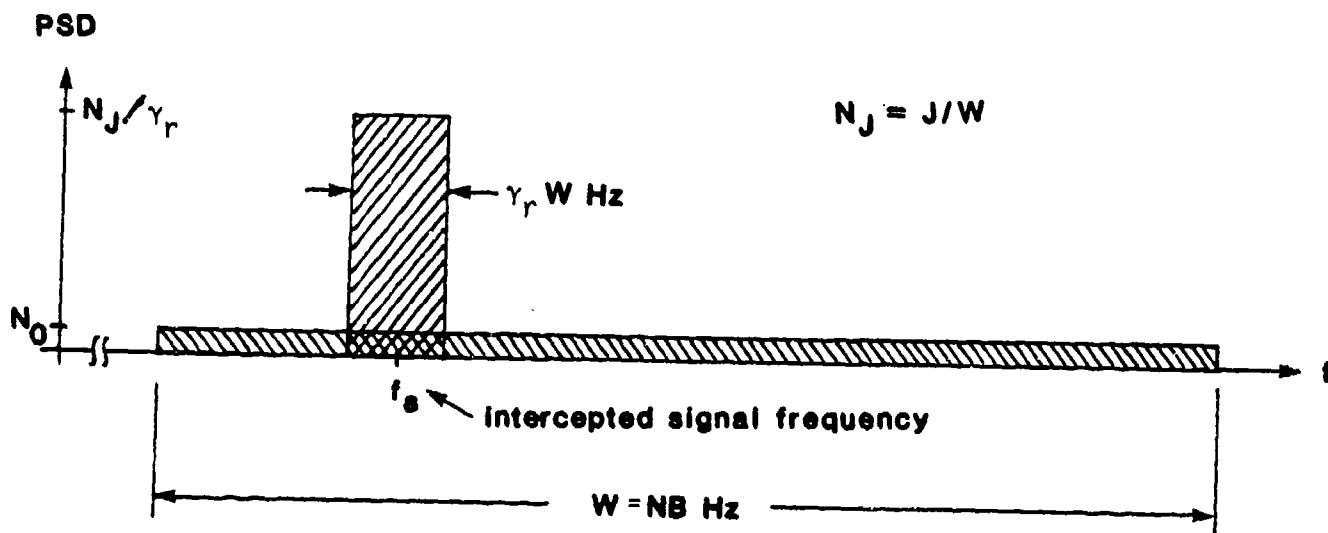
7.1.3 Performance of Conventional FH/MFSK in FNJ.

For conventional FH/MFSK, the M symbol frequency slots are contiguous at RF on each hop. Under simple FNJ, therefore, all M slots are jammed on each hop. That is,

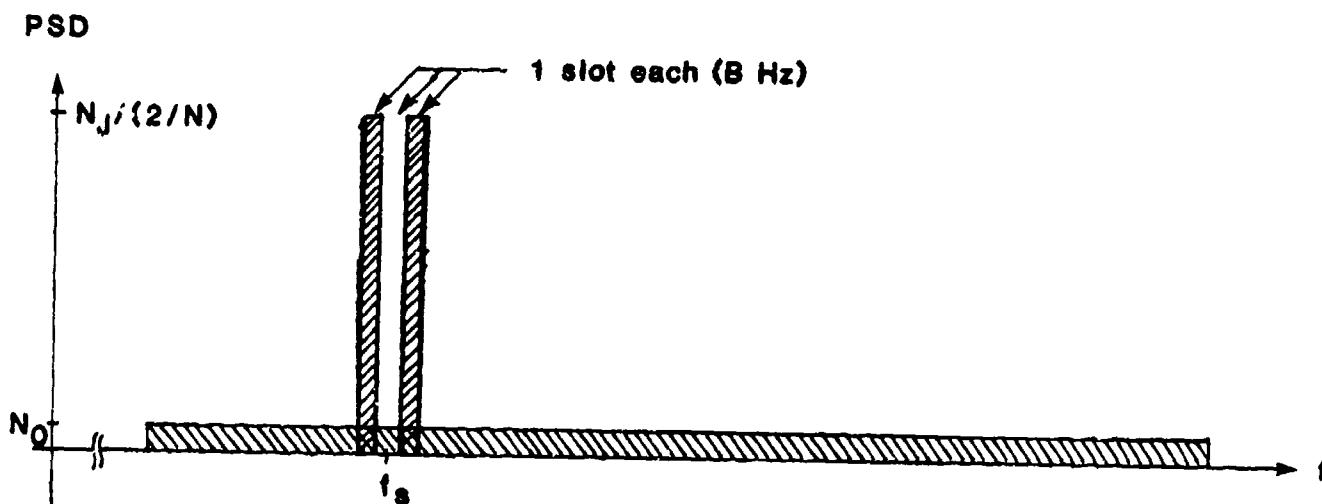
$$\ell_2 = \ell_3 = \dots = \ell_M = L. \quad (7.1-4)$$

The error probability for the system then is

$$P_b(e; \gamma_r) = \frac{M/2}{M-1} P_s(e; \gamma_r | \ell_m = L, \text{ all } m). \quad (7.1-5)$$



(a) "Simple" follow-on jammer.



(b) "Advanced" follow-on jammer.

FIGURE 7.1-1 THERMAL NOISE AND FOLLOW-ON JAMMER MODELS

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For the several receiver processing schemes we have the results using ρ_T given in (7.1-1),

$$P_b(e; \gamma_r) = \text{probability of error for broadband jamming } (\gamma=1), \\ \text{with } E_b/N_J \text{ replaced by } E_b/N_J + 10 \log_{10} \gamma_r. \quad (7.1-6)$$

Thus the error curves for roll-on jamming, $P_b(e)$ vs. E_b/N_J with $E_b/N_0 =$ constant, are those for full-band jamming, but moved to the right by $-10 \log_{10} \gamma_r$ dB.

Figure 7.1-2 illustrates the performance of conventionally-hopped FH/MFSK in follow-on noise jamming, assuming the follow-on jammer's bandwidth guarantees jamming of all M slots of the symbol. For example, if there are 2400 hopping slots and the jammer occupies 100 slots, then $\gamma_r = 100/2400 = -13.8$ dB, so that the effective E_b/N_J shown in the figure runs from about 14 dB to 64 dB. Something like 25 dB E_b/N_J gives an error rate of 10^{-2} for $L = 1$.

For the advanced FNJ, the effect on FH/MFSK is to produce the symbol error rate

$$P_s = \frac{2}{M} P_s(e | \ell_2=L; \ell_m=0, m \neq 2) \\ + (1 - \frac{2}{M}) P_s(e | \ell_2=\ell_3=L; \ell_m=0, m \neq 2, 3); \quad (7.1-7a)$$

with

$$\gamma_r = 2/N. \quad (7.1-7b)$$

Figure 7.1-3 illustrates the jammed BER for this case for $M=2, 4,$ and 8 . It is quite clear the the follow-on capability gives the jammer a tremendous advantage against FH/MFSK.

7.1.4 Performance of FH/RMFSK in Follow-on Jamming.

The FH/RMFSK hopping scheme is designed to defeat FNJ by making it difficult for the jammer to jam the nonsignal slots; the nonsignal slots are distributed randomly in the hopping band. The simple follow-on jammer

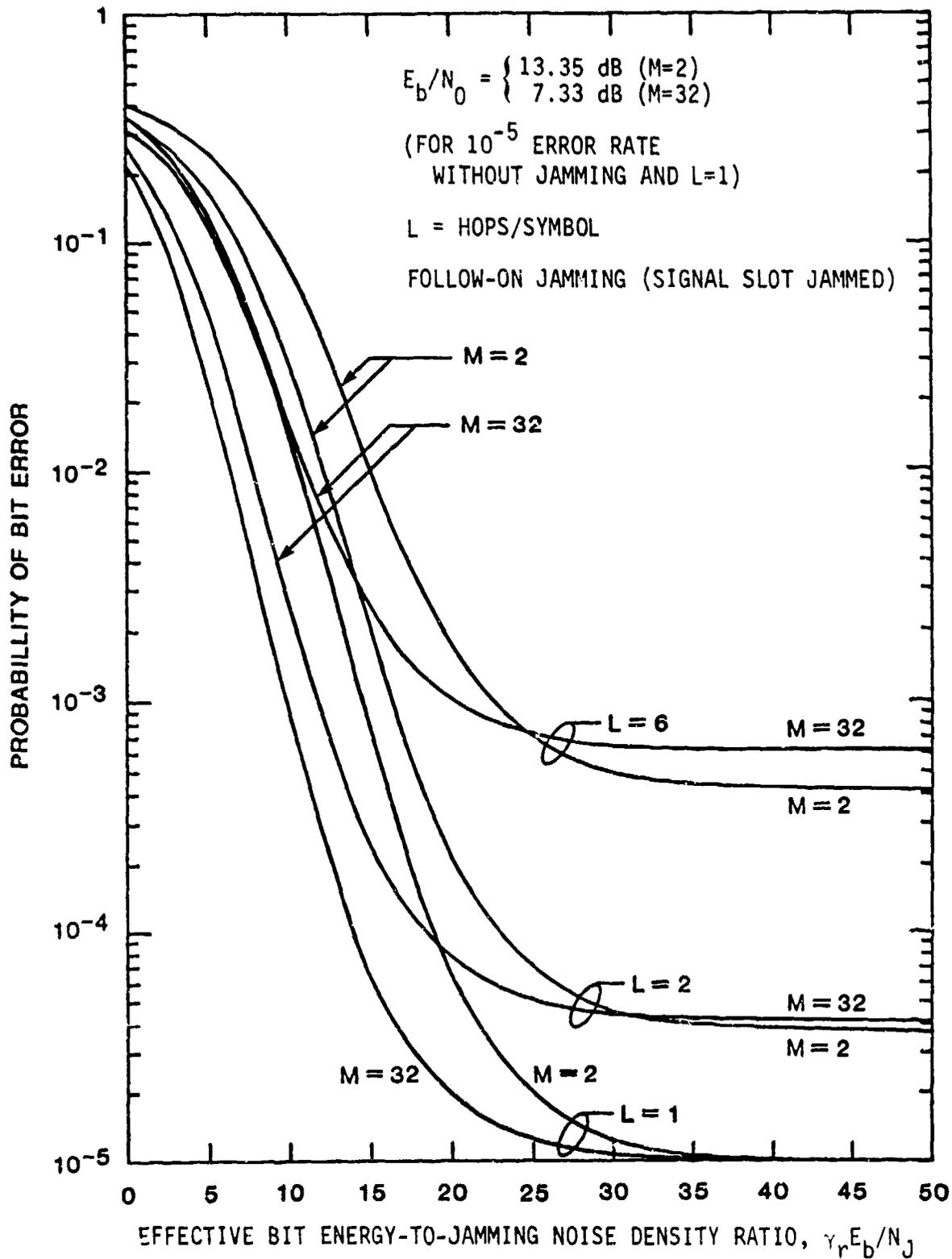


FIGURE 7.1-2 FOLLOW-ON JAMMING PERFORMANCE OF LCR AND ACJ-AGC RECEIVERS FOR FH/MFSK WITH $L=1,2,6$ AND $M=2,32$ WHEN E_b/N_0 YIELDS A 10^{-5} BER FOR $L=1$

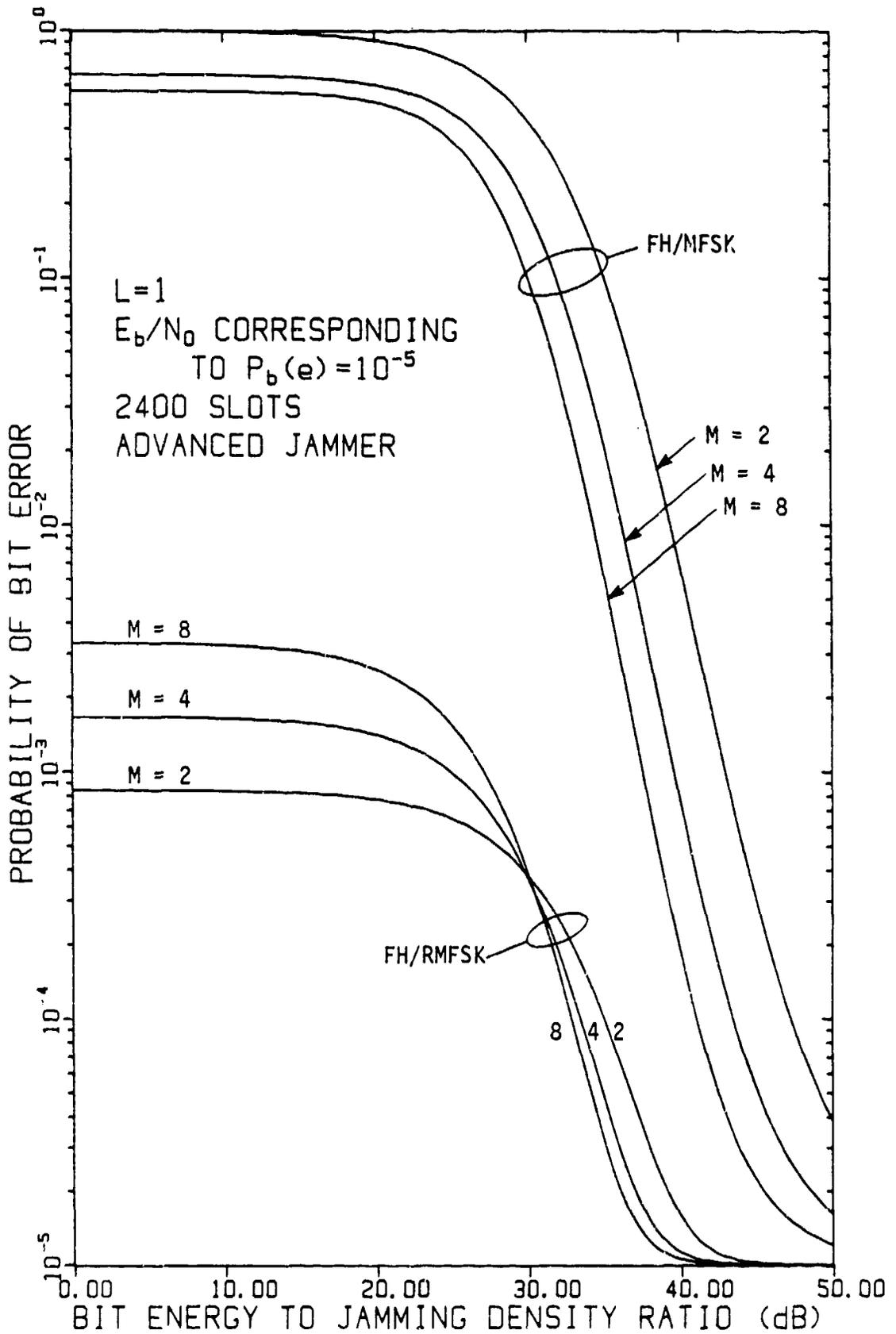


FIGURE 7.1-3 ADVANCED FOLLOW-ON NOISE JAMMING PERFORMANCE OF FH/RMFSK
 AND FH/MFSK FOR $L = 1$ HOP/SYMBOL AND UNJAMMED BER = 10^{-5}

can very likely help the receiver by placing more RF power into the signal's slot.

For $M \ll N$, that is, for a very wide hopping band compared to the symbol bandwidth, the effect of FNJ on FH/RMFSK is approximately to guarantee that

$$\begin{aligned} \ell_1 &= L \\ \ell_2 &= \ell_3 = \dots = \ell_M = 0. \end{aligned} \tag{7.1-8}$$

This has two effects. First, the per-hop SNR ρ_T is decreased since additional noise is inserted in the signal channel by the jammer. Second, the scaling of the square-law envelope samples increases for the same reason. Since the average value of a single sample is

$$\begin{aligned} E\{z_{1k}\} &= 2(\sigma_T^2 + S), \\ E\{z_{mk}\} &= 2\sigma_N^2, \quad m \geq 2, \end{aligned} \tag{7.1-9}$$

on the whole we anticipate that repeat jamming will increase the probability that the receiver will make a correct decision.

7.1.4.1 Soft-decision receivers.

For the various soft-decision receivers studied, under simple FNJ the decision variables are described as follows:

Linear combining (conventional) receiver

$$\begin{aligned} z_1 &= \sigma_T^2 \chi^2(2L; 2L\rho_T) \\ z_m &= \sigma_N^2 \chi^2(2L), \quad m \geq 2. \end{aligned} \tag{7.1-10}$$

AGC - individual channel normalization receiver (IC)

$$\begin{aligned} z_1 &= \chi^2(2L; 2L\rho_T) \\ z_m &= \chi^2(2L), \quad m \geq 2. \end{aligned} \tag{7.1-11}$$

AGC - any channel jammed normalization receiver (ACJ)

$$z_1 = x^2(2L, 2L\rho_T)$$

$$z_m = \frac{\sigma_N^2}{\sigma_T^2} x^2(2L), \quad m \geq 2. \quad (7.1-12)$$

Clipper receiver

$$z_1 = \sum_{k=1}^L [\sigma_T^2 x^2(2; 2\rho_T)] \text{ clip at } n$$

$$z_m = \sum_{k=1}^L [\sigma_N^2 x^2(2)] \text{ clip at } n. \quad (7.1-13)$$

Since multiplication of all channels by a constant factor does not affect the error probability, we observe that the conventional and AGC-ACJ receivers will achieve the same performance in FNJ:

$$P_b(e; \gamma_r)_{ACJ} = \frac{M/2}{M-1} \sum_{k=1}^{M-1} \binom{M-1}{k} \frac{(1)^{k+1}}{(1+kK)^L} \exp \left\{ -\frac{kK\rho_T}{1+kK} \right\} \\ \times \sum_{r=0}^{k(L-1)} C(k, r) \left(\frac{K}{1+kK} \right)^r \mathcal{L}_r^{L-1} \left(\frac{-L\rho_T}{1+kK} \right), \quad (7.1-14a)$$

where

$$K \triangleq \sigma_T^2 / \sigma_N^2, \quad (7.1-14b)$$

$$C(k, r) = \frac{1}{r} \sum_{n=1}^{\min(r, L-1)} \binom{r}{n} [(k+1)n-r] C(k, r-n), \quad (7.1-14c)$$

$$C(k, 0) = 1,$$

and $\mathcal{L}_r^{L-1}(\cdot)$ denotes the generalized Laguerre polynomial of degree r with parameter $L-1$.

The performance for the AGC receiver with individual channel normalization is the same as given by (7.1-14), but with $K=1$.

$$P_b(e; \gamma_r)_{IC} = P_b(e; \gamma_r)_{ACJ} \Big|_{K=1}. \quad (7.1-15)$$

The performance of the clipper receiver in follow-on jamming and FH/RMFSK is

$$P_b(e; \gamma_r)_C = 1 - \int_0^{L n^*} dx f_L(x; L) \left[G_L(x; 0) \right]^{M-1} - \left[Q(\sqrt{2\rho_T}, \sqrt{n^*/\sigma_T^2}) \right]^L \frac{e^{L n^*/2\sigma_T^2}}{M} \left[1 - \left(1 - e^{-L n^*/2\sigma_T^2} \right)^M \right], \quad (7.1-16)$$

where n^* is the optimum clipping threshold. For $L=1$, this threshold is infinite, causing the clipper receiver to have the same performance as the ACJ receiver under follow-on jamming for that case.

Figure 7.1-4 illustrates the performance of the randomly-hopped FH/RMFSK receivers against simple FJ for $L=1$. We observe that the error probability is maximized for a particular value of $\gamma_r E_b/N_j$; for the binary case, this value is slightly greater than 0 dB, while for $M=4$ and $M=8$, it is approximately -2.5 dB and -4.0 dB, respectively, for the assumed values of E_b/N_0 (chosen to achieve 10^{-5} error rate without jamming). It is interesting to note that for very strong jamming (to the left of the maximum error), the error rate increases with M as does the maximum error. Using the example of the last subsection, a 10^{-2} error rate is achieved for an E_b/N_j of about 17 dB for $\gamma_r = -14$ dB, an 8 dB improvement over conventional hopping.

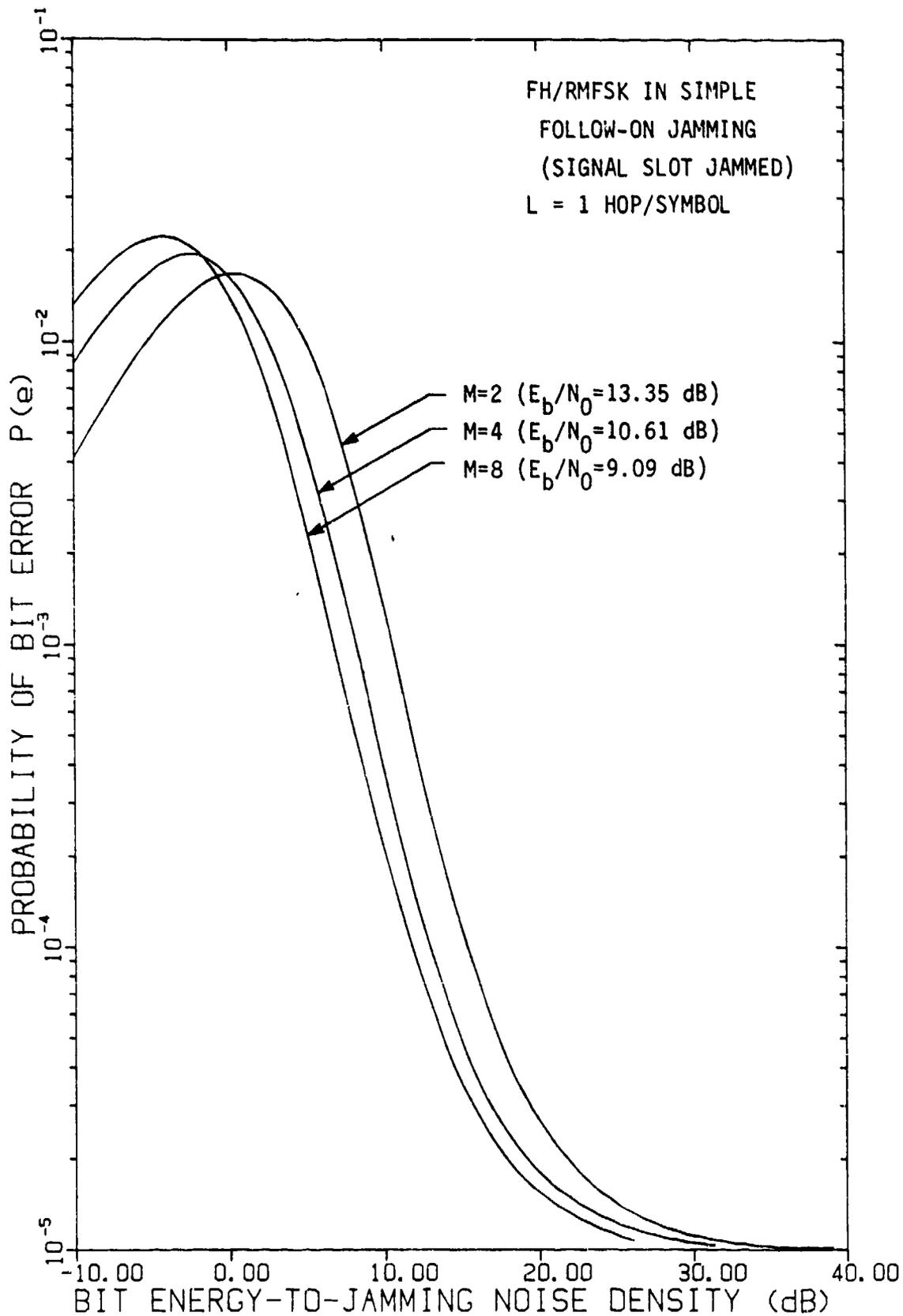


FIGURE 7.1-4 PERFORMANCE OF FH/RMFSK RECEIVERS IN SIMPLE FOLLOW-ON NOISE JAMMING VS. EFFECTIVE BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR L=1 HOP/SYMBOL AND M=2,4,8 WHEN E_b/N_0 YIELDS A 10^{-5} BER WITHOUT JAMMING

Figure 7.1-5 demonstrates that the same critical jammer power effect is observed for $L=2$, but at different values (approximately 3 dB higher), so that the jammer must know L in order to be effective. It is also evident from this figure that the maximum error rate is decreased by increasing L to 2.

For the advanced FNJ described above, using RMFSK hopping, there is only a slight chance of jamming the symbol on a given hop. The possible jamming events for a single hop are the vectors

$$\underline{v}_k = (0, v_{2k}, v_{3k}, \dots, v_{Mk}) \quad (7.1-17)$$

with probabilities

$$\begin{aligned} & \text{Pr}\{r \text{ nonsignal slots jammed}\} \\ &= \binom{N-M}{2-r} / \binom{N-1}{2} \binom{M-1}{r}. \end{aligned} \quad (7.1-18)$$

For $M=2$ the result is

$$P_b(e) = \sum_{k=0}^L \binom{L}{k} \left(\frac{2}{N-1}\right)^k \left(1 - \frac{2}{N-1}\right)^{L-k} P_b(e|\ell_1=0, \ell_2=k). \quad (7.1-19a)$$

$$= P_b(e|\ell_1=\ell_2=0) + \frac{2L}{N-1} P_b(e|\ell_1=0, \ell_2=1), \quad N-1 \gg 2. \quad (7.1-19b)$$

Thus the FH/RMFSK hopping scheme achieves very nearly the unjammed error performance of MFSK when the follow-on noise jammer is configured against FH/MFSK. As (7.1-19b) shows, for $L=1$ and $M=2$ the unjammed error rate is increased by, at most, $2/(N-1)$; this quantity equals 8.3×10^{-4} for $N=2400$, and 7.8×10^{-3} for $N=256$. For $L=1$, Figure 7.1-3 shows the performance of FH/RMFSK much improved over FH/MFSK in advanced FNJ.

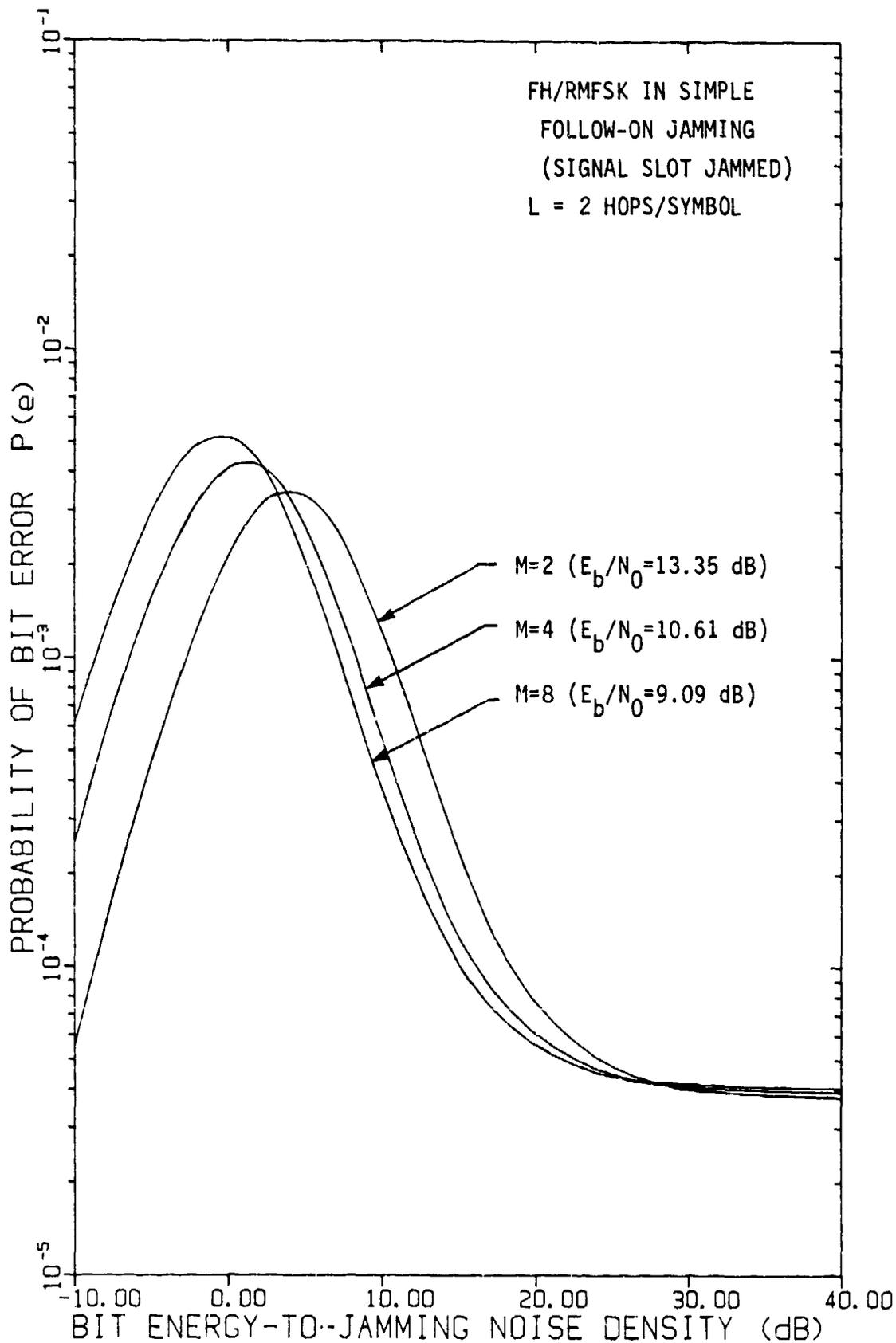


FIGURE 7.1-5 PERFORMANCE OF FH/RMFSK RECEIVERS IN SIMPLE FOLLOW-ON NOISE JAMMING VS. EFFECTIVE BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR L=2 HOPS/SYMBOL AND M=2,4,8 WHEN E_b/N_0 YIELDS A 10^{-5} BER WITHOUT JAMMING

An exception to Figure 7.1-3 for FH/RMFSK is the IC-AGC receiver, for which the "advanced" FNJ is completely nullified. The jammer in this case is "too smart," because the IC-AGC receiver is vulnerable to jamming only if the signal channel is jammed. This statement also holds for the case of conventional MFSK hopping if individual-channel normalization is employed.

7.1.4.2 Hard-decision receiver.

Performance of the hard-decision receiver in simple follow-on noise jamming is depicted in Figures 7.1-6 through 7.1-8 for values of $M=2, 4, \text{ and } 8$ respectively. The parameter E_b/N_0 was chosen to yield a 10^{-5} BER in the absence of jamming per respective M value for $L=1$ hop per symbol. It is clearly seen in each of the L $P(e)$ curves that as the jammer power is increased below that E_b/N_j value to cause maximum $P(e)$, a decrease in $P(e)$ takes place. Hence, for strong jamming the jammer is actually aiding the communicator by the addition of energy to the non-coherent FSK signal slot. We also have a diversity improvement for $L \geq 3$ hops per symbol in the strong jamming regions. Conversely, for weak jamming (beyond $E_b/N_j \approx 20$ dB) no diversity improvement is realized for $L \geq 2$ hops per symbol due to the dominance of the noncoherent combining loss existing for the stated thermal noise (E_b/N_0) values. Therefore, in order to be effective (maximum $P(e)$), the jammer must maintain E_b/N_j to within small ranges. For example, to ensure a minimum $P(e)$ of 10^{-2} for $M=2$ and $L=1$ (Figure 7.1-4), E_b/N_j must be held to values ranging from 4 to 14 dB.

The effects of decreased thermal noise levels ($E_b/N_0 = 20$ dB) for cases of $M=2, 4, \text{ and } 8$ are illustrated in Figures 7.1-9 through 7.1-11 respectively. Here we observe all of the L $P(e)$ curves exhibiting a

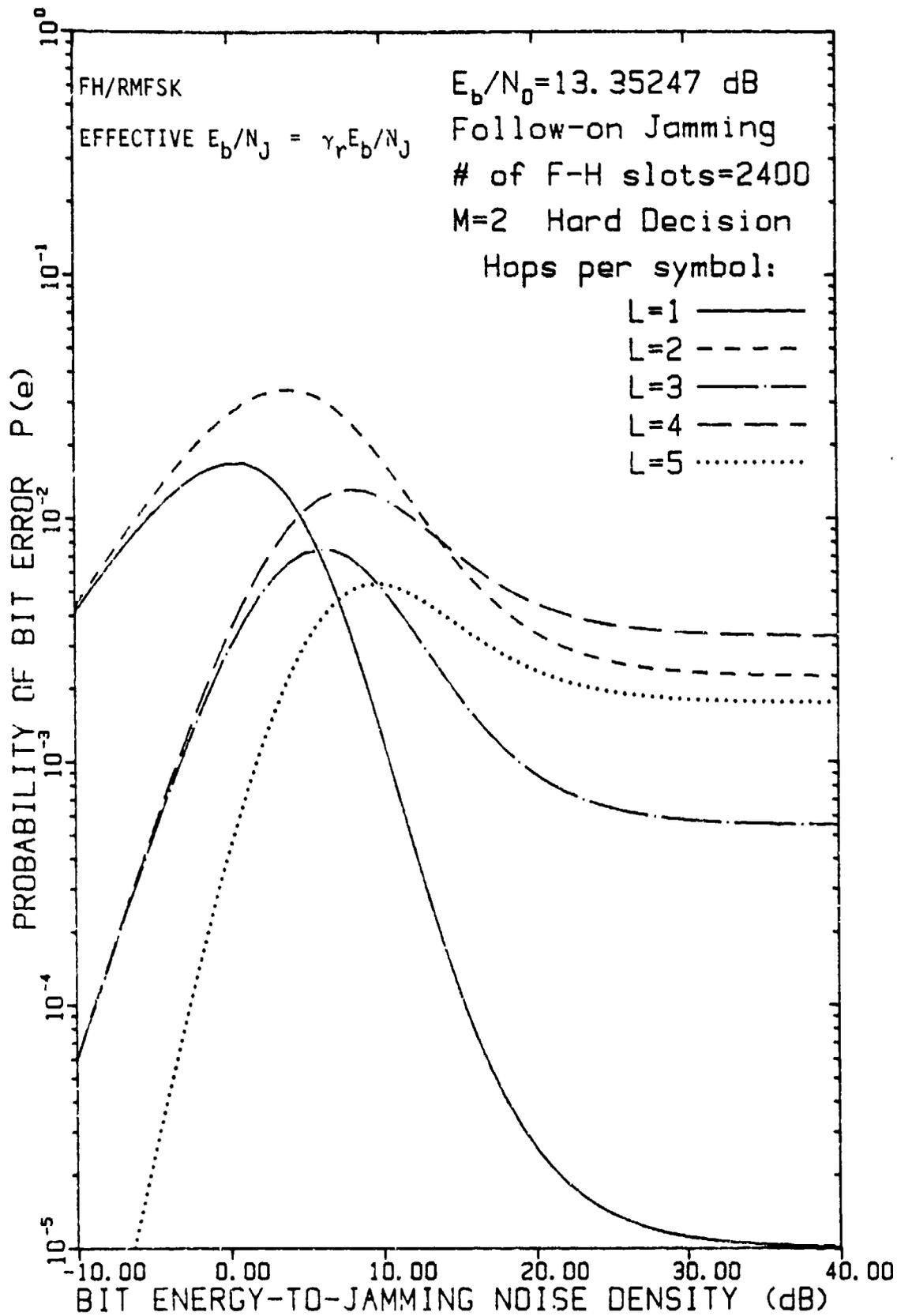


FIGURE 7.1-6 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNI VS. EFFECTIVE E_b/N_J FOR M=2 AND E_b/N_0 SUCH THAT UNJAMMED BER IS 10^{-5} WHEN L=1

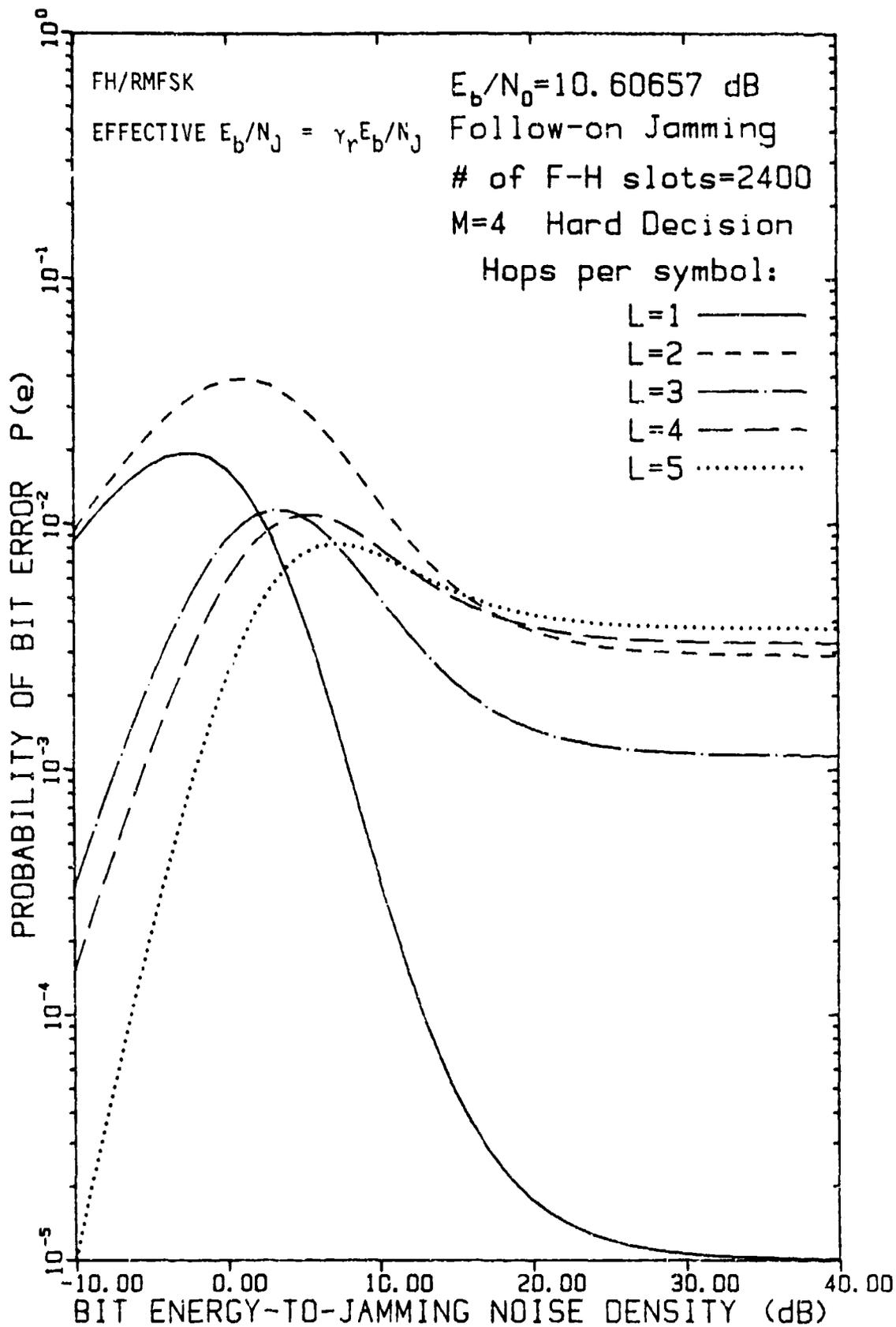


FIGURE 7.1-7 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNJ VS. EFFECTIVE E_b/N_J FOR $M=4$ AND E_b/N_0 SUCH THAT UNJAMMED BER IS 10^{-5} WHEN $L=1$

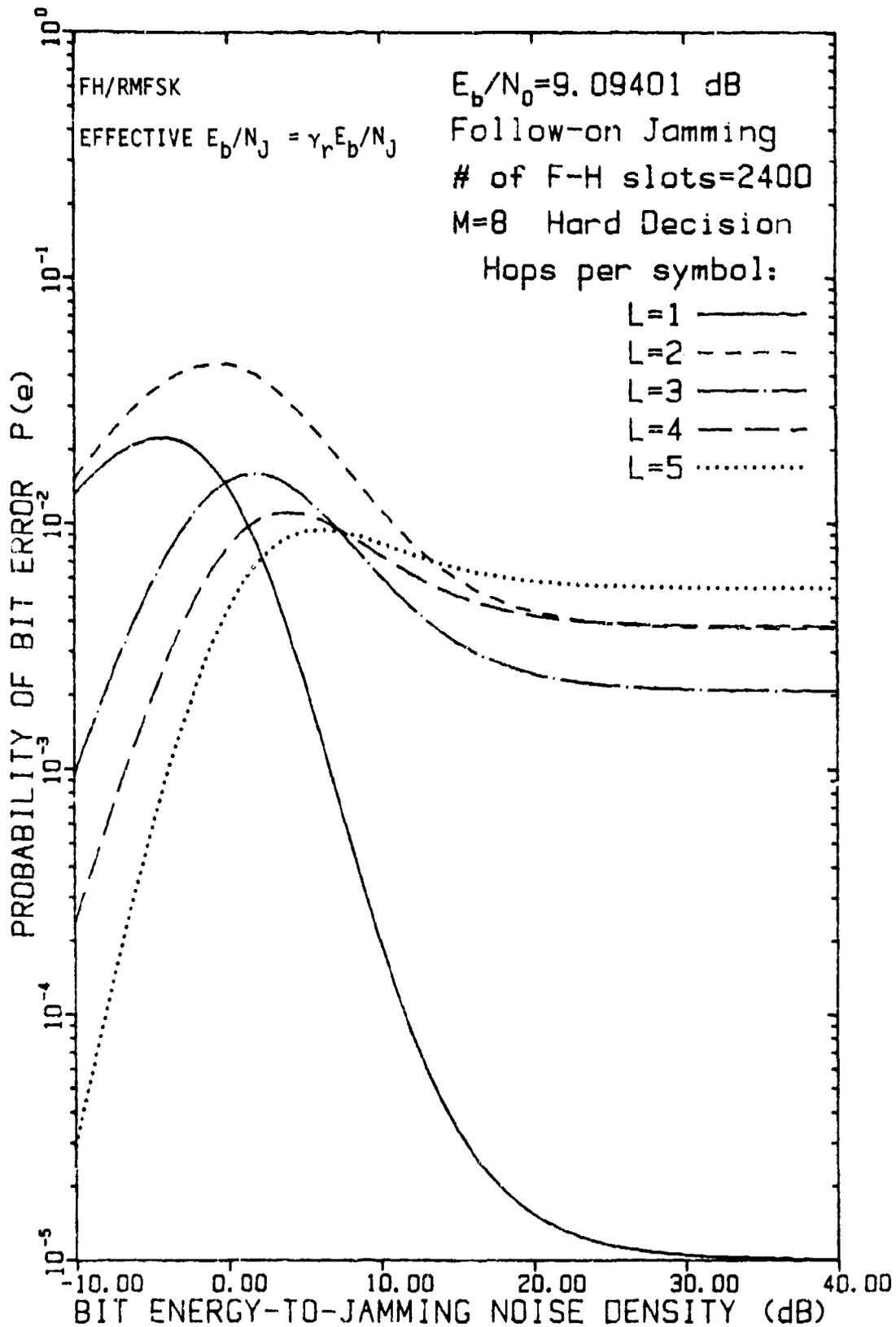


FIGURE 7.1-8 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNJ VS. EFFECTIVE E_b/N_J FOR $M=8$ AND E_b/N_0 SUCH THAT UNJAMMED BER IS 10^{-5} WHEN $L=1$

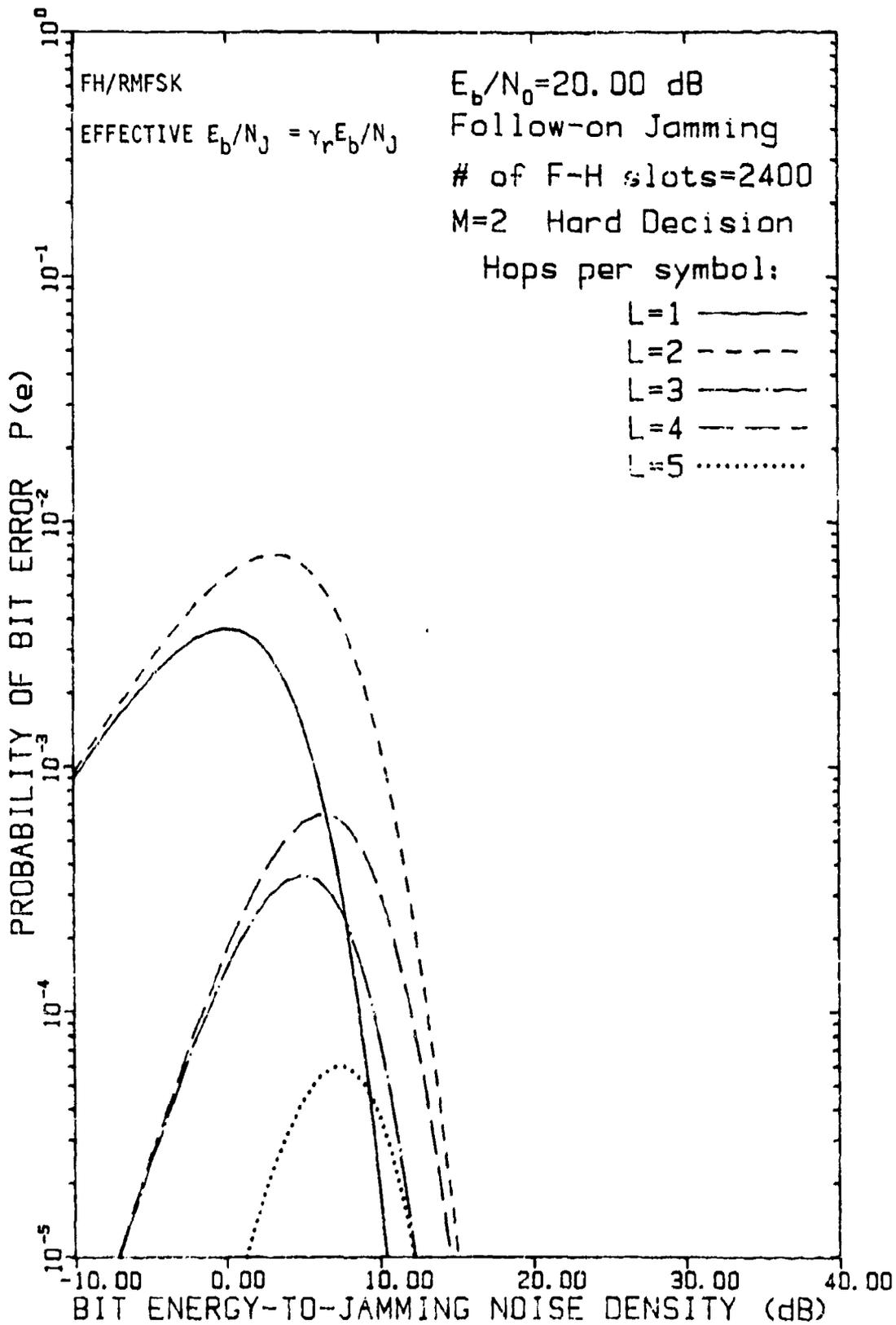


FIGURE 7.1-9 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNJ VS. EFFECTIVE E_b/N_J FOR $M = 2$ AND $E_b/N_0 = 20$ dB

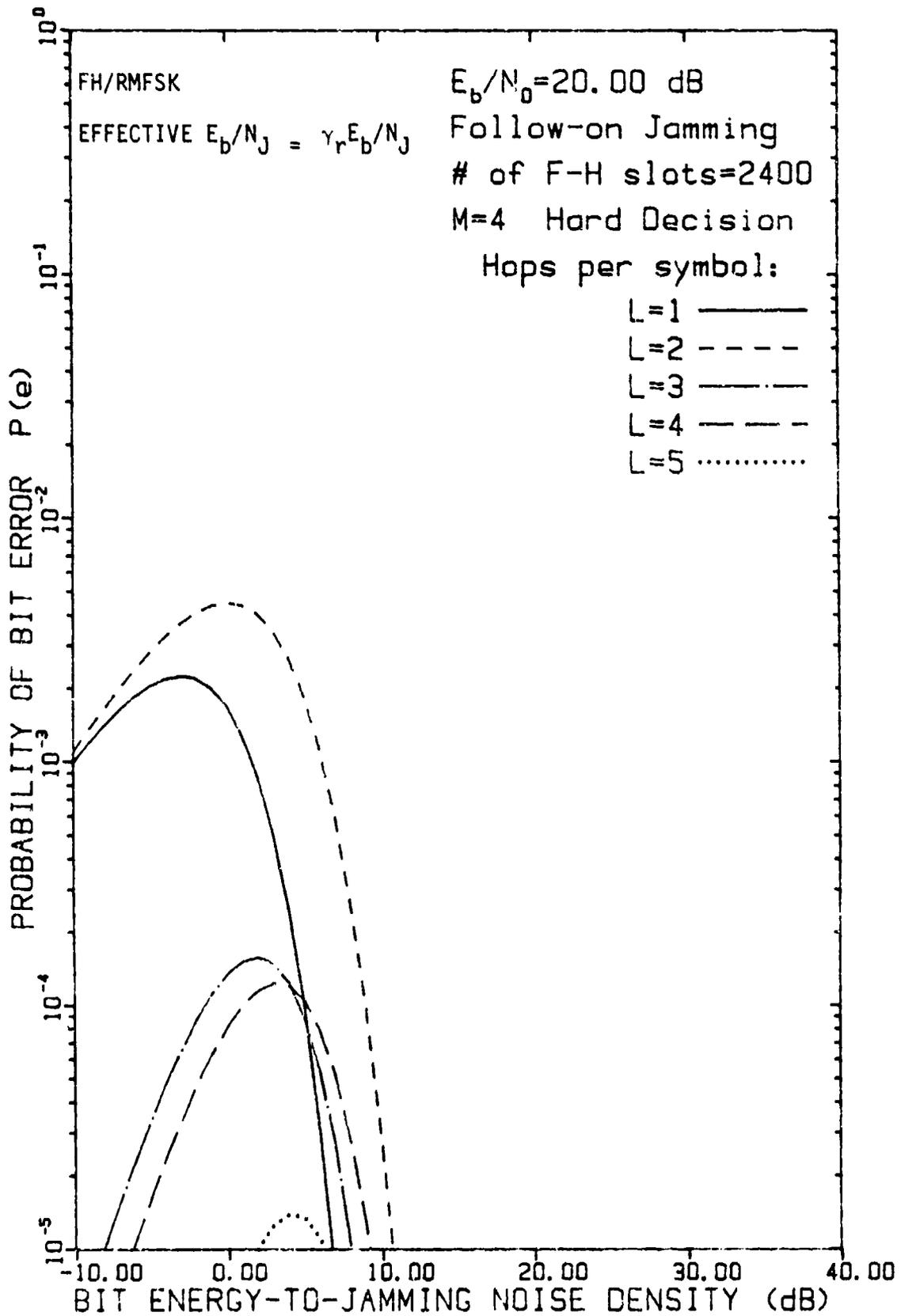


FIGURE 7.1-10 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNJ VS. EFFECTIVE E_b/N_J FOR M=4 AND $E_b/N_0 = 20$ dB

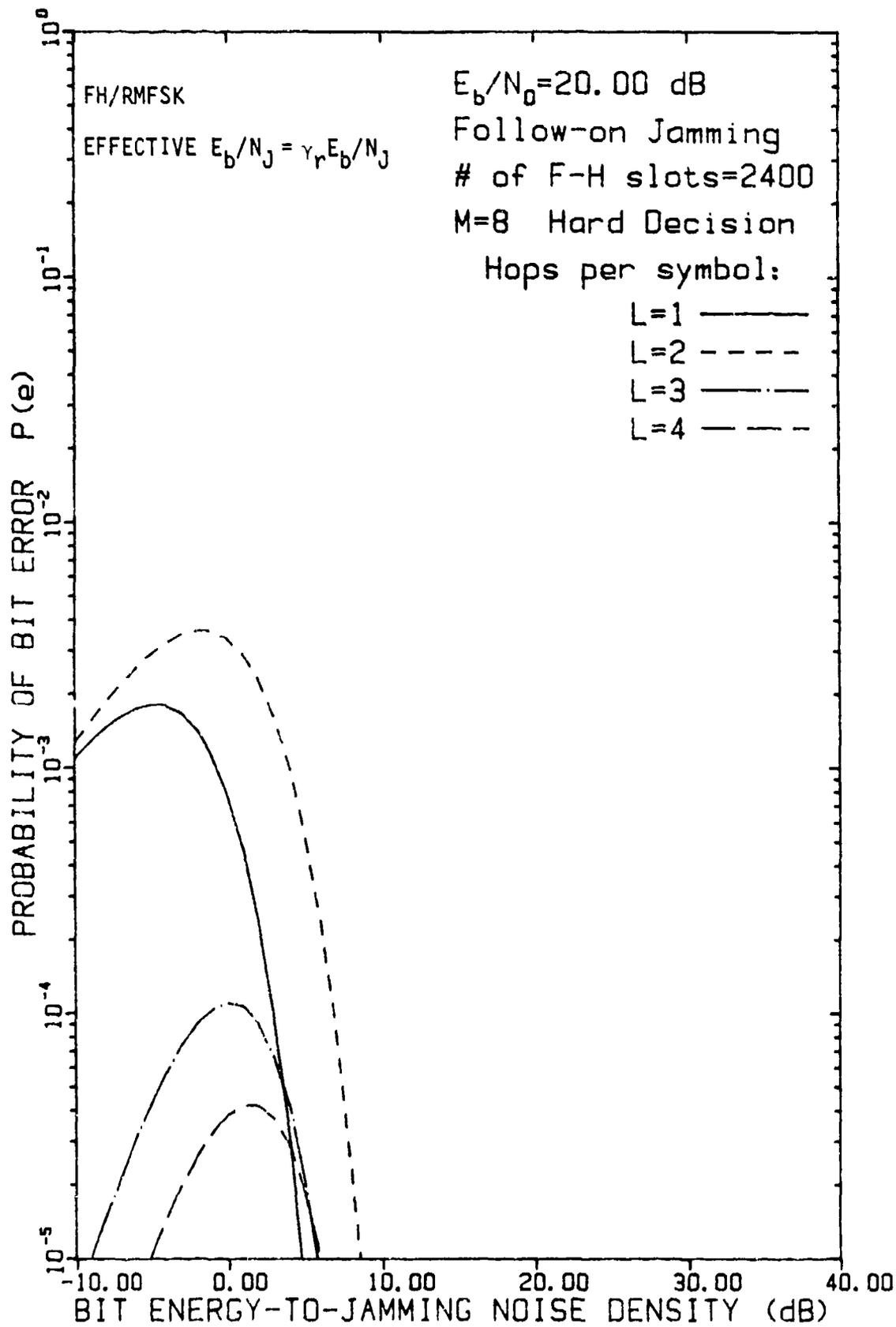


FIGURE 7.1-11 PERFORMANCE OF HARD-DECISION FH/RMFSK RECEIVER IN SIMPLE FNJ VS. EFFECTIVE E_b/N_J WHEN $M=8$ AND $E_b/N_0 = 20$ dB

"parabolic" type of behavior, i.e. steeply defined strong and weak jamming regions. Thus, the jammer appears to have quite a narrow window of E_b/N_J values to work within for attaining a maximum effect.

7.2 COMPARISONS OF FH/RMFSK RECEIVER PERFORMANCES IN WORST-CASE PARTIAL-BAND NOISE JAMMING (WCPBNJ)

It is understood that the motivation for using the proposed FH/RMFSK waveform is that it is less vulnerable to follow-on noise jamming (FNJ) or repeat noise jamming than is a conventional FH/MFSK block-hopping system. RMFSK is effective in that the FNJ is not able to place jamming power in the unused symbol frequency slots as is the case for MFSK where the M signalling frequencies are adjacent. At the FH/MFSK or FH/RMFSK receiver, the L hops comprising the MFSK symbol can be combined in a number of ways. Certain types of nonlinear combining soft-decision schemes, which weight the detected hops in some form to discriminate against jammed hops, are employed. Previous results [1] have shown that conventional FH/MFSK system performance with L -hop diversity in WCPBNJ is improved by nonlinear combining techniques. This study has addressed FH/RMFSK system performance in the less sophisticated, yet more pervasive, ECM tactic of PBNJ - a basic jamming threat which is inevitably encountered in an EW scenario. In Sections 3 through 6 it has been demonstrated that nonlinear combining yields improved RMFSK performance in this type of jamming.

In what follows, we compare performances of the different types of ECCM receivers for FH/RMFSK signals in the WCPBNJ environment. In Section 7.3 we also compare FH/MFSK and FH/RMFSK receiver performances in WCPBNJ, and in Section 7.4 consider the different effects of the RMFSK diversity combining

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techniques studied. Unless stated otherwise, we note that all E_b/N_0 values utilized in performance plots are chosen so as to yield a 10^{-5} BER in the absence of jamming when $L=1$ hop per symbol.

Comparisons among the different FH/RMFSK receivers analyzed are provided by Figures 7.2-1 and 7.2-2 for $M=2$ and $E_b/N_0 = 13.35$ dB, and by Figures 7.2-3 and 7.2-4 for $M=4$ at $E_b/N_0 = 10.61$ dB. The enormous amount of computer time required to obtain performance results for the clipper receiver at $L=3$ and the SNORM receiver for $M=4$ is beyond the scope of this study. Therefore, in some figures for comparison, these receivers are not represented. Explanations of the difficulties involved in such calculations were presented in the numerical results of Section 5 (clipper receiver) and Section 6 (SNORM receiver).

We can develop a performance ranking for these receivers with their respective parameter sets (M, L values) by assessing performances in the regions of strong, moderate, and weak jamming. These arbitrary regions are taken to mean the following: (1) strong jamming - usually full-band jamming with E_b/N_j values less than about 4 to 8 dB, (2) weak jamming - region of very small γ values with E_b/N_j usually greater than 35 to 40 dB, and (3) moderate jamming - area between strong and weak jamming.

For the case of $M=2, L=2$ (Figure 7.2-1), we find receiver performances asymptotically approaching two groups in the strong jamming region. These are (1) lower $P(e)$: IC-AGC, ACJ-AGC, clipper, and SNORM; (2) higher $P(e)$: hard-decision (HD), and square-law linear combining receiver (LCR). We observe that the first group is more effective due to their nonlinear weighting (normalization) schemes. In the second group, we have the LCR

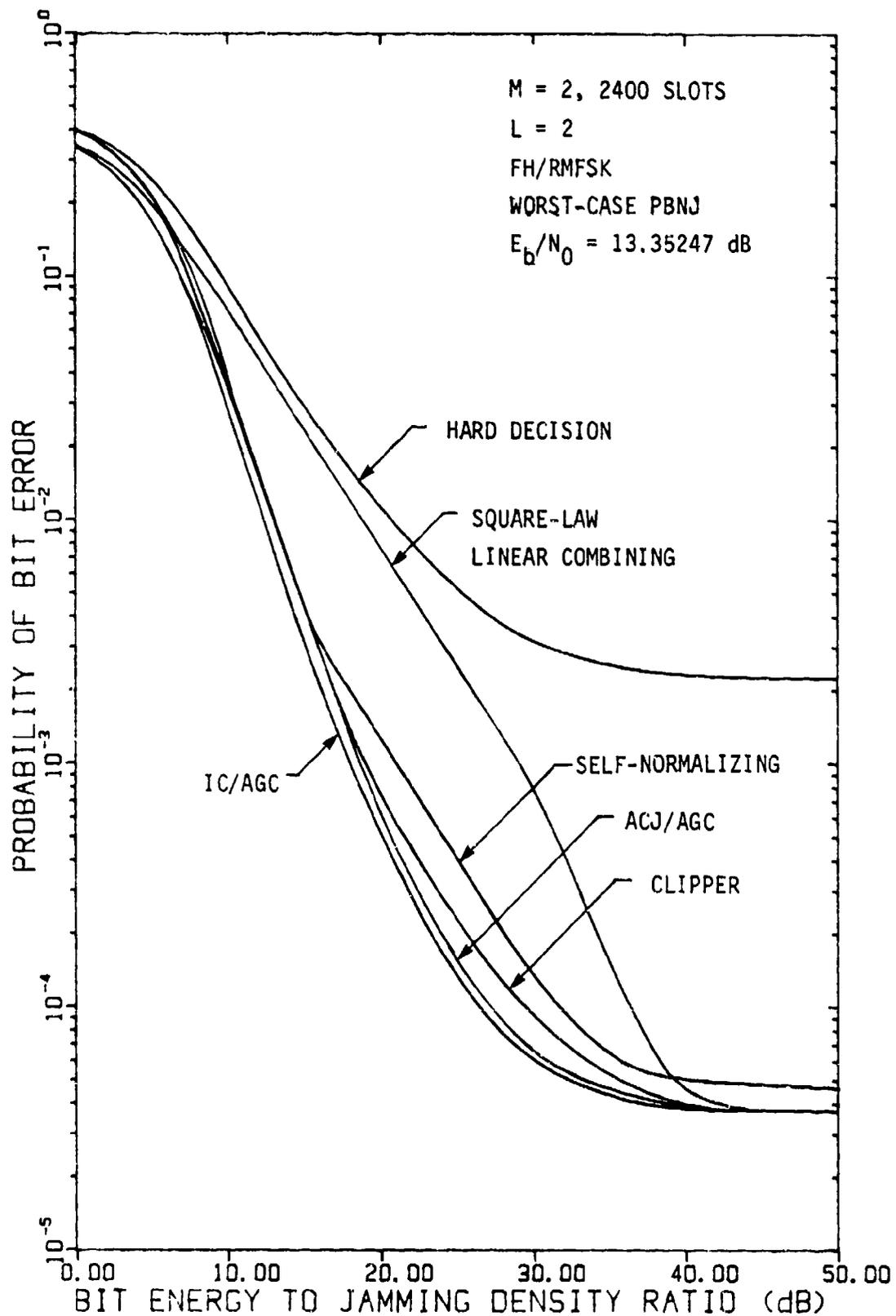


FIGURE 7.2-1 COMPARISON OF DIFFERENT RECEIVERS FOR FH/RMFSK WITH $M=2$ AND $L=2$ HOPS/SYMBOL WHEN $E_b/N_0 = 13.35247 \text{ dB}$

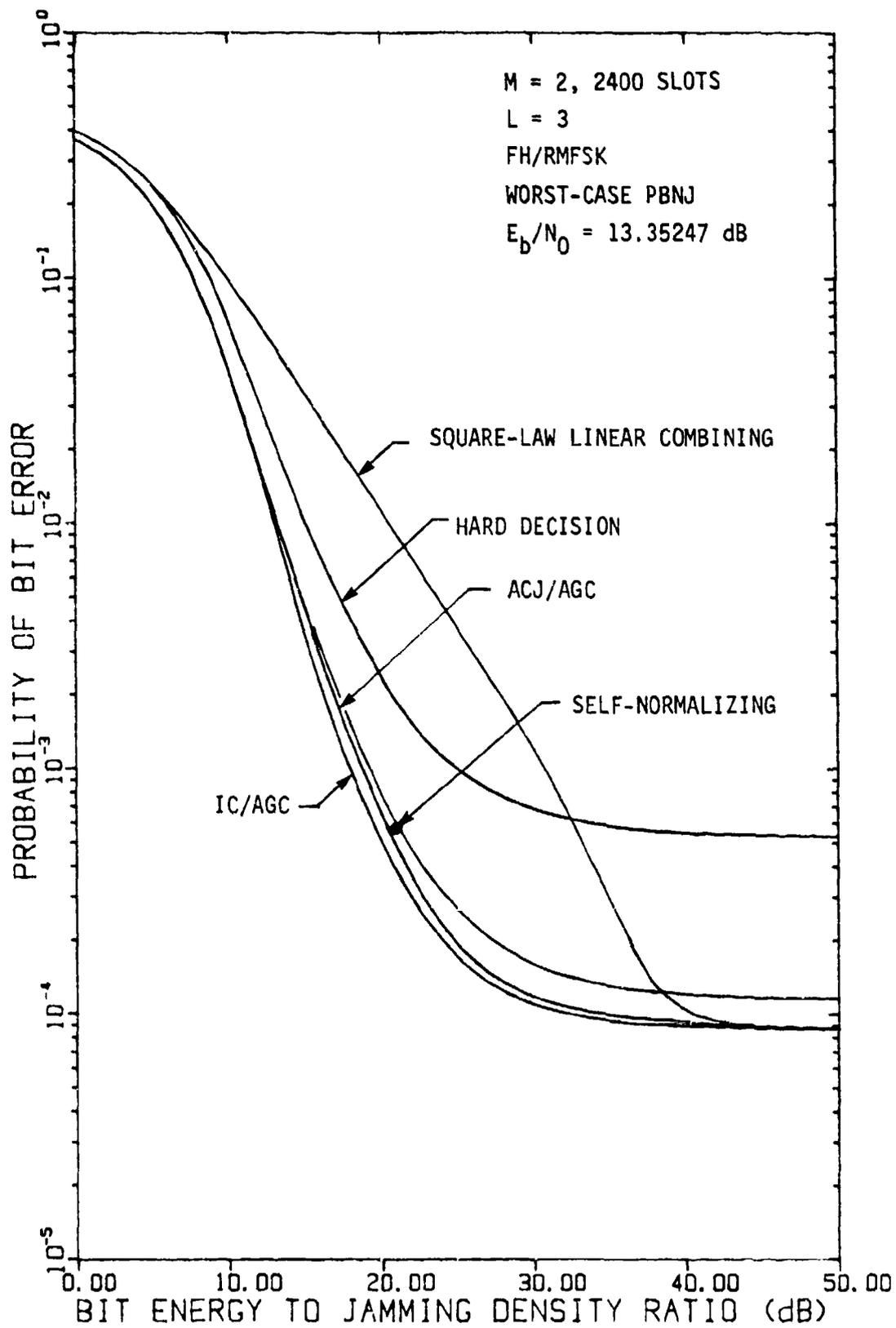


FIGURE 7.2-2 COMPARISON OF DIFFERENT RECEIVERS FOR FH/RMFSK WITH $M=2$ AND $L=3$ HOPS/SYMBOL WHEN $E_b/N_0 = 13.35247 \text{ dB}$

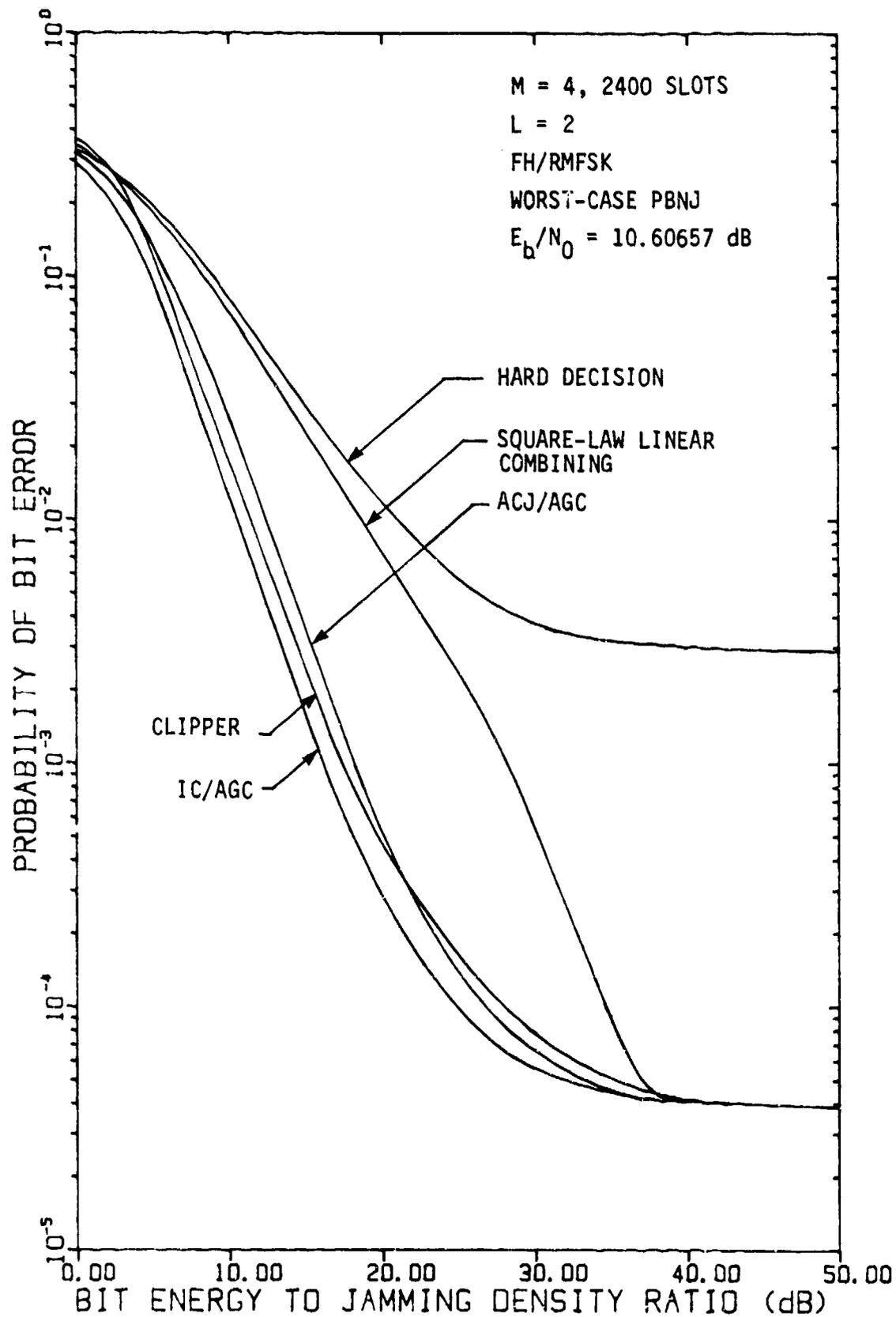


FIGURE 7.2-3 COMPARISON OF DIFFERENT RECEIVERS FOR FH/RMFSK WITH $M=4$ AND $L=2$ HOPS/SYMBOL WHEN $E_b/N_0 = 10.60657 \text{ dB}$

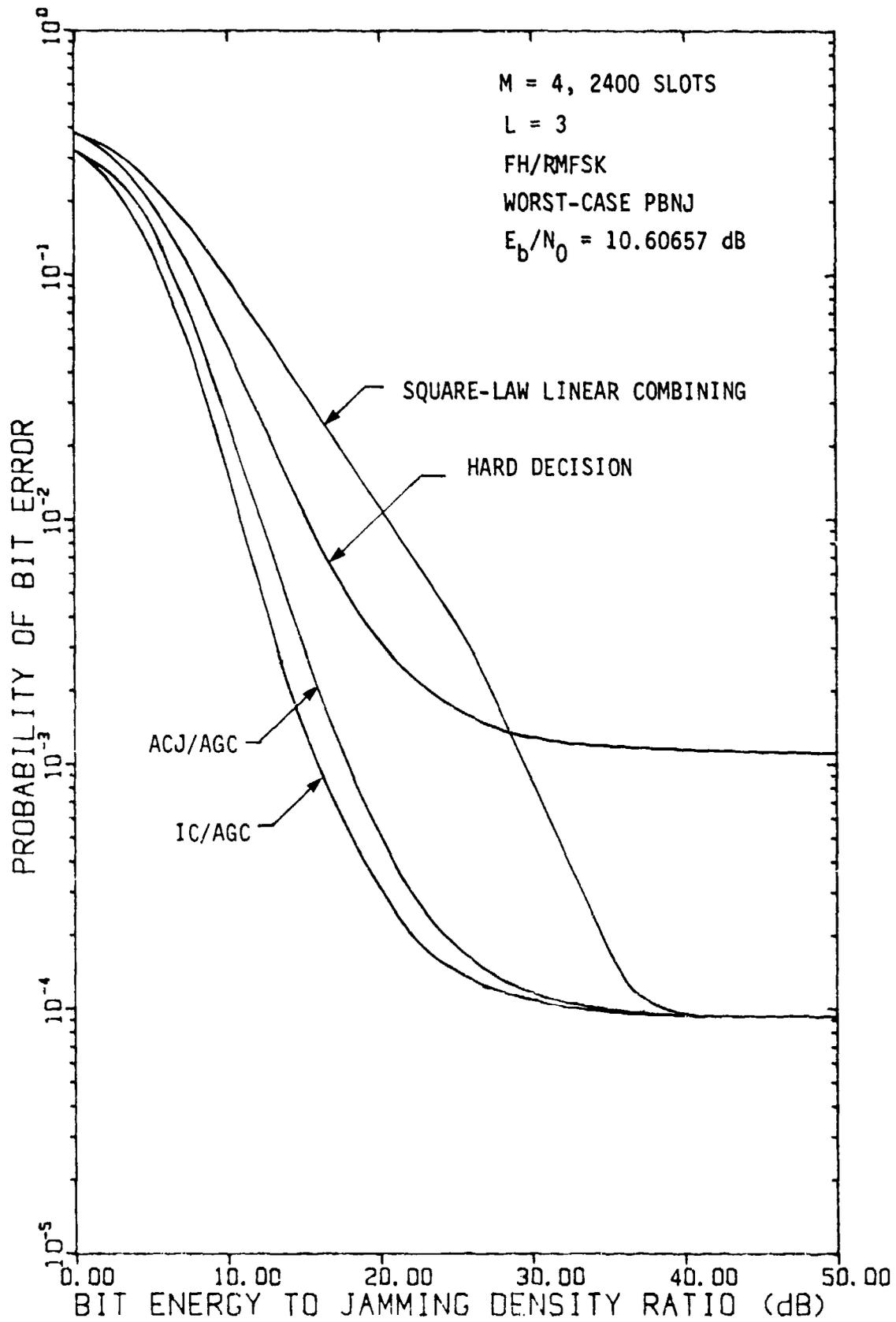


FIGURE 7.2-4 COMPARISON OF DIFFERENT RECEIVERS FOR FH/RMFSK WITH $M=4$ AND $L=3$ HOPS/SYMBOL WHEN $E_b/N_0 = 10.60657 \text{ dB}$

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which provides no weighting and the HD receiver which is ineffective for $L=2$ because it is subject to the possibility of a tie (for L even) in the final quantized decision variable values. In strong jamming, all receivers are experiencing full-band jamming at high power levels and discriminating against a jammed hop is not done by any receiver since all hops are jammed. It is only when the optimum γ values begin to fall below full-band jamming that we realize the performance improvement of the nonlinear combining techniques. In the moderate jamming region, we see a sub-division among the receivers which we termed "effective" for strong jamming: (a) "ideal" receivers (AGC) and (b) "practical" receivers (clipper, SNORM). In this region we notice performance results for the nonlinear combining types remaining within about 1 dB of each other up until around the point where $E_b/N_J > E_b/N_0$. That is, where thermal noise becomes more dominant than jamming noise. At these values, the SNORM and clipper receiver performances begin to degrade relative to the AGC types, yet still remain superior to the square-law LCR. We note the AGC receivers as maintaining a continuous and graceful transition in this moderate jamming region with the IC-AGC showing a slightly better performance. The worse performance for the ACJ-AGC is due to an imbalancing effect of the ACJ normalization scheme in which all receiver channels are inversely weighted by the largest of all the M channel noise powers. In contrast, the IC-AGC receiver normalizes each channel separately and thereby "balances" or "equalizes" the received noise powers.

The performance "breakaway" for $N_0 > N_J$ for the clipper and SNORM receivers reflects the dominance of the "unbalanced symbol" error mechanism

as γ becomes smaller; jamming in more than one M-ary channel becomes unlikely. In the case of the clipper, the receiver tends to limit the unbalancing contribution to the sum without affecting the signal channel sum. But for the SNORM receiver, any input noise power unbalance due to one channel being jammed reduces the signal channel sum. This is because the SNORM normalization weight is inversely proportional to the total noise power (i.e. sum of all M channels) measurement on a given hop without recognition of which individual channels are jammed.

In the weak or no jamming region, all receivers suffer degradation due to the noncoherent combining loss (NCL) when $L > 1$. As E_b/N_j approaches 50 dB (practically no jamming), the different performances of the receivers in the Gaussian channel are evident. All receivers suffer degradation relative to the $L=1$ result ($P(e) = 10^{-5}$) due to the NCL as Figure 7.2-1 demonstrates. We see that the SNORM and HD receivers are subject to higher NCL due, respectively, to inefficient combining and to the possibility of "tie votes" for $M=2$ and $L=2$, with the HD receiver being more severely affected because of its use of only two levels of quantization in the soft-decision.

Figure 7.2-2 compares receiver performances for the parameter set $M=2, L=3$. For strong jamming, we see a change in two groupings recognized for the case $M=2, L=2$. The HD receiver is now more effective than the square-law LCR, and provides a significant improvement in strong jamming. We attribute this to the fact that no ties exist in the majority logic decoding when L is odd. But in the weak or no jamming region the HD performance is worse than

LCR due to quantization noise effects. We likewise note that in moderate jamming, the SNORM performance has improved somewhat over the results for $M=2, L=2$. It is reasoned that for $L=3$, the multiple unbalancing effects predominant for decreasing γ values begin to become less probable. Performance rankings in the thermal-noise-limited region for SNORM remain unchanged from the case of $M=2, L=2$.

The effect of an increase in alphabet size ($M=4$) can be discerned from Figures 7.2-3 and 7.2-4. We notice that the parameter set $M=4, L=2$ (Figure 7.2-3) appears similar to the $M=2, L=2$ set where two distinct groupings are present in moderate jamming with the nonlinear combining soft-decision receiver group yielding superior performance. Considering the moderate jamming region to be from $E_b/N_j = 5$ to 39 dB, we observe that the clipper and ACJ-AGC receivers trade rankings around the regional midpoint of 22 dB; the clipper receiver showing ≤ 1 dB better performance over the range of 5 to 17 dB. However, in strong jamming we find the clipper's performance degrading to the point of being the overall worst performer at $E_b/N_j = 0$ dB. For the HD receiver, we find a worse performance than for $M=2, L=2$ because there are now two more channels allowing for the possibility of more tie decisions on the output decision variables.

In the case of $M=4, L=3$ (Figure 7.2-4), receiver performances appear similar to the behavior exhibited by the parameter set $M=2, L=3$ in that two distinguishable groups are presented. These are the AGC types (better performances) versus the square-law LCR and HD receivers. Throughout most of the strong and moderate jamming regions (2 to 35 dB), it is noticed that the ACJ-AGC performance is up to 2 dB worse than the IC-AGC; this again

being due to the unbalancing effect of the normalization weighting scheme of the ACJ-AGC receiver. The difference between the two AGC performances for $M=4$ is larger than for $M=2$ (see Section 7.3 for more discussion of this phenomenon). As for the HD receiver, it proves to be better than the HD cases for $L=2$ yet exhibits poorer performance than HD for $M=2, L=3$. Although there are no output decision variable ties for $L=3$, the HD receiver with more channels will now suffer increased quantization effects in approximating the LCR.

With regard to receiver performances in little thermal noise, Figure 7.2-5 depicts performance results of three candidates (IC-AGC, ACJ-AGC, SNORM) for the parameter set $M=2, L=2$ at $E_b/N_0 = 20$ dB. These receivers represent the previously shown most ideal performers (AGC types) and the more realizable SNORM receiver. The HD receiver, although a relatively simple ECCM diversity technique in practice, is not included in this comparative set due to the "tie" decision factor when $L=2$.

We observe for full-band jamming that the IC-AGC and ACJ-AGC receivers yield equivalent performances with the SNORM showing a slightly higher BER. Such behavior for the AGC receivers is to be expected in full-band jamming where a Gaussian channel performance is realized and discrimination against a jammed hop is nonexistent. But as γ becomes less than full-band, it is seen that the AGC receiver curves maintain the same negative slope for increasing E_b/N_j with the ACJ-AGC staying about 1 dB worse than the IC-AGC, this inferior performance being due to the previously described imbalancing effect of the ACJ-AGC normalization mechanism. For the SNORM receiver, we see its performance with respect to the AGC receivers as being: (1) about 0.5 dB

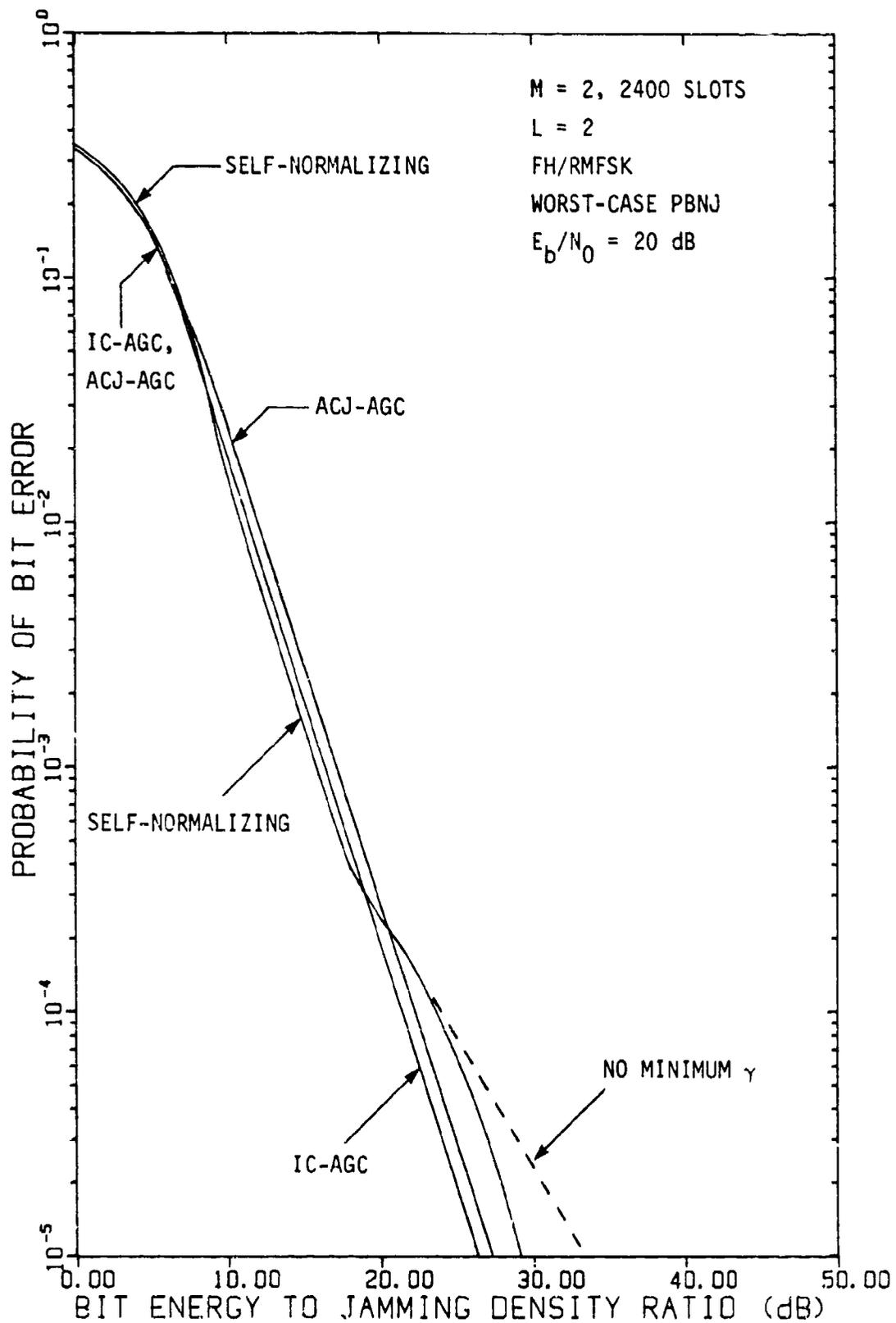


FIGURE 7.2-5 COMPARISON OF DIFFERENT RECEIVERS FOR FH/RMFSK WITH $M=2$ AND $L=2$ HOPS/SYMBOL WHEN $E_b/N_0 = 20 \text{ dB}$

worse in full-band jamming, (2) less than 1 dB better for $\gamma < 1$ (at $E_b/N_j \sim 10$ dB) up until $E_b/N_j > E_b/N_0$, and (3) more than either ACJ-AGC or IC-AGC from about 20 dB to the point where $P(e) = 10^{-5}$ is reached.

An empirical explanation of this phenomenon is obtained by comparing Figure 7.2-6 (IC-AGC) with Figure 7.2-7 (SNORM). These figures show each receiver's individual performance at $E_b/N_0 = 20$ dB for ten different values of γ ranging from $\gamma=0.001$ to $\gamma=1.0$ or full-band jamming. We note for the AGC receiver (Figure 7.2-6) that $\gamma=0.001$ and 0.002 curves produce $P(e) < 10^{-5}$ and thus do not appear on the performance plots. Upon observing the eight remaining γ -curves in these AGC plots, we see each of these $P(e)$ curves contributing the same smooth behavior toward producing an optimum γ -curve result which is a straight line for $\gamma < 1.0$, that is, a slope equal to $A/(E_b/N_j)^2$ where A is some constant defining the inverse-linear relationship existing between γ and available jamming power when $E_b/N_0 = 20$ dB.

However, for the SNORM case we find performance curves for $\gamma=0.005$ through $\gamma=0.2$ exhibiting behavior resulting in an optimum γ -curve which is not constant; over these γ ranges the SNORM performance is superior to the IC-AGC receiver. The upper envelope of the curves at first is proportional to $(E_b/N_j)^{-2}$, then transitions to a dependence on $(E_b/N_j)^{-1}$. For infinite E_b/N_0 , from [21] we expect the SNORM and IC-AGC BER's to be proportional to $(E_b/N_j)^{-2}$ indefinitely; for $L=3$ and MFSK the SNORM in [21] is shown to be dependent on $(E_b/N_j)^{-2}$ for no thermal noise, but the AGC is dependent on $(E_b/N_j)^{-3}$, as shown in Figure 7.2-8, taken from [21]. Therefore, the better performance of SNORM for high SNR is not to be expected in all cases of M and L . (See Section 7.3.3.5 for further discussion on the SNORM performance.)

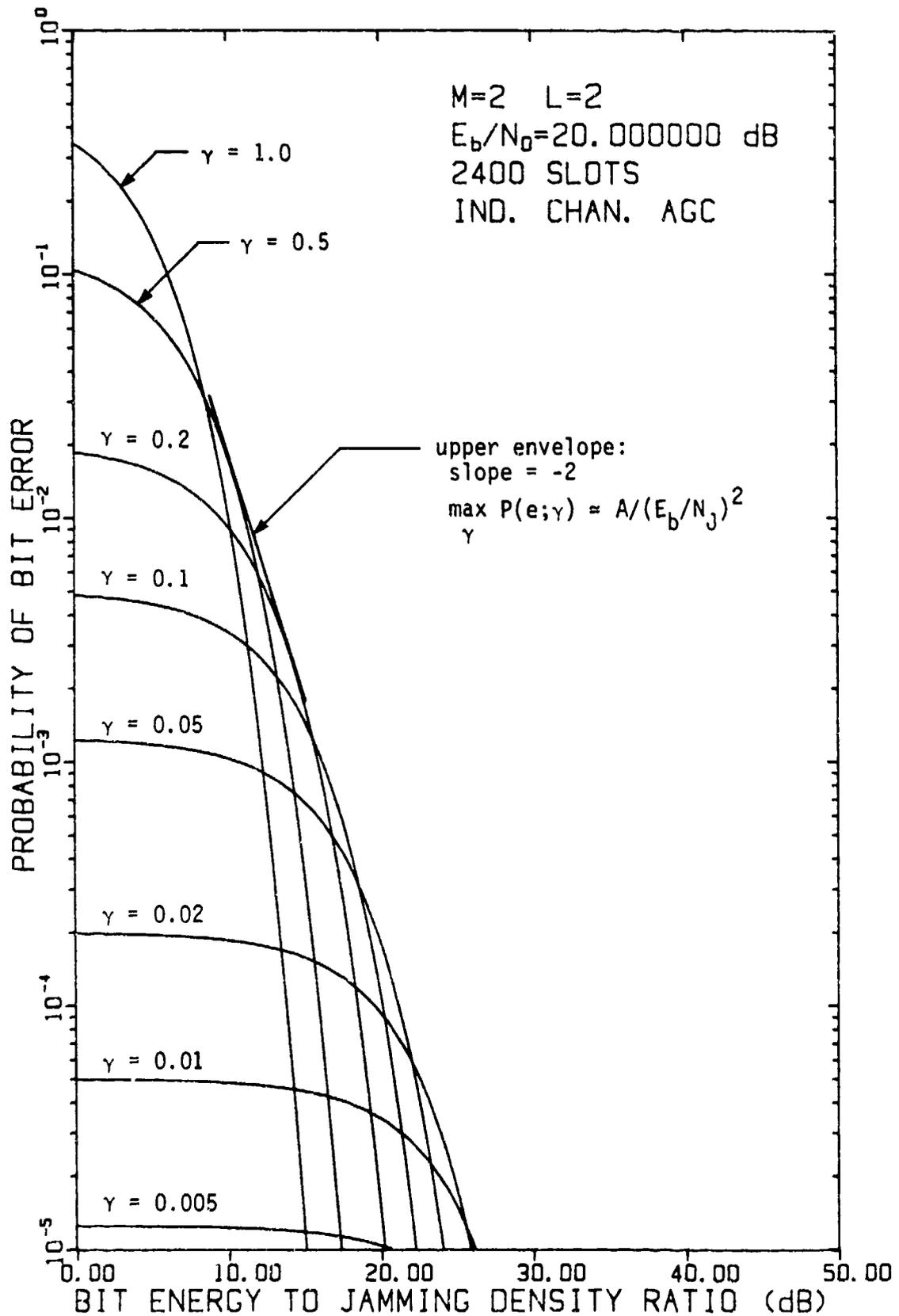


FIGURE 7.2-6 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR IC-AGC RECEIVER IN WORST-CASE PARTIAL-BAND NOISE FOR $M=2$, $L=2$ AT $E_b/N_0 = 20$ dB

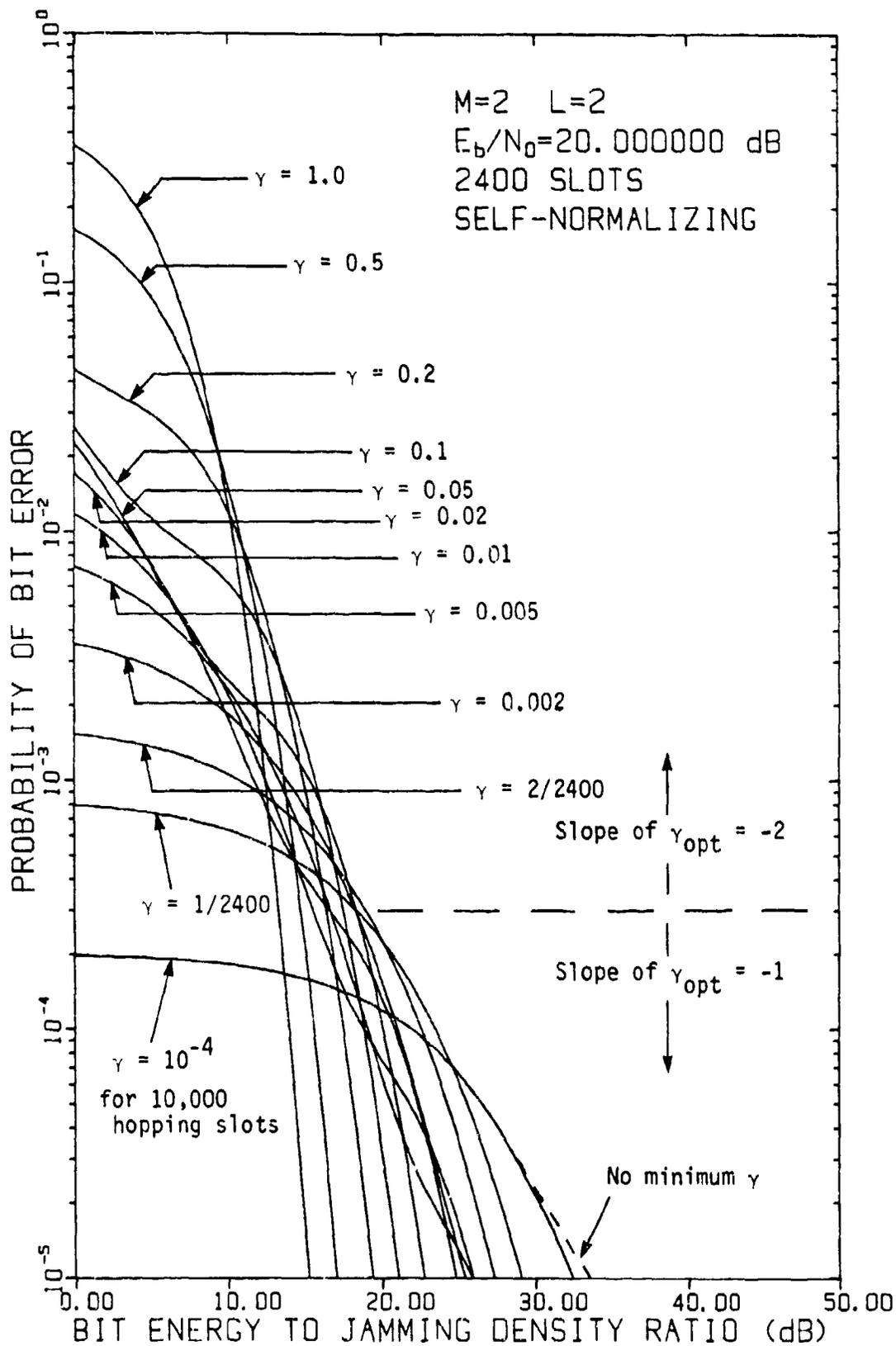


FIGURE 7.2-7 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR SELF-NORMALIZING RECEIVER IN WORST-CASE PARTIAL-BAND NOISE JAMMING FOR $M=2$, $L=2$ AT $E_b/N_0 = 20$ dB

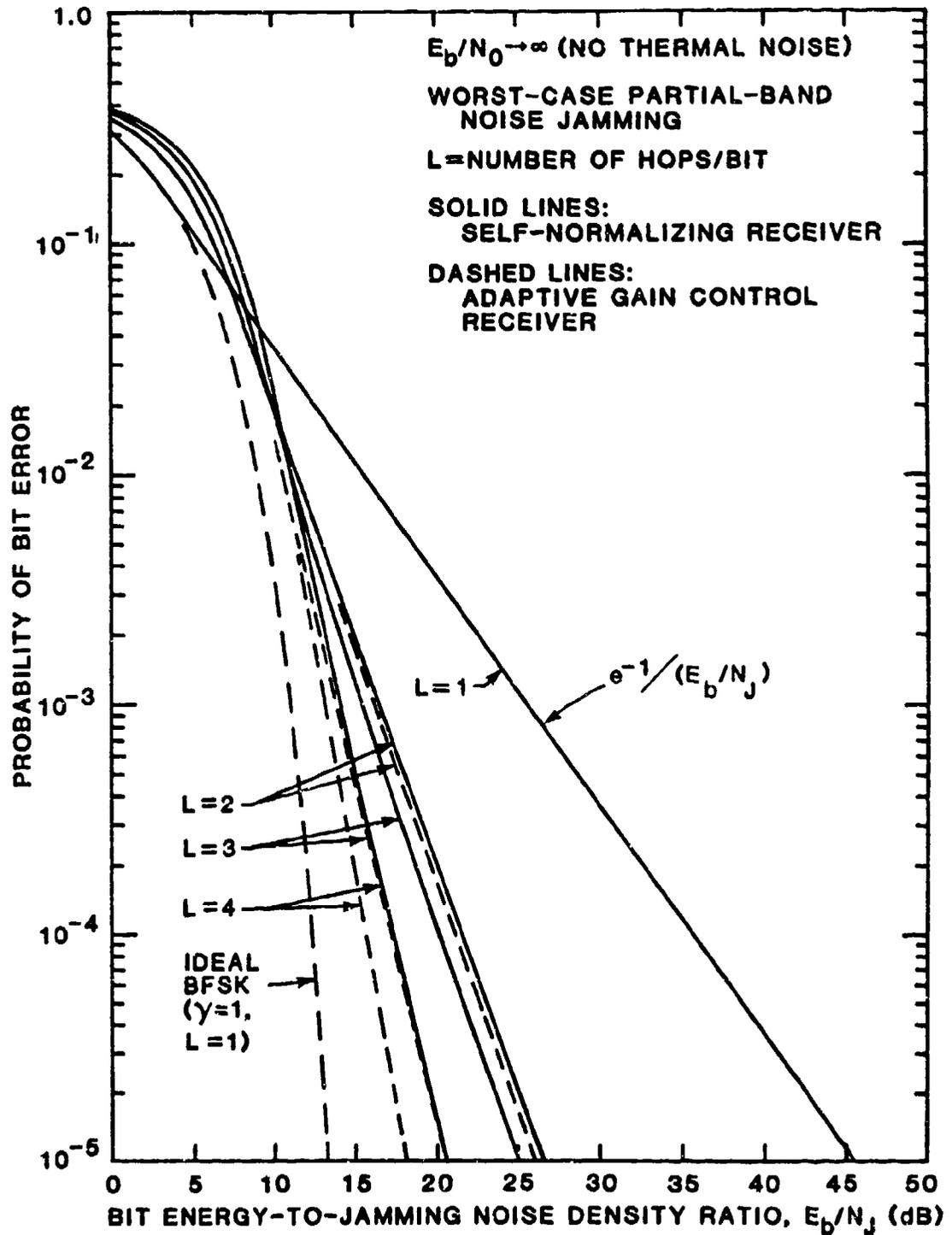


FIGURE 7.2-8 WORST-CASE PARTIAL-BAND NOISE JAMMING PERFORMANCE OF THE SELF-NORMALIZING FH/BFSK RECEIVER WHEN THERMAL NOISE IS ABSENT, WITH THE NUMBER OF HOPS/BIT (L) AS A PARAMETER

7.3 COMPARISONS OF FH/RMFSK AND FH/MFSK

Having compared the performances of the various FH/RMFSK receivers in worst-case partial-band noise jamming (WCPBNJ), we now consider the differences in performance to be expected between the random hopping MFSK system studied in this report (FH/RMFSK) and the conventional (adjacent or contiguous) hopping MFSK system (FH/MFSK) studied, for example, in [1].

7.3.1 The M=2, L=1 Case

It was found by Blanchard [6] that for M=2 and L=1 the two systems yield the same performance, at least for the $E_b/N_0 = 30$ dB case he studied, with the differences in possible jamming events accounted for by different optimum values of γ , the fraction of the system bandwidth which is jammed. (Typically, for high E_b/N_j the RMFSK γ_{opt} was found to be half that for MFSK.) Figure 7.3-1 displays the cases Blanchard considered, except that we use $E_b/N_0 = 13.35$ dB, corresponding to a 10^{-5} BER with no jamming. Our results indicate that the two systems do indeed perform the same for M=2 and L=1, except for certain differences for weak jamming (high E_b/N_j). What is the significance of the differences we observe in these computed results?

Curve A in Figure 7.3-1 is from [1] and represents the quantity

$$\max_{0 < \gamma \leq 1} \left[\frac{1}{2}(1-\gamma)e^{-E_b/2N_0} + \frac{1}{2}\gamma e^{-E_b/2N_T} \right], \quad (7.3-1a)$$

where

$$\frac{E_b}{N_T} = \frac{\frac{E_b}{N_0} \cdot \frac{\gamma E_b}{N_j}}{\frac{E_b}{N_0} + \frac{\gamma E_b}{N_j}} \quad (7.3-1b)$$

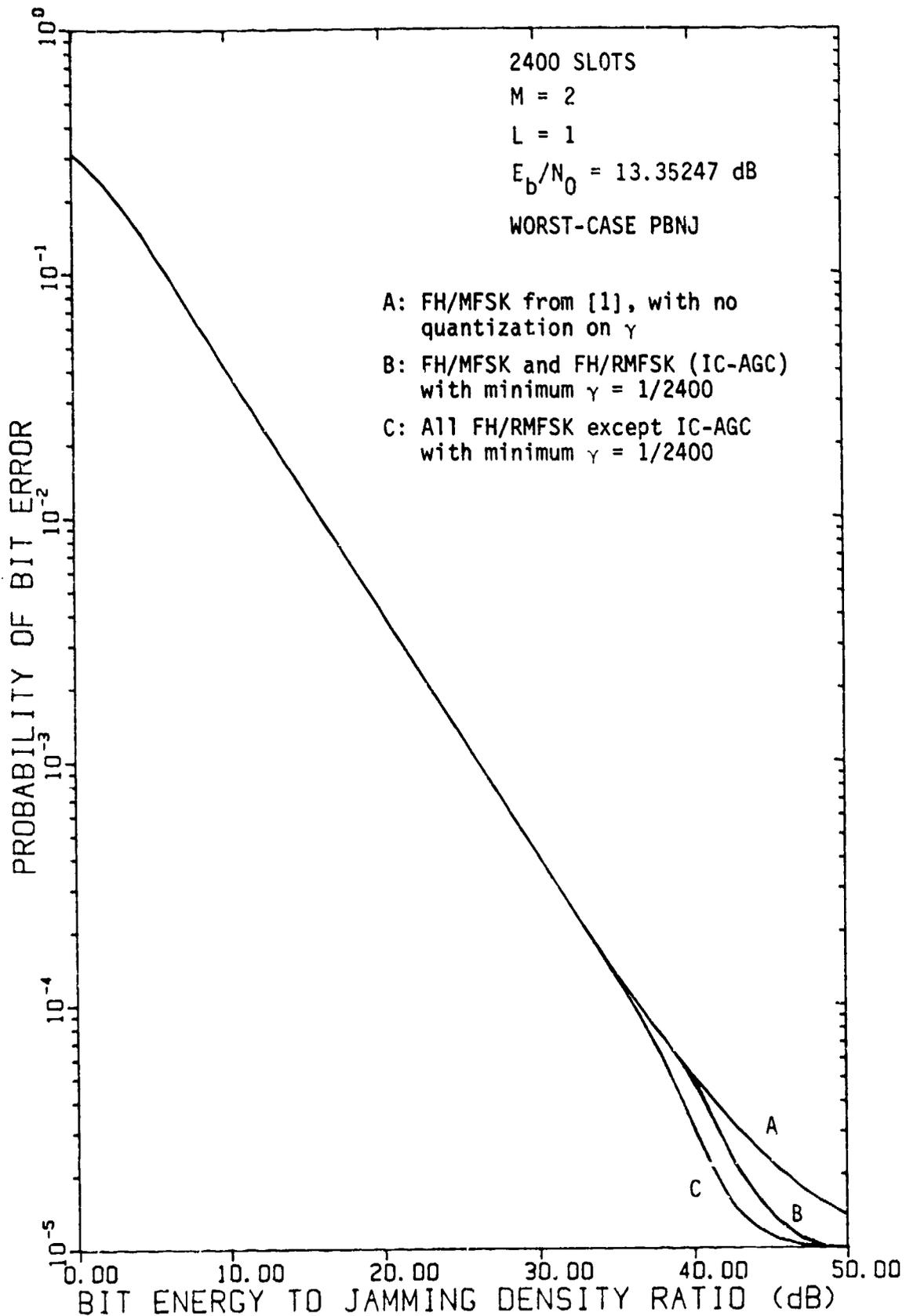


FIGURE 7.3-1 COMPARISON OF PERFORMANCE OF ALL RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH $M=2$ AND $L=1$ WHEN $E_b/N_0 = 13.35247$ dB

In this formulation for the binary case it is assumed that both slots are either jammed (with probability γ) or unjammed (with probability $1-\gamma$).

Curve B represents the quantity

$$\max_{1 \leq q \leq N} \left[\frac{1}{2} p_0 e^{-E_b/2N_0} + \frac{1}{2} p_1 e^{-E_b/2N_T} \right], \quad (7.3-2a)$$

where

$$p_1 = 1 - q/N = 1 - p_0. \quad (7.3-2b)$$

This formulation assumes that $\gamma = q/N$ is quantized - a discrete number (q) of the total number slots (N) are jammed, with the minimum γ equal to $1/N$. It further assumes that the system is FH/RMFSK with IC-AGC processing; a more explicit form of the error expression is

$$P(e; q) = (\pi_0 + \pi_1) \cdot \frac{1}{2} e^{-E_b/2N_0} + (\pi_1 + \pi_2) \cdot \frac{1}{2} e^{-E_b/2N_T}, \quad (7.3-3)$$

where

$$\pi_r = \text{prob. that } r \text{ slots are jammed}, \quad (7.3-4a)$$

and

$$\pi_0 = \frac{N-q}{N} \cdot \frac{N-q-1}{N-1} \quad (7.3-4b)$$

$$\pi_1 = \frac{N-q}{N} \cdot \frac{q}{N-1} \quad (7.3-4c)$$

$$\pi_2 = \frac{q}{N} \cdot \frac{q-1}{N-1}. \quad (7.3-4d)$$

Because of the individual channel normalization, the conditional BER depends only on whether the signal channel is jammed. Thus here are only two terms, with weights $p_0 = \pi_0 + \pi_1$ and $p_1 = \pi_1 + \pi_2$.

Now, the only difference between (7.3-2) and (7.3-1) is the quantization and minimum value of $\gamma = \frac{q}{N}$. Therefore in Figure 7.3-1 we identify curve B also with binary FH/MFSK, even though $q = 1$ violates the assumption that both channels are together jammed or unjammed.

Curve C in Figure 7.3-1 represents the quantity

$$\max_{1 \leq q \leq N} \left[\pi_0 \cdot \frac{1}{2} e^{-E_b/2N_0} + \pi_1 \cdot e^{-(E_b/N_0)/(K+1)} + \pi_2 \cdot \frac{1}{2} e^{-E_b/2N_T} \right], \quad (7.3-5a)$$

where

$$K = \frac{\sigma_T^2}{\sigma_N^2} = (E_b/N_0)/(E_b/N_T). \quad (7.3-5b)$$

This is the BER for all the FH/RMFSK receivers except the IC-AGC, and allows for only one of the two channels to be jammed.

Thus, in general, our results agree with Blanchard's conclusion that FH/RMFSK performs the same as FH/MFSK for $M=2$ and $L=1$, neglecting small asymptotic differences connected with assumptions on the quantization and minimum value of γ . Now we consider whether his conjecture that the two hopping systems perform the same for $M>2$ and $L=1$ is correct, and how the comparison is affected by $L>1$. In what follows, we shall use the fact that the IC-AGC FH/RMFSK receiver performs essentially the same as the AGC FH/MFSK receiver.

7.3.2 L=1 with Alphabet Size Varied.

In order to compare RMFSK and MFSK for $L=1$ and $M>1$, it is sufficient to consider Figures 7.3-2 and 7.3-3.

In Figure 7.3-2 the performances of the FH/MFSK and the IC-AGC FH/RMFSK receivers are shown for $L=1$ and $M=2,4,8$. The values of E_b/N_0 used were chosen

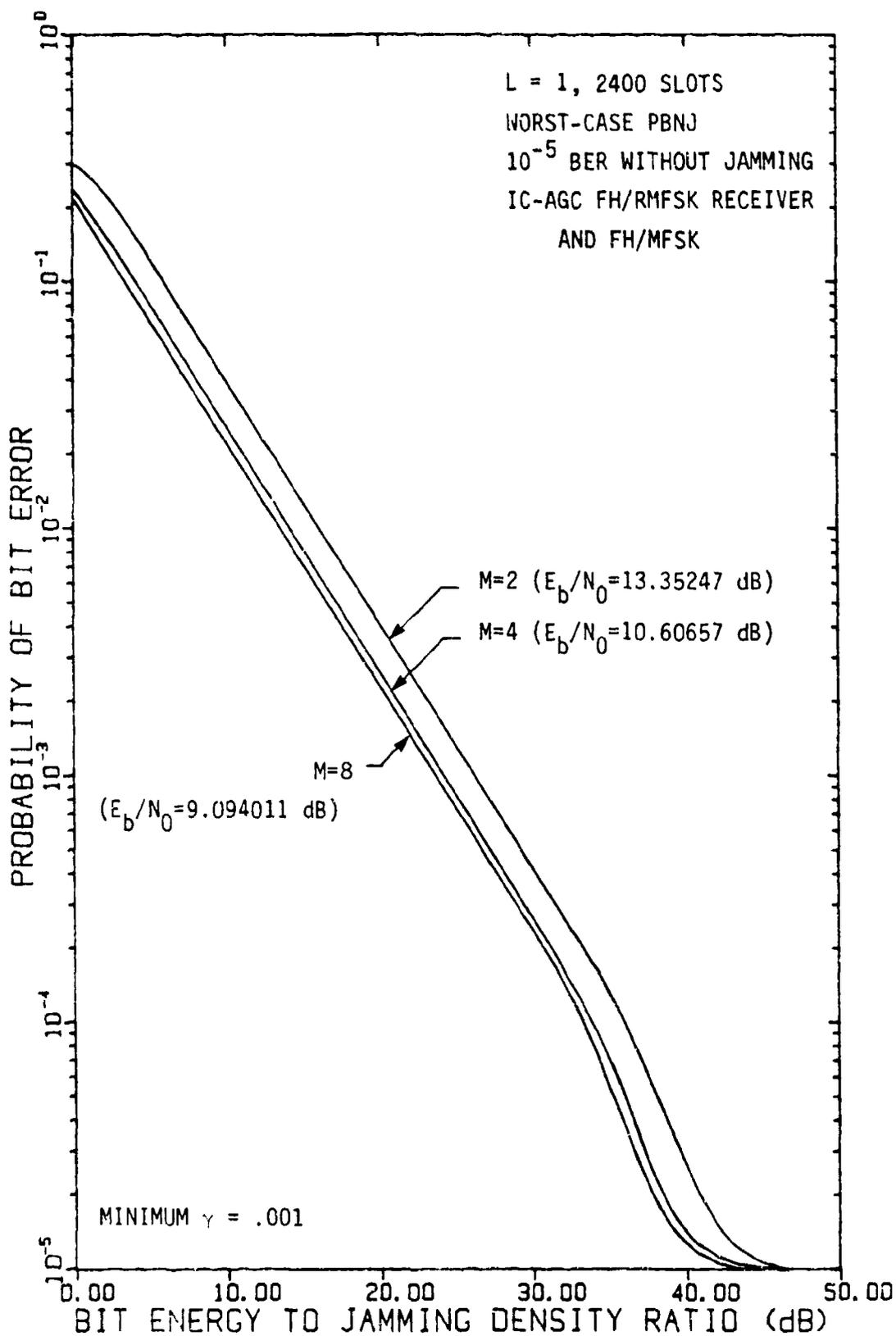


FIGURE 7.3-2 PERFORMANCE OF INDIVIDUAL CHANNEL AGC RECEIVER FOR FH/RMFSK WHEN $L=1$ HOP/SYMBOL WITH M AS A PARAMETER AND E_b/N_0 CORRESPONDING TO $P_b(e) = 10^{-5}$ IN THE ABSENCE OF JAMMING.

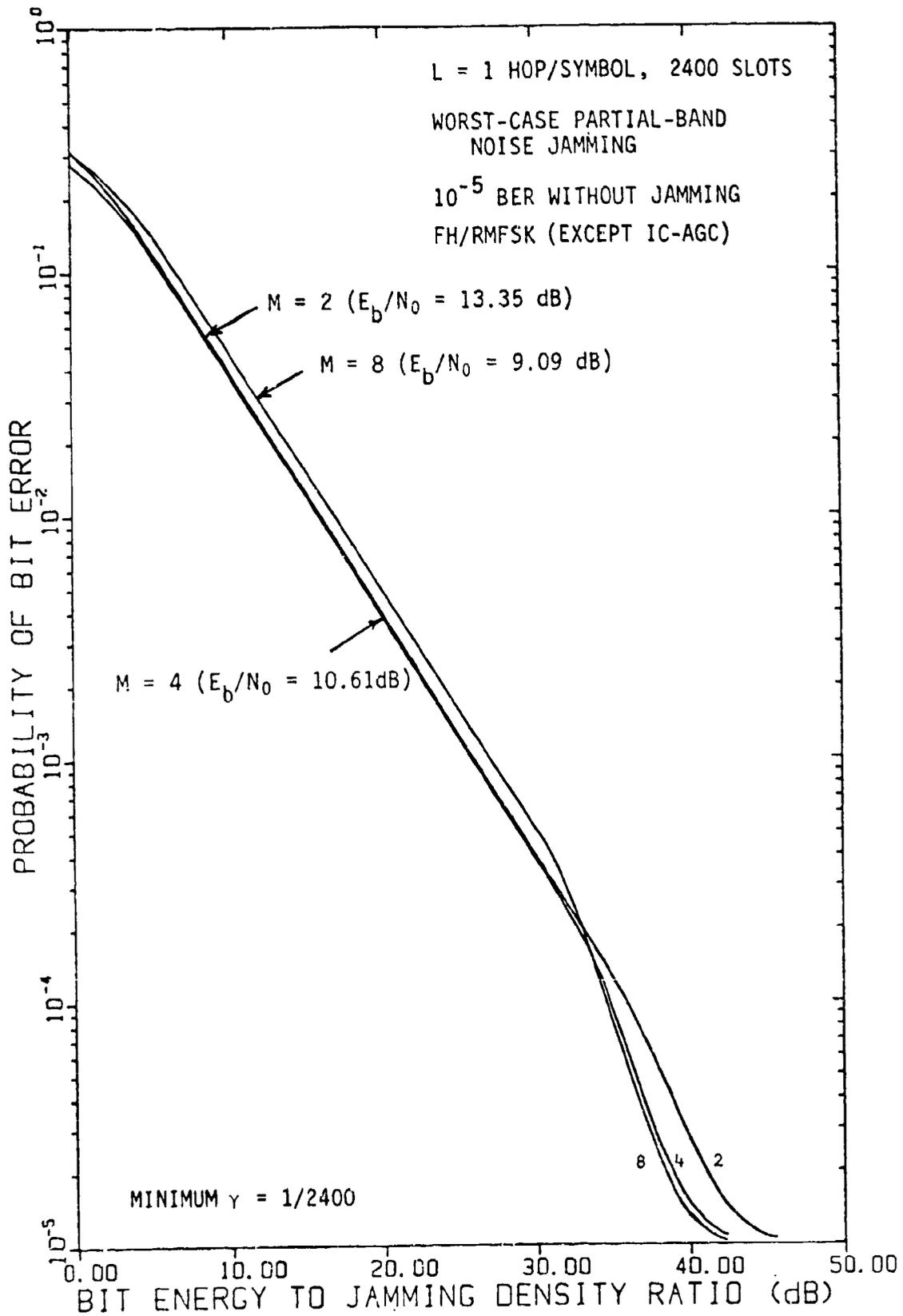


FIGURE 7.3-3 WORST-CASE PARTIAL-BAND NOISE JAMMING PERFORMANCE OF FH/RMFSK RECEIVERS FOR $L = 1$ HOP/SYMBOL AND $M = 2, 4, 8$ WHEN E_b/N_0 GIVES A 10^{-5} BER WITHOUT JAMMING

to give each example a 10^{-5} BER under no jamming. We observe from these results that for these systems the BER decreases as the alphabet size M increases, the conventional interpretation of which is that an "M-ary coding gain" is at work. This is the phenomenon usually observed for MFSK systems in the Gaussian interference channel.

In Figure 7.3-3 the same parameters are used as in Figure 7.3-2, but now the receivers are the FH/RMFSK receivers (except IC-AGC), which have identical performance for $L=1$. For these receivers we find that for strong jamming the system performance does not consistently improve as M increases, but instead improves very slightly for $M=4$ and degrades for $M=8$. Clearly this is the result of the increased probability, as M increases, of the most damaging jamming event: jamming power in a non-signal slot but not in the signal slot. Since the $M=2$ performances in the two figures are virtually the same, we conclude that FH/RMFSK is consistently more vulnerable to WCPBNJ than is FH/MFSK for $M>2$. The difference is about 3 dB for $M=4$ and 5 to 6 dB for $M=8$.

When the jamming is weak, we expect the relative performances for different M to approach the usual non-jammed behavior, and this is observable in Figure 7.3-3 for $E_b/N_j > 34$ dB.

7.3.3 Cases Where $L>1$ Hop/Symbol

Since the various FH/RMFSK receivers and their FH/MFSK counterparts begin to exhibit different performances when diversity is used ($L>1$), it is necessary to consider them separately. Most of the FH/MFSK results are taken

from [1]; however, when convenient we shall continue to utilize the fact that IC-AGC FH/RMFSK receiver performs essentially the same as the AGC FH/MFSK receiver.

7.3.3.1 Linear Combining Receiver.

We begin by comparing the performance of the square-law linear combining receiver for both FH/RMFSK and RH/MFSK signalling strategies. Figures 7.3-4 and 7.3-5 show these performances for $M=2$, $L=2$ and $M=4$, $L=2$ respectively. In both figures, it is apparent that MFSK is superior to RMFSK, ignoring the effects of the different minimum γ value used in the computations. This vulnerability of RMFSK can be attributed to what we term the "unbalancing" error mechanism inherent in partial-band jamming of RMFSK. Specifically, the random placement of M -ary slots over the hopping bandwidth W allows more chance for a jamming hit than does a block-hopping MFSK signal where it is assumed that either all M slots will be jammed or unjammed. This probability increases for greater values of M as evidenced by Figure 7.3-5.

7.3.3.2 AGC receivers.

Comparisons for $L>1$ among the AGC-type nonlinear combining receivers are exhibited in Figures 7.3-6 to 7.3-9. Recalling that RMFSK and MFSK hopping systems perform virtually the same for $M=2$ and $L=1$, it is instructive to observe in Figures 7.3-6 and 7.3-7 that the system performances differ by about 1 dB when $L=2$ or $L=3$. The approximately 3 dB difference noted for $M=4$ and $L=1$ continues to hold for $M=4$ and $L=2$ or 3, as shown in Figures 7.3-8 and 7.3-9.

7.3.3.3 Clipper receiver.

A very interesting consideration is brought to light by Figure 7.3-1 which shows the clipper receiver's FH/RMFSK performance for $L=2$ and several values of M . It was found analytically in Section 5 that the optimum clipping

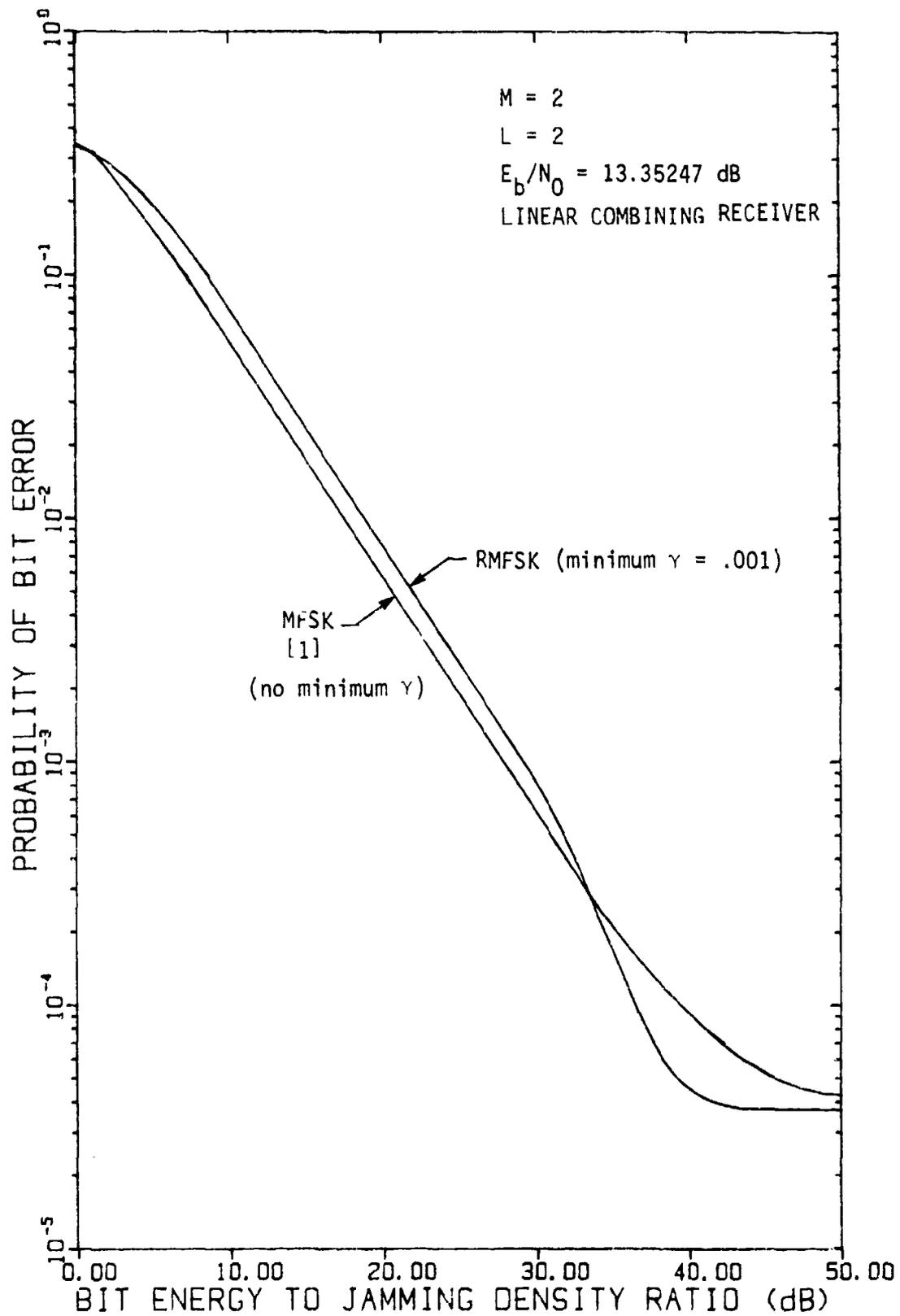


FIGURE 7.3-4 COMPARISON OF SQUARE-LAW LINEAR COMBINING RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH $M=2$ AND $L=2$ HOPS/SYMBOL WHEN $E_b/N_0 = 13.35247 \text{ dB}$

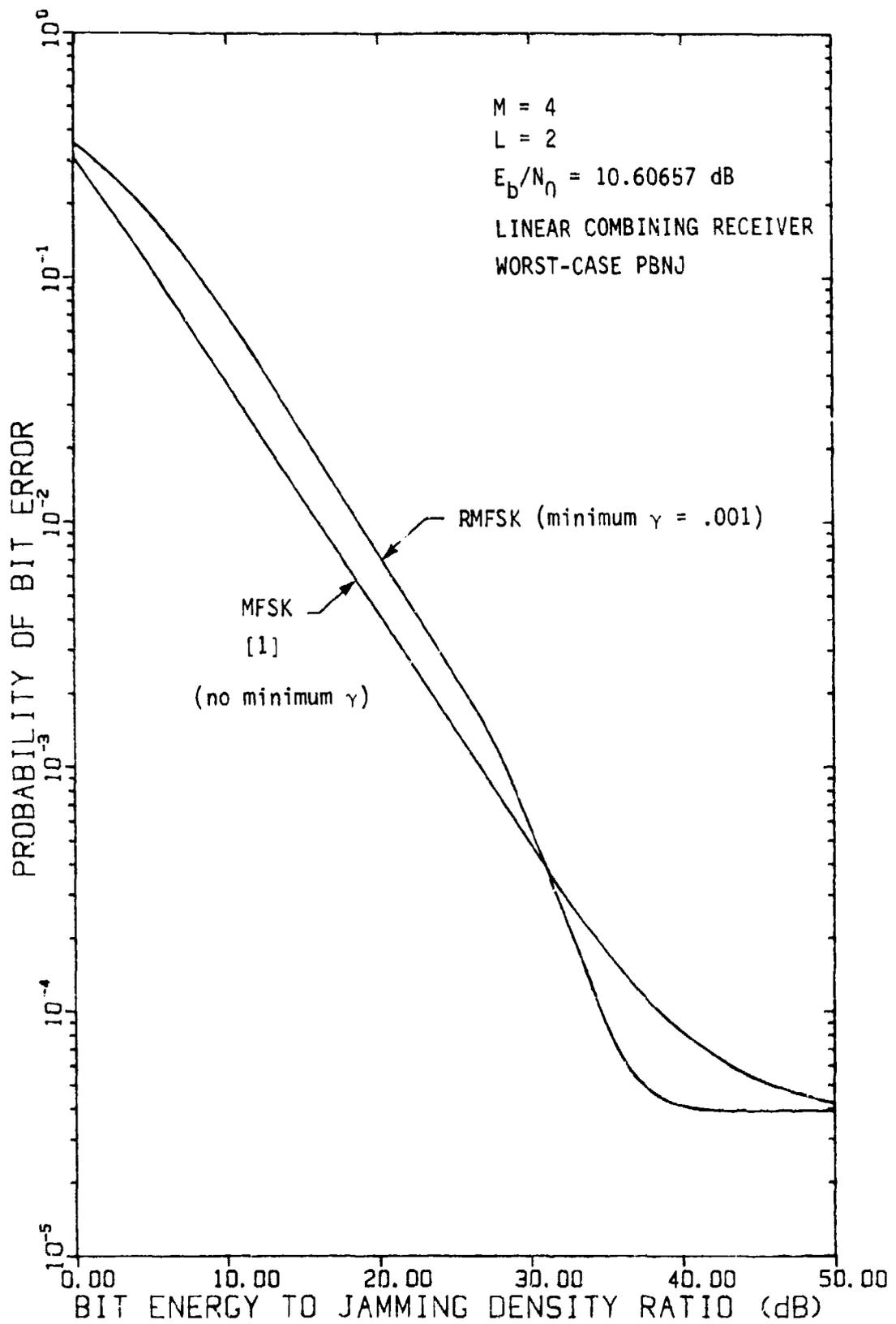


FIGURE 7.3-5 COMPARISON OF SQUARE-LAW LINEAR COMBINING RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH $M=4$ AND $L=2$ HOPS/SYMBOL WHEN $E_b/N_0 = 10.60657 \text{ dB}$

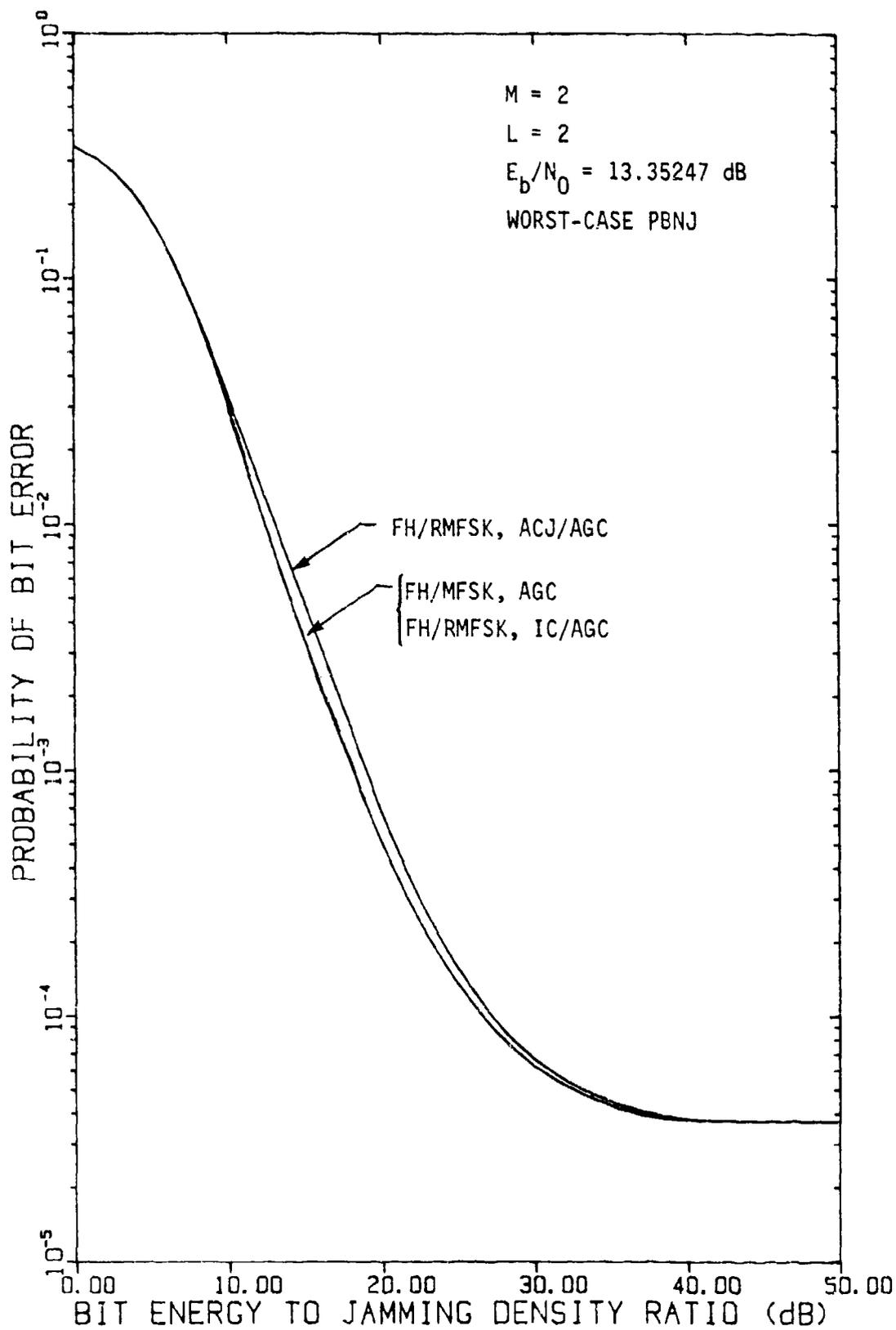


FIGURE 7.3-6 COMPARISON OF AGC RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH $M=2$ AND $L=2$ HOPS/SYMBOL WHEN $E_b/N_0 = 13.35247 \text{ dB}$

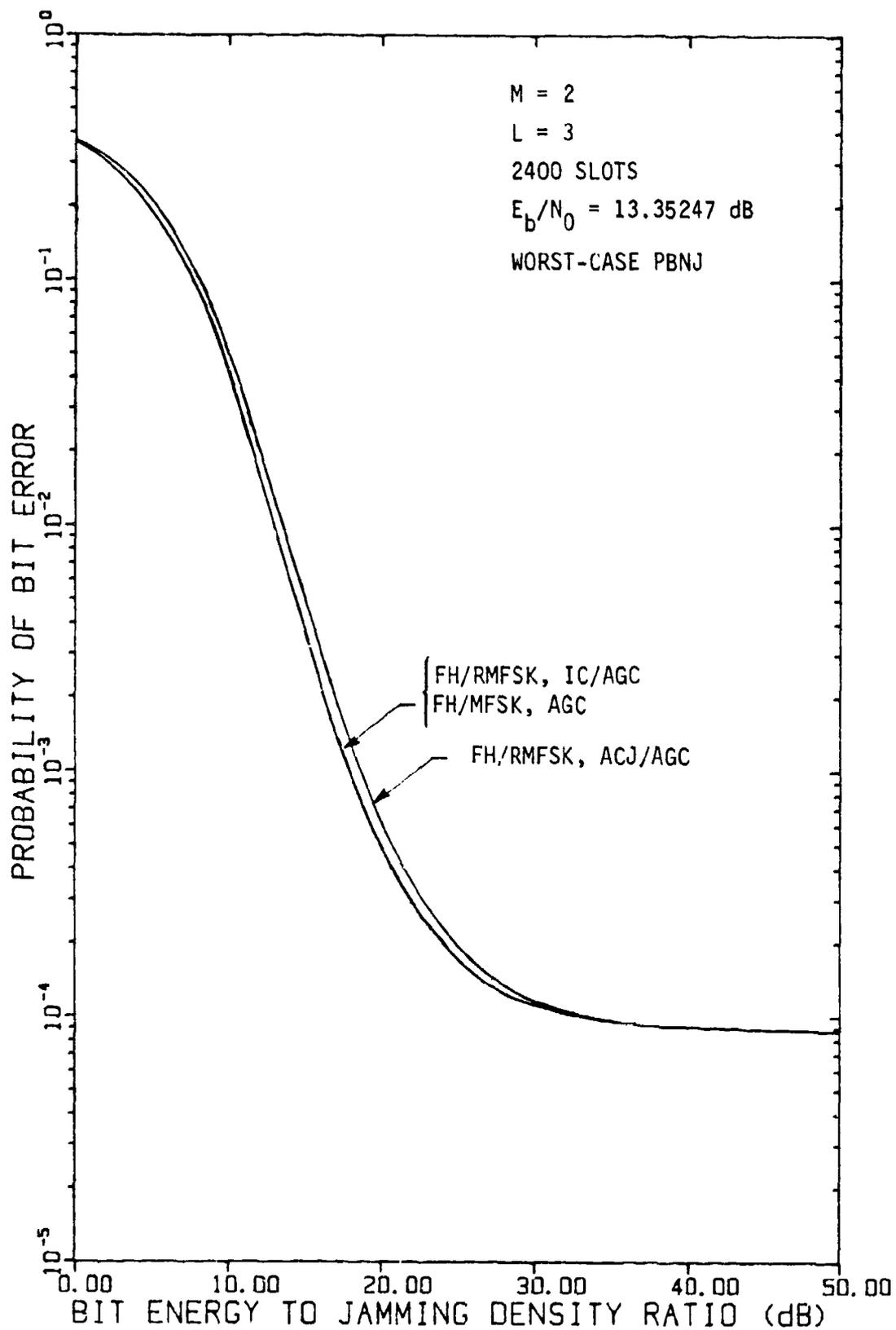


FIGURE 7.3-7 COMPARISON OF AGC RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH $M=2$ AND $L=3$ HOPS/SYMBOL WHEN $E_b/N_0 = 13.35247$ dB

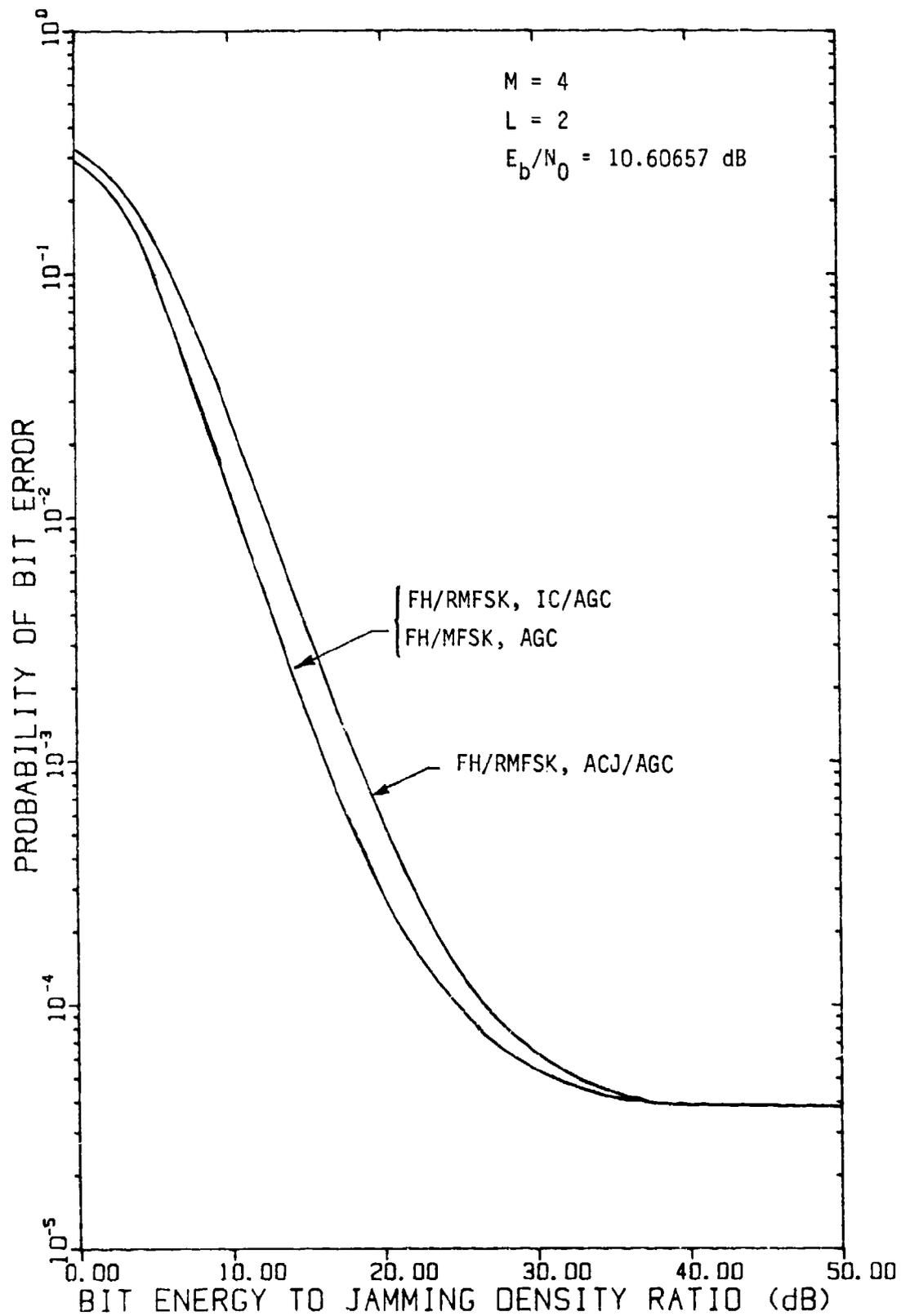


FIGURE 7.3-8 COMPARISON OF AGC RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH $M=4$ AND $L=2$ HOPS/SYMBOL WHEN $E_b/N_0 = 10.60657 \text{ dB}$

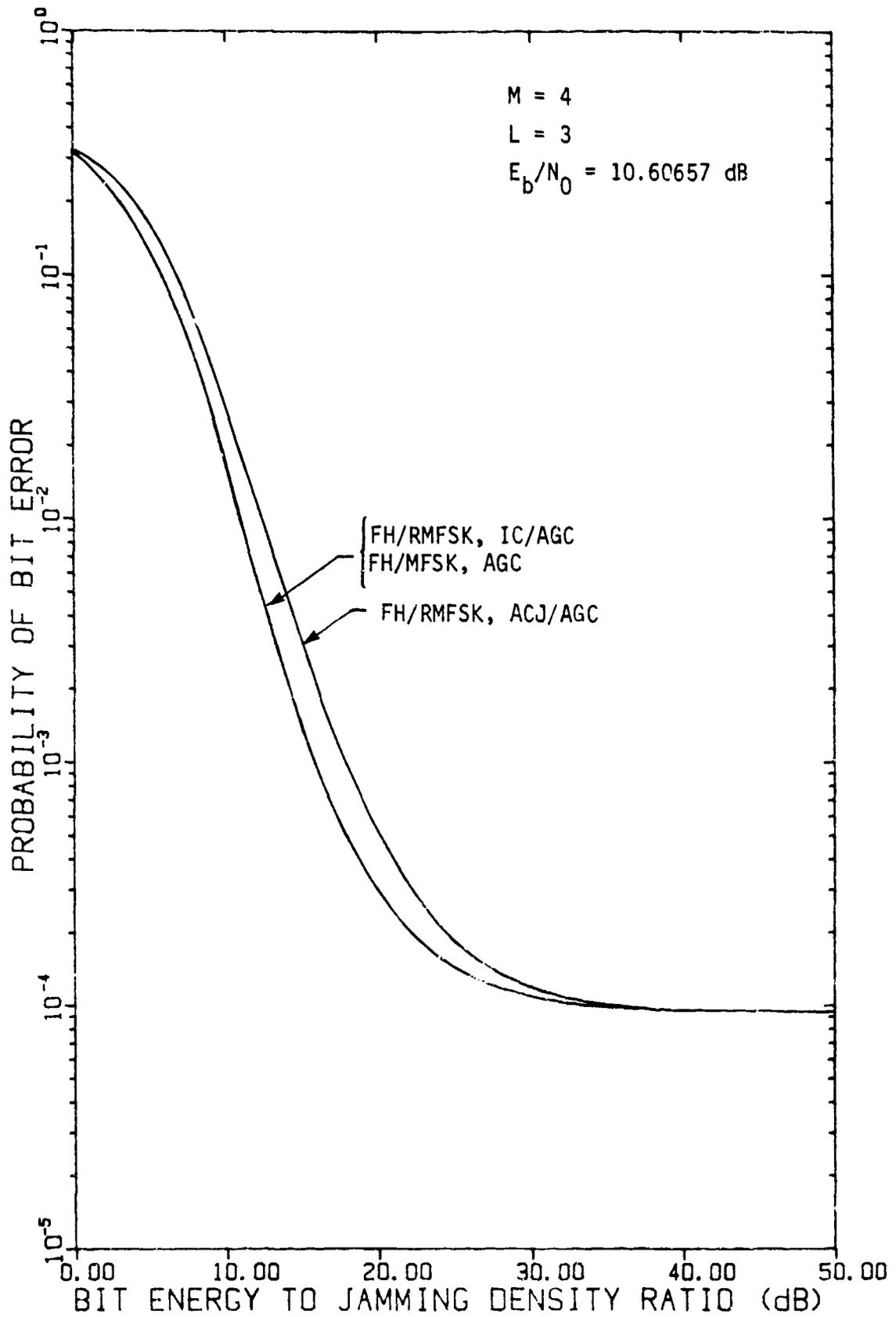


FIGURE 7.3-9 COMPARISON OF AGC RECEIVERS FOR FH/MFSK AND FH/RMFSK WITH $M=4$ AND $L=3$ HOPS/SYMBOL WHEN $E_b/N_0 = 10.60657 \text{ dB}$

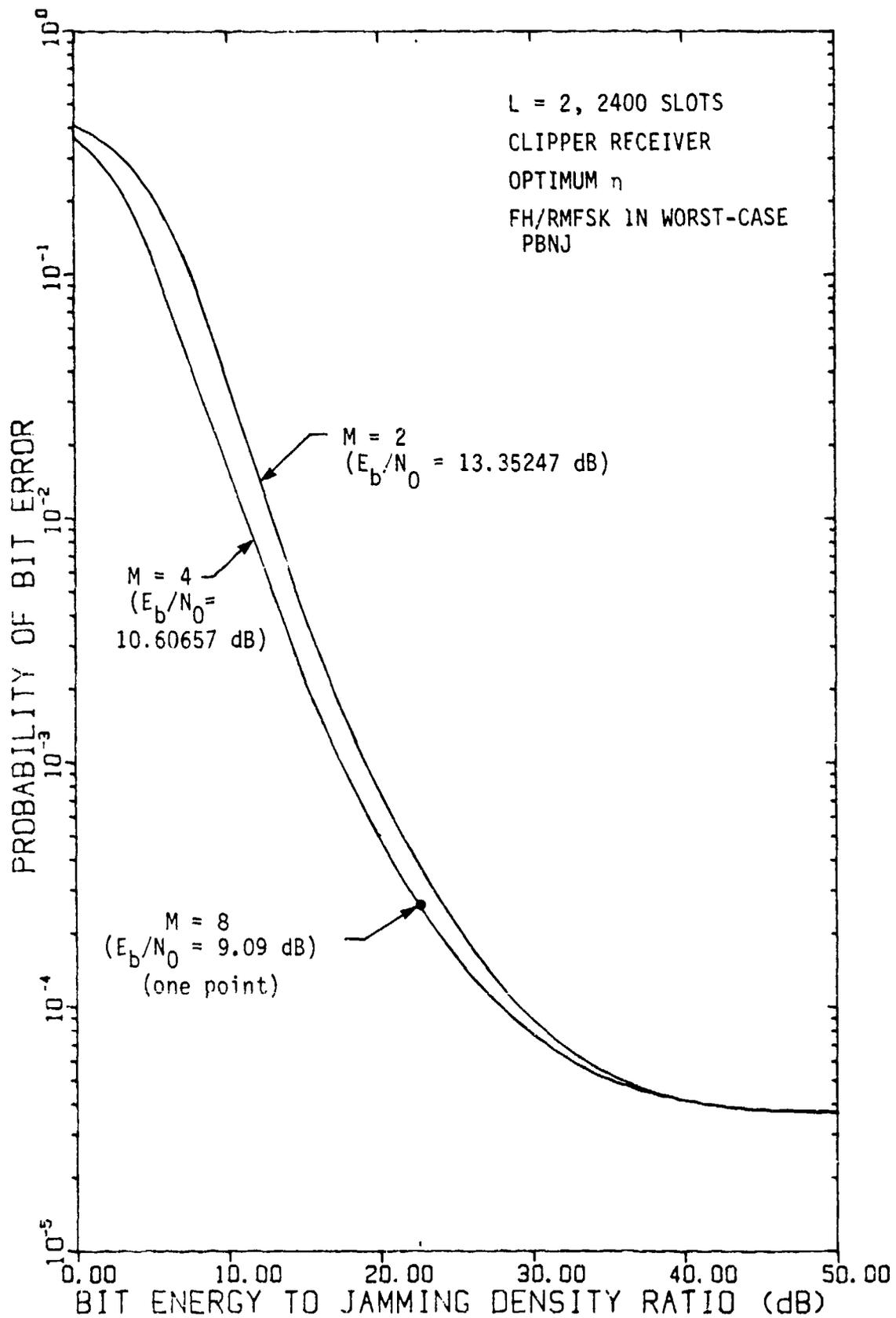


FIGURE 7.3-10 PERFORMANCE OF CLIPPER RECEIVER FOR FH/RMFSK WHEN L=2 HOPS/SYMBOL
 WITH M AS A PARAMETER AND E_b/N_0 CORRESPONDING TO $P_b(e) = 10^{-5}$ IN
 THE ABSENCE OF JAMMING (WHEN L=1 HOP/SYMBOL)

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threshold for $L=1$ is infinite (no clipping), whereas numerically it was determined that a finite threshold is optimum for $L>1$. Consequently, the $L=1$ "clipper" receiver is not a clipping receiver at all but one identical in operation to all the other RMFSK receivers for $L=1$ except IC-AGC, and its FH/RMFSK performance tends to get worse for increasing M as demonstrated previously in Figure 7.3.3. However, for $L=2$ we observe from Figure 7.3-10 that the clipper is performing in a manner similar to the IC-AGC, in that increasing M from 2 to 4 reduces the BER; however further increase to $M=8$ degrades performance. The reason for this similarity in behavior is that the clipper receiver, like the IC-AGC, operates to limit jamming input to the soft decision on an individual channel basis. The clipper is in this sense a crude version of the IC-AGC; but for higher values of M the losses become significant and the performance trend resembles the other RMFSK receivers more than the IC-AGC receiver. We then would expect clipping to be advantageous against jamming for $L=1$ as well; but the threshold was optimized for no jamming in order to avoid requiring the receiver to know or measure jamming parameters. If the threshold were jamming-dependent, the clipper receiver might follow the IC-AGC more closely for higher M . The tendency of the clipper receiver to "emulate" the IC-AGC was observed earlier in Figure 7.2-3, where we see that this tendency is more pronounced for strong jamming.

7.3.3.4 Hard-decision receiver.

Now if the clipper receiver can be thought of as a crude version of the IC-AGC, the hard-decision (HD) receiver can be considered a crude version of the ACJ-AGC because both act to limit or de-emphasize the entire set of M channels on a jammed hop, rather than operating on the channels separately.

Thus we observe in Figure 7.3-11 the tendency for the HD receiver's BER to increase with M (after $M > 4$) in strong jamming, just like the ACJ-AGC receiver's BER, and thereby to yield a worse performance for RMFSK than for MFSK. In weak or no jamming, the HD's BER for $L > 1$ gets worse for increasing M (unlike the other, soft-decision receivers) because noncoherent combining losses are in effect amplified by the quantization the HD uses.

7.3.3.5 Self-normalizing receiver.

In the previous examples, we have observed a consistent trend for RMFSK hopping to yield no better--and sometimes worse--performance than conventional MFSK hopping. This was explained as being due to the possibility of jamming being present in a non-signal channel but not in the signal channel for RMFSK but not for MFSK. It is also true that using RMFSK there can be jamming only in the signal channel, which tends to favor a correct decision. Apparently, using the LCR, AGC, clipper, and HD receivers, the jamming of one channel has a net effect of degrading the system performance for $L > 1$.

Now, we consider the comparison of RMFSK with MFSK using the self-normalizing receiver, and will see an exception to the trend previously observed. Figure 7.3-12 displays the SNORM error performances for FH/RMFSK and RH/MFSK in WCPBNJ for $M=L=2$ and $E_b/N_0 = 13.35$ dB and 20 dB. We see that the BER for RMFSK is better than for MFSK. This behavior seems to be connected with the jamming events in which both channels are either jammed or unjammed on a given hop, rather than those for which only one channel is jammed. This statement is supported by the fact that (1) the curves are roughly parallel for moderate-to-weak jamming (the portion of

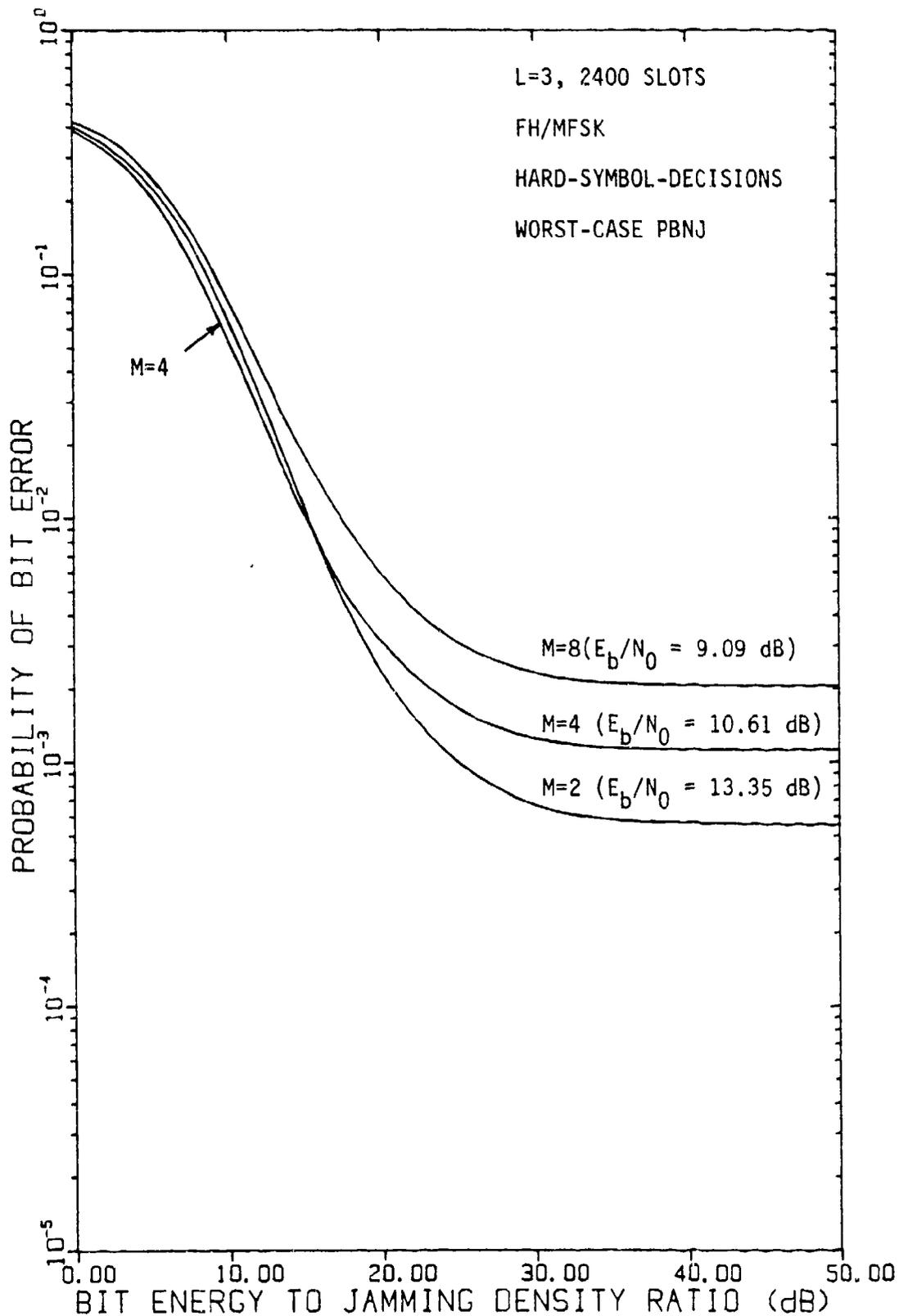


FIGURE 7.3-11 ERROR PERFORMANCE FOR HARD SYMBOL DECISION FH/RMFSK RECEIVER IN WORST-CASE PARTIAL-BAND NOISE JAMMING WHEN L=3 HOPS/SYMBOL; M=2,4,8; AND E_b/N_0 GIVES 10^{-5} ERROR WITHOUT JAMMING WHEN L=1

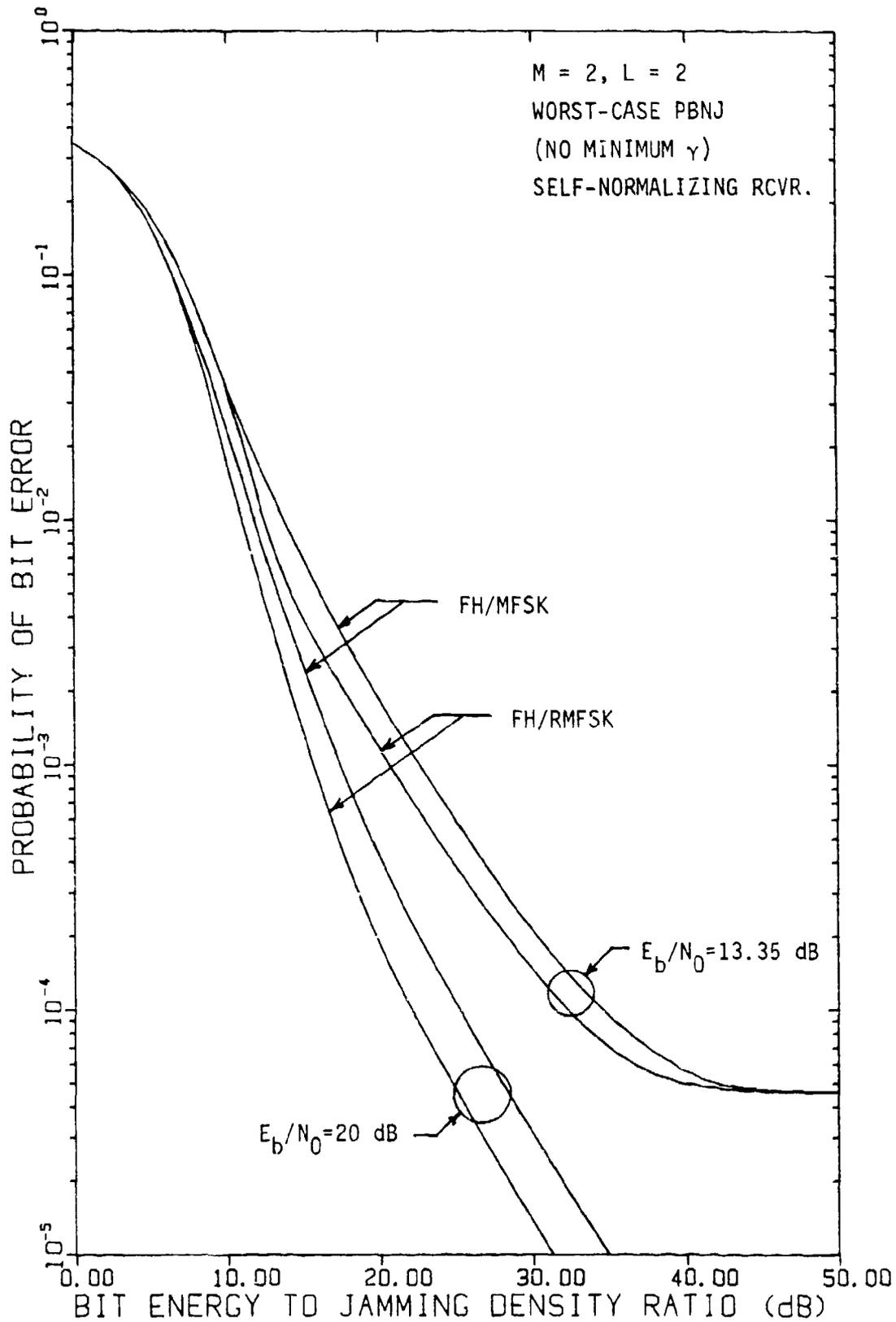


FIGURE 7.3-12 WORST-CASE PARTIAL-BAND NOISE JAMMING PERFORMANCES OF FH/RMFSK AND FH/MFSK USING THE SELF-NORMALIZING RECEIVER, FOR $M=2$ AND $L=2$

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the curve proportional to e^{-E_b/N_J} , and (2) the MFSK receiver is subject only to those particular jamming events, by assumption. The advantage of RMFSK, according to this interpretation, then lies in the smaller probability of both channels being jammed on a given hop.

Now if M were increased to M=4 or M=8, it would be expected that the effects of jamming in one channel only would tend to increase the RMFSK error, since then a damaging effect would be M-1 times as likely as a helping effect.

On the other hand, the improvement of the RMFSK error over that of MFSK can be explained in terms of how the SNORM receiver processes the jamming events for which only one channel is jammed, in contrast to the way the other RMFSK receivers process the events. When only the non-signal channel is jammed the hop statistics are ($K = \sigma_T^2/\sigma_N^2 \gg 1$)

$$z_{1k} = \frac{\chi^2(2, 2\rho_N)}{\chi^2(2, 2\rho_N) + K\chi^2(2)} \rightarrow 0 \quad (7.3-6a)$$

$$z_{2k} = \frac{K\chi^2(2)}{\chi^2(2, 2\rho_N) + K\chi^2(2)} \rightarrow 1. \quad (7.3-6b)$$

But when only the signal channel is jammed,

$$z_{1k} = \frac{K\chi^2(2, 2\rho_T)}{K\chi^2(2, 2\rho_T) + \chi^2(2)} \rightarrow 1 \quad (7.3-7a)$$

$$z_{2k} = \frac{\chi^2(2)}{K\chi^2(2, 2\rho_T) + \chi^2(2)} \rightarrow 0. \quad (7.3-7b)$$

That is, the SNORM per-hop processing is nearly equivalent to a hard symbol decision; the signal channel suppresses the nonsignal channel when the signal is jammed, but if the nonsignal channel is jammed, it is awarded a value of at most 1. Thus the receiver de-emphasizes jammed hops while at the same time distinguishing between "good" and "bad" jammed hops. The

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IC-AGC receiver, by contrast, penalizes the channel being jammed by normalizing its noise to the same variance as the other channel; this has the effect of suppressing the jammed channel when only one is jammed. This is a good thing to do when the non-signal channel is jammed; but it is not beneficial when the signal channel is jammed.

According to this second interpretation of the results the SNORM performance for RMFSK is better because it makes good use of the "favorable" jamming events, which do not occur for MFSK. However, we still would expect the RMFSK performance to degrade for higher M under this interpretation.

7.4 COMPARISON OF RECEIVER DIVERSITY EFFECTS

The FH/RMFSK receivers we have studied are distinguished by their methods of combining the L hops transmitted per MFSK symbol. The objective of the diversity transmission is to spread the signal on a symbol basis, making it less likely that the symbol is jammed for the entirety of its duration. The L pieces of the symbol transmission are then sequentially acquired noncoherently and accumulated after weighting or otherwise processing them individually. Since the combining is done noncoherently, the performance of the system without jamming or with full-band jamming (Gaussian channel) necessarily is degraded from that using one hop with same signal energy. However, when the system bandwidth is jammed partially, giving rise to a type of non-Gaussian interference channel, the system performance is improved using diversity, provided that the hop processing in some fashion limits or discriminates against those hops which are jammed.

The conventional diversity receiver for MFSK, which we have termed the linear combining square-law receiver (LCR), is known to be effective against signal fading, that is, when there exists a random-amplitude signal in a Gaussian channel. But against partial-band noise jamming (PBNJ), the LCR is not effective since jammed hops are not de-emphasized. Figure 7.4-1 illustrates for $M=2$ that LCR performance degrades in proportion to L .

One view of the individual-channel adaptive gain control receiver (IC-AGC), which normalizes each square-law detector sample by its a priori noise variance, is that it in effect renders the non-Gaussian PBNJ interference into a Gaussian interference with unit variance in each MFSK slot. The residual

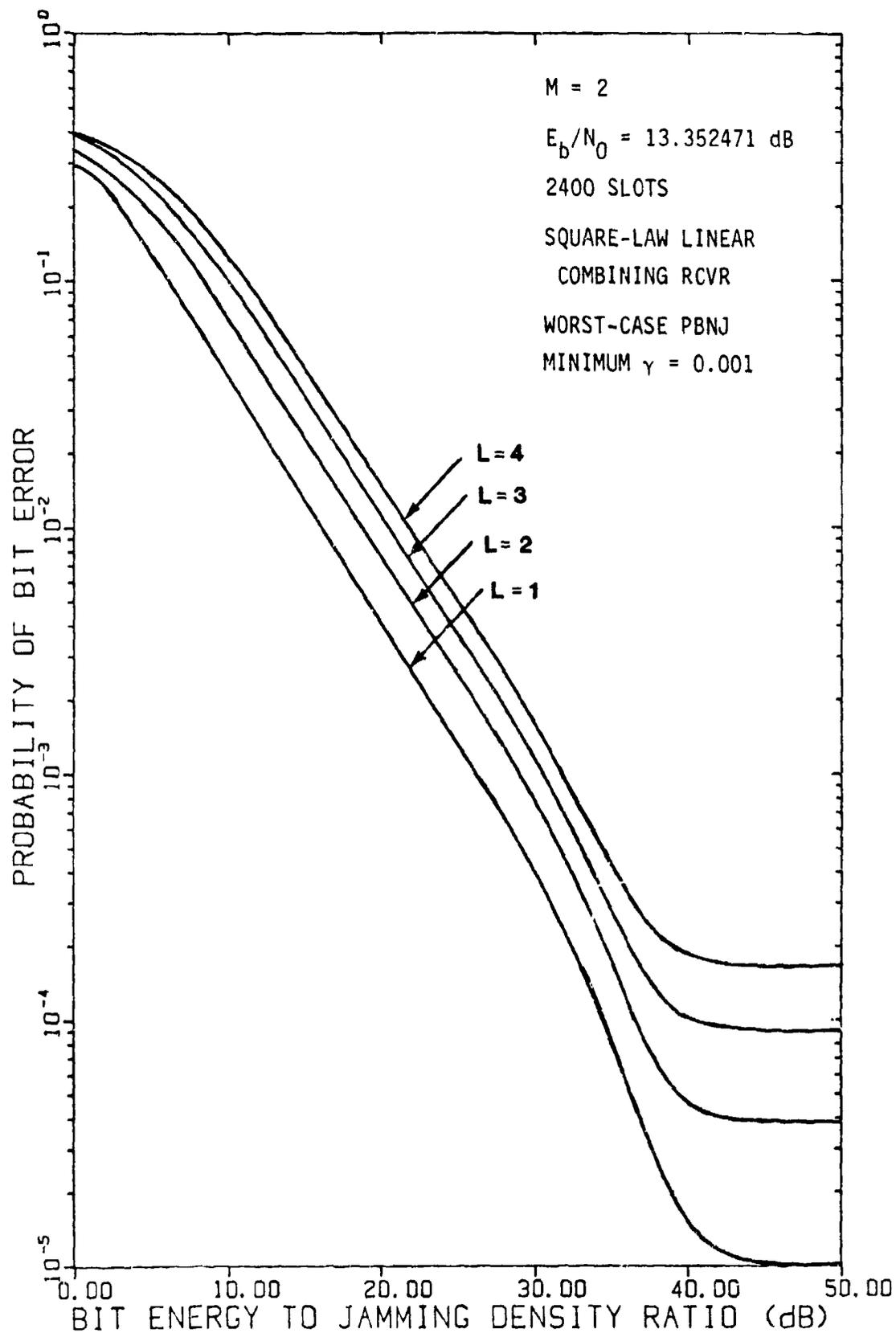


FIGURE 7.4-1 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR SQUARE-LAW COMBINING RECEIVER AND $M = 2$ WITH NUMBER OF HOPS/BIT AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING

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effect of the jamming after normalization is to reduce the SNR in the MFSK signal channel or slot by an amount which depends upon the random event of that channel's being jammed on k_1 of the L hops. Because the amount of reduction is inversely proportional to γ , the fraction of the system bandwidth which is jammed, while the probability of jamming is directly proportional to γ , there exists an optimum value of γ which maximizes the system error probability as a function of available jamming power; generally $\gamma_{opt} = \text{const}_1 / (E_b/N_J) = \text{const}_2 \cdot J$, that is, the optimum value of γ is directly proportional to jamming power, and for sufficient jammer power full-band jamming ($\gamma=1$) is optimum.

It is possible to reason without analysis that the IC-AGC receiver performs better than the LCR because, while the two receivers are subject to the same SNR degradation, the IC-AGC in effect "matches" the accumulator structure (soft-decision) to the channel. However, it is difficult to predict how any improvement would depend upon L , and whether an optimum value of L exists. Thus the analysis and computations of IC-AGC performance have been quite revealing. For example, Figure 7.4-2 shows that there is a tendency for increasing L to improve the IC-AGC performance for increasing E_b/N_J , but this tendency is by no means uniform for the value of E_b/N_0 shown. Figure 7.4-3 shows the IC-AGC performance which would be obtained if the best value of L were always used. This figure reveals that the effectiveness of the diversity depends on the degree of thermal noise present; when N_0 is not negligible, increasing L eventually gives rise to noncoherent combining losses which overcome the gains from diversity.

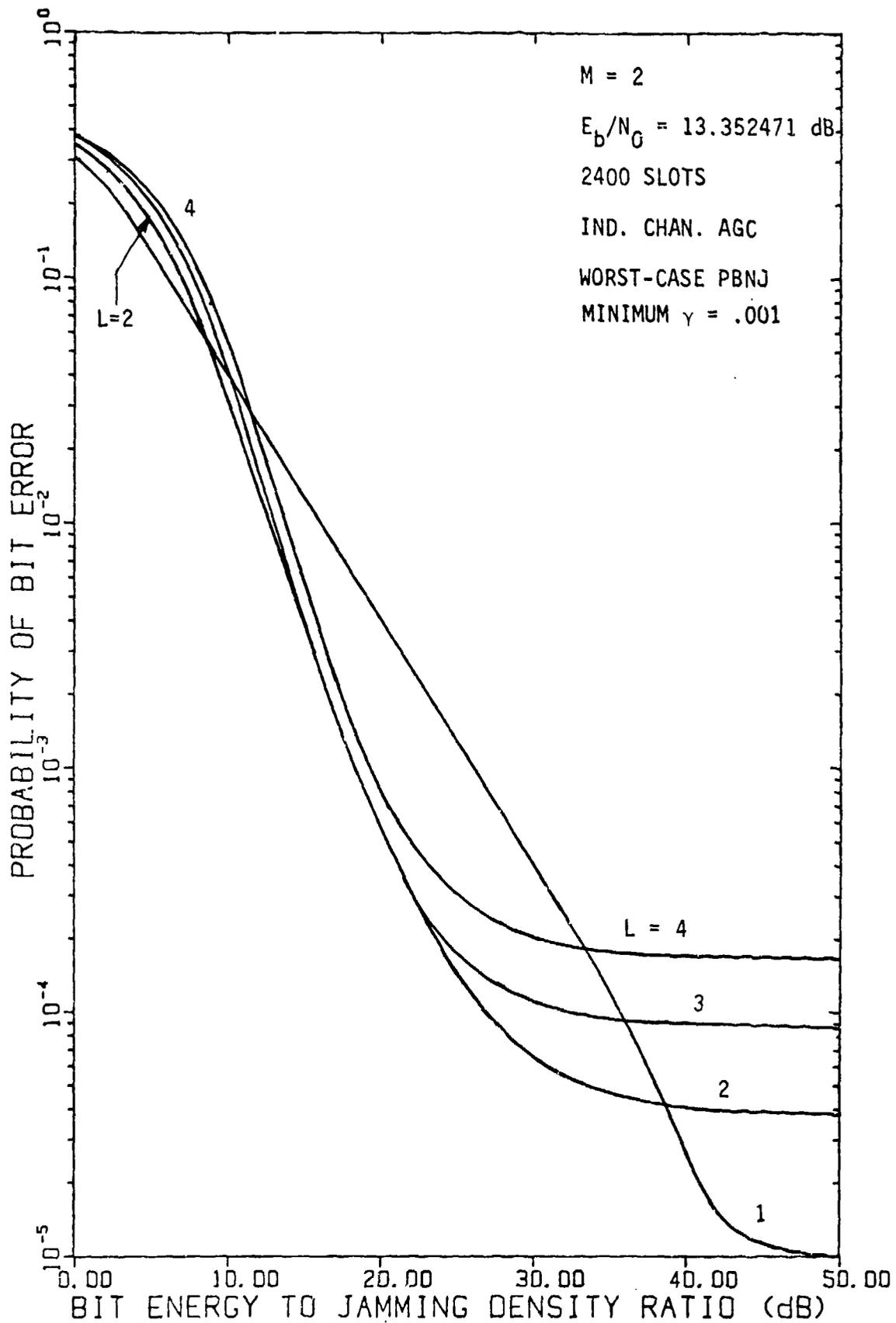


FIGURE 7.4-2 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR INDIVIDUAL CHANNEL AGC RECEIVER AND $M=2$ WITH NUMBER OF HOPS/SYMBOL AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING

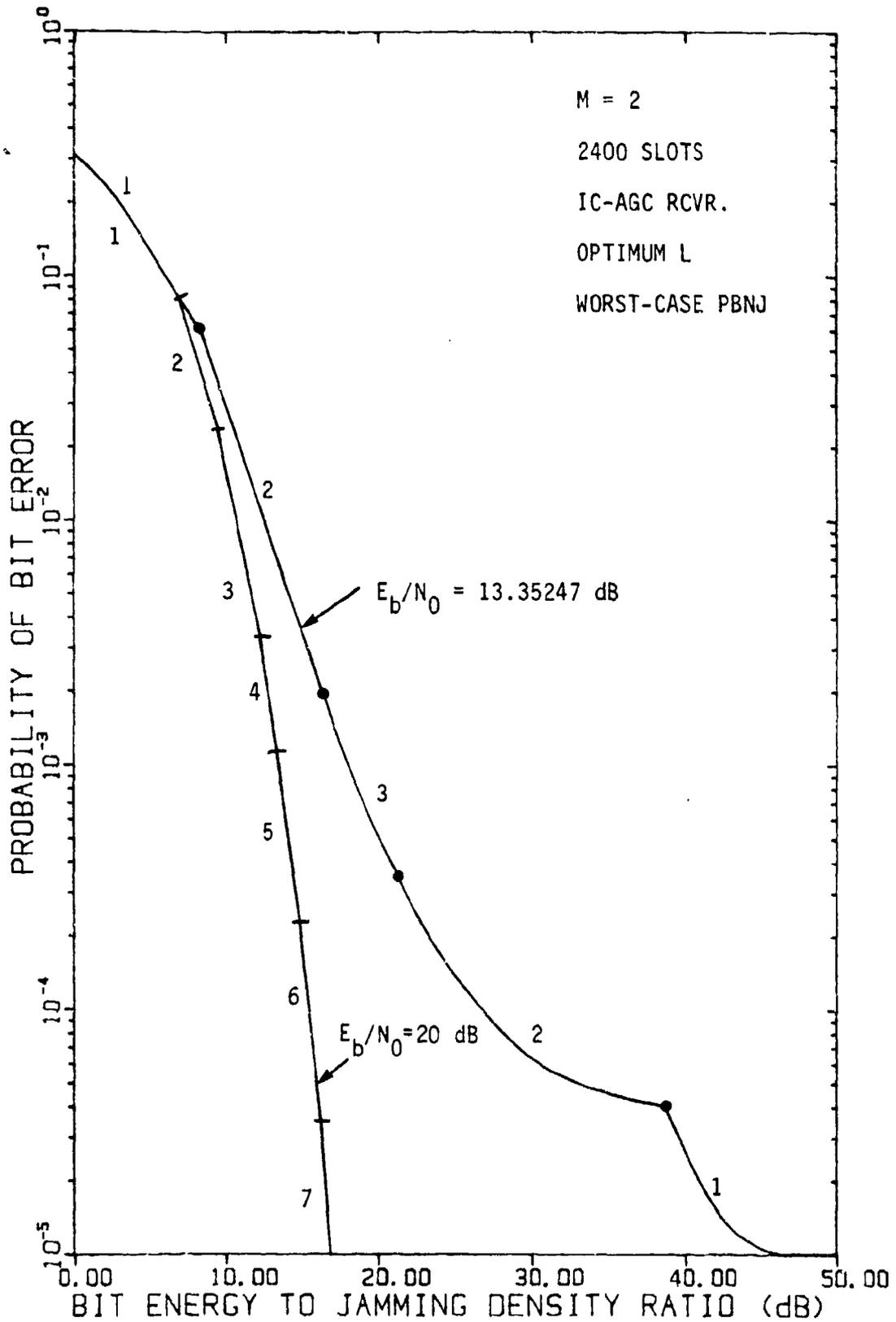


FIGURE 7.4-3 OPTIMUM DIVERSITY PERFORMANCE OF INDIVIDUAL CHANNEL AGC RECEIVER FOR FH/RMFSK WITH $M=2$ AND E_b/N_0 AS A PARAMETER

It is important to keep in mind that the performance in jamming can never be better than that without jamming, and that without jamming the best performance is for $L=1$. Therefore "optimum diversity" values may increase with E_b/N_j , but must eventually decrease again to $L=1$ as $E_b/N_j \rightarrow \infty$ (no jamming). However, as the figure demonstrates, for a desired performance of, say 10^{-5} , the optimum diversity value can be greater than one if the unjammed error is much smaller.

Examples of diversity effects for the "any channel jammed" receiver (ACJ) and the self-normalizing receiver are given by Figures 7.4-4 and 7.4-5. We observe that these receivers, in that their normalization techniques "approximate" that of the IC-AGC, achieve similar diversity gain effects.

Two of the receivers studied, the clipper and hard-symbol decision (HD) receivers, do not utilize normalization as such, yet accomplish a diversity gain effect by limiting the "contamination" that a jammed hop may bring to the symbol decision. It has noted previously that the clipper receiver's performance generally is close to that of the ACJ for stronger jamming, but that the HD receiver is considerably worse for the same amount of thermal noise. However, as Figure 7.4-6 shows, even this very simple HD approach can be considered effective in the diversity sense when thermal noise is low.

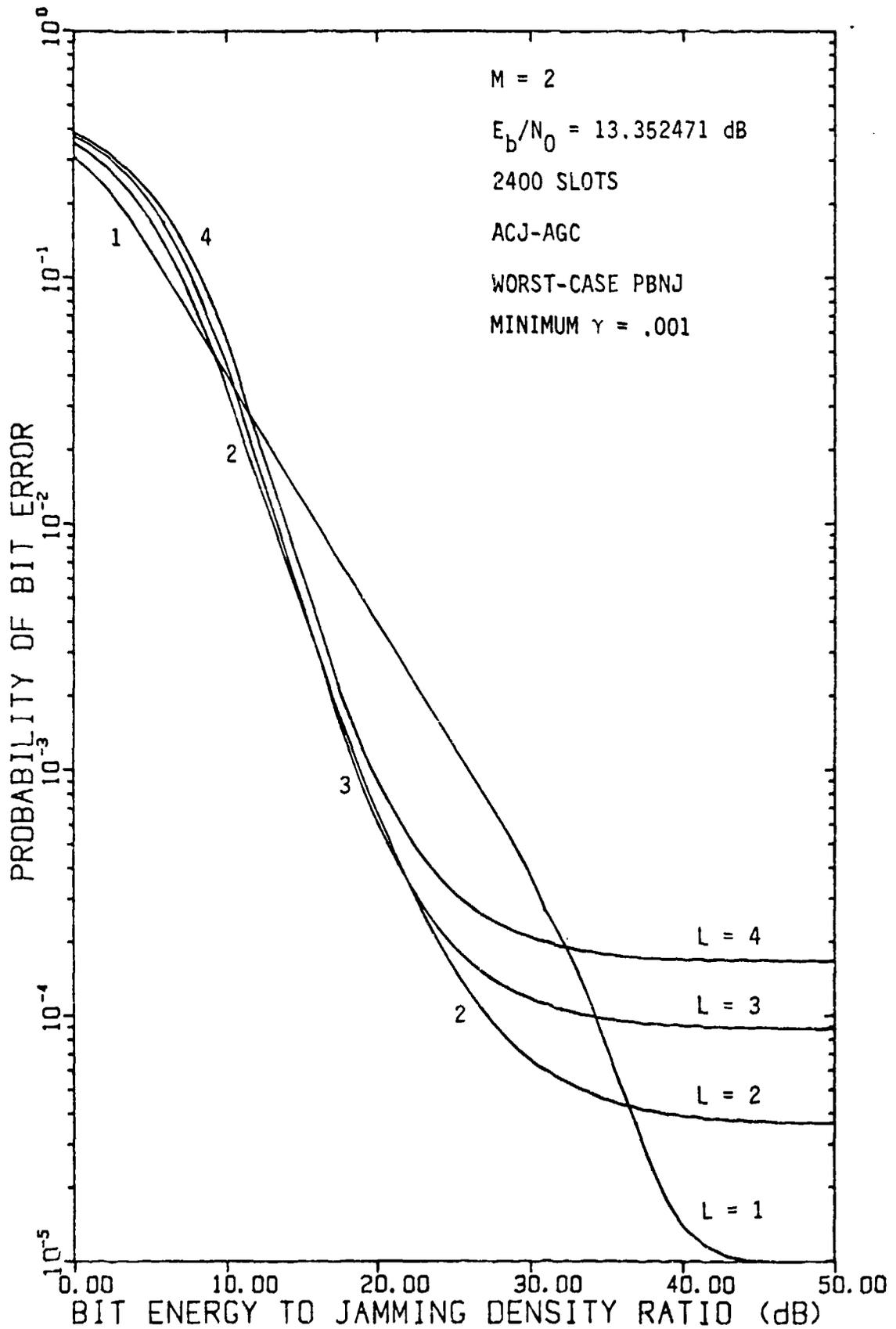


FIGURE 7.4-4 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING DENSITY RATIO FOR ANY-CHANNEL-JAMMED AGC RECEIVER AND $M=2$ WITH NUMBER OF HOPS/BIT AS A PARAMETER IN PRESENCE OF WORST-CASE PARTIAL-BAND NOISE JAMMING

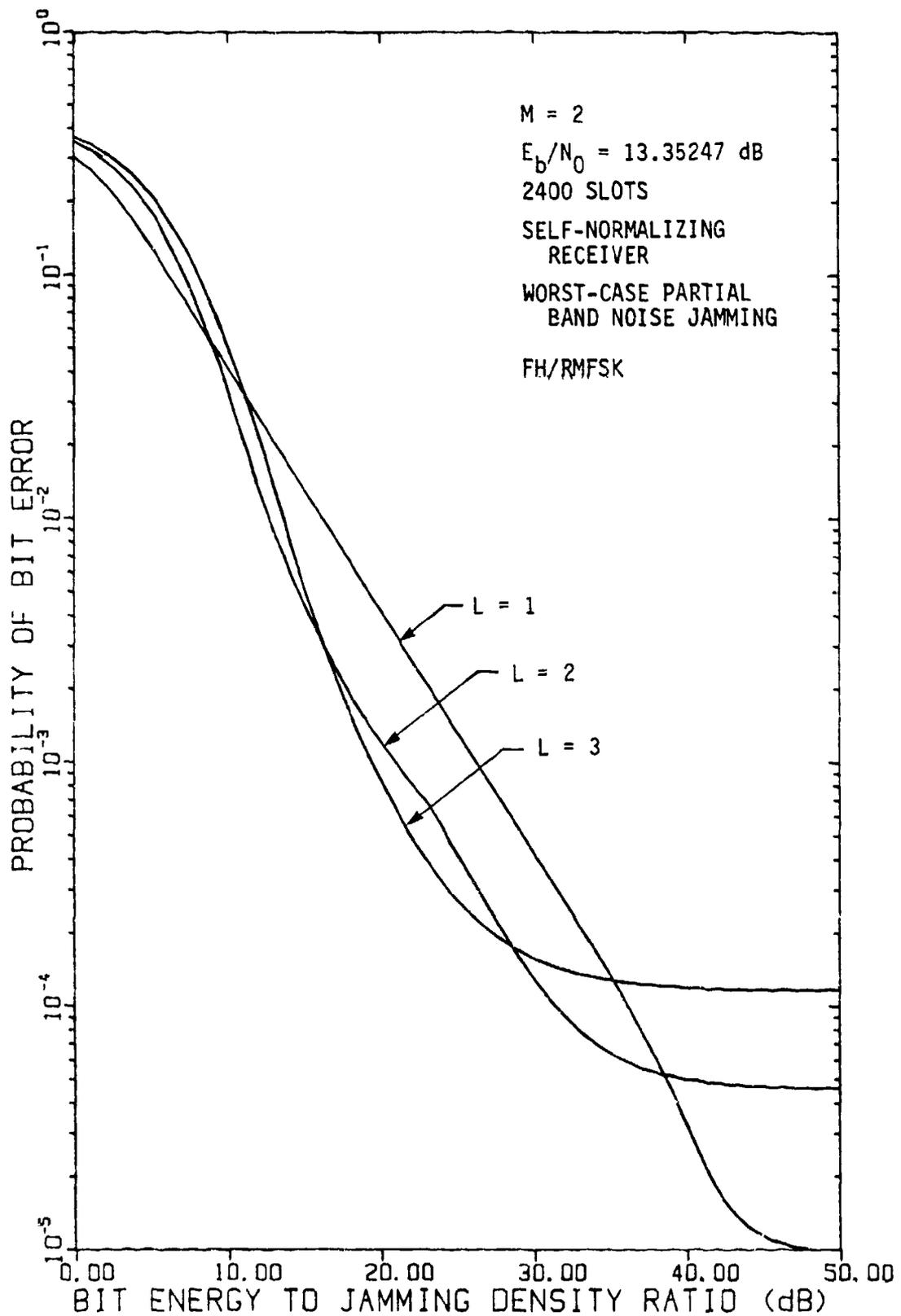


FIGURE 7.4-5 BIT ERROR PROBABILITY VS. BIT ENERGY TO JAMMING NOISE DENSITY RATIO FOR SELF-NORMALIZING RECEIVER IN WORST-CASE PARTIAL-BAND NOISE FOR $M=2$ AND $E_b/N_0=13.35 \text{ dB}$, WITH THE NUMBER OF HOPS/BIT (L) VARIED

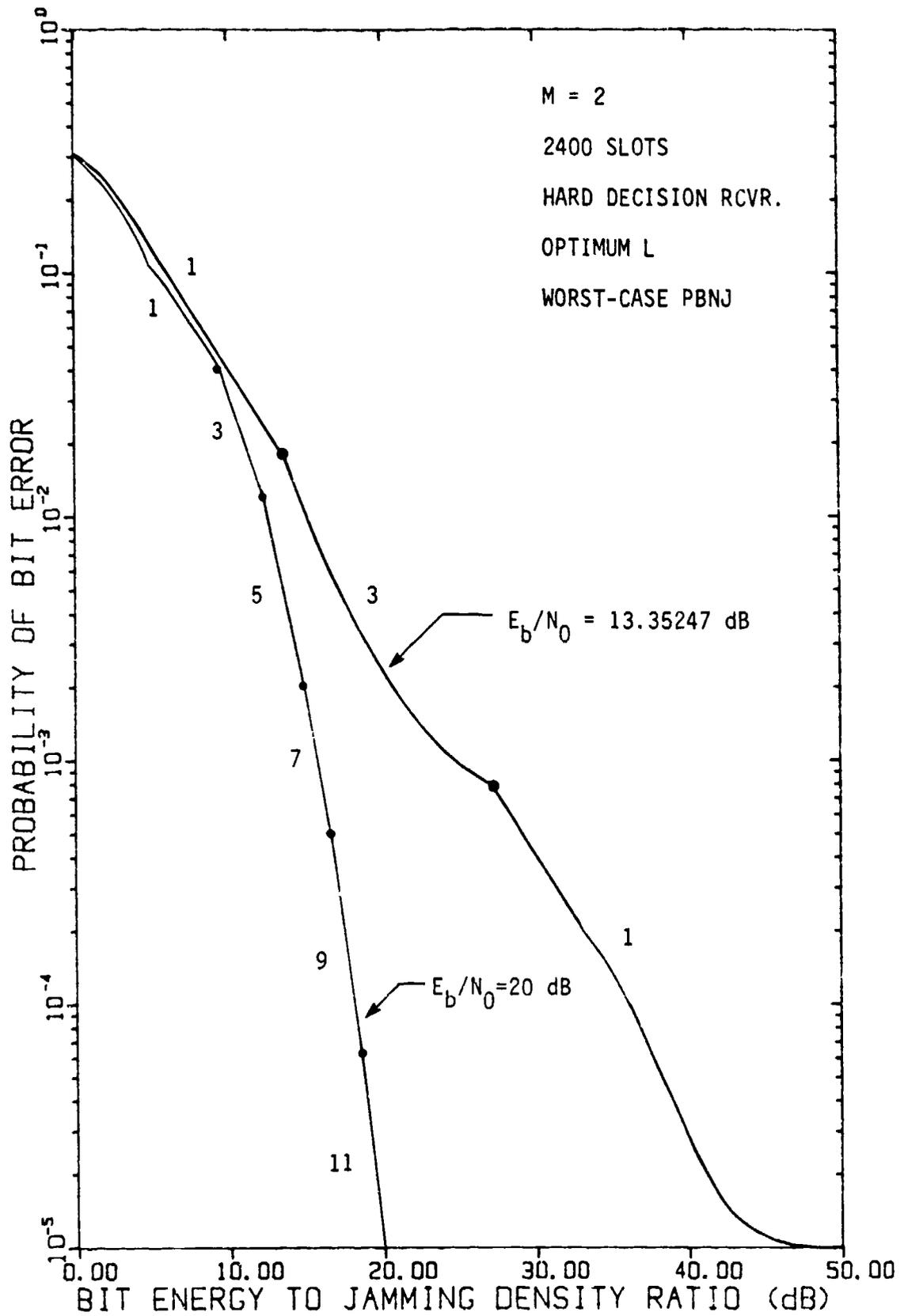


FIGURE 7.4-6 OPTIMUM DIVERSITY PERFORMANCE OF HARD DECISION RECEIVER FOR FH/RMFSK WITH $M=2$

8.0 ECCM RECEIVER IMPLEMENTATION STUDIES

In the previous sections, we analyzed the BER performance of various ECCM receiver processing schemes for uncoded FH/RMFSK radio systems in the presence of worst-case partial-band noise jamming (PBNJ).

Our objective has been to provide a comparison of these different systems, which vary extensively in their implementation complexities. These results will enable the ECCM system designer to weigh the engineering cost requirements of a particular receiver design versus the anti-jam effectiveness. Toward this end, we now explore practical issues related to the implementation of these different processing schemes along with an assessment of implementation effects.

8.1 IMPLEMENTATION ISSUES AND CONCEPTS

All receivers suppress the total noise jamming power by an amount equivalent to the system processing gain, defined as the ratio of the jammer bandwidth to the receiver bandwidth. Hopping the signal forces the jammer to spread its power over a wide bandwidth, but the jammer can maximize its effectiveness by selecting an optimum bandwidth, which is a certain fraction (γ) of the total hopping system bandwidth (W). This results in a BER which tends to be an inverse linear function of E_b/N_j , so that more than 40 dB of E_b/N_j is required to obtain BER's less than 10^{-4} . ECCM FH/MFSK and FH/MFSK systems counter this effect by using multiple hops per symbol, with the L hops per symbol combined at the receiver as in diversity transmission schemes.

We have demonstrated that effective jammer suppression is obtained by incorporating a nonlinear function in each M -ary channel prior to combining. The improvement in BER performance is realized by the fact that within a PBNJ environment, the nonlinear techniques (clipper, AGC, hard-decision, SNORM) mitigate the tendency of a jammed hop to dominate the symbol decision. Of these nonlinear

techniques we have studied, it was assumed that certain a priori information or perfect measurements are available to the clipper (SNR threshold) and AGC (noise powers) schemes; no such measuring tactics are necessary for the hard-decision or SNORM receivers. In what follows, we investigate the practicalities concerning the implementation and impact of non-ideal noise power measurements and threshold settings.

8.1.1 ECCM Receiver Information Requirements

In Table 8.1-1 we summarize the ECCM techniques used by the various receivers we have studied, including the information necessary for their implementation. The square-law linear combining receiver is presented as a baseline for comparison with the other, nonlinear combining types. Our specific interest here is to address the feasibility of implementation.

In the table, the nonlinear combining receivers are classified according to whether their anti-jam measures operate on a per-symbol basis (across all M channels) or on a per-channel basis. The per-symbol ECCM receivers include the ACJ-AGC, the hard-decision, and the SNORM receivers. Of these, the ACJ-AGC is seen as the only type utilizing a priori information on the received noise (thermal plus jamming), since it weights all channels on a given hop by

$$w_{mk} = w_k = \begin{cases} (\sigma_N^2)^{-1}, & \text{no channels jammed on hop } k \\ (\sigma_T^2)^{-2}, & \text{any channel jammed on hop } k \end{cases}$$
$$= \left[\max_m (\sigma_{mk}^2) \right]^{-1}. \quad (8.1-1)$$

TABLE 8.1-1 SUMMARY OF ECCM TECHNIQUES FOR THE RECEIVERS STUDIED

EXTENT OF HOP WEIGHTING OR ANTI-JAM-MEASURE	RECEIVER(S)	SUPPRESSION TECHNIQUE	INFORMATION REQUIRED
None	Linear Combining	None	None
Per-Symbol	Any-Channel-Jammed AGC (ACJ-AGC)	Scale Down Jammed Hops (Normalize Max Variance to 1)	$\text{Max}\{\sigma_{mk}^2\}$
	Hard-Symbol Decision Self-Normalizing (SNORM)	Limit Hops to One Vote Scale Down Jammed Hops (Normalize Channel Sum to 1)	None
Per-Channel	Individual Channel AGC (IC-AGC)	Scale Down Jammed Channels (Normalize All Variances to 1)	σ_{mk}^2
	Clipper	Limit Each Channel to Max Value of η	σ_N^2 and E_b/N_0 (to Set Value of η)

*Notes: (1) Hops numbered by k ($k=1,2,\dots,L$) and channels numbered by m ($m=1,2,\dots,M$)

(2) σ_{mk}^2 is the a priori variance of the total noise present in channel m at time k

(3) Since σ_{mk}^2 is assumed to be either σ_N^2 (unjammed) or σ_J^2 (jammed), "information required" may be interpreted as knowing whether a channel is jammed (IC-AGC) or a symbol is jammed (ACJ-AGC)

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Therefore, the information required is $w_k^{-1} = \max_m (\sigma_{mk}^2)$, which involves knowing σ_N^2, σ_T^2 , and whether any of the M channels is jammed. As a minimum, the ACJ receiver needs to know the ratio $\max_m (\sigma_{mk}^2) / \sigma_N^2$, since the operation of the receiver is unaffected if unjammed hops are left alone (weight = 1) and jammed hops are reduced by the factor σ_N^2 / σ_T^2 .

The hard-decision receiver is classified as a per-symbol ECCM receiver because, as we have observed in previous sections, its operation in effect limits each symbol piece (hop) to one vote in the M-ary majority logic decision, no matter how strongly a hop may have been jammed. Its operation does not require any a priori information or measurement.

The SNORM receiver derives its per-symbol weights from the M square-law envelope detector samples themselves:

$$w_k = \left(\sum_m x_{mk}^2 \right)^{-1} . \quad (8.1-2)$$

Therefore, it does not require a priori information or additional measurements in its operation.

The per-channel ECCM receivers include the IC-AGC and the clipper receivers, and both utilize a priori information. The IC-AGC weights each channel sample by the inverse of its a priori noise variance:

$$w_{mk} = (\sigma_{mk}^2)^{-1} . \quad (8.1-3)$$

In this manner, all channels on all hops are normalized to have unit noise variance; any channels which are jammed ($\sigma_{mk}^2 = \sigma_T^2$) are therefore suppressed. This technique involves knowing σ_N^2, σ_T^2 , and the jamming state or condition

of each channel. Alternately, the ratios σ_{mk}^2/σ_N^2 are needed, as a minimum.

The clipper receiver achieves an ECCM effect by "containing" any jammed channels; their contribution to the soft-decision sums cannot be any larger than the clipping threshold (η), no matter how strongly jammed. In order to set the threshold, both σ_N^2 and E_b/N_0 a priori values are needed since $\eta = \eta(\sigma_N^2, E_b/N_0)$ is chosen to minimize the error without jamming.

With respect to the additional receiver complexity required to develop information needed by the suppression technique, only the AGC schemes (jamming decision and normalization weights) and clipper (SNR levels) receivers need be addressed.

8.1.2 Measurement Approaches.

Implementation of the two AGC schemes requires differentiation between two zero-mean bandpass Gaussian noise processes with different variances which determine a jammed/unjammed channel state. In addition, to be useful as a quantity for a normalization weight in an AGC scheme, our measurement (noise-power estimate) must reflect as closely as possible in real-time the actual system state of the measured channel. This leads us to consider factors in both time and frequency domain representations of our measuring technique, i.e. the accuracy or quality of a band-limited channel noise-power measurement.

It is assumed throughout the measurement process that the data measures are sample records from a continuous stationary random process. Letting $x(t)$ be a single sample time history record from a zero-mean stationary (ergodic) Gaussian random process $\{x(t)\}$, the mean-square value (variance) of $\{x(t)\}$ can be estimated by time-averaging over a finite time interval τ as

$$\hat{\sigma}_x^2 = \frac{1}{\tau} \int_0^\tau x^2(t) dt \quad (8.1-4)$$

with the true mean-square value being

$$\sigma_x^2 = E [x^2(t)] \quad (8.1-5)$$

and is independent of t since $\{x(t)\}$ is stationary. Now the expected value of the estimate of σ_x^2 is

$$E[\hat{\sigma}_x^2] = \frac{1}{\tau} \int_0^\tau E[x^2(t)] dt = \frac{1}{\tau} \int_0^\tau \sigma_x^2 dt = \sigma_x^2. \quad (8.1-6)$$

Thus, $\hat{\sigma}_x^2$ is an unbiased estimate of σ_x^2 , independent of τ .

Regarding the quality of measurement, it is shown [19, p. 176] that the variance of a mean-square value estimate of band-limited white Gaussian noise with zero mean is

$$\text{Var}[\hat{\sigma}_x^2] = \frac{\sigma_x^2}{B\tau}. \quad (8.1-7)$$

Hence, we clearly see the inverse relationship of accuracy of our measurement to the measured channel time-bandwidth product; that is, for $B\tau \rightarrow \infty$ we would realize a "perfect" noise measurement.

8.1.2.1 A look-ahead measurement scheme.

Since the optimum power estimator uses a square-law detector [18] one could obtain a workable noise-power estimate by measuring the next slot to be hopped into by all M -ary channels; such a "look-ahead" scheme is illustrated in Figures 8.1-1 and 8.1-2.

Figure 8.1-1 shows the first part of the scheme in which we obtain the look-ahead receiver samples. Here we allow the receiver channel code-synchronized PN sequence generators to be delayed by one hop period (τ) in relationship to the respective measurement channel synchronization. In this manner we obtain the look-ahead samples at hop time k for use with

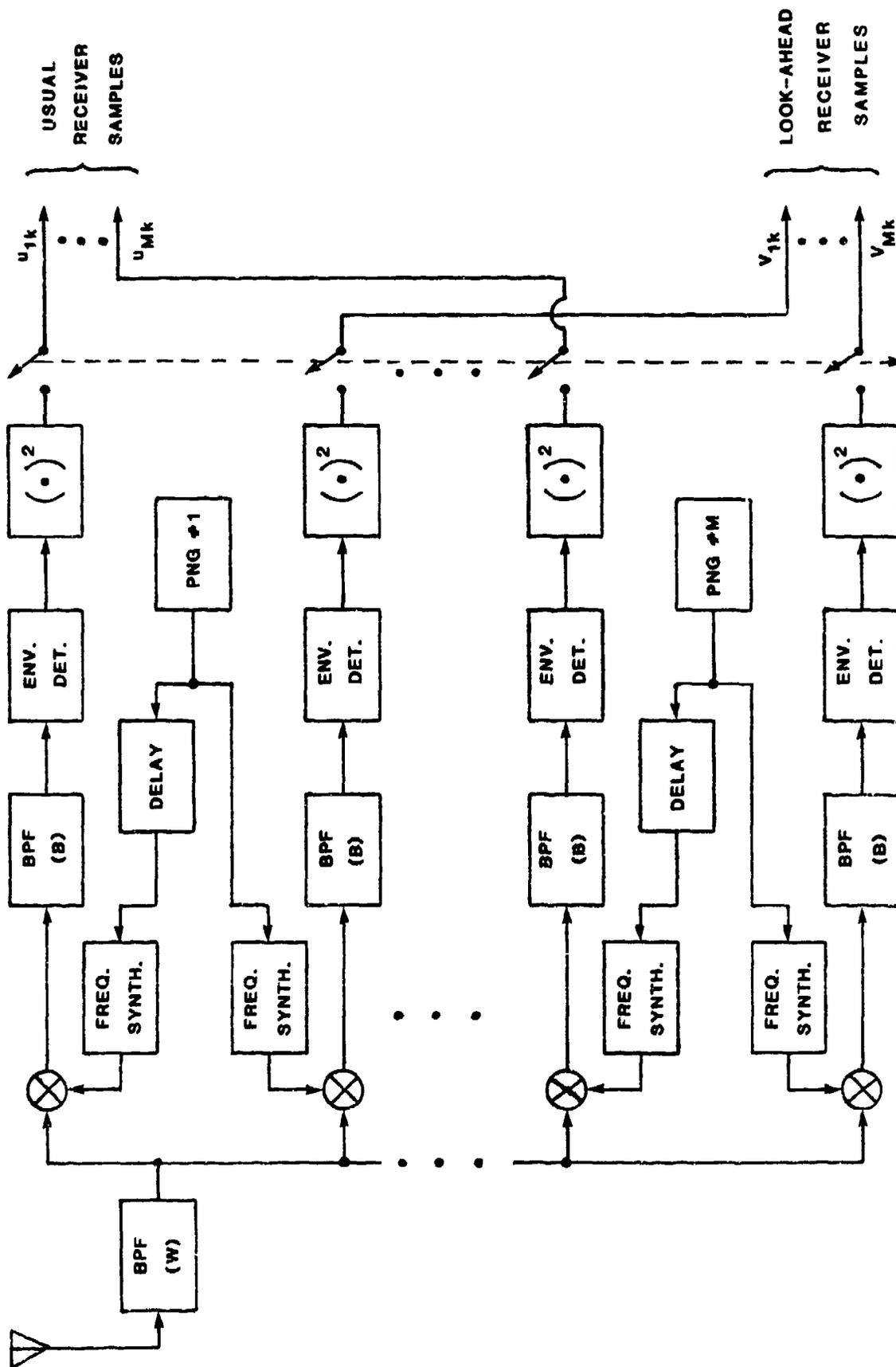


FIGURE 8.1-1 ACQUISITION OF JAMMING STATE DATA BY A "LOOK-AHEAD" SCHEME

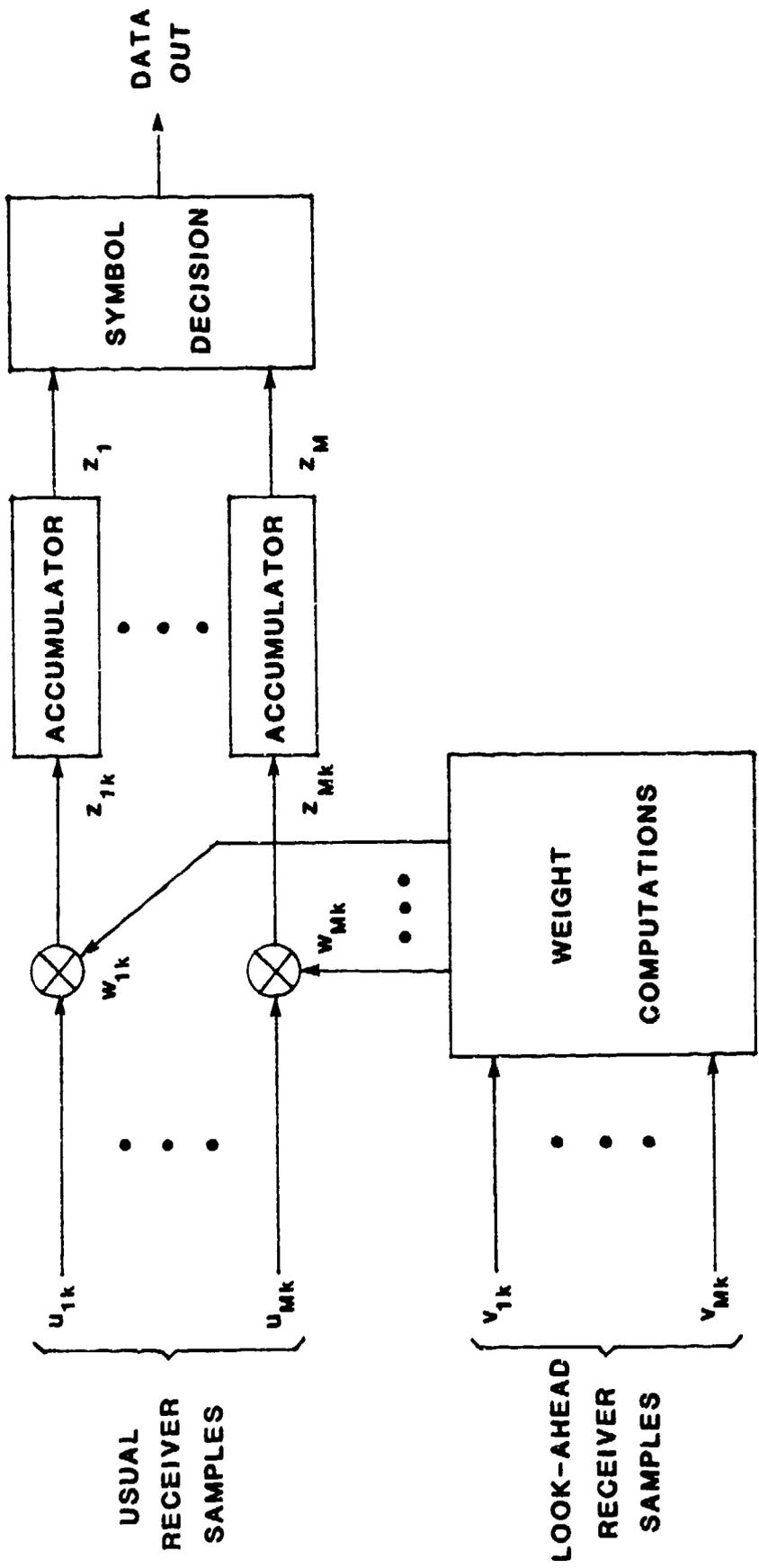


FIGURE 8.1-2 USE OF LOOK-AHEAD MEASUREMENTS TO DERIVE ECCM WEIGHTS

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the usual communicator receiver samples to be gathered at time $k+1$. We note that the look-ahead samples are assumed to be values of either σ_N^2 or σ_T^2 and that any type of spectral interference from other hoppers in the network is nonexistent. Also, this first part (Figure 8.1-1) of the look-ahead operation is the same for both IC- and ACJ-AGC receiver types.

In Figure 8.1-2 we show the use of the M-channel look-ahead measurements to derive the proper normalization weights. Thus, at time $k+1$ we obtain the following variables for accumulating in the L-hop diversity combining state for each of the M channels:

$$z_{mk} = x_{mk}^2 \cdot \left[\max_m (\hat{\sigma}_{mk}^2) \right]^{-1}, \text{ for ACJ} \quad (8.1-8)$$

$$z_{mk} = x_{mk}^2 \cdot (\hat{\sigma}_{mk}^2)^{-1}, \text{ for IC.} \quad (8.1-9)$$

Additionally, the weight computation network could incorporate a multi-hop/multi-channel processing stage (see Figure 8.1-3) to determine jamming state information (JSI) based upon multiple look-ahead measurements. Should the channel be jammed, we then have a one-sample estimate of the noise plus jamming power $\hat{\sigma}_T^2$. If unjammed, we obtain $\hat{\sigma}_N^2$ which is a one-sample measure of the channel's thermal noise power. The additional processing envisions each channel estimate $\hat{\sigma}_N^2$ as going into a smoothing operation, which would then output the improved σ_N^2 and σ_T^2 estimates. These values in turn are fed back to the jammed/unjammed decision blocks in the form of the detection threshold. The use of multiple-hop/multiple-channel measurements improves the quality of the noise power estimates by increasing the B_T product in (8.1-7). Details involved in both the jamming state detection and data smoothing of estimates are discussed later in this section.

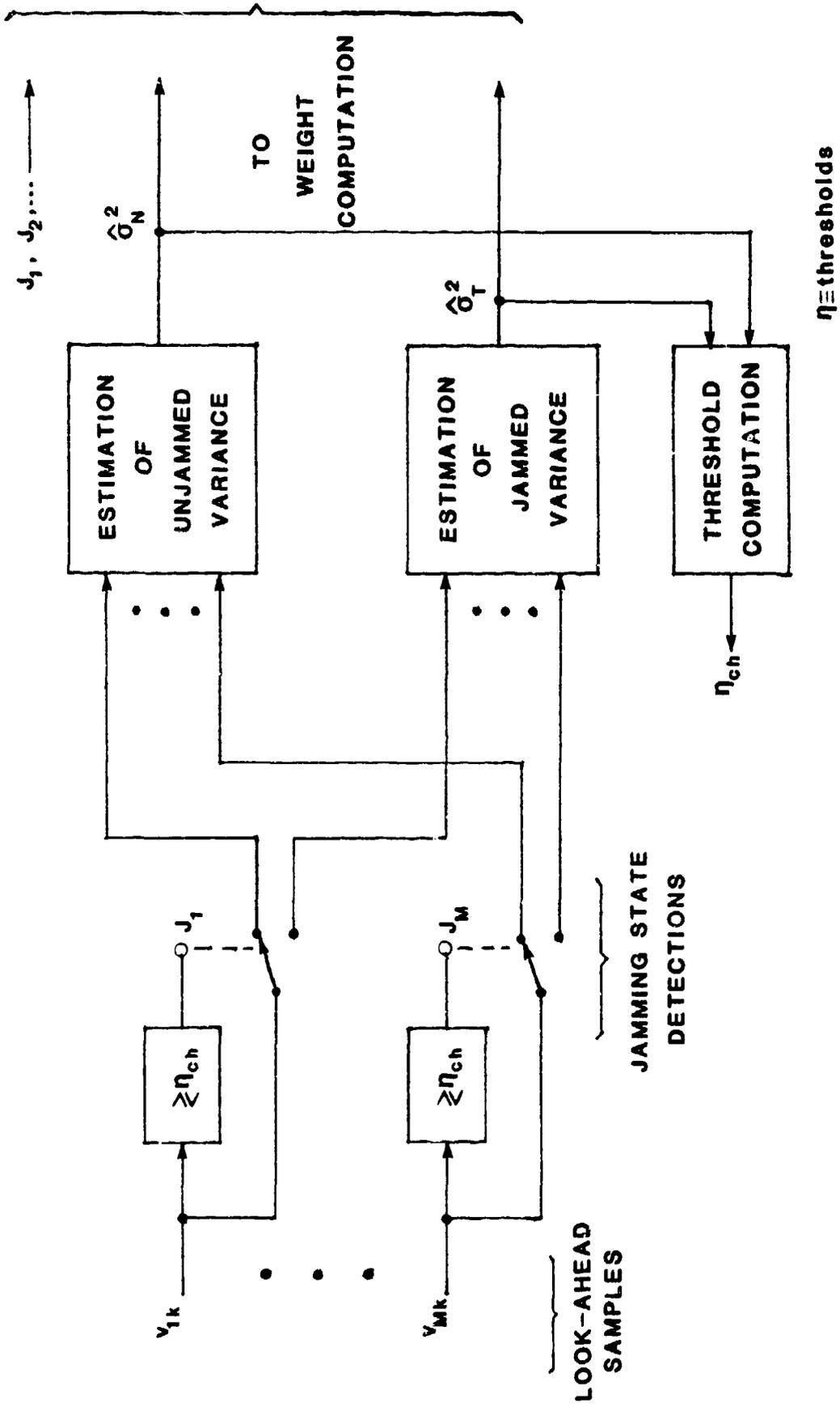


FIGURE 8.1-3 POSSIBLE USE OF MULTIPLE-HOP/MULTIPLE CHANNEL SAMPLES TO IMPROVE MEASUREMENTS

8.1.2.2 In-band measurement schemes.

One in-band measurement scheme assumes that $\hat{\sigma}_N^2$ and $\hat{\sigma}_T^2$ are to be measured between signal transmission times over the frequency-hopping bandwidth W . That is, a measurement process takes place during the communications "link idle" state, enabling the gathering of measurement samples over many hop periods. This procedure avoids the system complexity required by the look-ahead scheme, and is subject to the same caveats regarding stationarity of the noise (thermal, background, and jamming) environment across the hopping system bandwidth, as well as the corruption of purely noise estimates by the activity of other communications users during measurement intervals.

Receiver frame synchronization information or message preamble data could be used to decide when to cease or start taking measurement data samples. In concept, these measurements would enable high quality estimates of σ_N^2 and σ_T^2 to be developed, and/or perhaps even a stored estimate of the received noise power as a function of frequency.

Now, in order to detect the presence of jamming or to set the clipper's optimized threshold, we require knowledge of the received (average) thermal noise power $\sigma_N^2 = N_0 B$, where N_0 is the noise density in watts per hertz, assumed to be independent of frequency, and B is the channel bandwidth. In concept, an independent channel noise-power measurement system could be used to take several simultaneous measurements uniformly distributed within the thermal-noise-only (unjammed) portion of the hopping band W . The arithmetic average of these measurements would then be the estimate $\hat{\sigma}_N^2$. Assuming that σ_N^2 varies slowly, if at all, we would continuously correct the long-term moving average of $\hat{\sigma}_N^2$;

that is, a type of smoothing operation in which the processing scheme uses all measurements between times 0 and T to estimate the state of a system at a time t, where $0 < t < T$. This smoothed estimate of σ_N^2 over the time interval 0 and T can be denoted by $\sigma_N^2(t|T)$. Specifically, we envision a fixed-lag smoother in which a running smoothing solution lags the most recent measurement by a constant time delay Δ and is denoted as $\hat{\sigma}_N^2(T-\Delta|T)$. A reasonable value of Δ would be equal to the time for one symbol transmission.

Figure 8.1-4 illustrates a conceptual in-band measurement approach for the AGC type FH/RMFSK receivers. There are two modes: (a) between signal transmissions and (b) during signal transmissions. Between transmissions, the receiver continues to operate its synthesizers, detectors, and samplers to gather data for estimates of σ_N^2 and σ_T^2 , as mentioned above. During transmissions, jamming detection at the channel level (threshold η_{ch}) or symbol level (threshold η_{sym}) would furnish jamming state information (JSI) for selection of AGC weights, using thresholds based on the estimates of σ_N^2 and σ_T^2 . Possibly the samples received during transmissions could be used also for the variance estimation by feeding back the symbol decision to identify the channel with the signal, as suggested in the diagram.

With respect to the clipper receiver, Figure 8.1-5 depicts a scheme for setting the optimized clipping threshold η_0 . Toward this end, we need to obtain a current estimate of the clipper receiver's SNR. Similar to the previously described two-mode in-band noise-power measurement concept for the AGC receivers, an individual channel measurement system would likewise be used to estimate the received signal power S. Several measurements would be taken

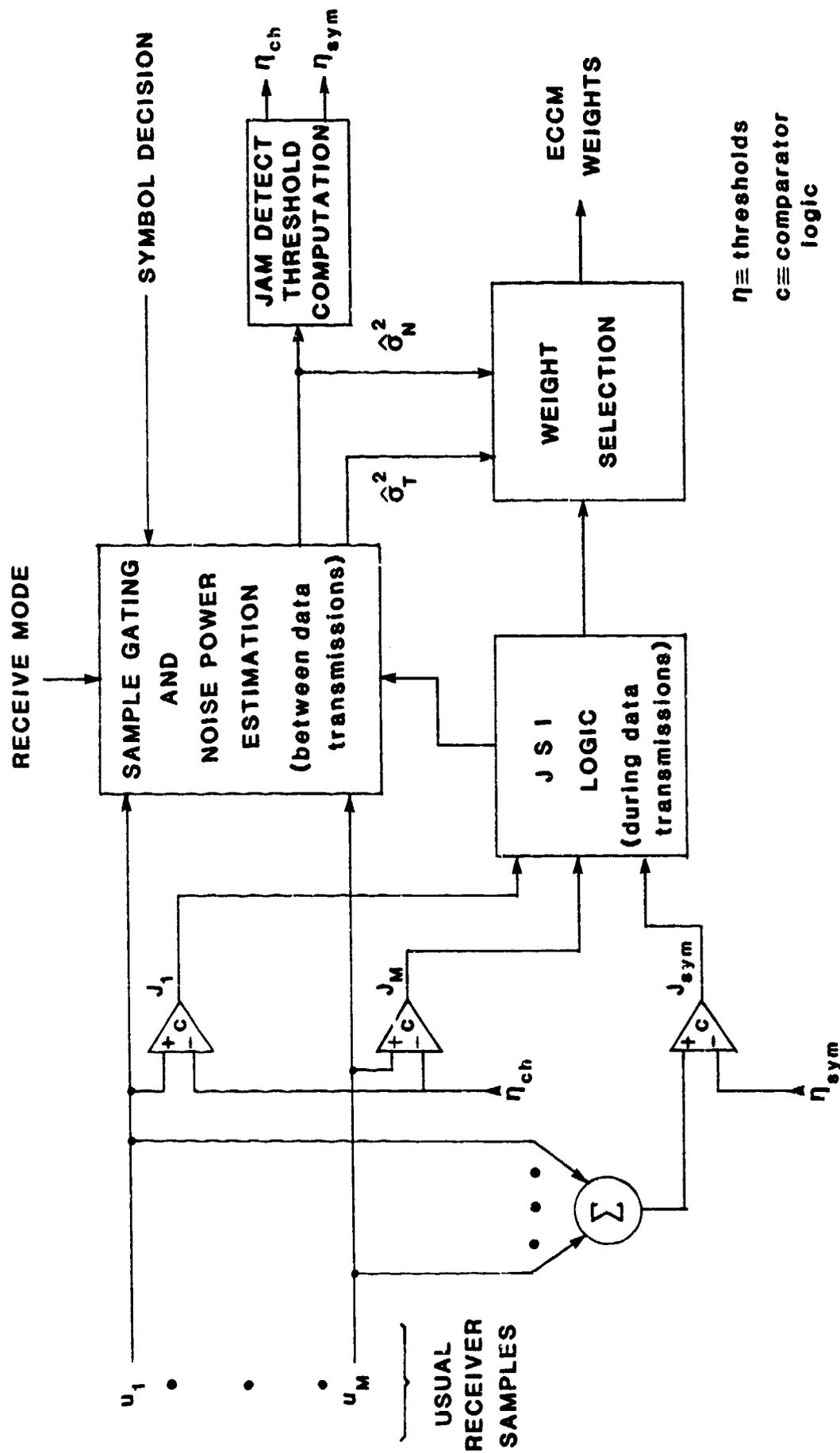


FIGURE 8.1-4 CONCEPTUAL PROCESSING OF IN-BAND MEASUREMENTS TO IMPLEMENT AGC RECEIVERS

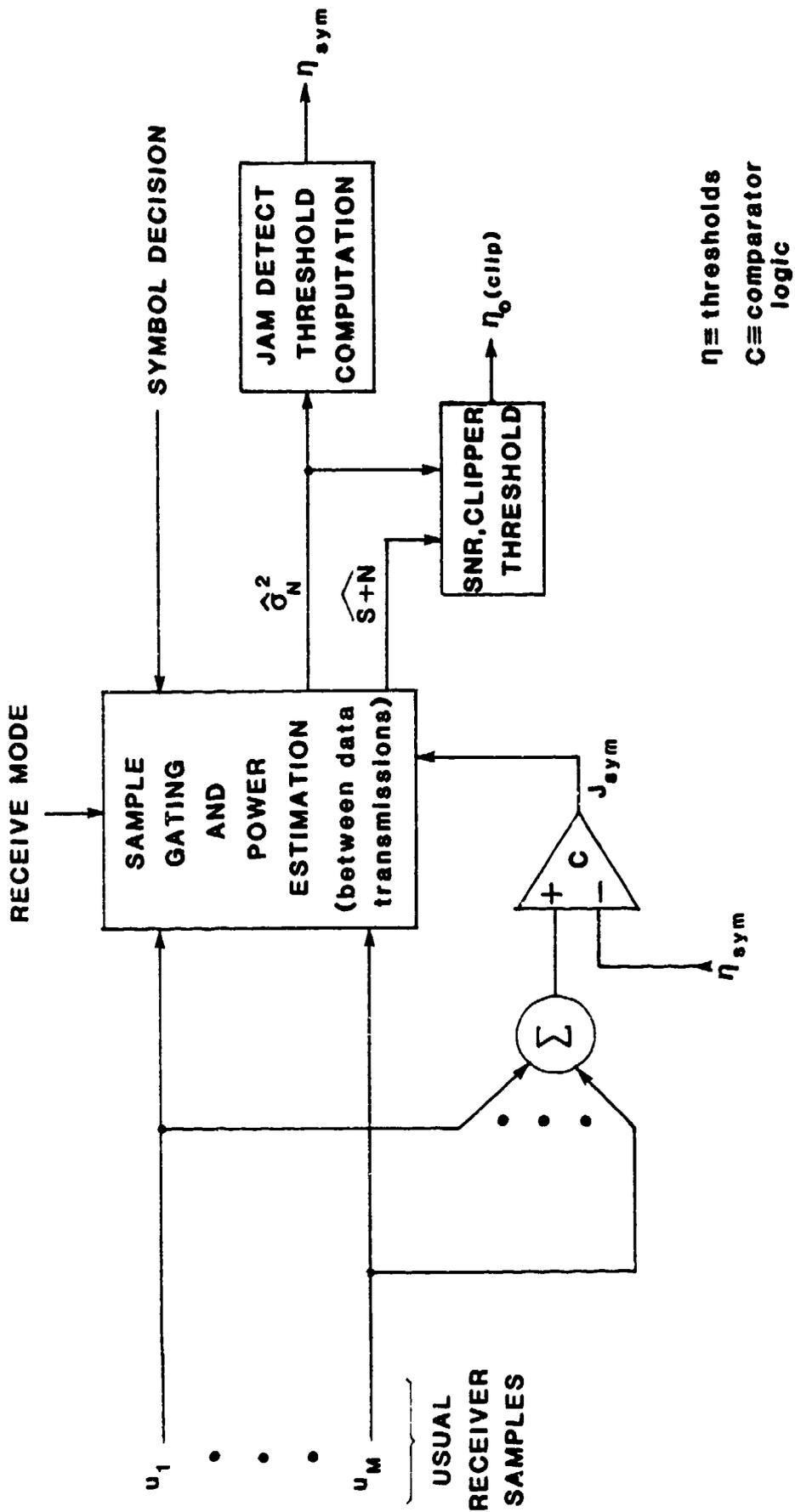


FIGURE 8.1-5 PROCESSING OF IN-BAND MEASUREMENTS TO IMPLEMENT CLIPPER RECEIVER

of the received signal plus thermal noise within the unjammed portion of the hopping band W , using symbol decision feedback to identify the signal channel. The arithmetic average of these measurements forms an estimate of signal plus noise power, \hat{P}_{sn} . This estimate could also be refined by a fixed-lag (per-symbol) smoothing operation. Hence, we would obtain

$$\hat{\rho}_N = \frac{\hat{S}}{\hat{\sigma}_N^2} = \frac{\hat{P}_{sn}}{\hat{\sigma}_N^2} - 1 = \frac{K}{L} \cdot \frac{\hat{E}_b}{N_0}, \quad (8.1-10)$$

the signal-to-noise ratio for a given hop dwell time, to be updated on a per-symbol basis.

Table 5.3-1 showed a summary of the clipper receiver values of η_0 for a given L , M , and E_b/N_0 . Note that these numeric values were obtained only for values of E_b/N_0 such that $P_b(e) = 10^{-5}$ for two MFSK systems in the absence of jamming. We point out that, in practice, new values of η_0 need to be computed for each different value of E_b/N_0 . Therefore the actual clipper receiver would require that matrices of η_0 values be stored in a read-only memory (ROM) look-up table.

8.1.3 Jamming State Decisions.

Implementation of a jamming state decision scheme would be based upon the following assumptions: (1) that a look-ahead or in-band measurement scheme is utilized, and (2) that the noise power estimates of $\hat{\sigma}_N^2$ and $\hat{\sigma}_T^2$ are readily available in time. Ultimately, the criterion for making a jammed/unjammed state detection is predicated upon a single observation (per channel) or a multiple observation (per symbol). The ACJ-AGC receiver requires detection of a jammed symbol hop whereas the IC-AGC requires detection of jamming in each of the M channels on each hop.

In both the symbol and the channel cases, the problem is one of deciding between Gaussian noise with variance σ_N^2 and Gaussian noise with variance σ_T^2 . Due to the lack of a priori jamming probabilities (γ) or cost functions, we utilize the Neyman-Pearson criterion as our hypothesis testing technique. Its test objective is to maximize the probability of detection (P_D) for a given probability of false alarm (α) and is accomplished by employing a likelihood ratio test.

8.1.3.1 Jammed channel detection.

A basic problem in determining a jammed/unjammed state with an individual channel look-ahead scheme is in accounting for the signal itself. Recalling that the look-ahead measurement channel is one hop period "ahead" of the normal receiver channel, the situation can arise in which the measurement channel is actually the present signal channel. However, we first analyze the case for a measured channel without a signal present.

Using the narrowband Gaussian process representation for the channel samples, the hypotheses to be considered are

$$H_0: p(\underline{n}_c, \underline{n}_s) = \frac{1}{2\pi\sigma_N^2} \exp\left\{-\frac{n_c^2 + n_s^2}{2\sigma_N^2}\right\} \quad (8.1-11)$$

$$H_1: p(\underline{n}_c, \underline{n}_s) = \frac{1}{2\pi\sigma_T^2} \exp\left\{-\frac{n_c^2 + n_s^2}{2\sigma_T^2}\right\} \quad (8.1-12)$$

where H_0 is the unjammed case and H_1 is the jammed case. For the likelihood ratio we obtain

$$\Lambda = \frac{\sigma_N^2}{\sigma_T^2} \exp \left\{ - \frac{(n_c^2 + n_s^2)}{2} \left(\frac{1}{\sigma_T^2} - \frac{1}{\sigma_N^2} \right) \right\} \quad (8.1-13)$$

and in comparing the log-likelihood function to a threshold we have

$$\log \Lambda = \log \left(\frac{\sigma_N^2}{\sigma_T^2} \right) + \frac{x^2}{2} \left(\frac{\sigma_T^2 - \sigma_N^2}{\sigma_N^2 \sigma_T^2} \right) \underset{H_0}{\overset{H_1}{>}} \eta \quad (8.1-14)$$

where $x^2 = n_c^2 + n_s^2$ is our estimated look-ahead noise-power measurement or sample test statistic for a single channel.

The value of x^2 to decide that jamming is present (H_1 true) is

$$x^2 \geq \frac{2\sigma_N^2 \sigma_T^2}{\sigma_T^2 - \sigma_N^2} \left[\eta - \log \left(\frac{\sigma_T^2}{\sigma_N^2} \right) \right] = n_{ch}, \quad \sigma_T^2 > \sigma_N^2. \quad (8.1-15)$$

For the false alarm probability we have

$$P_{FA} = \Pr \left\{ x^2 > n_{ch} | H_0 \right\} = \alpha = e^{-n_{ch}/2\sigma_N^2}. \quad (8.1-16)$$

Similarly, the probability of detection is

$$P_D = e^{-n_{ch}/2\sigma_T^2} \quad (8.1-17)$$

with a receiver operating characteristic for our jamming detector given by

$$P_D = \left[\alpha \right]^{\sigma_N^2 / \sigma_T^2} = \left[\alpha \right]^{1/K}. \quad (8.1-18)$$

Now in the case of the look-ahead channel sampling a signal channel slot, these samples (x^2) are independent noncentral chi-squared random variables with two degrees of freedom, multiplied by the total channel noise power σ_{ch}^2 , and with noncentrality parameters

$$\lambda = 2S/\sigma_{ch}^2 = \begin{cases} 2S/\sigma_N^2 = 2\rho_N, & \text{hop unjammed} \\ 2S/\sigma_T^2 = 2\rho_T, & \text{hop jammed.} \end{cases} \quad (8.1-19)$$

Therefore, the pdf of a given sample $u=x^2$ is

$$f_{sig}(u) = \frac{1}{2\sigma_{ch}^2} e^{-(u+2S)/2\sigma_{ch}^2} I_0(\sqrt{2Su}/\sigma_{ch}^2) \quad (8.1-20a)$$

$$= \begin{cases} (1/2\sigma_N^2) e^{-(u+2S)/2\sigma_N^2} I_0(\sqrt{2Su}/\sigma_N^2), & u \geq 0, \text{ hop unjammed} \\ (1/2\sigma_T^2) e^{-(u+2S)/2\sigma_T^2} I_0(\sqrt{2Su}/\sigma_T^2), & u \geq 0, \text{ hop jammed} \end{cases} \quad (8.1-20b)$$

where S signifies power in the signal itself. Consequently, the probability of a false alarm and the probability of detection can be written as

$$P_{FA} = Q(\sqrt{2\rho_N}, \sqrt{n_{ch}/\sigma_N^2}) \quad (8.1-21)$$

$$P_D = Q(\sqrt{2\rho_T}, \sqrt{n_{ch}/\sigma_T^2}) \quad (8.1-22)$$

where $Q(x,y)$ is Marcum's Q -function.

8.1.3.2 Jammed symbol detection.

In the jammed M-ary symbol detection case, we have the two hypotheses of: (1) no channel is jammed, and (2) any of the M channels is jammed, i.e. ACJ. Hence, the situation is one of multiple alternative hypothesis testing given as

$$H_0: \sigma_m^2 = \sigma_N^2, \quad m = 1, 2, \dots, M; \quad (8.1-23)$$

$$H_1: \sigma_m^2 = \sigma_J^2, \quad 2^M - 1 \text{ possible combinations of jammed channels};$$

where σ_m^2 is a parameter in the likelihood function of the measurement samples which is written as

$$p(\underline{n}_c, \underline{n}_s) = \prod_m \left(\frac{1}{2\pi\sigma_m^2} \right) e^{-x_m^2/2\sigma_m^2}. \quad (8.1-24)$$

As in the jammed channel detection case, we require a current estimate of σ_N^2 obtained from a look-ahead or in-band noise-power measurement scheme. Furthermore, an estimate of SNR ($\hat{\rho}_N$) is also needed for threshold determination when accounting for the signal itself being detected in a look-ahead scheme.

Since there are many alternatives to H_0 as expressed by (8.1-23), we consider the simplified test of whether the sum of the channel samples exceeds a threshold.

For the case of the signal being present in one of the channels, the sum of the samples ($u_m = x_m^2$) is distributed as, with no jamming,

$$\sum_{m=1}^M u_m = \sigma_N^2 \chi^2(2M, 2\rho_N) \quad (8.1-25)$$

i.e., a non-central chi-squared variable with $2M$ degrees of freedom, non-central parameter $2\rho_N$, and weighted by the current estimate of σ_N^2 . In the noise-only case, the distribution of the summed decision variables is a central chi-squared distribution, written as

$$\sum_{m=1}^M u_m = \sigma_N^2 \chi^2(2M) . \quad (8.1-26)$$

A scheme that may provide a workable jammed symbol detection when a signal slot can be jammed is realized by discarding the maximum of the u_m variables which results in the distributional assumption

$$\sum_{m=1}^{M-1} u_m - u_{\max} = \sigma_N^2 \chi^2(2M-2). \quad (8.1-27)$$

Note that the distribution of (8.1-27) is written as a central chi-square distribution which assumes that the signal itself is always stronger than the jamming. That is, the presence of weak jamming can easily be detected in the absence of the now discarded stronger signal. Should the jamming be strong, we have a situation in which the signal channel is included in the leftover sum. However, this is an acceptable scenario because the signal channel power will help in deciding if a jammed symbol condition exists.

The probability of a false alarm for this scheme is obtained by applying the methodology to formulate the non-signal channel probabilities for ACJ-AGC receivers as discussed in Section 4.3. Specifically, the probability of the sum of the non-signal channel measurement samples being less than the normalized symbol threshold on a given hop is written as

$$\Pr \left\{ \sum u_m < \frac{n_{\text{sym}}}{\sigma_N^2} \right\} = 1 - \frac{\Gamma(2M-2; n_{\text{sym}}/2\sigma_N^2)}{\Gamma(2M-2)} . \quad (8.1-28)$$

Hence, the probability of a false-alarm for jammed symbol detection is simply

$$P_{\text{FA}(\text{sym})} = \frac{\Gamma(2M-2; n_{\text{sym}}/2\sigma_N^2)}{\Gamma(2M-2)} . \quad (8.1-29)$$

Another method to implement jammed symbol detection in the presence of a signal would utilize a combination of the individual channel jammed detector outputs. For example, in the case of $M=4$ we can employ the combinatorial logic of any two individual channel detectors' outputs "ANDed" to produce a symbol jamming decision from the six possible combinations. The detection of the "any channel jammed" condition will be the logic variable

$$\begin{aligned} J_{\text{symbol}} = & (J_1 \cdot J_2) + (J_1 \cdot J_3) + (J_1 \cdot J_4) \\ & + (J_2 \cdot J_3) + (J_2 \cdot J_4) + (J_3 \cdot J_4) \end{aligned} \quad (8.1-30)$$

where it is assumed that detection of the signal itself has triggered the decision of a jammed channel (J_i) for a particular channel. Thus, we are able to realize a jammed symbol detection when a signal is present.

For the case of any M we have the overall false alarm probability (when no signal is present)

$$P_{\text{FA}(\text{sym})} = \sum_{n=2}^M \binom{M}{n} P_{\text{FA}(\text{CH})}^n (1 - P_{\text{FA}(\text{CH})})^{M-n} \quad (8.1-31)$$

which simplifies to

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$$P_{FA(sym)} \approx \binom{M}{2} P_{FA(CH)}^2 \quad (8.1-32)$$

as the overall false-alarm rate for a jammed symbol detection.

8.2 ASSESSMENT OF IMPLEMENTATION EFFECTS

In this second part of our implementation studies, we investigate the effects of the previously discussed implementation schemes on the performance of the ECCM receivers. We demonstrate the necessary adjustments involved in reformulating the probability of error expressions which are now conditioned on estimated (measured) parameters instead of assumed "perfect" measurement quantities. Our objective is to assess the "return on the investment" realized by resorting to the complex measurement schemes needed to implement the AGC receivers, which for ideal measurements achieve the best ECCM receiver performances in worst-case PBNJ. That is, we seek to answer the question, "Will practical implementations of the AGC receivers continue to outperform the simple SNORM and hard-decision receivers, which require no measurement?"

8.2.1 Methodology for Direct Assessment.

Analyses of the error performance of the implemented AGC schemes are extremely difficult for the following reasons. First, we must account for the measured (estimated) quantities $\hat{\sigma}_N^2$ and $\hat{\sigma}_T^2$ as random variables imbedded in the probability of error expressions. Second, the effect of errors in the JSI decisions upon the $P(e)$ expressions must also be considered. Previous analytical results in this report were derived on the assumption that perfect measurements were obtained for σ_N^2 and σ_T^2 ; this provided a lower bound on the error performance to be realized in practice. This ideal total error probability can be expressed

parametrically by

$$P(e) \equiv P(e; \underbrace{\gamma, E_b/N_0, E_b/N_J, L, M}_{\text{actual parameters}}; \underbrace{\sigma_N^2, \sigma_T^2, S}_{\text{parameters assumed by receivers}}) \quad (8.2-1)$$

and can be written as

$$P(e) = \sum \Pr\{[v]\} P_b(e|[v]), \quad (8.2-2)$$

where $[v]$ is a matrix describing the jamming event (see Section 2.2) over the L-hop diversity. The above expressions must now be restated as,

$$P(e) \equiv P(e; \gamma, E_b/N_0, E_b/N_J, L, M; \hat{\sigma}_N^2, \hat{\sigma}_T^2, \hat{S}) \quad (8.2-3)$$

$$P(e) = \sum \Pr\{[v], [\hat{v}]\} P(e|[v], [\hat{v}]) \quad (8.2-4)$$

in accounting for $\hat{\sigma}_N^2, \hat{\sigma}_T^2, \hat{S}$, as well as estimates in JSI.

For example, in the ideal situation the IC-AGC receiver decision variables are

$$z_m = \sum_{k=1}^L z_{mk}; \quad m=1, 2, \dots, M; \quad (8.2-5a)$$

where each z_{mk} is the square-law envelope detector sample x_{mk}^2 multiplied by the weight

$$w_{mk} = \begin{cases} 1/\sigma_N^2, & \text{channel } m \text{ not jammed on hop } k \\ 1/\sigma_T^2, & \text{channel } m \text{ jammed on hop } k. \end{cases} \quad (8.2-5b)$$

This ideal normalization results in the $\{z_{mk}\}$ being all unscaled chi-square random variables, as discussed in Section 4.

Now if the a priori quantities σ_N^2 and σ_T^2 are not available and the jamming condition of the channels is not known, we must use estimates $\hat{\sigma}_N^2$ and $\hat{\sigma}_T^2$, and also decide whether the channel is jammed. This results in the weights

$$\hat{w}_{mk} = \begin{cases} (\hat{\sigma}_N^2)^{-1} (1-P_{Fm}) + (\hat{\sigma}_T^2)^{-1} P_{Fm} = W_0, \text{ not jammed;} \\ (\hat{\sigma}_N^2)^{-1} (1-P_{Dm}) + (\hat{\sigma}_T^2)^{-1} P_{Dm} = W_1, \text{ jammed;} \end{cases} \quad (8.2-6)$$

where P_{Fm} and P_{Dm} are per-channel jamming false alarm and detection probabilities. In this description, it is assumed that the variance estimates are developed from look-ahead or in-band measurement data prior to the symbol being processed, and that the channel jamming decision is based on a one-hop look-ahead scheme. In the absence of signals in the look-ahead channels,

$$P_{Fm} = P_F(\hat{\sigma}_N^2, \hat{\sigma}_T^2; \sigma_N^2, \sigma_T^2) \text{ and } P_{Dm} = P_D(\hat{\sigma}_N^2, \hat{\sigma}_T^2; \sigma_N^2, \sigma_T^2);$$

that is, these probabilities are the same for each of the M symbol channels.

Thus, after accumulating the L hops, the decision statistics are, conditioned on jamming events and measurements,

$$z_1 = W_0 \sigma_N^2 \chi^2 [2(L-l_1); 2(L-l_1) \rho_N] \\ + W_1 \sigma_T^2 \chi^2 [2l_1; 2l_1 \rho_T]; \quad (8.2-7a)$$

$$z_m = W_0 \sigma_N^2 \chi^2 [2(L-l_m)] + W_1 \sigma_T^2 \chi^2 (2l_m), \quad m > 1; \quad (8.2-7b)$$

where $\underline{\ell} = (\ell_1, \ell_2, \dots, \ell_M)$ is the jamming event vector of the number of hops jammed in each channel. Recall now that the linear combining receiver (LCR) has the normalized decision statistics

$$u_1 = z_1 / \sigma_N^2 = \chi^2 [2(L - \ell_1); 2(L - \ell_1)\rho_N] + K\chi^2 [2\ell_1; 2\ell_1\rho_T]; \quad (8.2-8a)$$

$$u_m = z_m / \sigma_N^2 = \chi^2 [2(L - \ell_m)] + K\chi^2 (2\ell_m), \quad m > 1; \quad (8.2-8b)$$

where $K = \sigma_T^2 / \sigma_N^2$. We therefore recognize that, conditioned on the measurements, the implemented IC-AGC receiver's BER will have the same functional form as the LCR's with the new K value

$$K' = \frac{\sigma_T^2 W_1}{\sigma_N^2 W_0} = \frac{\sigma_T^2}{\sigma_N^2} \cdot \frac{\hat{\sigma}_T^2 (1 - P_D) + \hat{\sigma}_N^2 P_D}{\hat{\sigma}_T^2 (1 - P_F) + \hat{\sigma}_N^2 P_F}. \quad (8.2-9)$$

Evaluation of the effect of the measurements and JSI decisions then involves numerically averaging the LCR error probability (with K replaced by (8.2-9)) over the distributions of $\hat{\sigma}_N^2$ and $\hat{\sigma}_T^2$.

8.2.2 In Search of an Upper Bound: Simplified Measurement Models.

Equations (8.2-7) to (8.2-9) reveal that the implemented IC-AGC receiver statistics more or less tend toward those of the ineffective LCR, depending on the quality of the measurements. This fact underscores the important role of the a priori information utilized by the AGC receivers in

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their superior performance. We can take the position that the ideal AGC performances calculated in this report represent a lower bound on achievable BER, though perhaps not the lowest bound*, and then seek an upper bound instead of attempting the arduous and time-consuming direct analysis of implemented systems. Such an upper bound, if sufficiently tight, would be suitable for comparison with the BER results for the receivers not employing a priori information.

How shall we obtain an upper bound? Since the performance degradation associated with the receiver implementations is related to the quality of the measurements, we realize that any bound would be directly identified with a particular measurement approach, and parametric in the features of that approach (such as number of samples taken). We fully anticipate, for example, that an upper bound on the implemented system's BER would decrease as the number of samples used in the measurement increases. Therefore, it is reasonable to consider possible implementations utilizing one sample as candidates for systems whose performances represent an upper bound on what is achievable in the same manner as the ideal systems represent a lower bound.

For the ACJ-AGC FH/MFSK receiver, we consider the simplified version that we call the "practical ACJ" (PACJ) receiver. Since the ideal ACJ receiver uses the weights $w_k = (\max_m \sigma_{mk}^2)^{-1}$, we stipulate that the PACJ uses the weights

$$w_k = (\max_m x_{mk}^2). \quad (8.2-10)$$

*In Section 7 it was observed that the SNORM receiver can outperform the AGC receivers in some, limited circumstances.

This approach in effect utilizes the received square-law envelope detector samples themselves as (one sample) measurements of the noise power in each channel, and thus is a form of in-band measurement which is very simple indeed compared to schemes discussed in Section 8.1.

We note in passing that this PACJ receiver is related to the hard-decision receiver in the following way: the HD decision statistics $\{z_{mk}\}$ are the PACJ decision statistics after being subjected to a two-level quantization. That is,

$$z_{HD} = \begin{cases} 1, & z_{PACJ} \geq 1 \\ 0, & z_{PACJ} < 1. \end{cases} \quad (8.2-11)$$

For the IC-AGC receiver, we postulate that a one-hop look-ahead implementation yields one detector sample in each of the M channels just prior to the actual symbol's occupancy of those channels.* Since the ideal IC-AGC uses the weights $w_{mk} = (\sigma_{mk}^2)^{-1}$, we can treat the look ahead samples $\{v_{mk}\}$ as one-sample estimates of noise power and specify the "practical IC" (PIC) weights

$$w_{mk} = (v_{mk})^{-1}. \quad (8.2-12)$$

In what follows, we evaluate each of these practical receivers for the case of M=2, L=2 in order to compare them with their respective ideal receivers. We also can consider the SNORM and hard-decision receivers as

*Alternatively, the hopping and symbol rates could be reduced by one-half in order to sample the channel first, then receive the transmission; look-ahead is avoided in this way, at the expense of data rates.

"practical" AGC implementations, and will continue to exhibit their BER results for comparison purposes.

8.2.3 Example Evaluations of Practical AGC Receivers .

We now find the error probabilities for the PACJ and PIC receivers for M=2 and L=2.

8.2.3.1 Analysis of the PACJ receiver.

The PACJ receiver for M=2 is diagrammed in Figure 8.2-1. The decision statistics are

$$z_m = \sum_{k=1}^L \frac{x_{mk}^2}{\max_m(x_{mk}^2)}, \quad m = 1, 2. \quad (8.2-13)$$

In Appendix C it is shown that this receiver for L=2 and FH/RMFSK has the bit error probability

$$P_b(e) = 2 \int_0^1 dx f(x)G(x) + G^2(1) \quad (8.2-14a)$$

where $K = \sigma_T^2/\sigma_N^2$ and

$$\begin{aligned} f(x) = & \pi_0 \frac{1}{(x+1)^2} \exp \left\{ -\frac{\rho_N x}{(x+1)} \right\} \left[1 + \frac{\rho_N}{x+1} \right] \\ & + \pi_1 \frac{K}{(Kx+1)^2} \exp \left\{ -\frac{K\rho_T x}{(Kx+1)} \right\} \left[1 + \frac{\rho_T}{Kx+1} \right] \\ & + \pi_1 \frac{K}{(x+K)^2} \exp \left\{ -\frac{\rho_N x}{(x+K)} \right\} \left[1 + \frac{K\rho_N}{x+K} \right] \\ & + \pi_2 \frac{1}{(x+1)^2} \exp \left\{ -\frac{\rho_T x}{(x+1)} \right\} \left[1 + \frac{\rho_T}{x+1} \right]. \end{aligned} \quad (8.2-14b)$$

and

$$\begin{aligned} G(x) = & \pi_0 \frac{x}{x+1} e^{-\rho_N/(1+x)} + \pi_2 \frac{x}{x+1} e^{-\rho_T/(1+x)} \\ & + \pi_1 \frac{Kx}{Kx+1} e^{-\rho_N/(1+Kx)} + \pi_1 \frac{x}{x+K} e^{-K\rho_T/(x+K)}. \end{aligned} \quad (8.2-14c)$$

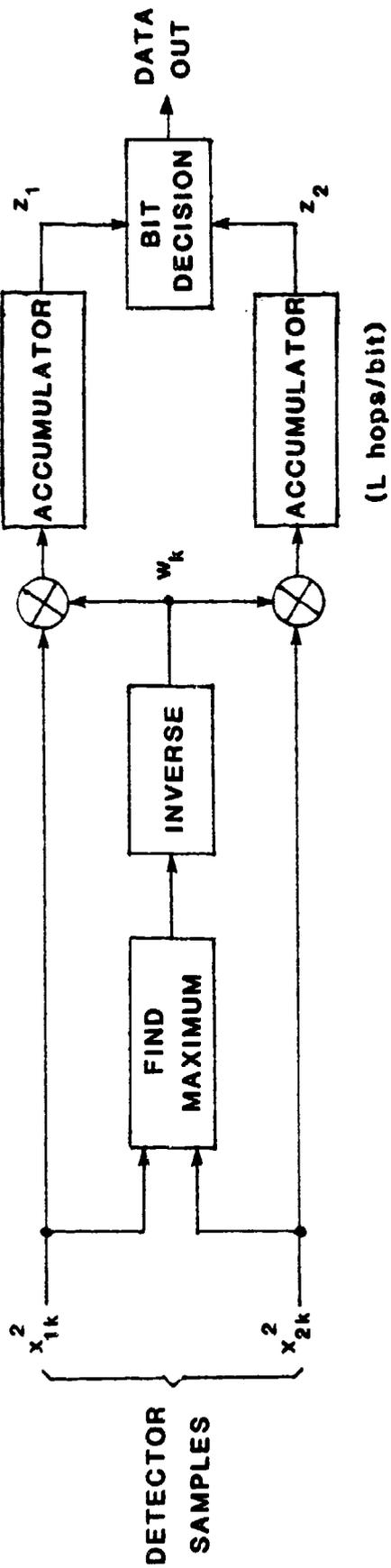


FIGURE 8.2-1 PRACTICAL ACJ-AGC(PACJ) RECEIVER FOR FH/RMFSK WHEN M=2

Numerical results for the L=2 PACJ binary receiver are shown for $E_b/N_0 = 13.35$ dB and 20 dB, respectively, in Figures 8.2-2 and 8.2-3. There is no difference between these results and those of the SNORM receiver that can be discerned from the figures - a close look at the data reveals a slight difference except for $\gamma=1$ (full-band jamming), for which analysis shows that the two receivers yield identical performance for the M=2, L=2 case.

8.2.3.2 Analysis of the PIC receiver.

The PIC receiver for M=2 is diagrammed in Figure 8.2-4. The decision statistics are

$$z_m = \sum_{k=1}^L \frac{u_{mk}}{v_{mk}}, \quad m=1,2; \quad (8.2-15)$$

where the $\{u_{mk}\}$ are the usual receiver samples and the $\{v_{mk}\}$ are look-ahead samples. These look-ahead samples are assumed to have the same noise powers as their corresponding "usual" samples, and not to have a signal present.

In Appendix C it is shown that the L=2 FH/RMFSK or FH/MFSK PIC receiver error probability in partial-band noise jamming is given by

$$\begin{aligned} P_b(e) &= (1-\gamma)^2 P_b(e|\rho_1=\rho_2=E_b/2N_0) \\ &+ 2\gamma(1-\gamma)P_b(e|\rho_1=E_b/2N_0, \rho_2=E_b/2N_T) \\ &+ \gamma^2 P_b(e|\rho_1=\rho_2=E_b/2N_T), \end{aligned} \quad (8.2-16a)$$

where γ is the jamming fraction and

$$\begin{aligned} P_b(e|\rho_1, \rho_2) &= \int_0^1 du \int_0^1 dv e^{-\rho_1 u - \rho_2 v} (1+\rho_1 - \rho_2 u)(1+\rho_2 - \rho_1 v) \\ &\times \frac{uv}{u+v} \left[2 + \frac{uv}{u+v} \ln\left(\frac{u+v}{uv} - 1\right) \right]. \end{aligned} \quad (8.2-16b)$$

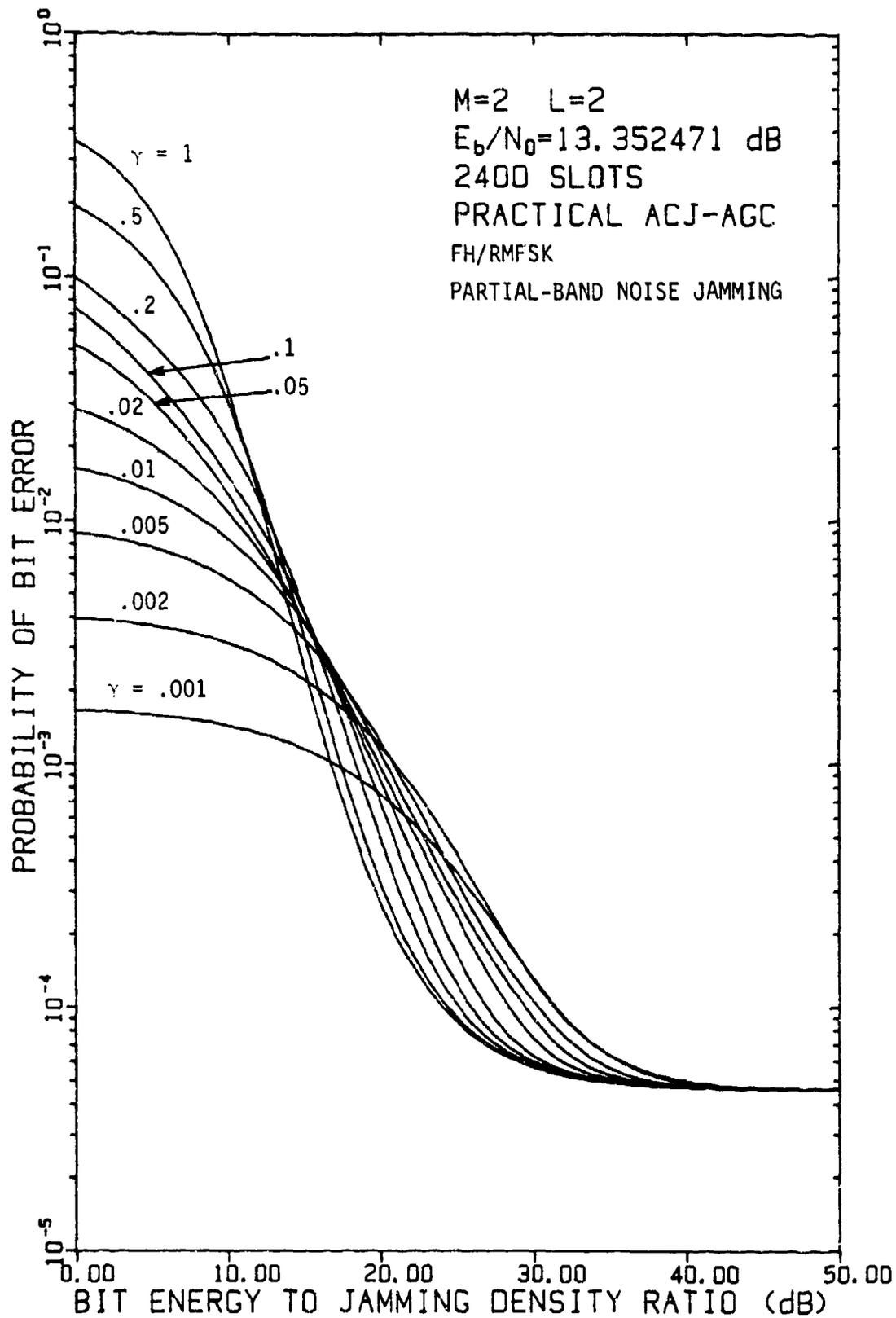


FIGURE 8.2-2 BER PERFORMANCE FOR PACJ FH/RMFSK RECEIVER IN PARTIAL-BAND NOISE JAMMING FOR $M=2$, $L=2$, AND $E_b/N_0 = 13.35 \text{ dB}$

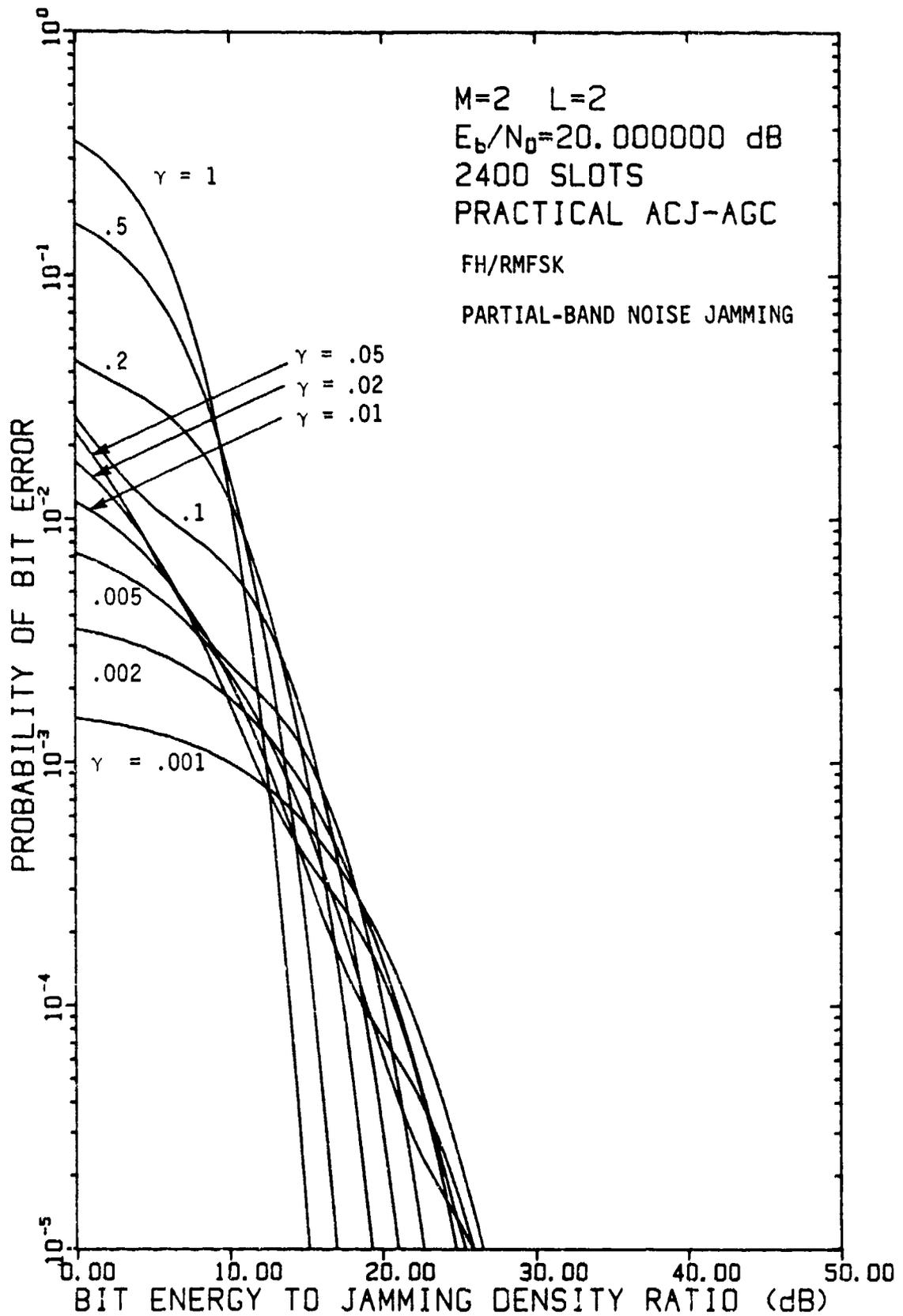


FIGURE 8.2-3 BER PERFORMANCE FOR PACJ FH/RMFSK RECEIVER IN PARTIAL-BAND NOISE JAMMING FOR $M=2$, $L=2$, AND $E_b/N_0 = 20$ dB

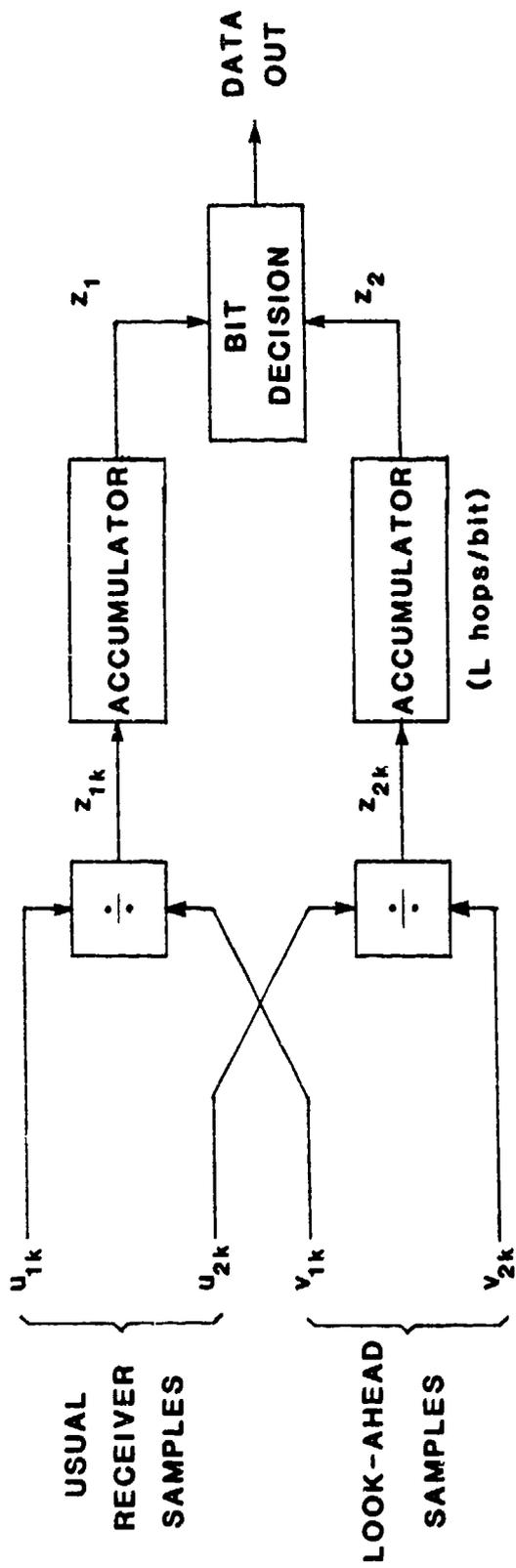


FIGURE 8.2-4 PRACTICAL IC-AGC(PIC) RECEIVER FOR FH/RMFSK WHEN $M=2$

Numerical results for the $L=2$ PIC binary receiver reveal that it performs very poorly. For example, for full-band noise jamming the BER varies with E_b/N_j as shown in Figure 8.2-5. We note that for high E_b/N_j (practically no jamming) the error probability is greater than 10^{-4} even for thermal noise so small that $E_b/N_0 = 40$ dB. Evidently the predictably poor quality of the one-sample estimate of noise variance is especially damaging when using it in the denominator of the ratios taken in (8.2-15).

With slightly more effort, we find that if the PIC receiver uses two look-ahead noise samples, summed to obtain a better variance estimate, the performance improves considerably. The decision statistics for this version of the PIC receiver are

$$z_m = \sum_{k=1}^L \frac{u_{mk}}{v_{mk1} + v_{mk2}}, \quad m=1,2. \quad (8.2-17)$$

Using the same analytical approach as in Appendix C, but with the sum of the look-ahead variables being $\sigma_{mk}^2 \chi^2(4)$ distributed, we find that the conditional $P(e)$ for $M=2$ and $L=2$ is

$$\begin{aligned} P_b(e|\rho_1, \rho_2) &= 4 \int_0^1 du \int_0^1 dv (1-u)(1-v) \exp\{-\rho_1(1-u) - \rho_2(1-v)\} \\ &\quad \times [1 + 2\rho_1^2 u + \rho_1^2 u^2/2] [1 + 2\rho_2^2 v + \rho_2^2 v^2/2] \\ &\quad \times g\left(\frac{u}{1-u} + \frac{v}{1-v}\right), \end{aligned} \quad (8.2-18a)$$

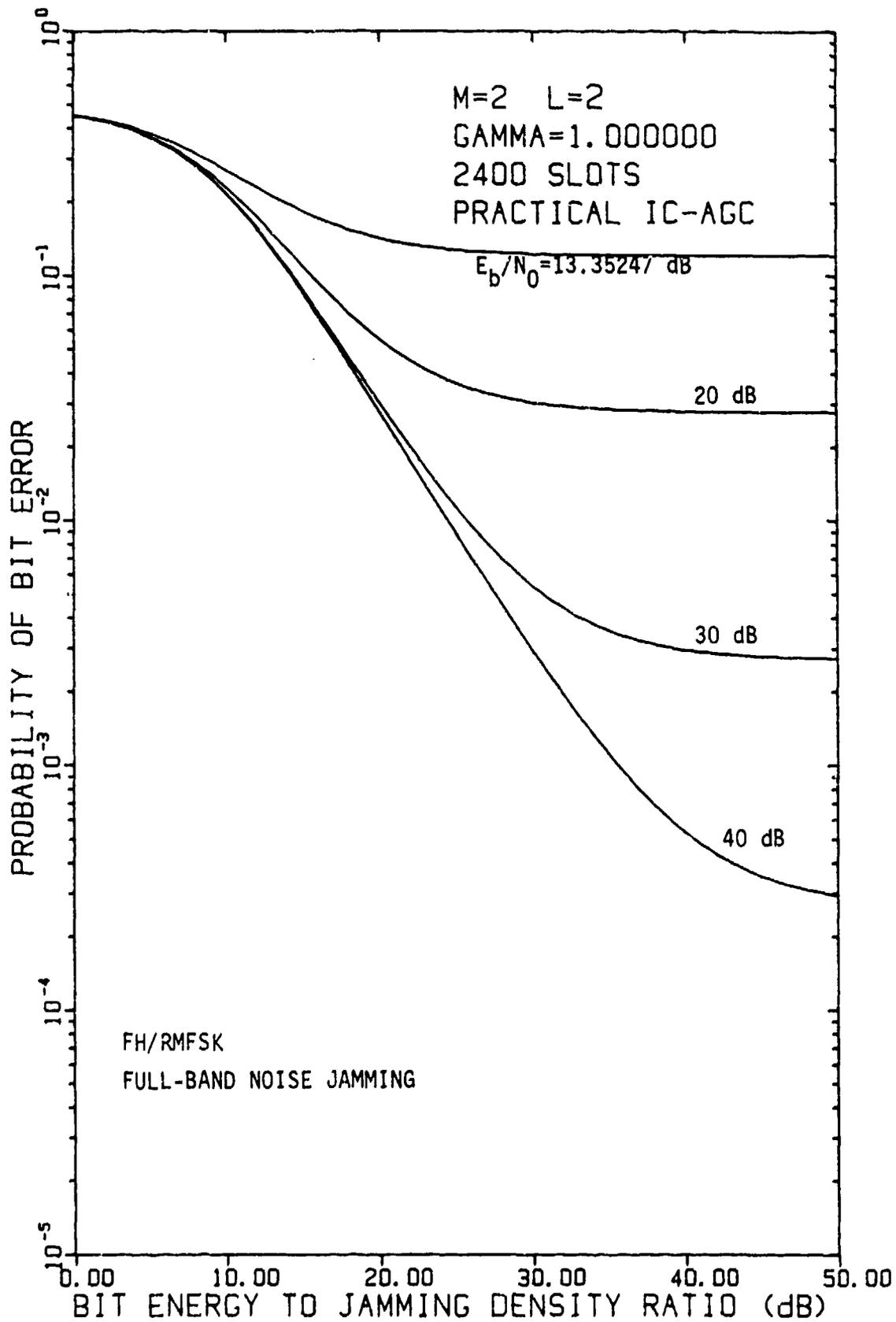


FIGURE 8.2-5 PERFORMANCE OF THE PRACTICAL IC-AGC FH/RMFSK RECEIVER IN FULL-BAND NOISE JAMMING FOR $M=L=2$.

where

$$g(x) = \frac{12}{(x+2)^4} \ln(x+1) + \frac{2(x^2+7x+4)}{(x+1)(x+2)^3} . \quad (8.2-18b)$$

Numerical results for full-band jamming are shown in Figure 8.2-6. For $E_b/N_0 = 13.35$ dB, there is not much of an improvement, but for 20 dB and higher E_b/N_0 , there is about a decade improvement over the asymptotic BER obtained using a one-sample noise estimate. Clearly as more look-ahead samples are used, the PIC receiver will act more like the IC-AGC. However, if more measurements are taken, the likelihood increases that the measurement will suffer from changes in the assumed noise environment over the measurement time. Perhaps with the RMFSK hopping scheme it is possible to gather noise samples from adjacent (unused) hopping slots in addition to (or in place of) using a look-ahead approach.

8.3 CONCLUSIONS AND RECOMMENDATIONS

Having considered practical implementations of ECCM receivers for FH/RMFSK in worst-case partial-band noise jamming (WCPBNJ), we are able to conclude our study with some "lessons learned," from which we also can recommend further studies.

8.3.1 Knowledge Gained from Study.

8.3.1.1 Ideal receiver performances.

The ideal receiver performances obtained in Sections 3-6 and compared in Section 7 show for the first time what the expected performance of random frequency-hopping MFSK is in WCPBNJ when L hops per symbol soft decisions are

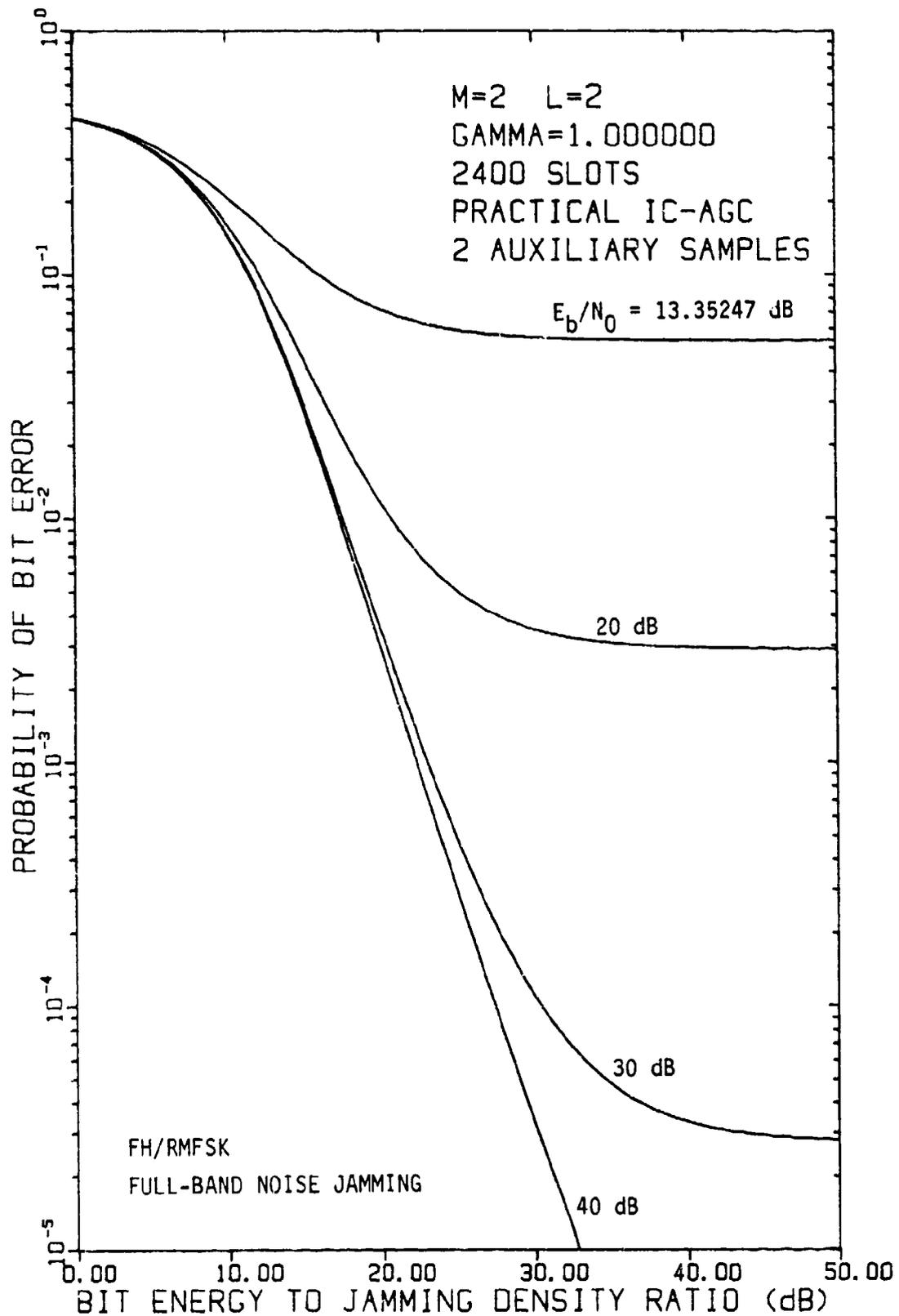


FIGURE 8.2-6 PERFORMANCE OF THE PRACTICAL IC-AGC FH/RMFSK RECEIVER USING A TWO-SAMPLE NOISE ESTIMATE IN FULL-BAND NOISE JAMMING FOR $M=L=2$.

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used and thermal noise is not neglected. We have learned that per-channel adaptive normalization, such as envisioned using the IC-AGC receiver weighting scheme, is effective in countering the jamming effects which are most damaging to the RMFSK performance: jamming power in non-signal symbol frequency channels and not in the signal channel. The IG-AGC scheme controls the channel gains in such a way as to equalize the a priori noise powers in the channels, in effect forcing the non-Gaussian input noise process to be a Gaussian process. The jamming then only affects the error through the reduction of relative signal power when jammed signal hops are normalized, and RMFSK using this scheme performs the same as the conventional MFSK hopping.

We have learned also that per-symbol adaptive normalization, typified by the ACJ-AGC receiver's weighting scheme, is effective in countering WCPBNJ, though not as effective as per-channel normalization. The per-symbol operation equalizes the maximum of the M channels' a priori noise powers to a constant value, but does not affect the relative powers among the M channels, so that to a certain extent the RMFSK system, unlike MFSK, is still subject to noise power imbalances on each hop and therefore is more vulnerable than MFSK. However, the per-symbol normalization does prevent jammed hops from dominating the soft-decision, and therefore achieves an ECCM or anti-jam diversity effect.

The per-channel and per-symbol AGC schemes perform the same for no jamming, and the performance achieved is sensitive to the amount of thermal and/or background noise present at the receiver, expressed relatively in our

study by the ratio E_b/N_0 . For any finite E_b/N_0 value, the receiver performances for weak or no jamming degrade in proportion to L , the number of hops/symbol, due to noncoherent combining losses. It cannot be emphasized too strongly that thermal noise should always be included in any study, because different implementations of the ideal receiver combining schemes in general are subject to different noncoherent combining losses plus any losses due to quantization effects. The ideal AGC performances we have obtained provide a lower bound on achievable performance in the sense that errorless noise power measurements are assumed.

8.3.1.2 Practical receiver performances.

The several "practical" FH/RMFSK receiver combining schemes we have studied may be classified as implementations of either per-channel (IC-AGC) or per-symbol (ACJ-AGC) ideal schemes.

The clipper receiver implements a per-channel ECCM scheme and thus to a certain extent achieves a performance in WCPBNJ whose parametric behavior follows that of the IC-AGC. However, its best clipping threshold value is parametric in received signal and thermal noise powers and in L . In our study we have calculated performances assuming that these powers are known a priori; it is expected that estimation of the correct threshold will degrade its performance. But if we can assume that a reasonably good estimate of unjammed SNR is available, the clipper receiver appears to be a viable candidate for a practical ECCM receiver.

It was shown that direct implementation of the IC-AGC using auxiliary noise power measurements has the potential for approaching the IC-AGC performance, but only if certain assumptions are made: (a) the receiver complexity can be further advanced economically to include the required noise power measurements (as many as possible per channel); (b) the noise processes

being measured are relatively stationary during the time of measurement and not subject to corruption from the signal or other signals. These assumptions are quite restrictive, so that we would choose to implement another, simpler scheme if its performance is satisfactory.

The per-symbol receiver ECCM schemes studied include the self-normalizing receiver (SNORM) and the so-called "practical ACJ" (PACJ). These receiver implementations are very simple, requiring only operations using the usual received envelope samples in the M symbol frequency channels. Somewhat surprisingly, these two schemes perform very well (nearly identically for M=2 and L=2), even better than the supposedly best ideal IC-AGC receiver under certain limited circumstances. Therefore, if the receivers using a priori noise and jamming information tend to represent a lower bound on system performance, and the small-sample size "practical" receivers, an upper bound--then we have observed a situation when lower and upper bounds converge to agree upon a predicted performance result. The implications are that we may regard the easily-calculated IC-AGC performance as representative of achievable system performance, with perhaps a slight implementation loss of a few dB when the simple practical receivers are employed, and when the system noise without jamming is small (high E_b/N_0).

8.3.1.3 RMFSK vs MFSK.

We have found that for smaller alphabet sizes (M=2 or 4), the error performance of FH/RMFSK in worst-case PBNJ is comparable to that of the conventional FH/MFSK, when appropriate receiver processing is employed, to the extent that we state that the price to be paid for the additional RMFSK system

complexity can be assessed against the threat of follow-on noise jamming. That is, if follow-on jamming is not considered a threat, MFSK should be used; but if it is a threat, RMFSK is an effective counter-countermeasure that also works satisfactorily in the worst-case partial-band noise jamming environment.

8.3.2 Recommendations.

With the perspective gained from our study we make the following recommendations for further research.

(a) Derivation of system error performances using the PACJ and similar "nonparametric" ECCM receivers; it is conjectured that analysis would yield BER expressions for $M > 2$ that are more feasible for computation than those for the clipper and self-normalizing receivers we have studied.

(b) Analysis of FH/RMFSK performance under multi-tone jamming; it has been asserted that FH/RMFSK "precludes" systematic tone jamming, but what, in quantitative terms, is its vulnerability to tone jamming, relative to FH/MFSK?

(c) Analysis of mutual interference effects in an FH/RMFSK system; these are considered to be more numerous than for FH/MFSK, and possibly more damaging - a more intricate setup for multiple users may be necessary to avoid mutual interference.

APPENDIX A

PROBABILITY DENSITY FUNCTIONS FOR SOFT DECISION
RECEIVER STATISTICS

For soft decision receivers with multiple hops per symbol, the M decision statistics are of the form

$$z_m = \sigma_{1m}^2 \chi^2(v_{1m}; \lambda_{1m}) + \sigma_{2m}^2 \chi^2(v_{2m}; \lambda_{2m}), \quad (A-1)$$

where $\chi^2(v; \lambda)$ denotes a noncentral chi-squared random variable with v degrees of freedom and noncentrality parameter λ , and σ_{1m}^2 and σ_{2m}^2 are different scalings. For the cases to be studied we can also write

$$u_m = z_m / \sigma_{1m}^2 = \chi^2(2L - 2\ell_m; \lambda_{1m}) + K_m \chi^2(2\ell_m; \lambda_{2m}), \quad (A-2)$$

since $v_{1m} + v_{2m} = 2L$, twice the number of hops per symbol, and $v_{2m} = 2\ell_m$, twice the number of jammed hops for symbol channel m . Also, without loss of generality we designate the first ($m = 1$) channel as the one containing the transmitted signal, and the others ($m \geq 2$) as containing noise only. Thus

$$\lambda_{11} = 2(L - \ell_1)S / \sigma_{11}^2, \quad \lambda_{21} = 2\ell_1 S / \sigma_{21}^2 \quad (A-3a)$$

and

$$\lambda_{1m} = \lambda_{2m} = 0, \quad m \geq 2. \quad (A-3b)$$

Previously it has been shown [1, Appendix 2E] that the probability density function (pdf) for u_m given by (A-2) is $p_u(\alpha; \ell_m, \lambda_{1m}, \lambda_{2m}, K_m)$, where

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$p_u(\alpha; \ell, \lambda_1, \lambda_2, K)$

$$= \frac{K-\ell}{2} \exp\left(-\frac{\lambda_1+\lambda_2+\alpha}{2}\right) \sum_{k=0}^{\infty} \sum_{r=0}^{\infty} \frac{\left(\frac{\lambda_1}{2}\right)^k \left(\frac{\lambda_2}{2K}\right)^r \left(\frac{\alpha}{2}\right)^{k+r+L-1}}{k! r! (k+r+L-1)!} \\ \times {}_1F_1(r+\ell; k+r+L; \frac{K-1}{K} \cdot \frac{\alpha}{2}), \quad (A-4)$$

where ${}_1F_1(a; b; x)$ is the confluent hypergeometric function. The computation of this expression is very time consuming; in this Appendix we consider alternative expressions that can be computed more quickly, or perhaps be amenable to approximations.

By expanding the confluent hypergeometric function in its series form,

$${}_1F_1\left(r+\ell; k+r+L; \frac{K-1}{K} \cdot \frac{\alpha}{2}\right) \quad (A-5)$$

$$= \sum_{n=0}^{\infty} \left(\frac{K-1}{K} \cdot \frac{\alpha}{2}\right)^n \frac{1}{n!} \frac{(r+\ell)_n}{(k+r+L)_n},$$

with

$$(a)_n = \Gamma(a+n)/\Gamma(a), \quad (A-6)$$

and summing over the index k , we obtain the expression

$$p_u(\alpha) = \frac{K-\ell}{2} \exp\left(-\frac{\lambda_1+\lambda_2+\alpha}{2}\right) \sum_{n=0}^{\infty} \left(\frac{K-1}{K}\right)^n \frac{1}{n!} \\ \times \sum_{r=0}^{\infty} \frac{\left(\frac{\lambda_2}{2K}\right)^r}{r!} (r+\ell)_n \left(\frac{\alpha}{\lambda_1}\right)^{(r+n+L-1)/2} I_{r+m+L-1}(\sqrt{\alpha\lambda_1}) \quad (A-7)$$

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This form is based on recognizing the series for $I_{r+n+L-1}(x)$, the modified Bessel function of the first kind of order $r+n+L-1$:

$$I_{r+n+L-1}(x) = \sum_{k=0}^{\infty} \frac{(x/2)^{2k+r+n+L-1}}{k! (k+r+n+L-1)!} \quad (A-8)$$

Further, we recognize that

$$\left(\frac{\alpha}{\lambda_1}\right)^{(q-1)/2} I_{q-1}(\sqrt{\alpha\lambda_1}) = 2e^{(\alpha+\lambda_1)/2} p_{\chi^2}(\alpha; 2q, \lambda_1), \quad (A-9)$$

where $p_{\chi^2}(\alpha; \nu, \lambda)$ is the pdf for a noncentral chi-squared random variable with ν degrees of freedom and noncentrality parameter λ . This allows us to write

$$p_u(\alpha) = K^{-L} e^{-\lambda_2/2} \sum_{n=0}^{\infty} \frac{\left(\frac{K-1}{K}\right)^n}{n!} \sum_{r=0}^{\infty} \frac{\left(\frac{\lambda_2}{2K}\right)^r}{r!} (r+L)_n \times p_{\chi^2}(\alpha; 2r+2n+2L; \lambda_1) \quad (A-10)$$

Now we concentrate on the summation over the indices r and n .

Since

$$\begin{aligned} \sum_{n=0}^{\infty} \sum_{r=0}^{\infty} f(n, r) g(n+r) \\ = \sum_{n=0}^{\infty} g(n) \sum_{r=0}^n f(n-r, r), \end{aligned} \quad (A-11)$$

we can express the pdf in the form

$$p_u(\alpha) = \sum_{n=0}^{\infty} c_n p_{\chi^2}(\alpha; 2L+2n, \lambda_1), \quad (\text{A-12})$$

where

$$c_n = e^{-\lambda_2/2} \left(\frac{K-1}{K}\right)^n \frac{1}{K^\ell} \sum_{r=0}^n \frac{\left(\frac{\lambda_2/2}{K-1}\right)^r}{r!} \frac{(r+\ell)_{n-r}}{(n-r)!}. \quad (\text{A-13})$$

Now,

$$\begin{aligned} \frac{(r+\ell)_{n-r}}{(n-r)!} &= \frac{(r+\ell+n-r-1)!}{(r+\ell-1)!(n-r)!} \\ &= \binom{n+\ell-1}{n-r}; \end{aligned} \quad (\text{A-14})$$

and

$$\sum_{r=0}^n \binom{n+a}{n-r} \frac{(-x)^r}{r!} = \mathcal{L}_n^a(x), \quad (\text{A-15})$$

with $\mathcal{L}_n^a(x)$ the generalized Laguerre polynomial. Thus,

$$c_n = e^{-\lambda_2/2} \left(\frac{K-1}{K}\right)^n \frac{1}{K^\ell} \mathcal{L}_n^{\ell-1} \left[-\frac{\lambda_2/2}{K-1} \right]. \quad (\text{A-16})$$

Since (A-12) is a pdf, it must integrate over α to unity. This requires that

$$\sum_{n=0}^{\infty} c_n = 1. \quad (\text{A-17})$$

In fact, this is so, since [3, eq. 8.975.1]

$$\begin{aligned} \sum_{n=0}^{\infty} b^n \mathcal{L}_n^a(x) &= (1-b)^{-a-1} \exp\left\{\frac{bx}{b-1}\right\} \\ &= K^\ell e^{\lambda_2/2}. \end{aligned} \quad (\text{A-18})$$

Approximation

The expression (A-12) for the pdf is in the form of a series of weighted chi-squared pdf's. This suggests an approximation based on truncating the series:

$$p_u(\alpha) \approx \frac{\sum_{n=0}^N c_n p_{\chi^2}(\alpha; 2L+2n, \lambda_1)}{\sum_{n=0}^N c_n}. \quad (\text{A-19})$$

Since, even for $\lambda_2 = 0$,

$$\frac{c_{n+1}}{c_n} = \frac{1}{n} \frac{\binom{K-1}{K}}{1+1/n} \cdot \frac{1+\ell/n}{1+(\ell-1)/n} \rightarrow \frac{\binom{K-1}{K}}{n} < \frac{1}{n}, \quad (\text{A-20})$$

the truncation is feasible but, depending on the value of K, the convergence of the weights $\{c_n\}$ is slow.

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APPENDIX B

COMBINATORIAL RELATIONS FOR JAMMING EVENT ENUMERATION

B.1 NUMBERS OF ORDERED VECTORS

We define

$$\begin{aligned} (L+1)S_M(L) &\triangleq \#\{\underline{\ell}' : \underline{\ell}' = (\ell_1; \ell_2 \leq \ell_3 \leq \dots \leq \ell_M)\} \\ &= (L-1) \sum_{\ell_M=0}^L \sum_{\ell_{M-1}=0}^{\ell_M} \dots \sum_{\ell_2=0}^{\ell_3} (1) . \end{aligned} \quad (\text{B.1-1})$$

By direct manipulation,

$$S_2(L) = \sum_{\ell_2=0}^L 1 = L+1 = \binom{L+1}{1} \quad (\text{B.1-2})$$

$$\begin{aligned} S_3(L) &= \sum_{\ell_3=0}^L \sum_{\ell_2=0}^{\ell_3} 1 = \sum_{\ell_3=0}^L S_2(\ell_3) \\ &= \sum_{\ell_3=0}^L \binom{\ell_3 + 1}{1} = \binom{L+2}{2} , \end{aligned} \quad (\text{B.1-3})$$

using [3, equation 0.151.1]. Assuming that

$$S_M(L) = \binom{L+M-1}{M-1} , \quad (\text{B.1-4})$$

we find that

$$\begin{aligned} S_{M+1}(L) &= \sum_{\ell_{M+1}=0}^L S_M(\ell_{M+1}) \\ &= \sum_{\ell_{M+1}=0}^L \binom{\ell_{M+1} + M - 1}{M-1} = \binom{L+M}{M} . \end{aligned} \quad (\text{B.1-5})$$

Thus (B.1-4) is proved by induction.

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B.2 NUMBERS OF PARTITIONED ORDERED VECTORS

We define $R_M(L;n)$ as $(L+1)^{-1}$ times the number of ordered vectors $\underline{\ell}$ such that there are exactly n partitions of the components $(\ell_2 \leq \ell_3 \leq \dots \leq \ell_M)$ such that the ℓ_m in the partition are equal. For example,

$$\begin{aligned} R_M(L;1) &= \sum_{\ell_M=0}^L \sum_{\ell_{M-1}=0}^{\ell_M} \dots \sum_{\ell_2=0}^{\ell_3} U(\ell_2 = \ell_3 = \ell_4 = \dots = \ell_M) \\ &= \sum_{\ell_M=0}^L 1 = (L+1), \end{aligned} \tag{B.2-1}$$

where $U(\cdot)$ is 1 if the relation in the argument is true and zero otherwise.

$R_M(L;1)$ is the number of ordered vectors where the components are all equal.

For two partitions, there are many cases, but they all produce the sum

$$R_M(L;2) = \sum_{r_2=1}^L \sum_{r_1=0}^{r_2-1} 1 = \sum_{r_2=0}^L r_2 = \binom{L+1}{2}, \tag{B.2-2}$$

from [3, equation 0.121.1]. We find that

$$R_M(L;n) = \sum_{r_n=0}^L R_{M-1}(r_n-1; n-1). \tag{B.2-3}$$

Asserting that

$$R_M(L;n) = \binom{L+1}{n} \equiv R(L;n) \tag{B.2-4}$$

and substituting (B.2-4) in (B.2-3) establishes this relation by inductive proof.

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B.3 TOTAL NUMBER OF VECTORS REPRESENTED BY ORDERED VECTORS

Since there are

$$\binom{M-1}{n_0, n_1, \dots, n_L} \quad (B.3-1)$$

$\underline{\ell}$ vectors represented by each $\underline{\ell}'$ vector (see Section 2.2), the total number of vectors represented is $(L+1)$ times the summation

$$\sum_{\ell_M=0}^L \sum_{\ell_{M-1}=0}^{\ell_M} \dots \sum_{\ell_2=0}^{\ell_3} \binom{M-1}{n_0, n_1, \dots, n_L} \triangleq T_M(L). \quad (B.3-2)$$

Using the partitioning relations, we can write

$$\begin{aligned} T_M(L) &= \sum \binom{M-1}{n_0, n_1, \dots, n_L} \times \left(\begin{array}{l} \text{number of partitioned } \underline{\ell}' \\ \text{which produce } n_0, n_1, \dots, n_L \end{array} \right) \\ &= \sum_{n=1}^{M-1} R(L;n) \sum_{\text{partitions}} \binom{M-1}{q_1, q_2, \dots, q_n} \cdot \binom{n}{r_1, r_2, \dots, r_{M-1}}, \end{aligned} \quad (B.3-3)$$

where

$$q_k \triangleq \text{number of equal } \ell\text{'s in partition } k \quad (B.3-4a)$$

$$r_s \triangleq \text{number of } q_k \text{ equal to } s. \quad (B.3-4b)$$

Thus

$$T_2(L) = \sum_{n=1}^1 R(L;n) \sum_{\text{part.}} \binom{1}{q_1, q_2, \dots, q_n} \binom{n}{r_1} = L+1, \quad (B.3-5)$$

and

$$\begin{aligned}
 T_3(L) &= \sum_{n=1}^2 R(L;n) \sum_{\text{part.}} \binom{2}{q_1, q_2, \dots, q_n} \binom{n}{r_1, r_2} \\
 &= \binom{L+1}{1} \binom{2}{2} \binom{1}{1} + \binom{L+1}{2} \binom{2}{1, 1} \binom{2}{2, 0} \\
 &= L + 1 + 2 \binom{L+1}{2} = (L+1)^2
 \end{aligned}
 \tag{B.3-6}$$

Calculations show that

$$\begin{aligned}
 T_4(L) &= (L+1)^3 \\
 T_5(L) &= (L+1)^4 \\
 \text{and } T_6(L) &= (L+1)^5.
 \end{aligned}$$

It can be shown [4, p. 106] that

$$T_M(L) = (L+1)^{M-1},
 \tag{B.3-7}$$

giving the total of $(L+1)^M$ \underline{x} vectors.

B.4 NUMBERS OF ACJ-AGC JAMMING EVENTS

From Section 4, the jamming events are described by the vector $\underline{\ell}$ and the number of hops with at least one channel jammed, ℓ_0 .

B.4.1 Number of $\{\ell_0, \underline{\ell}\}$ events

For a given ℓ_0 and $\underline{\ell}$, the number of events may be counted directly using

$$\#(\ell_0, \underline{\ell}) = \binom{L}{\ell_0} \sum_{\underline{v}_1 > \underline{0}} \dots \sum_{\underline{v}_{\ell_0} > \underline{0}} \delta(\sum_{k=1}^{\ell_0} \underline{v}_k, \underline{\ell}). \quad (\text{B.4-1})$$

Previously we have established that the number of $\underline{\ell}$ vectors for n hops is

$$\mathcal{S}(n) \triangleq \sum_{\underline{v}_1} \dots \sum_{\underline{v}_n} \delta(\sum_{k=1}^n \underline{v}_k, \underline{\ell}) \prod_{m=1}^M \binom{n}{\ell_m} \quad (\text{B.4-2})$$

Note that this quantity is zero for $n < \ell_x = \max_m \ell_m$. For notational convenience, let the sum in (B.4-1) be represented as

$$S(\ell_0) = \sum_{[\underline{v}:\ell_0] > \underline{0}} \delta(\cdot, \cdot) \equiv (\text{sum over all } [\underline{v}] \text{ with } \ell_0 \text{ non-zero columns and } M \text{ rows}). \quad (\text{B.4-3})$$

Then we find that

$$\sum_{[\underline{v}:\ell_0] > \underline{0}} = \sum_{[\underline{v}:\ell_0]} - \sum_{[\underline{v}:\ell_0]} \quad (\text{at least one zero column of } [\underline{v}:\ell_0]), \quad (\text{B.4-4a})$$

or

$$S(\ell_0) = \mathcal{S}(\ell_0) - (\text{sums with at least one zero column}). \quad (\text{B.4-4b})$$

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Now, a sum with exactly n non-zero columns of $[v]$ is equivalent to $S(n)$,
so that

$$S(\ell_0) = \mathcal{S}(\ell_0) - \sum_{n=1}^{\ell_0} \binom{\ell_0}{n} S(\ell_0 - n). \quad (\text{B.4-5})$$

For example,

$$S(1) = \mathcal{S}(1) - S(0) = \mathcal{S}(1) - S(0) \quad (\text{B.4-6a})$$

$$\begin{aligned} S(2) &= \mathcal{S}(2) - 2S(1) - \mathcal{S}(0) \\ &= \mathcal{S}(2) - 2\mathcal{S}(1) + \mathcal{S}(0) \end{aligned} \quad (\text{B.4-6b})$$

$$\begin{aligned} S(3) &= \mathcal{S}(3) - 3S(2) - 3S(1) - S(0) \\ &= \mathcal{S}(3) - 3\mathcal{S}(2) - 3\mathcal{S}(1) - \mathcal{S}(0). \end{aligned} \quad (\text{B.4-6c})$$

From these examples we conjecture that

$$S(n) = \sum_{k=0}^n \binom{n}{k} (-1)^k \mathcal{S}(n-k); \quad (\text{B.4-7})$$

substitution of (B.4-7) into (B.4-5) for $\ell_0 = n+1$ leads to an inductive proof.

Therefore, we obtain the result

$$\begin{aligned} \#(\ell_0, \underline{\ell}) &= S(\ell_0) \\ &= \binom{L}{\ell_0} \sum_{r=0}^{\ell_0} \binom{\ell_0}{r} (-1)^r \prod_{m=1}^M \binom{\ell_0 - r}{\ell_m} \end{aligned} \quad (\text{B.4-8a})$$

$$= \binom{L}{\ell_0} \sum_{r=0}^{\ell_0 - \ell_x} \binom{\ell_0}{r} (-1)^r \prod_{m=1}^M \binom{\ell_0 - r}{\ell_m}, \quad (\text{B.4-8b})$$

since terms of the sum are zero for $r > \ell_0 - \ell_x$.

B.4.2 Summation over ℓ_0 events

We now demonstrate that summation of (B.4-8a) over ℓ_0 gives the total number of $\underline{\ell}$ vectors. We use the fact that

$$\binom{\ell_0 - r}{k} = \frac{1}{k!} \frac{\partial^k}{\partial x^k} (1+x)^{\ell_0 - r} \Big|_{x=0} \quad (B.4-9)$$

Substituting this M times (once for each $\ell_m = k$) yields

$$\begin{aligned} \sum_{\ell_0=0}^L S(\ell_0) &= \left\{ \left(\prod_{m=1}^M \frac{1}{\ell_m!} \frac{\partial^{\ell_m}}{\partial x_m^{\ell_m}} \right) \sum_{\ell_0=0}^L \binom{L}{\ell_0} \sum_{r=0}^{\ell_0} \binom{\ell_0}{r} (-1)^r \right. \\ &\quad \left. \cdot \left[\prod (1+x_m) \right]^{\ell_0 - r} \right\} \Big|_{x=0} \\ &= \left\{ \prod_{m=1}^M \frac{1}{\ell_m!} \frac{\partial^{\ell_m}}{\partial x_m^{\ell_m}} (1+x_m)^L \right\} \Big|_{x=0} = \prod_{m=1}^M \binom{L}{\ell_m} \quad (B.4-10) \end{aligned}$$

B.4.3 Summation over $\underline{\ell}$ events

The number of ℓ_0 events can be found by summing (B.4-8a) over all possible $\underline{\ell}$ vectors. This is found to be

$$\begin{aligned} \#(\ell_0) &= \binom{L}{\ell_0} \sum_{r=0}^{\ell_0} \binom{\ell_0}{r} (-1)^r \sum_{\underline{\ell}} \prod_{m=1}^M \binom{\ell_0 - r}{\ell_m} \\ &= \binom{L}{\ell_0} \sum_{r=0}^{\ell_0} \binom{\ell_0}{r} (-1)^r (\ell_0 - r + 1)^M \quad (B.4-11) \end{aligned}$$

For example,

$$\#(\lambda_0=0) = 1 \quad (\text{B.4-12})$$

$$\#(\lambda_0=1) = L (2^M - 1) \quad (\text{B.4-13})$$

$$\#(\lambda_0=2) = \binom{L}{2} (3^M - 2 \cdot 2^M + 1) \quad (\text{B.4-14})$$

$$\#(\lambda_0=L) = (L+1)^M - L \cdot L^M + \binom{L}{2} (L-1)^M + \dots \quad (\text{B.4-15})$$

APPENDIX C

DERIVATION OF ERROR RATE EXPRESSIONS FOR "PRACTICAL
ACJ" AND "PRACTICAL IC" RECEIVERS

C.1 JOINT PDF FOR ONE PAIR OF SAMPLES (PACJ).

The error expression will be obtained for $M=2$ and $L=2$. For a single pair of square-law envelope detector samples the joint pdf is

$$p_0(x_1, x_2) = c_1 c_2 e^{-c_1 x_1 - c_2 x_2 - \rho_1} I_0(2\sqrt{\rho_1 c_1 x_1}), \quad (C.1-1a)$$

where

$$c_i = \begin{cases} 1/2\sigma_N^2, & \text{channel unjammed;} \\ 1/2\sigma_J^2, & \text{channel jammed,} \end{cases} \quad (C.1-1b)$$

and the signal is assumed to be in channel 1. The normalized variables (z_1, z_2) resulting from this pair have the pdf

$$p_1(z_1, z_2) = p_a(z_1, z_2; x_1 > x_2) + p_b(z_1, z_2; x_1 < x_2). \quad (C.1-2)$$

Now, when $x_1 > x_2$, z_1 is made equal to 1 and $z_2 = x_2/x_1$; thus

$$p_a(z_1, z_2; x_1 > x_2) = \delta(z_1 - 1) \int_0^{\infty} d\zeta \zeta p_0(\zeta, \zeta z_2), \quad 0 \leq z_2 \leq 1. \quad (C.1-3)$$

Similarly, when $x_2 < x_1$,

$$p_b(z_1, z_2; x_1 < x_2) = \delta(z_2 - 1) \int_0^{\infty} d\zeta \zeta p_0(\zeta z_1, \zeta), \quad 0 \leq z_1 \leq 1. \quad (C.1-4)$$

The result is that z_1 and z_2 , a single pair of normalized variables, have the joint pdf (conditioned on the possible jamming events) given by

$$p_1(z_1, z_2; c_1, c_2) = \frac{c_1 c_2}{(c_1 z_1 + c_2 z_2)^2} \exp \left\{ - \frac{c_2 z_2^{\rho_1}}{c_1 z_1 + c_2 z_2} \right\} \left[1 + \frac{\rho_1 c_1 z_1}{c_1 z_1 + c_2 z_2} \right] \\ \times [\delta(z_1 - 1) + \delta(z_2 - 1)], \quad 0 \leq z_1, z_2 \leq 1. \quad (C.1-5)$$

By direct integration it may be shown that

$$\Pr\{z_1 < z_2\} = \frac{c_1}{c_1 + c_2} \exp \left\{ - \frac{\rho_1 c_2}{c_1 + c_2} \right\} \quad (C.1-6a)$$

and that

$$\Pr\{z_1 > z_2\} = 1 - \Pr\{z_1 < z_2\}; \quad (C.1-6b)$$

thus the pdf integrates to unity as required.

Taking into account the four possible jamming events, the unconditional pdf may be written using

$$f(z_1, z_2) = \pi_0 \cdot \frac{1}{(z_1 + z_2)^2} \exp \left\{ - \frac{\rho_N z_2}{z_1 + z_2} \right\} \left[1 + \frac{\rho_N z_1}{z_1 + z_2} \right] \\ + \pi_1 \cdot \frac{K}{(z_1 + K z_2)^2} \exp \left\{ - \frac{K \rho_T z_2}{z_1 + K z_2} \right\} \left[1 + \frac{\rho_T z_1}{z_1 + K z_2} \right] \\ + \pi_1 \cdot \frac{K}{(K z_1 + z_2)^2} \exp \left\{ - \frac{\rho_N z_2}{K z_1 + z_2} \right\} \left[1 + \frac{\rho_N K z_1}{K z_1 + z_2} \right] \\ + \pi_2 \cdot \frac{1}{(z_1 + z_2)^2} \exp \left\{ - \frac{\rho_T z_2}{z_1 + z_2} \right\} \left[1 + \frac{\rho_T z_1}{z_1 + z_2} \right]; \\ 0 \leq z_1, z_2 \leq 1. \quad (C.1-7)$$

With this function the unconditioned pdf becomes

$$p_1(z_1, z_2) = f(z_1, z_2) [\delta(z_1-1) + \delta(z_2-1)]. \quad (C.1-8)$$

C.2 JOINT PDF FOR SUMS OF TWO SAMPLES (PACJ).

Using convolution, the joint pdf for L=2 is

$$p_2(z_1, z_2) = \int_{\max(0, z_1-1)}^{\min(1, z_1)} dv_1 \int_{\max(0, z_2-1)}^{\min(1, z_2)} dv_2 p_1(v_1, v_2) p_1(z_1-v_1, z_2-v_2), \quad (C.2-1)$$

which reduces to

$$\begin{aligned} p_2(z_1, z_2) = & \delta(z_1-2) \int_{\max(0, z_2-1)}^{\min(1, z_2)} dv_2 f(1, v_2) f(1, z_2-v_2) \\ & + 2f(1, z_2-1) f(z_1-1, 1) u(z_2-1) u(z_1-1) \\ & + \delta(z_2-2) \int_{\max(0, z_1-1)}^{\min(1, z_1)} dv_1 f(v_1, 1) f(z_1-v_1, 1), \end{aligned} \quad (C.2-2)$$

$0 \leq z_1, z_2 \leq 2.$

C.3 ERROR PROBABILITY FOR L=2 (PACJ).

The error probability is, using (C.2-2),

$$\begin{aligned} P(e) = \Pr\{z_2 > z_1\} = & 2 \int_1^2 dz_2 \int_1^{z_2} dz_1 f(1, z_2-1) f(z_1-1, 1) \\ & + \int_0^1 dz_2 \int_0^{z_2} dv_1 f(v_1, 1) f(z_1-v_1, 1) \\ & + \int_1^2 dz_2 \int_{z_2-1}^1 dv_1 f(v_1, 1) f(z_1-v_1, 1). \end{aligned} \quad (C.3-1)$$

Manipulation of the integrals yields

$$\begin{aligned}
 P(e) &= 2 \int_0^1 dx f(1,x) \int_0^x dy f(y,1) \\
 &+ \int_0^1 dx \int_0^x dy f(y,1) f(x-y,1) \\
 &+ \int_0^1 dx \int_x^1 dy f(y,1) f(x+1-y,1) \tag{C.3-2}
 \end{aligned}$$

$$\begin{aligned}
 &= 2 \int_0^1 dx f(1,x) G(x) \\
 &+ \int_0^1 dy f(y,1) G(1-y) \\
 &+ \int_0^1 dy f(y,1) [G(1) - G(1-y)] \tag{C.3-3}
 \end{aligned}$$

$$= 2 \int_0^1 dx f(1,x) G(x) + G^2(1), \tag{C.3-4}$$

where $G(x)$ is defined as

$$G(x) \triangleq \int_0^x du f(u,1) \tag{C.3-5}$$

$$\begin{aligned}
 &= \pi_0 \cdot \frac{x}{1+x} \exp \left\{ -\frac{\rho_N}{1+x} \right\} \\
 &+ \pi_1 \cdot \frac{x}{K+x} \exp \left\{ -\frac{K\rho_T}{K+x} \right\} \\
 &+ \pi_1 \cdot \frac{Kx}{1+Kx} \exp \left\{ -\frac{\rho_N}{1+Kx} \right\} \\
 &+ \pi_2 \cdot \frac{x}{1+x} \exp \left\{ -\frac{\rho_T}{1+x} \right\}. \tag{C.3-6}
 \end{aligned}$$

Analytically it can be shown that, for $\pi_0 = 1$ (no jamming) or $\pi_2 = 1$ (full-band jamming), the $P(e)$ equals

$$P(e) = \frac{1}{2} e^{-\rho} (1 + \rho/3), \quad (C.3-7)$$

where $\rho = \frac{1}{2} E_b/N_0$ for $\pi_0 = 1$ and $\rho = \frac{1}{2} E_b/N_T$ for $\pi_2 = 1$. This is precisely the same performance obtained by the self-normalizing receiver in Section 5, for the same conditions.

C.4 JOINT PDF FOR ONE PAIR OF SAMPLES (PIC).

The joint pdf of the usual and the look-ahead receiver square-law envelope detector samples is

$$p_0(u_1, u_2, v_1, v_2) = c_1^2 c_2^2 e^{-c_1(u_1+v_1) - c_2(u_2+v_2)} I_0(2\sqrt{\rho_1 c_1 u_1}), \quad (C.4-1)$$

where the constants are defined in (C.1-1b). The normalized variables are $z_{1k} = u_1/v_1$ and $z_{2k} = u_2/v_2$, with the joint pdf

$$\begin{aligned} p_{z_{1k}, z_{2k}}(\alpha, \beta) &= \int_0^\infty dv_1 \int_0^\infty dv_2 v_1 v_2 p_0(v_1^\alpha, v_2^\beta, v_1, v_2) \\ &= \frac{1}{(1+\beta)^2} \cdot \frac{1}{(1+\alpha)^2} \cdot \exp\left\{-\frac{\rho_1}{1+\alpha}\right\} \left[1 + \frac{\rho_1 \alpha}{1+\alpha}\right] \end{aligned} \quad (C.4-2a)$$

$$= p_1(\beta; 0) p_1(\alpha; \rho_1); \quad \alpha, \beta > 0. \quad (C.4-2b)$$

That is, the normalized variables are independent. Note that the jamming conditions are present only in the SNR, ρ_1 .

C.5 CDF FOR SUM OF TWO NON-SIGNAL VARIABLES (PIC).

Since the two channels are independent, we may first derive the probability that the non-signal sum z_2 is greater than the signal channel sum z_1 , given a specific value of z_1 , then later average over z_1 to get the error probability. Formally,

$$\Pr \left\{ z_2 > z_1 \mid z_1 = \alpha \right\} = 1 - F_2(\alpha), \quad (C.5-1)$$

where $F_2(\alpha)$ is the cumulative distribution function for z_2 . This function is found to be

$$\begin{aligned}
 F_2(\alpha) &= \Pr \left\{ z_2 = z_{21} + z_{22} \leq \alpha \right\} \\
 &= \int_0^\alpha \frac{dx}{(1+x)^2} \int_0^{\alpha-x} \frac{dy}{(1+y)^2} \\
 &= \int_0^\alpha \frac{dx}{(1+x)^2} \left[1 - \frac{1}{\alpha-x+1} \right] \\
 &= \frac{\alpha}{1+\alpha} - \int_0^\alpha \frac{dx}{(1+x)^2} \cdot \frac{1}{(\alpha-x+1)} \quad (C.5-2)
 \end{aligned}$$

Using a partial-fraction expansion results in

$$\begin{aligned}
 \Pr \{ z_2 > z_1 | z_1 = \alpha \} &= \frac{1}{(\alpha+2)^2} \left\{ \int_0^\alpha dx \frac{x+3+\alpha}{(1+x)^2} + \int_0^\alpha \frac{dx}{1+\alpha-x} \right\} + \frac{1}{1+\alpha} \\
 &= \frac{2}{(\alpha+2)^2} [\alpha + 2 + \ln(\alpha+1)]. \quad (C.5-3)
 \end{aligned}$$

C.6 ERROR PROBABILITY FOR L=2 (PIC).

The pdf for z_1 is the convolution

$$\begin{aligned}
 p_{z_1}(x) &= p_1(x; \rho_1) * p_1(x; \rho_2) \\
 &= \int_0^\alpha dx p_1(x; \rho_1) p_1(x-x; \rho_2), \quad \alpha > 0. \quad (C.6-1)
 \end{aligned}$$

Thus the error probability is, conditioned on ρ_1 and ρ_2 ,

$$\begin{aligned}
 P_b(e|\rho_1, \rho_2) &= 2 \int_0^\infty d\alpha \frac{\alpha+2+\ln(\alpha+1)}{(\alpha+2)^2} \int_0^\alpha dx p_1(x;\rho_1)p_1(\alpha-x;\rho_2) \\
 &= 2 \int_0^\infty d\alpha \int_0^\infty dx p_1(\alpha;\rho_1)p_1(x;\rho_2) \cdot \frac{\alpha+x+2+\ln(\alpha+x+1)}{(\alpha+x+2)^2} \quad (C.6-2)
 \end{aligned}$$

Transforming the integration variables by

$$u = \frac{1}{1+\alpha}, \quad v = \frac{1}{1+x} \quad (C.6-3)$$

results in the expression

$$\begin{aligned}
 P_b(e|\rho_1, \rho_2) &= 2 \int_0^1 du \int_0^1 dv e^{-\rho_1 u - \rho_2 v} (1+\rho_1-\rho_1 u)(1+\rho_2-\rho_2 v) \\
 &\quad \times \left(\frac{uv}{u+v} \right)^2 \left[\frac{u+v}{uv} + \ln \left(\frac{u+v}{uv} - 1 \right) \right]. \quad (C.6-4)
 \end{aligned}$$

Averaging over the jamming events (the number of hops jammed in the signal channel) yields the total error

$$\begin{aligned}
 P_b(e) &= (1-\gamma)^2 P_b(e|\rho_1=\rho_2=E_b/2N_0) \\
 &\quad + 2\gamma(1-\gamma) P_b(e|\rho_1=E_b/2N_0, \rho_2=E_b/2N_T) \\
 &\quad + \gamma^2 P_b(e|\rho_1=\rho_2=E_b/2N_T). \quad (C.6-5)
 \end{aligned}$$

APPENDIX D
COMPUTER PROGRAM FOR
SQUARE-LAW LINEAR COMBINING RECEIVER

The following pages contain the source code listing for the FORTRAN-77 program used for numerical computations of the performance of the square-law linear combining receiver for FH/RMFSK.


```

0068 1300 READ(4) MM,IN, LL,IN, EBNO,IN, NSL,IN, GAM,IN
0069 IF(MM,IN,NE.MM,OR,LL,IN,NE.LL,OR,EBNO,IN,NE,DEBNOL(10)
$ .OR,GAM,IN,NE,GAMMA,OR,NSL,IN,NE,NSLOTS)
$ STOP 'FILE SYNC ERROR OR CORRUPTED FILE'
JJ=0
0070 OPER(UNIT=6,FILE='FOR006.DAT',STATUS='OLD',FORM='FORMATTED',
0071 ACCESS='APPEND')
0072 JJ=JJ+1
0073 READ(4,END=742) DBSJR(JJ), PRLOG(JJ)
0074 WRITE(6,666) DBSJR(IJ),PE
0075 GOTO 740
0076 CLOSE(UNIT=4)
0077 CLOSE(UNIT=6)
0078 GOTO 755

C NO EXISTING FILE, THIS IS THE FIRST TIME: CREATE FILE HEADER RECORD
C
750 JJ=1
0080 OPEN(UNIT=4,FILE=FNAME,STATUS='NEW',FORM='UNFORMATTED')
0081 WRITE(4) MM,LL,DEBNOL(10),NSLOTS,GAMMA
0082 CLOSE(UNIT=4)
0083 WRITE(GNAME,735) MM,LL,IGOUT
0084 FORMAT('EV',11,11,14.4,'.DAT')
0085 OPEN(UNIT=3,FILE=GNAME,STATUS='OLD',FORM='UNFORMATTED',
$ READONLY,ERR=770)
0086 WRITE(5,3939)
0087 FORMAT(' READING EXISTING EVENT FILE')
0088 READ(3) D,IDSUB,NUSED,GOOD
0089 CLOSE(UNIT=3)
0090 GOTO 777

C IF FILE FOR EVENT PROBABILITIES DOES NOT EXIST, CALCULATE THEM
C AND CREATE A FILE.
770 CONTINUE
0091 WRITE(5,3938)
0092 FORMAT(' CREATING EVENT FILE')
0093 CALL GENPIE(11,MM,NO,NSLOTS,GOOD,MATRIX,MLON,MINC,MUP,PIE,
0094 D,IDSUB,NUSED)
$ OPEN(UNIT=3,FILE=GNAME,STATUS='NEW',FORM='UNFORMATTED')
0095 WRITE(3) D,IDSUB,NUSED,GOOD
0096 CLOSE(UNIT=3)
0097 DO 600 IJ=JJ,NJ
0098 IF(.NOT.GOOD) GOTO 700
0099 WRITE(5,601) IJ
C GIVE PROGRESS MESSAGE TO TI:
601 FORMAT(' IJ=',IJ)
0100 TRASH2=.TRUE.
0101 DEBNJ=START+(IJ-1)*OBINC
0102 IF(MM,LE.4) THEN
0103 HIGH=DEBNJ.GE.15.00
0104 ELSE
0105 HIGH=DEBNJ.GT.30.00
0106 END IF
0107
0108

```

```

0109 DBSJR(IJ)=DEBNJ
0110 R=10.DO**((DEBNJ/10.DO)
0111 RHOTS=GAMMA*R*EBNO/(GAMMA*R+EBNO)
0112 RHOT=K*RHOTS/FLL
C EVALUATE THE PROBABILITY
0113 CALL PSUBE(RHOT,LL,MM,PESYM,D,IDSUB,NUSED,PRERR,IPSUB)
0114 PE=RHOT*PESYM
0115 OPEN(UNIT=6,FILE='FOR006.DAT',STATUS='OLD',FORM='FORMATTED',
$ ACCESS='APPEND')
0116 WRITE(6,666) DBSJR(IJ),PE
0117 CLOSE(UNIT=6)
0118 FORMAT(1X,F7.3,5X,1PD12.5)
0119 PRLOG(IJ)=DLOG10(PE)
0120 OPEN(UNIT=4,FILE=FNAME,STATUS='OLD',ACCESS='APPEND',
$ FORM='UNFORMATTED')
0121 WRITE(4) DBSJR(IJ), PRLOG(IJ)
0122 CLOSE(UNIT=4)
0123 CONTINUE
0124 OPEN(UNIT=4,FILE=FNAME,STATUS='NEW',FORM='UNFORMATTED')
0125 WRITE(4) MM,LL,DEBNOL(10),NSLOTS,GAMMA,DBSJR,PRLOG
0126 CLOSE(UNIT=4)
0127 CONTINUE
0128 CONTINUE
0129 CONTINUE
0130 STOP 'PLEASE PURGE DATA FILES'
0131 END

```

```

0001 SUBROUTINE GET(NJ,START,DBINC)
C
C INTERACTIVE INPUT OF PARAMETERS FOR RUN
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 CHARACTER*9 FIELD,BLANK9
0004 COMMON /INPUTS/ DEBNOL(5),LLIST(4),NSLOTS,GAMLST(31),K,MM
0005 COMMON /SIZE/ NO,NL,NG
C DEFAULT LISTS TEMPORARILY NEEDED ARE IN SHARED STORAGE WITH
C THE LARGE CONVOLUTION WORKING ARRAYS
0006 COMPC'N /SHARE/ DG(31),DSHR(5,4)
0007 DATA BLANK9/'
0008 WRITE(5,33)
0009 FORMAT(' BITS/SYMBOL (K) [2]: ',5)
0010 READ(5,3)K
0011 IF((K.EQ.0)K=2
MM=2**K
0012 WRITE(5,2)
0013 FORMAT(' HOW MANY EB/NO? [1]: ',5)
0014 READ(5,3)NO
0015 IF(NO.EQ.0)NO=1
0016 DO 7 IN=1,NO
0017 IF(K.LE.4) THEN
0018 DO=DSHR(IN,K)
0019 ELSE
0020 DO=0.00
0021 END IF
0022 WRITE(5,5)IN,DO
0023 FORMAT(' EB/NO(' ,I2,' ) [',F9.6,' ]: ',5)
0024 READ(5,6)FIELD
0025 FORMAT(A9)
0026 IF(FIELD.EQ.BLANK9) THEN
0027 DEBNOL(IN)=DO
0028 ELSE
0029 DECODE(9,61,FIELD)DEBNOL(IN)
0030 FORMAT(F9.6)
0031 END IF
0032 CONTINUE
0033 WRITE(5,16)
0034 FORMAT(' HOW MANY L? [4]: ',5)
0035 READ(5,3)NL
0036 IF(NL.EQ.0)NL=4
0037 DO 21 IN=1,NL
0038 WRITE(5,19)IN,IN
0039 FORMAT(' L(' ,I1,' ) [',I1,' ]: ',5)
0040 READ(5,3)LLIST(IN)
0041 IF(LLIST(IN).EQ.0)LLIST(IN)=IN
0042 CONTINUE
0043 WRITE(5,23)
0044 FORMAT(' HOPPING SLOTS? [2400]: ',5)
0045 READ(5,24)NSLOTS
0046
0047

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0033 IF(NONE) THEN
0034 IF(HIGH) THEN
0035 CALL PSEL1(JAM,LL,M,RHOM,RHOT,PROB)
0036 ELSE
0037 CALL PSEL2(JAM,LL,M,RHOM,RHOT,PROB)
0038 END IF
C ... AND SAVE IT FOR POSSIBLE FUTURE RE-USE
0039 CALL PUTIN(PROB,PRERR,IPSUB,MPS,625,ISUB,KODE,STORE)
0040 IF(KODE.NE.0) STOP 2
0041 END IF
C SUM UP UNCONDITIONAL ERROR PROBABILITY
0042 PE=PE+PIE*PROB
0043 DO 197 I=2,M
0044 ITER=M+2-I
0045 JAM(ITER)=JAM(ITER)+1
0046 IF(JAM(ITER).LE.LL) GOTO 100
0047 CONTINUE
0048 CONTINUE
0049 RETURN
0050 END
  
```

```

0001 SUBROUTINE PSUBE(RHOM,RHOT,LL,M,PE,D,IDSUB,MUSED,PRERR,IPSUB)
C COMPUTE UNCONDITIONAL ERROR PROBABILITY
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 INTEGER JAM(8),LUP(8),JSUB(8)
0004 LOGICAL*1 GO,NONE,STORE
0005 LOGICAL TRASH1,TRASHZ,HIGH
C WE DO WANT TO STORE ZERO ELEMENTS OF THE DENSITY FUNCTION,
C SINCE IT SAVES TIME TO AVOID REPEATING THE UNDERFLWS
0006 VIRTUAL PRERR(625),IPSUB(625)
0007 VIRTUAL D(625),IDSUB(625)
0008 COMMON /REGION/ HIGH
0009 COMMON /RESET/ TRASH1,TRASHZ
0010 COMMON /SHAREZ/ LOW(8),LINC(8)
0011 DATA STORE/.TRUE./
0012 PE=0.00
0013 MPS=0
0014 JAM1=-1
0015 DO 199 I1=0,LL
0016 JAM(I1)=11
0017 ITER=1
0018 DO 100 I=ITER+1,M
C
C OUTERMOST NONSIGMAL LOOP ALWAYS STARTS FROM 0, BUT
C THE OTHERS START FROM THE CURRENT VALUE OF THE
C NEXT OUTER MORE LOOP TO PRODUCE THE SORTED EVENTS
C
0019 IF(I.GT.2) THEN
0020 JAM(I)=JAM(I-1)
0021 ELSE
0022 JAM(I)=0
0023 END IF
0024 CONTINUE
0025 IF(JAM1.NE.JAM(1)) THEN
0026 JAM1=JAM(1)
0027 TRASH1=.TRUE.
0028 END IF
0029 CALL EVENT(LL,M,JAM,PIE,D,IDSUB,MUSED)
C BYPASS CONDITIONAL ERROR PROBABILITY CALCULATION IF EVENT
C PROBABILITY IS ZERO. THIS SAVES MUCH TIME
0030 IF(PIE.EQ.0.00)GOTO 198
C STORE AS ELEMENT OF A SPARSE ARRAY TO SAVE MEMORY
C EVEN THROUGH WE STORE ZEROS, THE SORTING OF SUBSCRIPTS
C CUTS OUT MANY ELEMENTS.
0031 CALL LOGN(M,LOW,LUP,JAM,ISUB)
C TRY TO FIND CONDITIONAL ERROR PROBABILITY IN STORED ARRAY
0032 CALL LOOKUP(PROB,PRERR,IPSUB,MPS,625,ISUB,STORE,NONE)
C IF IT IS NOT THERE, WE MUST COMPUTE IT
  
```

```

0001 $ SUBROUTINE GENPJE(LL,MM,NQ,NSLOTS,GOOD,MATRIX,MLOW,MINC,
      MUP,PIE,O,IDSUB,MUSED)
C SUBROUTINE TO GENERATE EVENT PROBABILITIES
C
C IMPLICIT DOUBLE PRECISION (A-H,O-Z)
C LOGICAL*1 GO,STORE,NOME,GOOD
C DIMENSION MATRIX(LL,MM),MLOW(LL,MM),MINC(LL,MM),MUP(LL,MM),
      PIE(O:MM),IMORK(8),LOWRK(8),LUPRK(8)
C STORE=.FALSE. => DON'T STORE ZERO ELEMENTS OF SPARSE ARRAY.
      STORE=.FALSE.
      GOOD=.TRUE.
0002 IF(NQ.LE.O) THEN
0003   GOOD=.FALSE.
0004   RETURN
0005 END IF
0006 DO 90 I=0,MM
0007   CALL PRIHOP(I,MM,NQ,NSLOTS,A)
0008   PIE(I)=A
0009   DO 95 J=1,MM
0010     DO 95 J=1,LL
0011       MLOW(I,J)=0
0012       MUP(I,J)=1
0013       MINC(I,J)=1
0014     CONTINUE
0015   NUSED=N+1
0016   DO 98 J=1,MM
0017     LOWRK(J)=0
0018     LUPRK(J)=LL
0019   CONTINUE
0020 CALL MLINIT(MATRIX,MLOW,LL,MM)
0021 CONTINUE
0022 C FORM COLUMN SUMS AND COMPUTE P(EVENT)
0023 P=1.DO
0024 DO 101 I=1,LL
0025   K=0
0026   DO 100 J=1,MM
0027     K=K+MATRIX(I,J)
0028   CONTINUE
0029   P=P*PIE(K)
0030 CONTINUE
0031 C FORM JAMMING EVENT VECTOR
0032 DO 102 J=1,MM
0033   IMORK(J)=0
0034   DO 102 I=1,LL
0035     IMORK(J)=IMORK(J)+MATRIX(I,J)
0036   CONTINUE
0037 CONTINUE
0038 IMORK(J)=IMORK(J)+MATRIX(I,J)
0039 CONTINUE
0040 CONTINUE
0041

```

```

0001 SUBROUTINE EVENT(LL,M,JAM,PIE,O,IDSUB,MUSED)
C SUBROUTINE TO LOOK UP EVENT PROBABILITY FROM STORED ARRAY
C
C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
C LOGICAL*1 STORE,NOME
C DIMENSION JAM(8),LUP(8)
C VIRTUAL D(625),IDSUB(625)
C COMMON /SHAREZ/ LOW(8),LINC(8)
C DATA STORE/.FALSE./
C SET UP ARRAY DESCRIPTION D(O:LL,...,O:LL) WITH M DIMENSIONS
      DO 1 I=1,M
0002   LUP(I)=LL
0003   CONTINUE
0004 C COMPUTE LINEAR EQUIVALENT SUBSCRIPT FOR JAMMING EVENT
0005 CALL LOCN(M,LOW,LUP,JAM,ISUB)
0006 C LOOK UP THE VALUE, GET 0.DO IF NOT THERE
0007 CALL LOOKUP(PIE,O,IDSUB,MUSED,625,ISUB,STORE,NOME)
0008 RETURN
0009 END
0010
0011
0012
0013
0014

```

```

0042 C SORT NONSIGNAL CHANNELS
0043 DO 103 I=2,MM-1
0044 DO 103 J=I+1,MM
0045 IF(IWORK(J).LT.IWORK(I)) THEN
0046 ITEMP=IWORK(I)
0047 IWORK(I)=IWORK(J)
0048 IWORK(J)=ITEMP
0049 END IF
0050 CONTINUE
0051 CALL LOGN(MM,LMMNRK,LUPNRK,IWORK,ISUB)
0052 CALL LOOKUP(DOUT,D,IDSUB,NUSED,625,ISUB,STORE,NONE)
0053 DOUT=DOUT+P
0054 CALL PUTIN(DOUT,D,IDSUB,NUSED,625,ISUB,IERR,STORE)
0055 IF(IERR.NE.0) STOP 'TOO MANY EVENTS'
0056 CALL MLITER(MATRIX,MLON,MUP,MINC,LL,MM,GO)
0057 RETURN
0058 END
  
```

```

0001 SUBROUTINE PUTIN(CIN,C,ICSUB,MUSE,NMAX,K,IERR,STORE)
0002 C THIS SUBROUTINE INSERTS AN ELEMENT INTO A SPARSE ARRAY FOR
0003 C WHICH ONLY THE NON-ZERO ELEMENTS ARE KEPT IN STORAGE IF
0004 C THE SWITCH STORE IS .TRUE.
0005 C
0006 C THE DOUBLE PRECISION VALUE CIN IS STORED AS C(K) WHERE
0007 C THE AUXILIARY ARRAY ICSUB IS USED TO KEEP TRACK OF THE
0008 C SUBSCRIPTS OF THE CORRESPONDING ENTRIES IN C.
0009 C LONG (4-BYTE) INTEGERS ARE USED TO ACCOMMODATE LARGE
0010 C SUBSCRIPT VALUES FOR THE SPARSE ARRAY C.
0011 C USAGE:
0012 C LOGICAL*I STORE
0013 C DOUBLE PRECISION C,CIN
0014 C VIRTUAL ICSUB(NMAX),C(NMAX)
0015 C CALL PUTIN(CIN,C,ICSUB,MUSE,NMAX,K,IERR,STORE)
0016 C WHERE
0017 C CIN = VALUE OF ELEMENT TO STORE
0018 C C = ARRAY IN WHICH NON-ZERO VALUES ARE ACTUALLY STORED
0019 C ICSUB = AUXILIARY ARRAY FOR ACTUAL SUBSCRIPT VALUES
0020 C MUSE = NUMBER OF ELEMENTS OF C CURRENTLY OCCUPIED
0021 C NMAX = SIZE OF ARRAY C
0022 C IERR = ERROR RETURN CODE, 0 IF NO ERROR OR 1 IF THERE IS
0023 C NO ROOM AVAILABLE IN C
0024 C STORE = .TRUE. TO STORE ZEROS EXPLICITLY, ELSE .FALSE.
0025 C NOTE: IF CIN=0 AND THE SUBSCRIPT K IS FOUND IN ICSUB, THEN
0026 C THE ELEMENT IS DELETED BY SHIFTING DOWNWARD ALL
0027 C FOLLOWING ELEMENTS OF THE ARRAY
  
```

C PROGRAMMER: ROBERT H. FRENCH DATE: 11 JANUARY 1984

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 VIRTUAL ICSUB(NMAX),C(NMAX)
0004 LOGICAL*I STORE
0005 IERR=0
0006 IF(STORE) GOTO 5
0007 IF(CIN.EQ.0.DO) GOTO 30
0008 IF(MUSE.EQ.0)GOTO 20
0009 DO 10 I=1,MUSE
0010 IF(ICSUB(I).NE.K) GOTO 10
0011 C(I)=CIN
0012 RETURN
0013 CONTINUE
0014 IF(MUSE.LT.NMAX) GOTO 20
0015 IERR=1
0016 RETURN
0017 MUSE=MUSE+1
0018 ICSUB(MUSE)=K
0019 C(MUSE)=CIN
0020 RETURN
  
```

```

0021 30 DO 40 I=1,MUSE
0022 J=1
0023 IF(ICSUB(I).EQ.K) GOTO 50
0024 CONTINUE
0025 RETURN
  
```

```

C REMOVE THE ZEROED ELEMENT AND BUMP COUNT OF ENTRIES USED
  
```

```

0026 DO 60 I=J,MUSE-1
0027 ICSUB(I)=ICSUB(I+1)
0028 C(I)=C(I+1)
0029 CONTINUE
0030 MUSE=MUSE-1
0031 RETURN
0032 END
  
```

```

0001 SUBROUTINE LOOKUP(COUT,C,ICSUB,M,MMAX,K,STORE,NOME)
C THIS SUBROUTINE RETRIEVES AN ELEMENT OF A SPARSE ARRAY WHICH
C HAS BEEN STORED COMPACTLY BY STORING ONLY NON-ZERO ELEMENTS.
C THE ARRAY IS DOUBLE PRECISION.
  
```

```

C USAGE:
C VIRTUAL ICSUB(MMAX), C(MMAX)
C LOGICAL*1 STORE, NOME
C DOUBLE PRECISION COUT
C CALL LOOKUP(COUT,C,ICSUB,M,MMAX,K,STORE,NOME)
C WHERE
C COUT = VALUE OF C(K) (OUTPUT FROM SUBROUTINE)
C C = ARRAY USED TO STORE NON-ZERO ELEMENTS
C ICSUB = AUXILIARY ARRAY TO STORE ACTUAL SUBSCRIPTS
C N = NUMBER OF ELEMENTS OF C CURRENTLY IN USE
C MMAX = SIZE OF C
C K = SUBSCRIPT OF SPARSE ARRAY TO LOOK UP
C STORE = .TRUE. IF ZEROS STORED EXPLICITLY, ELSE .FALSE.
C NOME = .FALSE. IF ZEROS NOT STORED OR ZEROS STORED AND
C ELEMENT IS FOUND IN THE STORED ARRAY
C .TRUE. IF ZEROS ARE STORED AND THE ELEMENT IS
C NOT FOUND (OUTPUT QUANTITY)
  
```

```

C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
  
```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 VIRTUAL ICSUB(MMAX),C(MMAX)
0004 LOGICAL*1 STORE, NOME
0005 NOME=.FALSE.
0006 DO 10 I=1,M
0007 IF(ICSUB(I).NE.K)GOTO 10
0008 COUT=C(I)
0009 RETURN
0010 CONTINUE
0011 IF(STORE) THEN
0012 NOME=.TRUE.
0013 ELSE
0014 COUT=0.
0015 END IF
0016 RETURN
0017 END
  
```

```
0001 SUBROUTINE LOCN(NDIM,ILOW,IUP,ISUB,LINEAR)
C THIS SUBROUTINE COMPUTES THE EQUIVALENT LINEAR SUBSCRIPT FOR
C A MULTIDIMENSIONAL ARRAY OF NDIM DIMENSIONS
C IF THE ARRAY A IS DEFINED AS
C DIMENSION A(ILOW(1):IUP(1),.....,ILOW(NDIM):IUP(NDIM))
C AND ISUB(1),.....,ISUB(NDIM) IS A SET OF SUBSCRIPTS FOR A,
C THEN THIS SUBROUTINE RETURNS IN LINEAR THE OFFSET FROM THE
C ORIGIN OF A TO THE ELEMENT A(ISUB(1),.....,ISUB(NDIM)), ASSUMING
C THE FIRST SUBSCRIPT VARIES MOST RAPIDLY.
C USAGE:
C DIMENSION ILOW(NDIM),IUP(NDIM),ISUB(NDIM)
C DATA ILOW/lower limits of defined subscripts of array/
C DATA IUP/upper limits of defined subscripts of array/
C ...SET ISUB TO DESIRED SEQUENCE OF SUBSCRIPTS...
C CALL LOCN(NDIM,ILOW,IUP,ISUB,LINEAR)
C WHERE
C NDIM = NUMBER OF DIMENSIONS THE ARRAY HAS
C ILOW = ARRAY OF LOWER SUBSCRIPT BOUNDS
C IUP = ARRAY OF UPPER SUBSCRIPT BOUNDS
C ISUB = ARRAY CONTAINING SUBSCRIPT FOR WHICH LOCATION IS
C TO BE COMPUTED
C LINEAR = RETURNED VALUE OF OFFSET INTO ARRAY IN MEMORY
```

```
0002 DIMENSION ILOW(NDIM),IUP(NDIM),ISUB(NDIM)
0003 LINEAR=0
0004 DO 10 I=1,NDIM-1
0005 J=NDIM-I+1
0006 LINEAR=(LINEAR+(ISUB(J)-ILOW(J)))*(IUP(J-1)-ILOW(J-1)+1)
0007 CONTINUE
0008 LINEAR=LINEAR+ISUB(1)-ILOW(1)
0009 RETURN
0010 END
```

```
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
```

```
0001 SUBROUTINE MLINIT(LMAT,LLOW,LMAXC,LMAXR)
C THIS SUBROUTINE INITIALIZES A "MATRIX DO-LOOP" STRUCTURE
C DEFINED BY THE FOLLOWING PSEUDO-FORTRAN CODE:
C DO 100 LMAT(1,1)=LLOW(1,1),LUP(1,1),LINC(1,1)
C
C
C DO 100 LMAT(LMAXC,1)=LLOW(LMAXC,1),LUP(LMAXC,1),LINC(LMAXC,1)
C DO 100 LMAT(1,2)=LLOW(1,2),LUP(1,2),LINC(1,2)
C
C
C DO 100 LMAT(LMAXC,2)=LLOW(LMAXC,2),LUP(LMAXC,2),LINC(LMAXC,2)
C DO 100 LMAT(1,LMAXR)=LLOW(1,LMAXR),LUP(1,LMAXR),LINC(1,LMAXR)
C
C
C DO 100 LMAT(LMAXC,LMAXR)=LLOW(LMAXC,LMAXR),LUP(LMAXC,LMAXR),
C LINC(LMAXC,LMAXR)
C $
C (STATEMENTS IN RANGE OF LOOP)
C
C 100 CONTINUE
C
C THE COMPANION ROUTINE MLITER HANDLES THE LOOP CONTROL AT THE
C CONTINUE STATEMENT IN THE ABOVE STRUCTURE
C
C USAGE:
C LOGICAL*1 GO
C DIMENSION LMAT(LMAXC,LMAXR),LLOW(LMAXC,LMAXR),LUP(LMAXC,LMAXR)
C DIMENSION LINC(LMAXC,LMAXR)
C (INITIALIZE MATRIX LLOW TO STARTING VALUES OF THE NESTED LOOPS)
C (INITIALIZE MATRIX LUP TO STOPPING VALUES OF THE NESTED LOOPS)
C (INITIALIZE MATRIX LINC TO INCREMENTS OF THE LOOPS)
C CALL MLINIT(LMAT,LLOW,LMAXC,LMAXR)
C 100 CONTINUE
C
C : (STATEMENTS IN RANGE OF LOOPS)
C
C CALL MLITER(LMAT,LLOW,LUP,LINC,LMAXC,LMAXR,GO)
C IF(GO)GOTO 100
C
C WHERE
C LMAT = ARRAY FOR STORAGE OF LOOP INDICES. LMAT(1,1) IS THE
C OUTER-MOST LOOP; LMAT(LMAXC,LMAXR), THE INNER-MOST LOOP.
C LLOW = ARRAY FOR STORAGE OF LOOP STARTING VALUES, IN SAME
C SEQUENCE AS LMAT
C LUP = ARRAY FOR STORAGE OF LOOP FINAL VALUES, IN SAME
C SEQUENCE AS LMAT
C LINC = ARRAY FOR STORAGE OF LOOP INCREMENTS, IN SAME
C SEQUENCE AS LMAT
C LMAX = NUMBER OF LOOPS NESTED
C GO = LOGICAL VARIABLE, .TRUE. IF JUMP BACK TO BEGINNING OF
C STATEMENTS IN THE RANGE OF THE LOOP SHOULD OCCUR,
C .FALSE. OTHERWISE (I.E. OUTER-MOST LOOP TERMINATED)
```

```

C PROGRAMMER: ROBERT H. FRENCH DATE: 10 MARCH 1986
C
0002 DIMENSION LMAT(LMAXC,LMAXR),LLOW(LMAXC,LMAXR)
0003 DO 1 N=1,LMAXR
0004 DO 1 M=1,LMAXC
0005 LMAT(M,N)=LLOW(M,N)
0006 CONTINUE
0007 RETURN
0008 END
  
```

```

0001 SUBROUTINE MLITER(LMAT,LLOW,LUP,LINC,LMAXC,LMAXR,GO)
C LOOP ITERATION LOGIC FOR A "MATRIX DO-LOOP"
C SEE DETAILED COMMENTS IN SUBROUTINE MLINIT FOR USAGE AND
C PARAMETER DEFINITIONS
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 10 MARCH 1986
C
0002 LOGICAL *1 GO
0003 DIMENSION LMAT(LMAXC,LMAXR),LLOW(LMAXC,LMAXR),LUP(LMAXC,LMAXR)
0004 DIMENSION LINC(LMAXC,LMAXR)
0005 GO=.TRUE.
0006 DO 100 MDX=1,LMAXR
0007 NSUB=LMAXR+1-MDX
0008 DO 100 MDY=1,LMAXC
0009 NSUB=LMAXC+1-MDY
0010 LMAT(MSUB,NSUB)=LMAT(MSUB,NSUB)+LINC(MSUB,NSUB)
0011 IF((LINC(MSUB,NSUB).GE.O.AND.LMAT(MSUB,NSUB).LE.LUP(MSUB,NSUB))
    .OR.
    $ (LINC(MSUB,NSUB).LT.O.AND.LMAT(MSUB,NSUB).GE.LUP(MSUB,NSUB)))
    $ RETURN
    $ LMAT(MSUB,NSUB)=LLOW(MSUB,NSUB)
    $ CONTINUE
    $ GO=.FALSE.
    $ RETURN
    $ END
  
```

```

0001 SUBROUTINE PRIHOP(KJAM,KM,KQ,KN,AIN)
C THIS SUBROUTINE COMPUTES THE EVENT PROBABILITY FOR ALL
C POSSIBLE JAMMING PATTERNS WITH NON-ZERO PROBABILITY FOR
C L=1 HOP/SYMBOL FOR RMFSK/FH IN PBNQ
C
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 AIN=0.00
0004 IF(KJAM.GT.MIND(KQ,KN)) RETURN
0005 KPMAX=KJAM-1
0006 LPMAX=KM-KJAM-1
0007 JPMAX=KN-1
0008 IMAX=MAX0(KPMAX,LPMAX,JPMAX)
0009 PROB=1.00
0010 Q=KQ
0011 DIFFNQ=KN-KQ
0012 EN=KN
0013 DO 100 LOOP=0,IMAX
0014 F=LOOP
0015 IF(LOOP.LE.KPMAX) PROD=PROD*(Q-F)
0016 IF(LOOP.LE.JPMAX) PROD=PROD/(EN-F)
0017 IF(LOOP.LE.LPMAX) PROD=PROD*(DIFFNQ-F)
0018 CONTINUE
0019 AIN=PROD
0020 RETURN
0021 END
100

```

```

0001 BLOCK DATA
C INITIALIZE SHARED CONSTANTS
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 COMMON /SHARE/ DG(31),DSMR(5,4)
0004 COMMON /SHARE2/ LHM(8),LINC(8)
C DEFAULT LISTS FOR INTERACTIVE PARAMETER INPUTS
C ARE SHARED WITH LARGE WORKING STORAGE ARRAYS SINCE THEY
C MAY BE DESTROYED ONCE THE INPUT PARAMETERS ARE SET UP
0005 DATA DG / .00100, .00200, .00500,
$ .100, .200, .500, 1.00, 21*0.00/
$ DATA DSMR /13.3524700, 12.313300, 10.9444300, 0.00, 0.00,
$ 10.60657200, 9.628400, 8.3524800, 0.00, 0.00,
$ 9.0940100, 8.169000, 6.97199500, 0.00, 0.00,
$ 8.0783500, 7.199600, 6.06964600, 0.00, 0.00/
C FREQUENTLY NEEDED CONSTANT ARRAYS AND SCALARS
0007 DATA LHM/8*0/,LINC/8*1/
0008 END

```

```

0001 SUBROUTINE PSEL1(JSUB,LLL,MM,RHOM,RHOT,PROB)
C
C RANDOM MFSK/FP IN PARTIAL BAND TONE JAMMING,
C GIVEN A JAMMING EVENT
C
C JSUB - JAMMING EVENT VECTOR
C LL - NUMBER OF HOPS/SYMBOL
C MM - ALPHABET SIZE
C PROB - RESULTING CONDITIONAL ERROR PROBABILITY
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 EXTERNAL PGRAND,DGAU20
0004 DIMENSION WORK(75), STACK(75), SAVE(75)
0005 INTEGER JSUB(8)
0006 INTEGER NCHAN(O:6)
0007 COMMON /PARMS/ BIGK, LL
C AJ = NONCENTRAL PARAMETER FOR JAMMED HOPS
C ANJ = NONCENTRAL PARAMETER FOR NON-JAMMED HOPS
C NNJ = NUMBER OF NON-JAMMED HOPS
C COMMON /PUIPAR/ AJ, ANJ, NJ, NNJ
C COMMON /JAMCNT/ NCHAN
C
0008 BICK-RHOM/RHOT
0009 NNJ=LL-NNJ
0010 AJ=2.DO*NNJ*RHOT
0011 ANJ=2.DO*NNJ*RHOM
0012 ASHIFT=DMAX1(AJ,ANJ)
0013 SUBTRACT 1 FOR BENEFIT OF RAPID CHI-SQUARE DENSITY CALCULATION
0014 NJ=NNJ-1
0015 NNJ=NNJ-1
C
C COUNT NUMBER OF NON-SIGNAL CHANNELS WITH L HOPS JAMMED
C
0019 DO 1 I=0,LL
0020 NCHAN(I)=0
0021 CONTINUE
0022 DO 2 I=2,MM
0023 JSUB=JSUB(I)
0024 NCHAN(I SUB)=NCHAN(I SUB)+1
0025 CONTINUE
0026 CALL DLG515(PGRAND,ASHIFT,TAIL)
0027 CALL ADQUAD(O,ASHIFT,BODY,DGAU20,PGRAND,1,D-10,
0028 WORK,STACK,75,KODE)
0029 IF(KODE.NE.0) THEN
0030 WRITE(5,3) KODE
0031 FORMAT(' ADQUAD ERROR CODE = ',I1)
0032 STOP ' FATAL ERROR.'
0033 END IF
0034 PROB=TAIL+BODY
0035 RETURN
0036 END

```

```

0001 DOUBLE PRECISION FUNCTION PGRAND(BETA)
C
C INTEGRAND FUNCTION
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 LOGICAL TRASH1, TRASH2
0004 VIRTUAL PUJV(2053), PUIT(2053), IPUIT(2053),
0005 FLV(8191), FLT(8191), IFLT(8191)
0006 INTEGER NCHAN(O:6)
0007 COMMON /JAMCNT/ NCHAN
0008 COMMON /PARMS/ BIGK, LL
C AJ = NONCENTRAL PARAMETER FOR JAMMED HOPS
C ANJ = NONCENTRAL PARAMETER FOR NON-JAMMED HOPS
C NJ = NUMBER OF JAMMED HOPS
C NNJ = NUMBER OF NON-JAMMED HOPS
C COMMON /PUIPAR/ AJ, ANJ, NJ, NNJ
C IF(TRASH1) THEN
C TRASH THE SIGNAL CHANNEL DENSITY TABLES
0009 DO 1 I=1,2053
0010 PUJV(I)=0.DO
0011 DO 2 I=1,2053
0012 PUIT(I)=0.DO
0013 DO 3 I=1,2053
0014 IPUIT(I)=0
0015 TRASH1=.FALSE.
0016 END IF
0017 IF(TRASH2) THEN
0018 TRASH THE NON-SIGNAL CHANNEL DENSITY TABLES
0019 DO 4 I=1,8191
0020 FLV(I)=0.DO
0021 DO 5 I=1,8191
0022 FLT(I)=0.DO
0023 DO 6 I=1,8191
0024 IFLT(I)=0
0025 TRASH2=.FALSE.
0026 END IF
0027 PROD=1.DO
0028 DO 10 I=0,LL
0029 IF(NCHAN(I).NE.0) THEN
0030 JSUB=ISUB
0031 I=FLT(I SUB)
0032 IT=IFLT(I SUB)
0033 IF(IT.EQ.0.DO .AND. IT.EQ.0) THEN
0034 NOT FOUND, COMPUTE IT AND ENTER INTO TABLE
0035 X=FL(BETA,I)
0036 FLV(I SUB)=X
0037 FLT(I SUB)=X
0038 IFLT(I SUB)=I+1
0039

```

```

0040 ELSE IF(T.NE.BETA .OR. IT.NE.I+1) THEN
0041 NOT THIS ENTRY, TRY NEXT ONE
0042 ISUB=ISUB+1
0043 IF(ISUB.GT.8191) ISUB=ISUB-8191
0044 IF(ISUB.NE.JSUB) THEN
0045 GOTO 20
0046 ELSE
0047 HASH TABLE OVERFLOW, MUST COMPUTE, CAN NOT STORE
0048 X=FL(BETA,I)
0049 END IF
0050 ELSE IF(T.EQ.BETA .AND. IT.EQ.I+1) THEN
0051 GOT IT!
0052 X=FLV(ISUB)
0053 END IF
0054 PROD=PROD*DXI(X,NCHAN(I))
0055 END IF
0056 ISUB=IHASH(BETA,2053)
0057 JSUB=ISUB
0058 T=PUIT(ISUB)
0059 IT=IPUIT(ISUB)
0060 IF(T.EQ.O.DO .AND. IT.EQ.O) THEN
0061 NOT FOUND, COMPUTE IT AND ENTER INTO TABLE
0062 Y=PUJ(BETA)
0063 PUIV(ISUB)=Y
0064 PUIT(ISUB)=BETA
0065 IPUIT(ISUB)=NJ+2
0066 ELSE IF(T.NE.BETA .OR. IT.NE.NJ+2) THEN
0067 NOT THIS ENTRY, TRY NEXT
0068 ISUB=ISUB+1
0069 IF(ISUB.GT.2053) ISUB=ISUB-2053
0070 IF(ISUB.NE.JSUB) THEN
0071 GOTO 30
0072 ELSE
0073 HASH TABLE OVERFLOWED
0074 Y=PUJ(BETA)
0075 END IF
0076 ELSE IF(T.EQ.BETA .AND. IT.EQ.NJ+2) THEN
0077 GOT IT!
0078 Y=PUIV(ISUB)
0079 END IF
0080 PGRAND=Y*(1.OO-PROD)
0081 RETURN
0082 END

```

```

0001 DOUBLE PRECISION FUNCTION PUI(Y)
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 DIMENSION WORK(75), STACK(75), SAVE(75)
0004 EXTERNAL DGXVI, PUIG
0005 COMMON /PUICOM/ Y
0006 AJ = NONCENTRAL PARAMETER FOR JAMMED HOPS
0007 ANJ = NONCENTRAL PARAMETER FOR NON-JAMMED HOPS
0008 NJ = NUMBER OF JAMMED HOPS
0009 NNJ = NUMBER OF NON-JAMMED HOPS
0010 COMMON /PUIPAR/ AJ, ANJ, NJ, NNJ
0011 COMMON /PARMS/ BTGK, LL
0012 IF(Y.LE.O.DO) THEN
0013 PUI=O.DO
0014 RETURN
0015 END IF
0016 C NJ HAS ALREADY HAD 1 SUBTRACTED FOR RAPID CHI-SQUARE EVALUATION...
0017 IF(NJ.NE.-1 .AND. NJ.NE.LL-1) THEN
0018 WE MUST CONVOLVE TWO NONCENTRAL CHI-SQUARE DENSITIES
0019 YY=Y
0020 CALL ADQUA2(O.,Y,VALUE,DGXVI,PUIG,1.D-10,WORK,STACK,
0021 $ SAVE,75,KODE)
0022 IF(KODE.NE.O) THEN
0023 WRITE(5,1) KODE
0024 FORMAT(' PUI ADQUA2 ERROR: KODE=',12)
0025 STOP 'FATAL ERROR'
0026 END IF
0027 PUI=VALUE/BTGK
0028 ELSE
0029 WE ONLY HAVE ONE NONCENTRAL CHI-SQUARE DENSITY
0030 WITH 2*LL DEGREES OF FREEDOM
0031 IF(NJ.EQ.-1) THEN
0032 ALL HOPS UNJAMMED
0033 CALL CHISQE(Y,LL-1,ANJ,F,KODE)
0034 ELSE IF(NJ.EQ.LL-1) THEN
0035 ALL HOPS JAMMED
0036 CALL CHISQE(Y/BTGK,LL-1,AJ,F,KODE)
0037 F=F/BTGK
0038 END IF
0039 IF(KODE.NE.O) THEN
0040 WRITE(5,11) KODE
0041 FORMAT(' BESSEL FUNCTION ERROR CODE = ',11)
0042 STOP 'FATAL ERROR'
0043 END IF
0044 PUI=F
0045 RETURN
0046 END

```

```
0001 C DOUBLE PRECISION FUNCTION PUIG(X)
      C INNER INTEGRAND FUNCTION
      C
      C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /PUICOM/ YY
      AJ = NONCENTRAL PARAMETER FOR JAMMED HOPS
      ANJ = NONCENTRAL PARAMETER FOR NON-JAMMED HOPS
      NJ = NUMBER OF JAMMED HOPS
      NNJ = NUMBER OF NON-JAMMED HOPS
      COMMON /PUIPAR/ AJ, ANJ, NJ, NNJ
      COMMON /PARMS/ BIGK, LL
      CALL CHISOE(X,BIGK,NJ,AJ,F1,KODE)
      IF(KODE.NE.0) THEN
        WRITE(5,1) KODE
        FORMAT(' BESSEL FUNCTION ERROR CODE: ',I2)
        STOP 'FATAL IN JAMMED HOP DENSITY'
      END IF
      CALL CHISOE(YY-X,NNJ,ANJ,F2,KODE)
      IF(KODE.NE.0) THEN
        WRITE(5,1) KODE
        STOP 'FATAL IN UNJAMMED HOP DENSITY'
      END IF
      PUIG=F1*F2
      RETURN
      END
```

```
0001 C DOUBLE PRECISION FUNCTION FL(ALPHA,LL)
      C NONSIGNAL CHANNEL CUMULATIVE DISTRIBUTION
      C
      C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /PARMS/ BIGK, LL
      IF(L.NE.0.AND.L.NE.LL) THEN
        R=(BIGK-1.)/BIGK
        BIGLOG=DL06(BIGK)
        ARG=L*BIGLOG+ALPHA/2.
        FL=L-1
      DO 100 N=0,150
        FN=N
        IF(N.EQ.0) THEN
          PART=1.DO
          TERM=DIEXP(ALPHA/2.DO,START,LL-1)
          SUM=TERM
        ELSE
          PART=PART*R*(FN+LL)/FN
          TERM=PART*DIEXP(ALPHA/2.DO,START,LL+N-1)
          DUMMY=SUM+TERM
          SUM=DUMMY
        END IF
        IF(DABS(TERM).LE.1.D-11*DBS(SUM)) GOTO 125
        CONTINUE
      STOP 'FL SUM DID NOT CONVERGE'
      ELSE
        IF(L.EQ.0) THEN
          A=ALPHA/2.DO
        ELSE IF(L.EQ.LL) THEN
          A=ALPHA/(2.DO*BIGK)
        END IF
        START=DEXP(-A)
        SUM=DIEXP(A,START,LL-1)
      END IF
      FL=1.DO-SUM
      RETURN
      END
```

```
0001      DOUBLE PRECISION FUNCTION DIEXP(X, START, IUP)
          C INCOMPLETE EXPONENTIAL FUNCTION (DOUBLE PRECISION)
          C
          C
          IMPLICIT DOUBLE PRECISION(A-H, O-Z)
          TERM=START
          DIEXP=TERM
          IF(IUP.EQ.0) RETURN
          DO 100 I=1, IUP
             F=1
             TERM=TERM**X/F
             DIEXP=DIEXP+TERM
          100 CONTINUE
          RETURN
          END
```

```
0001      SUBROUTINE CHISQE(X, N, A, DEN, KODE)
          C NON-CENTRAL CHI-SQUARE DENSITY FOR EVEN DEGREES OF FREEDOM
          C DEGREES OF FREEDOM (M) IS M=2*N+2
          C
          IMPLICIT DOUBLE PRECISION(A-H, O-Z)
          B=DSORT(X**A)
          CALL DXBESI(B, N, BESSEL, KODE)
          R=X/A
          IF(R.NE.0.D0) THEN
             POWER=R**(N/2.D0)
          ELSE
             IF(N.NE.0) THEN
                POWER=0.D0
             ELSE
                POWER=1.D0
             END IF
          END IF
          DEN=0.5D0*POWER*DEXP(B-0.5D0*(X**A))*BESSEL
          RETURN
          END
```

```

0001 SUBROUTINE ADQUA2(XL,XU,Y,OR,F,TOL,WORK,STACK,SAVE,M,KODE)
C ADAPTIVE QUADRATURE ALGORITHM
C XL - LOWER LIMIT OF INTEGRAL (IN)
C XU - UPPER LIMIT OF INTEGRAL (IN)
C Y - VALUE OF INTEGRAL (OUT)
C QR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
C WITH CALLING SEQUENCE
C CALL QR(XL,XU,F,Y)
C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
C WORK - WORK ARRAY OF SIZE N (IN)
C STACK - SECOND WORK ARRAY OF SIZE N, MUST NOT BE
C SAME ARRAY AS WORK (IN)
C SAVE - THIRD WORK ARRAY OF SIZE N, MUST NOT BE
C SAME ARRAY AS WORK NOR SAME AS STACK (IN)
C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (IN)
C KODE - ERROR INDICATOR (OUT)
C 0 -- NO ERROR
C 1 -- WORK ARRAYS TOO SMALL
C 2 -- EPS DIVIDED TO ZERO, EITHER ASKING FOR TOO
C TIGHT A TOLERANCE OR ROUND-OFF PREVENTS
C ATTAINING REQUIRED ACCURACY
C R. H. FRENCH, 14 AUGUST 1984
    
```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 EXTERNAL F
0004 DIMENSION WORK(N),STACK(N),SAVE(N)
0005 KODE=0
0006 Y=0.00
0007 WORK(1)=XU
0008 A=XL
0009 MPTS=1
0010 EPS=TOL
0011 STACK(1)=EPS
0012 CALL QR(XL,XU,F,T)
0013 SAVE(1)=T
0014 B=WORK(NPTS)
0015 XM=(A+B)*0.500
0016 CALL QR(A,XM,F,P1)
0017 CALL QR(XM,B,F,P2)
0018 IF(DABS(T-P1-P2).LE.EPS) GOTO 20
C SPLIT IT
MPTS=MPTS+1
IF(MPTS.GT.N) THEN
    KODE=1
    RETURN
END IF
WORK(MPTS)=XM
EPS=EPS/2.00
    
```

```

0026 IF(EPS.EQ.0.00) THEN
0027 KODE=2
0028 RETURN
0029 EMD IF
0030 STACK(MPTS)=EPS
0031 T=P1
0032 SAVE(MPTS)=P2
0033 GOTO 10
C FINISHED A PIECE
20 Y=Y+P1+P2
EPS=STACK(MPTS)
T=SAVE(MPTS)
MPTS=MPTS-1
A=B
IF(MPTS.EQ.0) RETURN
GOTO 10
END
    
```

```

0001 SUBROUTINE ADQUA2(XL,XU,Y,OR,F,TOL,WORK,STACK,SAVE,M,KODE)
C ADAPTIVE QUADRATURE ALGORITHM
C XL - LOWER LIMIT OF INTEGRAL (IN)
C XU - UPPER LIMIT OF INTEGRAL (IN)
C Y - VALUE OF INTEGRAL (OUT)
C QR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
C WITH CALLING SEQUENCE
C CALL QR(XL,XU,F,Y)
C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
C WORK - WORK ARRAY OF SIZE N (IN)
C STACK - SECOND WORK ARRAY OF SIZE N, MUST NOT BE
C SAME ARRAY AS WORK (IN)
C SAVE - THIRD WORK ARRAY OF SIZE N, MUST NOT BE
C SAME ARRAY AS WORK NOR SAME AS STACK (IN)
C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (IN)
C KODE - ERROR INDICATOR (OUT)
C 0 -- NO ERROR
C 1 -- WORK ARRAYS TOO SMALL
C 2 -- EPS DIVIDED TO ZERO, EITHER ASKING FOR TOO
C TIGHT A TOLERANCE OR ROUND-OFF PREVENTS
C ATTAINING REQUIRED ACCURACY
C R. H. FRENCH, 14 AUGUST 1984
    
```

0001 FUNCTION IHASH(BETA, ISIZE)

C AD HOC HASHING FUNCTION

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)

IF(BETA.LT.1.D0) THEN

IF(BETA.GT.0.500) THEN

B=1.D0/(1.0100-BETA)

ELSE

B=50.D0+1.500/(BETA+0.01100)

END IF

ELSE

R=100000.D0*(BETA-DINT(BETA*1000.D0)/1000.D0)+187.D0

END IF

B=B*23.D0

SIZE=ISIZE

I=DM00(R+0.500,SIZE)+0.500

ISHASH=I+1

RETURN

END

0001 SUBROUTINE DG16(A,B,F,ANSWER)

C 16-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL

C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4

C R. H. FRENCH, 28 FEBRUARY 1986

C

0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)

0003 DIMENSION X(8),M(8)

0004 DATA X / 0.09501250983763744018500,

0.28160355077925891323000,

0.45801677765722738634200,

0.61787624440264374844700,

0.75540440835500303389500,

0.86563120238783174388000,

0.94457502307323257607800,

0.98940093499164993259600 /

0005 DATA W / 0.18945061045506849628500,

0.18260341504492358886700,

0.16915651939500253818900,

0.14959598881657673208100,

0.12462897125553387205200,

0.09515851168249278481000,

0.06225323938647892863000,

0.02715245941175409485200 /

ANSWER=0.D0

BMA02=(B-A)/2.D0

BPA02=(B+A)/2.D0

DO 10 I=1,8

C=X(I)*BMA02

Y1=BPA02+C

Y2=BPA02-C

ANSWER=ANSWER+W(I)*(F(Y1)+F(Y2))

CONTINUE

10 ANSWER=ANSWER*BMA02

RETURN

END

0001 SUBROUTINE DGXYI(A,B,F,ANSWER)
C 16-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
C C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
C R. H. FRENCH, 28 FEBRUARY 1986

0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 DIMENSION X(8),M(8)
0004 DATA X/ 0.09501250983763744018500,
\$ 0.28160355077925891323000,
\$ 0.4580167765722738634200,
\$ 0.61787624440264374844700,
\$ 0.7540440835500303389500,
\$ 0.86563120238783174388000,
\$ 0.94457502307323257607800,
\$ 0.98940093499164993259600 /
DATA M/ 0.18945061045506849628500,
\$ 0.18260341504492358886700,
\$ 0.16915651939500253818900,
\$ 0.1495959881657673208100,
\$ 0.12462897125553387205200,
\$ 0.09515851168249278481000,
\$ 0.05225352393864789286300,
\$ 0.02715245941175409485200 /
ANSWER=0.00
BPA02=(B-A)/2.D0
BPA02=(B+A)/2.D0
DO 10 I=1,8
C=X(I)*BPA02
Y1=BPA02+C
Y2=BPA02-C
ANSWER=ANSWER+W(I)*(F(Y1)+F(Y2))
CONTINUE
ANSWER=ANSWER*BPA02
RETURN
END

10
0013
0014
0015
0016
0017

0001 SUBROUTINE DLGS15(F,A,RESULT)
C DOUBLE PRECISION 15-POINT LAGUERRE INTEGRATION
C FOR THE SHIFTED SEMI-INFINITE INTERVAL:
C INFINITY
C / / F(X) DX = RESULT
C / / /

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 DIMENSION X(15),W(15)
0004 DATA X/ 0.09330781201700,
\$ 0.49269174030200,
\$ 1.21559541207100,
\$ 2.26994952620400,
\$ 3.66762272175100,
\$ 5.42533662741400,
\$ 7.56591622661300,
\$ 10.12022856801900,
\$ 13.13028248217600,
\$ 16.65440770833000,
\$ 20.77647889944900,
\$ 25.62389422672900,
\$ 31.40751916975400,
\$ 38.53068330648600,
\$ 48.02608557268600 /
DATA W/ 0.23957817031100,
\$ 0.56010084279300,
\$ 0.88700826291900,
\$ 1.2236644021500,
\$ 1.5744487216300,
\$ 1.9447519765300,
\$ 2.3415020566400,
\$ 2.7740419268300,
\$ 3.2556433464000,
\$ 3.8063117142300,
\$ 4.458477538400,
\$ 5.2700177844300,
\$ 6.3595634697300,
\$ 8.0317876321200,
\$ 11.527772100900 /

0005
RESULT=0.D0
DO 10 J=1,15
RESULT=RESULT+W(J)*F(A+X(J))
CONTINUE
RETURN
END
10
0006
0007
0008
0009
0010
0011

```

PDP-11 FORTRAN-77 V4.0-1 09:43:11 16-Jul-86 /F77/TR:BLOCKS/MR
RMFSKMH8.FIN;13

0001 SUBROUTINE PSEL2(JSUB,LLL,MM,RHOM1,RHOT1,PROB)
C RANDOM WFSK/FH IN PARTIAL BAND TONE JAMMING,
C GIVEN A JAMMING EVENT
C
C JSUB - JAMMING EVENT VECTOR
C LL - NUMBER OF HOPS/SYMBOL
C MM - ALPHABET SIZE
C PROB - RESULTING CONDITIONAL ERROR PROBABILITY
C
C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
C INTEGER RO,RL
C INTEGER JSUB(8)
C DIMENSION N(0:9),CO(0:24),CL(0:24),COEFK(10),COEFP(10),
C $ COEFR(10),COEFK(10),INDUP(20)
C $ INDEX(20),INDUP(20)
C SYMBOLIC NAMES FOR INDEX-IDENTIFICATION SUBSCRIPTS
C (LOOP NUMBER MODULO 3)
C PARAMETER(LOOPP=1, LOOPK=2, LOOPP=0)
C COMMON /PARMS/ BIGK, LL
C COMMON /RRRCOM/ RHOT, RHON
C
C COMPUTE PARAMETERS
C
C RHON=RHOM1
C RHOT=RHOT1
C LL=LLL
C BIGK=RHON/RHOT
C LM1=LL-1
C FKLL1 = DXI(BIGK-1,DO,LMI)
C
C COMPUTE POWERS OF FL
C
C DO 10 I=0,LL
C N(I)=0
C 10 CONTINUE
C DO 12 I=2,MM
C N(JSUB(I))=N(JSUB(I))+1
C 12 CONTINUE
C
C OVER-ALL SUMMATION INITIALIZATION
C
C SUM=0.00
C
C ---- START LOOP ON RO ----
C
C DO 9000 RO=0,N(O)
C FRO=RO
C IF (PO.EQ.O) THEN
C COEFO=1.00
C ELSE
C COEFO=-COEFO*((N(O)-FRO+1.00)/FRO)
C END IF
0002
0003
0004
0005
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PDP-11 FORTRAN-77 V4.0-1 09:43:11 16-Jul-86 /F77/TR:BLOCKS/MR
RMFSKMH8.FIN;13

C ---- START LOOP ON KO ----
C
C KOMAX=RO*LMI
C
C PRE-COMPUTE THE J.C.P. MILLER COEFFICIENTS OVER KO:
C
C CALL JCPMF(CO,KOMAX,RO,LMI)
C DO 8000 KO=0,KOMAX
C
C ---- START LOOP ON RL ----
C
C DO 7000 RL=0,N(LL)
C FRL=RL
C IF (RL.EQ.O) THEN
C COEFL=1.00
C ELSE
C COEFL=-COEFL*((N(LL)-FRL+1.00)/FRL)
C END IF
C
C ---- START LOOP ON KL ----
C
C KLMAX=RL*LMI
C
C PRE-COMPUTE THE J.C.P. MILLER COEFFICIENTS OVER KL:
C
C CALL JCPMF(CL,KLMAX,RL,LMI)
C DO 6000 KI=0,KLMAX
C IF (KI.EQ.O) THEN
C POWKL=1.00
C ELSE
C POWKL=POWKL/BIGK
C COEFKL=CL(KI)*POWKL
C
C ---- START THE VARIABLE-LEVEL NESTED LOOPS
C
C LOOPS=3*LMI
C IF (LOOPS.EQ.O) GOTO 555
C SET UP INDEX OF OUTERMOST LOOP
C INDEX(I)=0
C SET UP THE NON-VARYING UPPER LIMITS
C DO 22 I=1,LMI
C INDUP(3*I-2)=N(I)
C 22 CONTINUE
C
C MARK OUTER-MOST LOOP AS JUST ITERATED
C
C NSUB=1
C
C PERFORM INITIALIZATION CODE FOR LOOPS FROM JUST-ITERATED LOOP
C INWARD

```

```

0079      555      A0=RO+R1/BIGK
0080      IBO=KO+KL
0081      IF (LOOPS.GT.0) THEN
0082          DO 60 J=1,LMI
0083              A0=A0+INDEX(3*J-1)+(-INDEX(3*J-2)-INDEX(3*J-1))/B)BK
0084              IBO=IBO+INDEX(3*J)
0085              CONTINUE
0086      END IF
0087      PART=GRAL(A0,IBO,JSUB(1))
C --- ADD A TERM TO THE OVERALL SUM ---
C
0088      PROD=COEF*CO(KO)*COEFL*COEFL*COEFL
0089      IF (LOOPS.GT.0) THEN
0090          DO 70 J=1,LMI
0091              PROD=PROD*COEFL(1)*COEFL(1)*COEFL(1)
0092          CONTINUE
0093      END IF
0094      PROD=PROD*PART
0095      SUM=SUM+PROD
C --- ITERATE THE LOOPS ---
C
0096      IF (LOOPS.GT.0) THEN
0097          DO 80 NSUB=LOOPS,1,-1
0098              INDEX(NSUB)=INDEX(NSUB)+1
0099              IF (INDEX(NSUB).LE.INDUP(NSUB)) THEN
0100                  GOTO 1000
0101              END IF
0102              CONTINUE
C
0103      END IF
C --- ITERATE THE NORMAL DO LOOPS ---
C
0104      CONTINUE
0105      CONTINUE
0106      CONTINUE
0107      CCONTINUE
0108      PROB=1.DO-SUM
0109      RETURN
0110      END
  
```

```

C FIRST, RESET INDICES OF ALL LOOPS WITHIN ONE JUST ITERATED
C
0055      1000      DO 25 J=NSUB+1,LOOPS
0056          INDEX(J)=0
0057      25      CONTINUE
C SECOND, PERFORM FRONT-OF-LOOP CODE FOR EACH LOOP FROM ITERATED
C LOOP ON INWARD
C
0058      DO 30 J=NSUB,LOOPS
0059          IRKP=MOD(I,3)
0060          IF (IRKP.EQ.0) THEN
0061              MEST=(1+2)/3
0062              IF (IRKP.EQ.1) THEN
0063                  FIRST TIME THROUGH:
0064                  (A) INITIALIZE THE COEFFICIENT
0065                      COEFL(NEST)=1.DO
0066              ELSE
0067                  NOT FIRST TIME THROUGH:
0068                  UPDATE THE COEFFICIENT
0069                      COEFL(NEST)=-((COEFL(NEST)/FK(1))+(1-DO)/INDEX(1))
0070              END IF
0071              SET UP VARIABLE UPPER LIMITS FOR THE ASSOCIATED
0072              K LOOP
0073              INDUP(I+1)=INDEX(1)
0074              ELSE IF (IRKP.EQ.1) THEN
0075                  WE ARE ITERATING A K LOOP
0076                  IF (INDEX(1).EQ.0) THEN
0077                      FIRST TIME THROUGH:
0078                      INITIALIZE THE COEFFICIENT
0079                      COEFL(NEST)=1.DO
0080              ELSE
0081                  NOT FIRST TIME THROUGH:
0082                  UPDATE THE COEFFICIENT
0083                      COEFL(NEST)=COEFL(NEST)*
0084                      ((INDEX(1)-1)-INDEX(1)+1.DO)/INDEX(1))
0085              END IF
0086              SET UP THE UPPER LIMIT FOR THE ASSOCIATED P LOOP
0087              INDUP(I+1)=ICAPP(LL,NEST,INDEX(1-1),INDEX(1))
0088              ELSE IF (IRKP.EQ.2) THEN
0089                  WE ARE ITERATING A P LOOP
0090                  COEFL(NEST)=DEE(LL,NEST,INDEX(1-2),INDEX(1-1),INDEX(1),BIGK)
0091              END IF
0092      30      CONTINUE
C --- DO THE INTEGRAL ---
C
  
```

```

0001      DOUBLE PRECISION FUNCTION DEE(LL,NEST,IR,K,IP,BIGK)
0002      C THE FUNCTION D(P) FOR 'BNJ/RMFSK
0003      C
0004      C
0005      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0006      DIMENSION G(0:8)
0007      IF(LL.EQ.2) THEN
0008          DEE=DXI(BIGK,IR-K)
0009          IF(NEST.EQ.1) THEN
0010              DEE=DXI(BIGK,IR-K)
0011          ELSE IF(LL.EQ.3) THEN
0012              IF(NEST.EQ.1) THEN
0013                  DEE=DXI(BIGK*8IGK,IR-K)*DBINCO(K,IP)*
0014                      DXI(2.00*BIGK-1.00,K-IP)*DXI(BIGK-1.00,IP)
0015              ELSE IF(NEST.EQ.2) THEN
0016                  DEE=DBINCO(IR-K,IP)*DXI(BIGK*(BIGK-2.00),IR-K-IP)*
0017                      DXI(BIGK-1.00,IP)
0018              END IF
0019          ELSE IF(LL.EQ.4) THEN
0020              IF(NEST.EQ.1) THEN
0021                  DENOM=3.00*BIGK*8IGK-3.00*BIGK+1.00
0022                  COEF1=(2.00*BIGK*8IGK-3.00*BIGK+1.00)/DENOM
0023                  G(0)=1.00
0024                  G(1)=K*COEF1
0025                  BK1=8IGK-1.00
0026                  COEF2=0.500*BK1*BK1/DENOM
0027                  DO 41 N=2,IP
0028                      G(N)=(K+1-N)*COEF1+G(N-1) + (2*K+2-N)*COEF2*G(N-2)/N
0029                  CONTINUE
0030                  DEE=DXI(BIGK,3*(IR-K))*DXI(DENOM,K)*G(IP)
0031                  IF(MOD(K,2).EQ.1) DEE=-DEE
0032                  ELSE IF(NEST.EQ.2) THEN
0033                      IUP=MINO(IP,K)
0034                      ILOW=MAXO(0,IP-IR+K)
0035                      BK3=8IGK-3.00
0036                      TK1=3.00*BIGK-1.00
0037                      BASE=BK3*BIGK/TK1
0038                      SUMQ=0.00
0039                      DO 42 IQ=ILOW,IUP
0040                          SUMQ=SUMQ+DBINCO(IR-K,IP-IQ)*DBINCO(K,IQ)
0041                          *DXI(BK3,IR-K-IP)*DXI((TK1,K)*DXI(BASE,IQ)
0042                          G(0)=1.00
0043                          ELSE IF(NEST.EQ.3) THEN
0044                              DENOM=DXI(BIGK,3)-3.00*BIGK*8IGK+3.00*BIGK
0045                              COEF1=(BIGK*8IGK-3.00*BIGK+2.00)/DENOM
0046                              G(1)=(IR-K)*COEF1
0047                              BK1=8IGK-1.00
0048                              COEF2=0.500*BK1*BK1/(BIGK*DENOM)
    
```

```

0047      DO 43 N=2,IP
0048          G(N)=[(IR-K+1-N)*COEF1+G(N-1)+
0049              (2*(IR-K+1-N)*COEF2*G(N-2)]/N
0050      CONTINUE
0051      DEE=DXI(DENOM,(IR-K)*G(IP)
0052      IF(MOD(K,2).EQ.1) DEE=-DEE
0053      END IF
0054      RETURN
0055      END
    
```

```

0001      FUNCTION ICAPP(LL,NEST,IR,K)
          C
          C COMPUTE UPPER SUMMATION LIMIT PL
          C
          IF(LL.LE.2) THEN
0002             ICAPP=0
0003          ELSE IF(LL.EQ.3) THEN
0004             IF(NEST.EQ.1) THEN
0005                 ICAPP=K
0006             ELSE IF(NEST.EQ.2) THEN
0007                 ICAPP=IR-K
0008             END IF
0009          ELSE IF(LL.EQ.4) THEN
0010             IF(NEST.EQ.1) THEN
0011                 ICAPP=2*K
0012             ELSE IF(NEST.EQ.2) THEN
0013                 ICAPP=IR
0014             ELSE IF(NEST.EQ.3) THEN
0015                 ICAPP=2*(IR-K)
0016             END IF
0017          END IF
0018          RETURN
0019          END
0020
  
```

```

0001      DOUBLE PRECISION FUNCTION GRAL(A0,I80,L1)
          C
          C THE INNER INTEGRAL FOR CONDITIONAL ERROR PROBABILITY
          C
          IMPLICIT DOUBLE PRECISION (A-H,O-Z)
          COMMON /RRRCOM/ RHOT,RHOM
          COMMON /PARMS/ BIGK, LL
          AI=1.DO+A0
          IF(LL.GT.0 .AND. L1.LT.LL) THEN
0002             BK1=BIGK-1.DO
0003             BASE=BK1/(1.DO+BIGK*A0)
0004             ARG1=(L1-LL)*RHOM/AI
0005             ARG2=-L1*RHOM+BIGK/(1.DO+BIGK*A0)
0006             ARG2=ARG2*AI/BK1
0007             POWER=1.DO
0008             SUM=0.DO
0009             DO 100 I=0,I80
0010                 IF(IQ.GT.0) POWER=POWER*BASE
0011                 PM=DSORT(POWER)
0012                 SUM=GNLGP(M(IQ+LL-1,I80-IQ,ARG1,PM)*GNLGP(M(L1-1,I0,ARG2,PM))+SUM
0013                     F=1.DO
0014                     DO 110 J=1,I80
0015                         F=F*(I/AI)
0016                     CONTINUE
0017                 F=(F/DXI(AI,LL-LL))/DXI(1,I0+BIGK*A0,L1)
0018                 GRAL=SUM+F*DEXP(I/ARG1+ARG2)*A0
0019             ELSE IF(L1.EQ.0) THEN
0020                 ARG1=-LL*RHOM/AI
0021                 START=DEXP(ARG1*A0-LL*DLOG(AI))
0022                 DO 200 I=1,I80
0023                     START=START*(I/AI)
0024                 CONTINUE
0025                 GRAL=GNLGP(M(LL-1,I80,ARG1,START)
0026                     ELSE IF(L1.EQ.LL) THEN
0027                 A2=1.DO+BIGK/A2
0028                 ARG1=-LL*RHOM/A2
0029                 BASE=BIGK/A2
0030                 START=DEXP(ARG1*BIGK*A0-LL*DLOG(A2))
0031                 DO 300 J=1,I80
0032                     START=START*(I*BASE)
0033                 CONTINUE
0034                 GRAL=GNLGP(M(LL-1,I80,ARG1,START)
0035                     END IF
0036             RETURN
0037             END
0038
  
```

```

0001      DOUBLE PRECISION FUNCTION GMLGPM(IALFA,N,X,PREMUL)
C
C   GENERALIZED LAGUERRE POLYNOMIALS WITH PREMULIPLICATION BY
C   A WEIGHTING FACTOR (TO DEFER THE INEVITABLE OVERFLOWS)
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      TERM=DBINCOI(N+IALFA,N)*PREMUL
      SUM=TERM
      IF(N.EQ.0) GOTO 200
      DO 100 N=1,N
        TERM=TERM*((X/M)*((N-M+1.DO)/(IALFA+M)))
        SUM=SUM+TERM
      CONTINUE
      GMLGPM=SUM
      RETURN
      END
  
```

```

0001      SUBROUTINE JCPMF(C,KMAX,R,LMI)
C
C   COMPUTE J.C.P. MILLER COEFFICIENTS DIVIDED BY K!
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      INTEGER R
      DIMENSION C(0:KMAX)
      C(0)=1.DO
      IF(LMI.EQ.0) RETURN
      DO 100 I=1,KMAX
        SUM=0.DO
        MUP=MINO(K,LMI)
        BC=1.DO
        DO 90 N=1,MUP
          BC=BC*((K-N+1.DO)/N)
          SUM=SUM+BC*((R+1.DO)*N-K)*C(K-N)
        CONTINUE
        C(K)=SUM/K
      CONTINUE
      FAC=1.DO
      DO 110 K=1,KMAX
        FAC=FAC*K
        C(K)=C(K)/FAC
      CONTINUE
      RETURN
      END
  
```

APPENDIX E
COMPUTER PROGRAM FOR
INDIVIDUAL CHANNEL ADAPTIVE GAIN CONTROL RECEIVER

The following pages contain the source code listing for the FORTRAN-77 program used for numerical computations of the performance of the individual channel adaptive gain control receiver for FH/RMFSK.

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```
0001      PROGRAM AGCIC
C THIS PROGRAM COMPUTES THE ERROR PROBABILITY FOR RANDOM M-ARY
C FSK WITH MULTIPLE HOPS/BIT IN THE PRESENCE OF PARTIAL-BAND
C NOISE JAMMING WITH INDIVIDUAL CHANNEL AGC RECEIVER
C BY NUMERICAL INTEGRATION
C
C ANALYSIS: L. E. MILLER, R. H. FRENCH
C PROGRAM: R. H. FRENCH
C
C V 1.0.0 - COMPUTATIONS ONLY
C
C      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
C      PARAMETER (LJ=126)
C      CHARACTER*13 FNAME, GNAME
C      CHARACTER*1 REPLY, YES, NO
C      REAL*4 PLOG(LJ), DBSJR(LJ)
C COMMON /INPUTS/ PASSES, PARAMETERS OF THE RUN
C COMMON /SIZE/ DEBMDL(5), LL, LST(4), NSLOTS, GAMLST(31), K, MH
C COMMON /SIZE/ NO, ML, NG
C DATA YES, NO, Y, N, /
C CALL GET(NJ, START, DBINC)
C CALL ERSET(29, .TRUE., .FALSE., .TRUE., .FALSE., 15)
C WOBRT=0.500*(MH)/(MH-1.00)
C DO 900 IL=1, ML
C   LL=LLST(IL)
C   FLL=LL
C   DO 800 IO=1, MO
C     ERNO=10.00*(DEBMDL(IO)/10.00)
C     RHOH=K*EBNO/FLL
C     IOOUT=DEBMDL(IO)
C     DO 700 IG=1, MG
C       GAMMA=GAMLST(IG)
C       MQ=GAMMA*NSLOTS*0.500
C     OPEN DATA FILE
C
C     IGOUT=GAMMA*1000.00+0.500
C     WRITE(FNAME,730) MH, LL, IOOUT, IGOUT
C     FORMAT('I', I1, I1, I2.2, I4.4, ' DAT')
C     WRITE(6,776) MH, LL, DEBMDL(IO), GAMMA
C     FORMAT(/, ' INDEPENDENT CHANNEL AGC RECEIVER'//
C   ' M= ', I2.5X, ' L= ', I2.5X, ' EB/NO= ', F8.4, 5X, ' GAMMA= ', I10D10.3/
C   ' SWR(DB)', I0X, '(PIE)')
C   WRITE(5,733) FNAME
C   FORMAT(' WORKING ON FILE ', A13)
C   OPEN(UNIT=4, FILE=FNAME, STATUS='OLD', FORM='UNFORMATTED',
C     ERR=750)
C
C   HAVE AN EXISTING FILE, READ TO SEE HOW FAR WE GOT BEFORE
C
C
```

```
0031      WRITE(5,1100)
0032      FORMAT(' IS THIS A FIX-UP RE-RUN? (Y/N): ', $)
0033      READ(5,1101) REPLY
0034      FORMAT(A1)
0035      IF(REPLY.EQ.NO) GOTO 1300
C
C WE ARE TO FIX UP AN EXISTING FILE WHICH HAS GARBAGE AT THE HIGH END
C
C ... FIRST READ EVERYTHING
C      READ(4) MHIN, LLIN, EBMDIN, NSLIN, GAMIN, DBSJR, PRLOG
C      IF(MHIN.NE.MH .OR. LLIN.NE.LL .OR. EBMDIN.NE.DEBMDL(10)
C        .OR. GAMIN.NE.GAMMA .OR. NSLIN.NE.NSLOTS)
C        $      STOP 'FILE SYNC ERROR OR CORRUPTED FILE'
C        $
C      WRITE(5,1102)
C      FORMAT(' HOW MANY POINTS ARE GOOD? ', $)
C      READ(5,1103) JGOOD
C      FORMAT(I3)
C      CLOSE(UNIT=4)
C      JJ=JGOOD+1
C CREATE A NEW FILE WITH THE GOOD PART OF THE DATA IN IT
C IN PROGRESS. FORMAT
C      OPEN(UNIT=4, FILE=FNAME, STATUS='NEW', FORM='UNFORMATTED')
C      WRITE(4) MH, LL, DEBMDL(10), NSLOTS, GAMMA
C      DO 1120 IGOOD=1, JGOOD
C        WRITE(4) DBSJR(IGOOD), PRLOG(IGOOD)
C      CONTINUE
C      CLOSE(UNIT=4)
C AND GO FINISH UP THE CALCULATIONS
C      GOTO 755
C
C NOT A FIX-UP RUN, MUST BE A CONTINUATION RUN
C
C      READ(4) MHIN, LLIN, EBMDIN, NSLIN, GAMIN
C      IF(MHIN.NE.MH .OR. LLIN.NE.LL .OR. EBMDIN.NE.DEBMDL(10)
C        .OR. GAMIN.NE.GAMMA .OR. NSLIN.NE.NSLOTS)
C        $      STOP 'FILE SYNC ERROR OR CORRUPTED FILE'
C        $
C      JJ=0
C      JJ=JJ+1
C      READ(4, END=742) DBSJR(JJ), PRLOG(JJ)
C      GOTO 740
C      CLOSE(UNIT=4)
C      GOTO 755
C
C NO EXISTING FILE, THIS IS THE FIRST TIME: CREATE FILE HEADER RECORD
C
C      JJ=1
C      OPEN(UNIT=4, FILE=FNAME, STATUS='NEW', FORM='UNFORMATTED')
C      WRITE(4) MH, LL, DEBMDL(10), NSLOTS, GAMMA
C      CLOSE(UNIT=4)
C      DO 600 IJ=JJ, NJ
C      GIVE PROGRESS MESSAGE TO TI:
C      WRITE(5, 601) IJ
C      FORMAT(' IJ= ', I3)
C
```

```
0066 TRASH2=.TRUE.  
0067 DEBNJ=START*(I,J-1)+DBINC  
0068 HIGH=DEBNJ.GE.15.DO  
0069 DBSJR(I,J)=DEBNJ  
0070 R=10.DO*(DEBNJ/10.DO)  
0071 RHOTS=GAMMA*R+EBNO/(GAMMA+R+EBNO)  
0072 RHOT=R+RHOTS/FLL  
C EVALUATE THE PROBABILITY  
0073 CALL PSUBE(RHON,RHOT,LL,MM,GAMMA,PESYM)  
0074 PE=WORBIT+PESYM  
0075 WRITE(6,666) DBSJR(I,J),PE  
0076 FORMAT(1X,F7.3,5X,1PD12.5)  
0077 PRL0G(I,J)=DLOG10(PE)  
0078 $ FORM=UNFORMATTED',  
FORM=UNFORMATTED',  
ACCESS='APPEND',  
FILE=FILENAME,STATUS='OLD',  
PRLOG(I,J)  
0079 WRITE(4) DBSJR(I,J), PRL0G(I,J)  
0080 CLOSE(UNIT=4)  
0081 CONTINUE  
0082 OPEN(UNIT=4,FILE=FILENAME,STATUS='NEW',FORM='UNFORMATTED')  
0083 WRITE(4) MM,LL,DEBNOL(10),NSLOTS,GAMMA,DBSJR,PRL0G  
0084 CLOSE(UNIT=4)  
0085 CONTINUE  
0086 CONTINUE  
0087 STOP 'PLEASE PURGE DATA FILES'  
0088 $  
0089 END
```

```
0001 C SUBROUTINE GET(MJ,START,DBINC)  
C INTERACTIVE INPUT OF PARAMETERS FOR RUN  
C  
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)  
0003 CHARACTER*9 FIELD,BLANK9  
0004 COMMON /INPUTS/ DEBNOL(5),LLIST(4),NSLOTS,GAMLST(31),K,MM  
0005 COMMON /SIZE/ NO,NL,NG  
0006 DIMENSION DG(31),DSNR(5,4)  
0007 DATA DG / .00100, .00200, .00500,  
$.0100, .0200, .0500,  
$.100, .200, .500, 1.00, 21*0.00/  
0008 DATA DSNR /13.3524700, 12.313300, 10.9444300, 0.00, 0.00,  
$.10, 60657200, 9.628400, 8.3524800, 0.00, 0.00,  
$.0940100, 8.169000, 6.97199500, 0.00, 0.00,  
$.8, 0.0783500, 7.199600, 6.06964600, 0.00, 0.00/  
0009 DATA BLANK9/'  
0010 WRITE(5,33)  
0011 FORMAT(' BITS/SYMBOL (K) [2]: ', $)  
0012 READ(5,3)K  
0013 IF(K.EQ.0)K=2  
0014 MM=2**K  
0015 WRITE(5,2)  
0016 FORMAT(' HOW MANY EB/MO? [1]: ', $)  
0017 READ(5,3)NO  
0018 FORMAT(I2)  
0019 IF(NO.EQ.0)NO=1  
0020 DO 7 IN=1,NO  
0021 IF(K.LE.4) THEN  
0022 DO=DSNR(IN,K)  
0023 ELSE  
0024 DO=0.DO  
0025 END IF  
0026 WRITE(5,5)IN,DO  
0027 FORMAT(' EB/MO(' ,12,') [' ,F9.6,'] : ', $)  
0028 READ(5,6)FIELD  
0029 FORMAT(A9)  
0030 IF(FIELD.EQ.BLANK9) THEN  
0031 DEBNOL(IN)=DO  
0032 ELSE  
0033 DECODE(9,61,FIELD)DEBNOL(IN)  
0034 FORMAT(F9.6)  
0035 END IF  
0036 CONTINUE  
0037 WRITE(5,16)  
0038 FORMAT(' HOW MANY L? [4]: ', $)  
0039 READ(5,3)NL  
0040 IF(NL.EQ.0)NL=4  
0041 DO 21 IN=1,NL  
0042 WRITE(5,19)IN,IN  
0043 FORMAT(' L(' ,11,') [' ,11,'] : ', $)  
0044 READ(5,3)LLIST(IN)
```

```

0001 C SUBROUTINE PSUBE(RHOM,RHGT,LL,M,GAMMA,PE)
0002 C COMPUTE UNCONDITIONAL ERROR PROBABILITY
0003 C
0004 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0005 PE=0.00
0006 DMG=1.00-GAMMA
0007 DO 100 LL=0,LL
0008 ALAM=2.00*(LL-1)*RHOM+2.00*1)*RHOT
0009 IF(DMG.NE.0.00 .OR. (DMG.EQ.0.00 .AND. LL.EQ.LL)) THEN
0010 CALL PSL1(ALAM,M,LL,PECON)
0011 PE=PE+DBINCO(LL,1)*DXI(GAMMA,LL)*DXI(DMG,LL-1)*PECON
0012 END IF
0013 CONTINUE
0014 RETURN
0015 END
    
```

```

0045 IF(LLIST(IN).EQ.0)LLIST(IN)=IN
0046 CONTINUE
0047 WRITE(5,23)
0048 FORMAT(' HOOPING SLOTS? [2400]: ',5)
0049 READ(5,24)NSLOTS
0050 FORMAT(I5)
0051 IF(NSLOTS.EQ.0)NSLOTS=2400
0052 WRITE(5,26)
0053 FORMAT(' HOW MANY GAMMA? [10]: ',5)
0054 READ(5,3)NG
0055 IF(NG.EQ.0)NG=10
0056 DO 31 IN=1,NG
0057 WRITE(5,29)IN,DG(IN)
0058 FORMAT(' GAMMA(',12,') [',1P08.1,']: ',5)
0059 READ(5,30)GAMST(IN)
0060 FORMAT(D15.8)
0061 IF(GAMST(IN).EQ.0.00)GAMST(IN)=DG(IN)
0062 CONTINUE
0063 WRITE(5,39)
0064 FORMAT(' HOW MANY EB/NJ? [1]: ',5)
0065 READ(5,34,ERR=38) NJ
0066 FORMAT(I3)
0067 IF(NJ.EQ.0) NJ=1
0068 IF(NJ.LT.0 .OR. NJ.GT.126) GOTO 32
0069 WRITE(5,41)
0070 FORMAT(' STARTING VALUE FOR EB/NJ (DB) [0.1]: ',5)
0071 READ(5,42,ERR=40) START
0072 FORMAT(F6.3)
0073 IF(NJ.EQ.1) RETURN
0074 WRITE(5,36)
0075 FORMAT(' DB INCREMENT FOR EB/NJ [5.1]: ',5)
0076 READ(5,37,ERR=35) DBINC
0077 FORMAT(F6.3)
0078 IF(DBINC.EQ.0.) DBINC=5.
0079 RETURN
0080 END
    
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0001 SUBROUTINE PSL1(ALAM,MM,LL,PE)
      C COMPUTE CONDITIONAL ERROR PROBABILITY
      C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      DIMENSION C(0:50)
      PE=0.00
      POWER1=-1.00
      DO 900 K=1,MM-1
        FK=K
        POWER1=-POWER1*(MM-FK)/FK
        COEFK=POWER1/DXI(FK+1,DO,LL)
        IRU=K*(LL-1)
        IF(IRU.GT.50) STOP 'C ARRAY TOO SMALL'
        FK1=K+1
        AX=-FK*ALAM*0.500/FK1
        ARG=-ALAM*0.500/FK1
        SUMR=0.00
        DO 800 IR=0,IRU
          R=IR

```

```

      C BUILD UP THE JCP MILLER COEFFICIENT
      C
      IF(IR.EQ.0) THEN
        C(O)=1.00
      ELSE
        NUP=MINO(IR,MAXD(1,LL-1))
        SEE=0.00
        DO 100 N=1,NUP
          SEE=SEE+DBINCO(IR,N)*((K+1)*N-IR)+C(IR-N)
        CONTINUE
        C(IR)=SEE/R
      END IF
      GL=GENLAG(LL-1,IR,ARG)
      Y=DEXP(AX+DLOG(GL))
      TERM=C(IR)*Y/DXI(FK1,IR)
      SUMR=SUMR+TERM
    800 CONTINUE
      PE=PE+COEFK*SUMR
    900 CONTINUE
      RETURN
      END

```

```

0001 DOUBLE PRECISION FUNCTION GENLAG(IALFA,N,X)
      C GENERALIZED LAGUERRE POLYNOMIALS
      C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      TERM=DBINCO(N+IALFA,N)
      SUM=TERM
      IF(N.EQ.0) GOTO 200
      DO 100 M=1,N
        TERM=-TERM*((X/M)*((N-M+1.00)/(IALFA+M)))
        SUM=SUM+TERM
      CONTINUE
      GENLAG=SUM
      RETURN
      END

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APPENDIX F
COMPUTER PROGRAM FOR
NUMERICAL COMPUTATIONS FOR THE
ANY-CHANNEL-JAMMED ADAPTIVE GAIN CONTROL RECEIVER

The following pages contain the source code listing for the FORTRAN-77 program used for numerical computations of the performance of the any-channel-jammed adaptive gain control receiver for FH/RMFSK.

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```
0001      PROGRAM CRNHOP
C
C THIS PROGRAM COMPUTES THE ERROR PROBABILITY FOR RANDOM M-ARY
C FSK/FH WITH MULTIPLE HOPS/BIT IN THE PRESENCE OF PARTIAL-BAND
C NOISE JAMMING USING THE ANY-CHANNEL-JAMMED AGC RECEIVER
C
C ANALYSIS: L. E. MILLER, R. H. FRENCH
C PROGRAM: R. H. FRENCH
C
C V 5.1.0 - COMPUTATIONS ONLY
C
C
C      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
C      PARAMETER (LJ=126)
C      CHARACTER*13 FNAME, GNAME
C      LOGICAL TRASH1, TRASH2
C      LOGICAL*1 GOOD
C      DIMENSION MATRIX(4,8),MLOW(4,8),MUP(4,8),MINC(4,8),
C      $      PTE(0:8)
C      REAL*4 PRLOG(LJ),DBSRJ(LJ)
C      VIRTUAL A(100),IASUB(100),C(625),IGSUB(625)
C      VIRTUAL D(625),IDSUB(625),PRERR(625),IPSUB(625)
C
C      COMMON /RESET/ TRASH1, TRASH2
C      COMMON /INPUTS/ PASSES PARAMETERS OF THE RUN
C      COMMON /SIZE/ PASSES NUMBERS OF PARAMETERS
C      COMMON /SIZE/ NO,ML,NG
C      COMMON /PARMS/ BTBK, LL
C
C      CALL ERRET(29, .TRUE.,.FALSE.,.TRUE.,.FALSE.,.15)
C      CALL GET(NJ,START,DBTNC)
C      BITS=K
C      NORBIT=0.500*MM/(MM-.1:.00)
C      DO 900 JL=1,ML
C      LL=LLIST(JL)
C      FLL=LL
C      DO 800 IO=1,NO
C      EBNO=IO.DO**((DEBNOL(IO)/10.DO)
C      RHO=BITS*EBNO/FLL
C      IOOUT=DEBNOL(IO)
C      DO 700 IG=1,NG
C      GAMMA=GAMLST(IG)
C      NO-GAMMA*NSLOTS=0.500
C
C      OPEN DATA FILE
C
C      IGOUT=GAMMA*1000.DO+0.500
C      WRITE(FNAME,730) MM,LL,IOOUT,IGOUT
C      FORMAT('A',I1,I1,I2.2,I4.4,'.DAT')
C      WRITE(6,776) MM,LL,DEBNOL(IO),GAMMA
C      FORMAT('M=',I2.5X,'L=',I2.5X,'EB/NO=',F8.4.5X,
C      $      'GAMMA=',IP010.3)
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0034      WRITE(5,733) FNAME
0035      FORMAT(' WORKING ON FILE ',A13)
0036      OPEN(UNIT=4,FILE=FNAME,STATUS='OLD',FORM='UNFORMATTED',
C      $      ERR=750)
C
C HAVE AN EXISTING FILE, READ TO SEE HOW FAR WE GOT BEFORE
C
C
C      READ(4) MMIM, LLIM, EBNOJM, NSLIM, GAMIM
C      IF(MMIM.NE.MM.OR.LLIM.NE.LL.OR.EBNOJM.NE.DEBNOL(IO)
C      $      .OR.GAMIM.NE.GAMMA.OR.NSLIM.NE.NSLOTS)
C      $      STOP 'FILE SYNC ERROR OR CORRUPTED FILE'
C      JJ=0
C      JJ=JJ+1
C      READ(4,END=742) DBSRJ(JJ), PRLOG(JJ)
C      GOTO 740
C      CLOSE(UNIT=4)
C      GOTO 755
C
C NO EXISTING FILE, THIS IS THE FIRST TIME:
C      CREATE FILE HEADER RECORD
C
C      JJ=1
C      OPEN(UNIT=4,FILE=FNAME,STATUS='NEW',FORM='UNFORMATTED')
C      WRITE(4) MM,LL,DEBNOL(IO),NSLOTS,GAMMA
C      CLOSE(UNIT=4)
C      WRITE(GNAME,735) MM,LL,IGOUT
C      FORMAT('JV',I1,I1,I4.4,'.DAT')
C      OPEN(UNIT=3,FILE=GNAME,STATUS='OLD',FORM='UNFORMATTED',
C      $      READONLY,ERR=770)
C      WRITE(5,3939)
C      FORMAT(' READING EXISTING EVENT FILE')
C      READ(3) D,IDSUB,MUSED,GOOD
C      CLOSE(UNIT=3)
C      GOTO 777
C
C IF FILE FOR EVENT PROBABILITIES DOES NOT EXIST,
C CALCULATE THEM AND CREATE A FILE.
C
C      CONTINUE
C      WRITE(5,3938)
C      FORMAT(' CREATING EVENT FILE')
C      CALL GENPTE(LL,MM,NO,NSLOTS,GOOD,MATRIX,MLOW,MUP,PIE,
C      $      D,IDSUB,MUSED)
C      OPEN(UNIT=3,FILE=GNAME,STATUS='NEW',FORM='UNFORMATTED')
C      WRITE(3) D,IDSUB,MUSED,GOOD
C      CLOSE(UNIT=3)
C      IF(.NOT.GOOD) GOTO 700
C      DO 600 IJ=JJ,HJ
C      C GIVE PROGRESS MESSAGE TO TI:
C      WRITE(5,601) IJ
C      FORMAT(' IJ=',I3)
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0068 TRASH2=.TRUE.  
0069 DEBNJ=START+(I,J-1)*DBINC  
0070 DBSJR(I,J)=DEBNJ  
0071 R=I0.00**((DEBNJ/10.00)  
0072 RHOIS=GAMMA**EBNO/(GAMMA**R*EBNO)  
0073 RHOI=BITS*RHOIS/FLL  
C EVALUATE THE PROBABILITY  
0074 CALL PSUBE(RHOM,RHOI,I,MM,BITS,PESYM,D,IDSUB,MUSED,PRERR,  
$ IPSUB,DEBNJ)  
0075 PE=HOBBIT*PESYM  
0076 WRITE(6,666) DBSJR(I,J),PE  
0077 FORMAT(1X,F7.3,5X,IPO12.5)  
0078 PRLOG(I,J)=DLOG10(PE)  
0079 OPEN(UNIT=4,FILE=FNAME,STATUS='OLD',ACCESS='APPEND',  
$ FORM='UNFORMATTED',  
WRITE(4) DBSJR(I,J), PRLOG(I,J)  
0080 CLOSE(UNIT=4)  
0081 CONTINUE  
0082 OPEN(UNIT=4,FILE=FNAME,STATUS='NEW',FORM='UNFORMATTED')  
0083 WRITE(4) MM,I,DEBNOL(I0),NSLOTS,GAMMA,DBSJR,PRLOG  
0084 CLOSE(UNIT=4)  
0085 CONTINUE  
0086 CONTINUE  
0087 CONTINUE  
0088 CONTINUE  
0089 STOP 'PLEASE PURGE DATA FILES'  
0090 END
```

77 3

```
0001 SUBROUTINE GET(NJ,START,DBINC)  
C  
C C INTERACTIVE INPUT OF PARAMETERS FOR RUN  
C  
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)  
0003 CHARACTER*9 FIELD,BLANK9  
0004 COMMON /INPUTS/ DEBNOL(5),LLIST(4),NSLOTS,GAMLST(31),K,MM  
0005 COMMON /SIZE/ NO,NL,NG  
C DEFAULT LISTS TEMPORARILY NEEDED ARE IN SHARED STORAGE WITH  
C THE LARGE CONVOLUTION WORKING ARRAYS  
0006 COMMON /SHARE/ DG(31),DSNR(5,4)  
0007 DATA BLANK9/'  
0008 WRITE(5,33)  
0009 FORMAT(' BITS/SYMBOL (K) [2]: ',S)  
0010 READ(5,3)K  
0011 IF(K.EQ.0)K=2  
0012 MM=2**K  
0013 WRITE(5,2)  
0014 FORMAT(' HOW MANY EB/NO? [1]: ',S)  
0015 READ(5,3)NO  
0016 FORMAT(' [2]: ',S)  
0017 IF(NO.EQ.0)NO=1  
0018 DO 7 IN=1,NO  
0019 IF(K.LE.4) THEN  
0020 DO=DSNR(IN,K)  
0021 ELSE  
0022 DO=0.00  
0023 END IF  
0024 WRITE(5,5)IN,DO  
0025 FORMAT(' EB/NO(' ,I2,') [' ,F9.6,'] : ',S)  
0026 READ(5,6)FIELD  
0027 FORMAT(A9)  
0028 IF(FIELD.EQ.BLANK9) THEN  
0029 DEBNOL(IN)=DO  
0030 ELSE  
0031 DECODE(9,61,FIELD)DEBNOL(IN)  
0032 FORMAT(F9.6)  
0033 END IF  
0034 CONTINUE  
0035 FORMAT(' HOW MANY L? [17]: ',S)  
0036 READ(5,3)NL  
0037 IF(NL.EQ.0)NL=1  
0038 DO 21 IN=1,NL  
0039 WRITE(5,19)IN  
0040 FORMAT(' L? [17,'] [4]: ',S)  
0041 READ(5,3)LLIST(IN)  
0042 IF(LLIST(IN).EQ.0)LLIST(IN)=4  
0043 CONTINUE  
0044 WRITE(5,23)  
0045 FORMAT(' HOPPING SLOTS? [2400]: ',S)  
0046 READ(5,24)NSLOTS  
0047
```

```

0001 SUBROUTINE PSUBE(RHON,RHOT,LL,M,BITS,
      $ PE,D,IDSUB,MUSED,PRERR,IPSUB,EBNJ)
C COMPUTE UNCONDITIONAL ERROR PROBABILITY
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      INTEGER JAM(8),JAML(0:8)
      LOGICAL TRASHI, TRASH2
      VIRTUAL PRERR(625),IPSUB(625)
      VIRTUAL D(625),IDSUB(625)
      COMMON /RESET/ TRASHI, TRASH2
      PE=0.00
      DO 199 LTHJ=0,LL
      JAML(0)=LTHJ
      JAM(1)=-1
      DO 199 I1=0,LTHJ
      JAM(I1)=I1
      ITER=1
      DO 101 I=ITER+1,M
    
```

```

100 DO 101 I=ITER+1,M
      FLAG TO MAKE SURE TRASHI IS TRUE FOR FIRST TIME AROUND
      JAM(I)=LTHJ
      JAM(1)=-1
      DO 199 I1=0,LTHJ
      JAM(I1)=I1
      ITER=1
      DO 101 I=ITER+1,M
    
```

```

C OUTERMOST MONSIGNAL LOOP ALWAYS STARTS FROM 0,
C OTHERS FROM THE CURRENT VALUE OF NEXT OUTER MORE LOOP INDEX
C
      IF(I.GT.2) THEN
      JAM(I)-JAM(I-1)
      ELSE
      JAM(I)=0
      END IF
      CONTINUE
      CALL EVENT(LL,M,JAML,PIE,D,IDSUB,MUSED)
      IF(PIE.EQ.0.00) GOTO 190
    
```

```

0048 FORMAT(I5)
0049 IF(NSLOTS.EQ.0)NSLOTS=2400
0050 WRITE(5,26)
0051 FORMAT(' HOW MANY GAMMA? I3): ', $)
0052 READ(5,3)NG
0053 IF(NG.EQ.0)NG=3
0054 DO 31 I=1,NG
0055 WRITE(5,29)IN,DG(IN)
0056 FORMAT(' GAMMA('I2: ') ['I,1PDB.1,']: ', $)
0057 READ(5,30,ERR=28)GAMLST(IN)
0058 FORMAT(D15.8)
0059 IF(GAMLST(IN).EQ.0.00)GAMLST(IN)=DG(IN)
0060 CONTINUE
0061 WRITE(5,39)
0062 FORMAT(' HOW MANY EB/NJ? I126): ', $)
0063 READ(5,34,ERR=38) NJ
0064 FORMAT(I3)
0065 IF(NJ.EQ.0) NJ=126
0066 IF(NJ.LT.0 .OR. NJ.GT.126) GOTO 32
0067 WRITE(5,41)
0068 FORMAT(' STARTING VALUE FOR EB/NJ (DB) [0.]: ', $)
0069 READ(5,42,ERR=40) START
0070 FORMAT(F6.3)
0071 IF(NJ.EQ.1) RETURN
0072 WRITE(5,36)
0073 FORMAT(' DB INCREMENT FOR EB/NJ [.4]: ', $)
0074 READ(5,37,ERR=35) DBINC
0075 FORMAT(F6.3)
0076 IF(DBINC.EQ.0.) DBINC=0.400
0077 RETURN
0078 END
    
```

```

C TRANSFORM TO EQUIVALENT LINEAR RECEIVER CONDITIONAL PROBABILITY
C
      RHOTIO=RHOTI
      IF(JAML(0)-JAML(1).NE.LL) THEN
      RHOTI=RHON+(JAML(1)*(RHOT-RHON))/(LL-JAML(0)+JAML(1))
      ELSE
      RHOTI=RHOT
      END IF
      IF(RHOTIO.NE.RHOTI) TRASH2=.TRUE.
      JAMI=JAM(I)
      DO 105 I=1,M
      JAM(I)=LL-JAML(0)+JAML(I)
      CONTINUE
      IF(JAMI.NE.JAM(I)) TRASHI=.TRUE.
      UCK=RHON/RHOT
      IF(RHOTI.NE.RHON) THEN
      EBNJI=LL*(RHON*(RHOTI/(BITS*(RHOTI-RHON)))
      ELSE
      EBNJI=LL*(RHON/BITS)
      END IF
    
```

```

0024 RHOTIO=RHOTI
0025 IF(JAML(0)-JAML(1).NE.LL) THEN
0026 RHOTI=RHON+(JAML(1)*(RHOT-RHON))/(LL-JAML(0)+JAML(1))
0027 ELSE
0028 RHOTI=RHOT
0029 END IF
0030 IF(RHOTIO.NE.RHOTI) TRASH2=.TRUE.
0031 JAMI=JAM(I)
0032 DO 105 I=1,M
0033 JAM(I)=LL-JAML(0)+JAML(I)
0034 CONTINUE
0035 IF(JAMI.NE.JAM(I)) TRASHI=.TRUE.
0036 UCK=RHON/RHOT
0037 IF(RHOTI.NE.RHON) THEN
0038 EBNJI=LL*(RHON*(RHOTI/(BITS*(RHOTI-RHON)))
0039 ELSE
0040 EBNJI=LL*(RHON/BITS)
0041 END IF
    
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```

C EMPIRICAL SWITCH-OVER POINT FOR COMPUTATIONAL METHOD
C TO AVOID ROUND-OFF AND TRUNCATION ERRORS
$ IF(EBNJ.GT.31.6 .OR. EBNJ.GE.40.DO .OR.
$ (RHOM.GE.30.DO .AND. EBNJ.GE.25.DO)
$ .OR. UCK.LT.1.01DO) THEN
CALL PSEL1(JAM,LL,M,RHOM,RHOT1,UCK,PROB)
ELSE
CALL PSEL2(JAM,LL,M,RHOM,RHOT1,UCK,PROB)
TRASH1=.TRUE.
TRASH2=.TRUE.
END IF
PE=PE+PROB*PIE
DO 190 I=2,M
ITER=ITER+1
JAML(ITER)=JAML(ITER)+1
IF(JAML(ITER).LE.LTHJ) GOTO 100
CONTINUE
RETURN
END
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SUBROUTINE EVENT(LL,M,JAML,PIE,D,IDSUB,MUSED)
C SUBROUTINE TO LOOK UP EVENT PROBABILITY FROM STORED ARRAY
C
IMPLICIT DOUBLE PRECISION(A-H,O-Z)
LOGICAL*1 STORE,MOME
DIMENSION JAML(9),LUP(9)
VIRTUAL D(625),IDSUB(625)
DIMENSION LOW(9)
DATA STORE/.FALSE./, LOW(9)*0/
C SET UP ARRAY DESCRIPTION D(0:LL,....,0:LL) WITH M+1 DIMENSIONS
DO I I=1,M+1
LUP(I)=LL
CONTINUE
1
C COMPUTE LINEAR EQUIVALENT SUBSCRIPT FOR JAMMING EVENT
CALL LDCN(M+1,LOW,LUP,JAML,ISUB)
C LOOK UP THE VALUE, GET 0 DO IF NOT THERE
CALL LGOKUP(PIE,D,IDSUB,MUSED,625,ISUB,STORE,MOME)
RETURN
END
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0045 C SORT NONSIGNAL CHANNELS
0046 DO 103 I=2,MM-1
0047 DO 103 J=I+1,MM
0048 IF(IWORK(J).LT.IWORK(I)) THEN
0049 ITEMP=IWORK(I)
0050 IWORK(I)=IWORK(J)
0051 ITEMP=IWORK(J)
0052 END IF
0053 DO 103
0054 CALL LOCN(MM+1,LOMRK,LUPIRK,IWORK,ISUB)
0055 CALL LOOKUP(DOUT,D,IDSUB,MUSED,625,ISUB,STORE,MOME)
0056 DOUT=DOUT+P
0057 CALL PUTIN(DOUT,D,IDSUB,MUSED,625,ISUB,IERR,STORE)
0058 IF(IERR.NE.0) STOP 'TOO MANY EVENTS'
0059 C ITERATE MATRIX-INDEX LOOP
0060 CALL MLITER(MATRIX,MLON,MUP,MINC,LL,MM,GO)
0061 IF(GO) GOTO 999
0062 RETURN
0063 END
    
```

```

0001 S SUBROUTINE GENPIE(LL,MM,NQ,NSLOTS,GOOD,MATRIX,MLON,MINC,
0002 MUP,PIE,D,IDSUB,MUSED)
0003 C SUBROUTINE TO GENERATE EVENT PROBABILITIES
0004 C
0005 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0006 LOGICAL I GO,STORE,MOME,GOOD
0007 DIMENSION MATRIX(LL,MM),MLON(LL,MM),MINC(LL,MM),MUP(LL,MM),
0008 PIE(O:MM),IWORK(O:8),LOMRK(O:8),LUPRK(O:8)
0009 $ VIRTUAL O(625),IOSUB(625)
0010 C STORE=.FALSE. => DON'T STORE ZERO ELEMENTS OF SPARSE ARRAY.
0011 STORE=.TRUE.
0012 GOOD=.FALSE.
0013 IF(NQ.LE.O) THEN
0014 GOOD=.FALSE.
0015 RETURN
0016 END IF
0017 DO 90 I=0,MM
0018 CALL PRIHOP(I,MM,NQ,NSLOTS,A)
0019 PIE(I)=A
0020 CONTINUE
0021 DO 95 J=1,MM
0022 DO 95 I=1,LL
0023 DO 95 I=1,LL
0024 MLOM(I,J)=0
0025 MUP(I,J)=1
0026 MINC(I,J)=1
0027 CONTINUE
0028 MUSED=0
0029 DO 98 I=0,MM
0030 LOMRK(I)=0
0031 LUPRK(I)=LL
0032 CONTINUE
0033 C INITIALIZE MATRIX-INDEX LOOP
0034 CALL MLINIT(MATRIX,MLON,LL,MM)
0035 999 CONTINUE
0036 C FORM COLUMN SUMS AND COUNT JAMMED HOPS AND COMPUTE P(EVENT)
0037 LTHJ=0
0038 P=1.00
0039 DO 101 I=1,LL
0040 K=0
0041 DO 100 J=1,MM
0042 K=K+MATRIX(I,J)
0043 CONTINUE
0044 IF(K.NE.O) LTHJ=LTHJ+1
0045 P=P*PIE(K)
0046 CONTINUE
0047 C FORM JAMMING EVENT VECTOR
0048 I=WORK(O)=LTHJ
0049 DO 102 J=1,MM
0050 IWORK(J)=0
0051 DO 102 I=1,LL
0052 IWORK(J)=IWORK(J)+MATRIX(I,J)
0053 CONTINUE
0054 102
    
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0001 SUBROUTINE PUTIN(CIN,C,ICSUB,MUSE,MMAX,K,IEIR,STORE)
C THIS SUBROUTINE INSERTS AN ELEMENT INTO A SPARSE ARRAY FOR
C WHICH ONLY THE NON-ZERO ELEMENTS ARE KEPT IN STORAGE IF
C THE SWITCH STORE IS .TRUE.
C
C THE DOUBLE PRECISION VALUE CIN IS STORED AS C(K) WHERE
C THE AUXILIARY ARRAY ICSUB IS USED TO KEEP TRACK OF THE
C SUBSCRIPTS OF THE CORRESPONDING ENTRIES IN C.
C LONG (4-BYTE) INTEGERS ARE USED TO ACCOMMODATE LARGE
C SUBSCRIPT VALUES FOR THE SPARSE ARRAY C.
C
C USAGE:
C LOGICAL*1 STORE
C DOUBLE PRECISION C,CIN
C VIRTUAL ICSUB(MMAX),C(MMAX)
C CALL PUTIN(CIN,C,ICSUB,MUSE,MMAX,K,IEIR,STORE)
C WHERE
C CIN = VALUE OF ELEMENT TO STORE
C C = ARRAY IN WHICH NON-ZERO VALUES ARE ACTUALLY STORED
C ICSUB = AUXILIARY ARRAY FOR ACTUAL SUBSCRIPT VALUES
C MUSE = NUMBER OF ELEMENTS IN C CURRENTLY OCCUPIED
C MMF = SIZE OF ARRAY C
C I = ERROR RETURN CODE, 0 IF NO ERROR OR 1 IF THERE IS
C J = ROOM AVAILABLE IN C
C S = .E = .TRUE. TO STORE ZEROES EXPLICITLY, ELSE .FALSE.
C NOTE: THE ELEMENT IS DELETED BY SHIFTING DOWNWARD ALL
C FOLLOWING ELEMENTS OF THE ARRAY
C
C PROGRAMMER: ROBERT H. FRENCH DATE: 11 JANUARY 1984

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0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 VIRTUAL ICSUB(MMAX),C(MMAX)
0004 LOGICAL*1 STORE
0005 IERR=0
0006 IF(STORE) GOTO 5
0007 IF(CIN.EQ.0.0) GOTO 30
0008 IF(MUSE.EQ.0) GOTO 20
0009 DO 10 I=1,MUSE
0010 IF(ICSUB(I).NE.K) GOTO 10
0011 C(I)=CIN
0012 RETURN
0013 CONTINUE
0014 IF(MUSE.LT.MMAX) GOTO 20
0015 IERR=1
0016 RETURN
0017 MUSE=MUSE+1
0018 ICSUB(MUSE)=K
0019 C(MUSE)=CIN
0020 RETURN

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```

0021 DO 40 I=1,MUSE
0022 J=I
0023 IF(ICSUB(I).EQ.K) GOTO 50
0024 CONTINUE
0025 RETURN
C
C REMOVE THE ZEROED ELEMENT AND BUMP COUNT OF ENTRIES USED
C
0026 DO 60 J=J,MUSE-1
0027 ICSUB(J)=ICSUB(J+1)
0028 C(J)=C(J+1)
0029 CONTINUE
0030 MUSE=MUSE-1
0031 RETURN
0032 END

```

```

0001 SUBROUTINE LOOKUP(COUT,C,ICSUB,N,MMAX,K,STORE,MOME)
C
C THIS SUBROUTINE RETRIEVES AN ELEMENT OF A SPARSE ARRAY WHICH
C HAS BEEN STORED COMPACTLY BY STORING ONLY NON-ZERO ELEMENTS.
C
C THE ARRAY IS DOUBLE PRECISION.
C
C USAGE:
C   VIRTUAL ICSUB(MMAX), C(MMAX)
C   LOGICAL*1 STORE, MOME
C   DOUBLE PRECISION COUT
C   CALL LOOKUP(COUT,C,ICSUB,N,MMAX,K,STORE,MOME)
C WHERE
C   COUT = VALUE OF C(K) (OUTPUT FROM SUBROUTINE)
C   C = ARRAY USED TO STORE NON-ZERO ELEMENTS
C   ICSUB = AUXILIARY ARRAY TO STORE ACTUAL SUBSCRIPTS
C   N = NUMBER OF ELEMENTS OF C CURRENTLY IN USE
C   MMAX = SIZE OF C
C   K = SUBSCRIPT OF SPARSE ARRAY TO LOOK UP
C   STORE = .TRUE. IF ZEROES STORED EXPLICITLY, ELSE .FALSE.
C   MOME = .FALSE. IF ZEROES NOT STORED OR ZEROES STORED AND
C           ELEMENT IS FOUND IN THE STORED ARRAY
C           .TRUE. IF ZEROES ARE STORED AND THE ELEMENT IS
C           NOT FOUND (OUTPUT QUANTITY)
C
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
    
```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 VIRTUAL ICSUB(MMAX),C(MMAX)
0004 LOGICAL*1 STORE, MOME
0005 MOME=.FALSE.
0006 DO 10 I=1,N
0007 IF(ICSUB(I).NE.K)GOTO 10
0008 COUT=C(I)
0009 RETURN
0010 CONTINUE
0011 IF(STORE) THEN
0012 MOME=.TRUE.
0013 ELSE
0014 COUT=0.
0015 END IF
0016 RETURN
0017 END
    
```

TI
 00

```

0001 SUBROUTINE LOCN(MDIM,ILOW,IUP,ISUB,LINEAR)
C
C THIS SUBROUTINE COMPUTES THE EQUIVALENT LINEAR SUBSCRIPT FOR
C A MULTIDIMENSIONAL ARRAY OF MDIM DIMENSIONS
C
C IF THE ARRAY A IS DEFINED AS
C   DIMENSION A(ILOW(1):IUP(1),...,ILOW(MDIM):IUP(MDIM))
C   AND ISUB(1),...,ISUB(MDIM) IS A SET OF SUBSCRIPTS FOR A,
C THEN THIS SUBROUTINE RETURNS IN LINEAR THE OFFSET FROM THE
C ORIGIN OF A TO THE ELEMENT A(ISUB(1),...,ISUB(MDIM)), ASSUMING
C THE FIRST SUBSCRIPT VARIES MOST RAPIDLY.
C
C USAGE:
C   DIMENSION ILOW(MDIM),IUP(MDIM),ISUB(MDIM)
C   DATA ILOW/lower limits of defined subscripts of array/
C   DATA IUP/upper limits of defined subscripts of array/
C   ...SET ISUB TO DESIRED SEQUENCE OF SUBSCRIPTS...
C   CALL LOCN(MDIM,ILOW,IUP,ISUB,LINEAR)
C WHERE
C   MDIM = NUMBER OF DIMENSIONS THE ARRAY HAS
C   ILOW = ARRAY OF LOWER SUBSCRIPT BOUNDS
C   IUP = ARRAY OF UPPER SUBSCRIPT BOUNDS
C   ISUB = ARRAY CONTAINING SUBSCRIPT FOR WHICH LOCATION IS
C           TO BE COMPUTED
C   LINEAR = RETURNED VALUE OF OFFSET INTO ARRAY IN MEMORY
C
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
    
```

```

0002 DIMENSION ILOW(MDIM),IUP(MDIM),ISUB(MDIM)
0003 LINEAR=0
0004 DO 10 I=1,MDIM-1
0005 J=MDIM-I+1
0006 LINEAR=(LINEAR+(ISUB(J)-ILOW(J)))*(IUP(J-1)-ILOW(J-1)+1)
0007 CONTINUE
0008 LINEAR=LINEAR+ISUB(1)-ILOW(1)
0009 RETURN
0010 END
    
```

```

0001 SUBROUTINE PR1HOP(KJAM,KM,KQ,KN,AIN)
      C THIS SUBROUTINE COMPUTES THE EVENT PROBABILITY FOR ALL
      C POSSIBLE JAMMING PATTERNS WITH NON-ZERO PROBABILITY FOR
      C L=1 HOP/SYMBOL FOR RMFSK/FH IN PBNJ
      C
      C IMPLICIT DOUBLE PRECISION (A-H,O-Z)
      AIN=0.DO
      IF (KJAM.GT.MING(KQ,KM)) RETURN
      KPMAX=KJAM-1
      LPMAX=KM-KJAM-1
      JPMAX=KN-KJAM-1
      IMAX=MAX0(KPMAX,LPMAX,JPMAX)
      PROD=1.DO
      Q=KQ
      DIFFNO=KN-KQ
      EN=KN
      DO 100 LOOP=0,IMAX
      F=L.OOP
      IF (LOOP.LE.KPMAX) PROD=PROD*(Q-F)
      IF (LOOP.LE.JPMAX) PROD=PROD/(EN-F)
      IF (LOOP.LE.LPMAX) PROD=PROD*(DIFFNO-F)
      CONTINUE
      AIN=PROD
      RETURN
      END
    100
  
```

```

0001 BLOCK DATA
      C INITIALIZE SHARED CONSTANTS
      C IMPLICIT DOUBLE PRECISION (A-H,O-Z)
      COMMON /SHARE/ DG(31),DSMR(5,4)
      COMMON /SHAREZ/ LOM(4),LINC(4)
      C DEFAULT LISTS FOR INTERACTIVE PARAMETER INPUTS
      C ARE SHARED WITH LARGE WORKING STORAGE ARRAYS SINCE THEY
      C MAY BE DESTROYED ONCE THE INPUT PARAMETERS ARE SET UP
      DATA DG / .00100, .00200, .00500, .100, .200, .500, 1.00, 21*0.DO/
      DATA DSMR /13.3524700, 12.313300, 10.9444300, 0.00, 0.00,
      $ 10.60657200, 9.628400, 8.3524800, 0.00, 0.00,
      $ 9.0940100, 8.169000, 6.97199500, 0.00, 0.00,
      $ 8.0783500, 7.199600, 6.06964600, 0.00, 0.00/
      C FREQUENTLY NEEDED CONSTANT ARRAYS AND SCALARS
      DATA LOM/4*0/,LINC/4*1/
      END
    0007
    0008
  
```

```

0001 DOUBLE PRECISION FUNCTION PGRAND(BETA)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 LOGICAL TRASH1, TRASH2
0004 VIRTUAL PUIV(2053), PUIT(2053), IPUIT(2053),
    $ FLV(8191), FLT(8191), IFLT(8191)
0005 INTEGER NCHAN(0:6)
0006 COMMON /JAMCNT/ NCHAN
0007 COMMON /RESET/ TRASH1, TRASH2
0008 COMMON /PARMS/ BIGK, LL
    C AJ = NONCENTRAL PARAMETER FOR JAMMED HOPS
    C ANJ = NONCENTRAL PARAMETER FOR NON-JAMMED HOPS
    C NJ = NUMBER OF JAMMED HOPS
    C MNJ = NUMBER OF NON-JAMMED HOPS
    C COMMON /PUIPAR/ AJ, ANJ, NJ, MNJ
    C IF (TRASH1) THEN
    C TRASH THE SIGNAL CHANNEL DENSITY TABLES
    C DO 1 I=1,2053
    C PUIV(I)=0.00
    C DO 2 I=1,2053
    C PUIT(I)=0.00
    C DO 3 I=1,2053
    C IPUIT(I)=0
    C TRASH1=.FALSE.
    C END IF
    C IF (TRASH2) THEN
    C TRASH THE NON-SIGNAL CHANNEL DENSITY TABLES
    C DO 4 I=1,8191
    C FLV(I)=0.00
    C DO 5 I=1,8191
    C FLT(I)=0.00
    C DO 6 I=1,8191
    C IFLT(I)=0
    C TRASH2=.FALSE.
    C END IF
    C PROD=1.00
    C DO 10 I=0,LL
    C IF (NCHAN(I).NE.0) THEN
    C ISUB=THASH(BETA,8191)
    C JSUB=ISUB
    C T=FLT(ISUB)
    C IT=IFLT(ISUB)
    C IF (T.EQ.0.00 .AND. IT.EQ.0) THEN
    C NOT FOUND, COMPUTE IT AND ENTER INTO TABLE
    C X=FL(BETA,I)
    C FLV(ISUB)=X
    C FLT(ISUB)=BETA
    C IFLT(ISUB)=I+1
    C ELSE IF (T.NE.BETA .OR. IT.NE.I+1) THEN
    C NOT THIS ENTRY, TRY NEXT ONE
    C ISUB=ISUB+1
    C IF (ISUB.GT.8191) ISUB=ISUB-8191
    
```

```

0001 SUBROUTINE PSELL(JSUB,LLL,MM,RHOM,RHOT,UCK,PROB)
0002 C RANDOM MFSX/FH IN PARTIAL BAND TONE JAMMING,
0003 C GIVEN A JAMMING EVENT
0004 C JSUB - JAMMING EVENT VECTOR
0005 C LL - NUMBER OF HOPS/SYMBOL
0006 C MM - ALPHABET SIZE
0007 C PROB - RESULTING CONDITIONAL ERROR PROBABILITY
0008 C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0009 C EXTERNAL PGRAND, DGAU20
0010 C DIMENSION WORK(75), STACK(75), SAVE(75)
0011 C INTEGER JSUB(4)
0012 C INTEGER NCHAN(0:6)
0013 C COMMON /PARMS/ BIGK, LL
0014 C AJ = NONCENTRAL PARAMETER FOR JAMMED HOPS
0015 C ANJ = NONCENTRAL PARAMETER FOR NON-JAMMED HOPS
0016 C NJ = NUMBER OF JAMMED HOPS
0017 C MNJ = NUMBER OF NON-JAMMED HOPS
0018 C COMMON /PUIPAR/ AJ, ANJ, NJ, MNJ
0019 C LL=LLL
0020 C BIGK=UCK
0021 C NJ=JSUB(1)
0022 C MNJ=LL-MJ
0023 C AJ=2.00*MJ*RHOT
0024 C ANJ=2.00*MNJ*RHOM
0025 C ASHIFT=DMAX1(AJ,ANJ)
0026 C SUBTRACT 1 FOR BENEFIT OF RAPID CHI-SQUARE DENSITY CALCULATION
0027 C NJ=NJ-1
0028 C MNJ=MNJ-1
0029 C COUNT NUMBER OF NON-SIGNAL CHANNELS WITH L HOPS JAMMED
0030 C DO 1 I=0,LL
0031 C NCHAN(I)=0
0032 C CONTINUE
0033 C DO 2 I=2,MM
0034 C ISUB=JSUB(I)
0035 C NCHAN(ISUB)=NCHAN(ISUB)+1
0036 C CONTINUE
0037 C CALL DLGSLIS(PGRAND,ASHIFT,TAIL)
0038 C CALL ADQUAD(0,ASHIFT,BODY,DGAU20,PGRAND,1,D-10,
    $ WORK,STACK,75,MODE)
0039 C IF (MODE.NE.0) THEN
0040 C WRITE(5,3) MODE
0041 C FORMAT(' ADQUAD ERROR CODE = ',11)
0042 C STOP 'FATAL ERROR'
0043 C END IF
0044 C PROB=TAIL+BODY
0045 C RETURN
0046 C END
    
```

```

0043 IF((ISUB.NE.JSUB) THEN
0044   GOT0 20
0045 ELSE
0046   HASH TABLE OVERFLOW, MUST COMPUTE, CAN NOT STORE
0047   X=FL(BETA,I)
0048   ELSE IF(T.EQ.BETA .AND. IT.EQ.I+1) THEN
0049     GOT IT!
0050     X=FLV(ISUB)
0051     END IF
0052     PROD=PROD*DX(I,X,NCHAN(I))
0053     END IF
0054     10 CONTINUE
0055     ISUB=IMASH(BETA,2053)
0056     JSUB=ISUB
0057     T=PUIT(ISUB)
0058     IT=IPUIT(JSUB)
0059     IF(T.EQ.G.DO .AND. IT.EQ.O) THEN
0060       NOT FOUND, COMPUTE IT AND ENTER INTO TABLE
0061       Y=PUIT(BETA)
0062       PUIV(ISUB)=Y
0063       PUIT(ISUB)=BETA
0064       IPUIT(ISUB)=NJ+2
0065       ELSE IF(T.NE.BETA .OR. IT.NE.NJ+2) THEN
0066         NOT THIS ENTRY, TRY NEXT
0067         ISUB=ISUB+1
0068         IF(ISUB.GT.2053) ISUB=ISUB-2053
0069         IF((ISUB.NE.JSUB) THEN
0070           GOT0 30
0071         ELSE
0072           HASH TABLE OVERFLOWED
0073           Y=PUIT(BETA)
0074           END IF
0075           ELSE IF(T.EQ.BETA .AND. IT.EQ.NJ+2) THEN
0076             GOT IT!
0077             Y=PUIV(ISUB)
0078           END IF
0079           PERAND=Y*(1.DO-PROD)
0080           RETURN
0081           END

```

```

0001 DOUBLE PRECISION FUNCTION PUI(Y)
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 DIMENSION WORK(75), STACK(75), SAVE(75)
0004 EXTERNAL DGXVI, PUIG
0005 COMMON /PUICOM/ Y
0006   AJ = NONCENTRAL PARAMETER FOR JAMMED HOPS
0007   ANJ = NONCENTRAL PARAMETER FOR NON-JAMMED HOPS
0008   NJ = NUMBER OF JAMMED HOPS
0009   NNJ = NUMBER OF NON-JAMMED HOPS
0010   COMMON /PUIPAR/ AJ, ANJ, NJ, NNJ
0011   COMMON /PARMS/ BIGK, LL
0012   IF(Y.LE.0.DO) THEN
0013     PUI=0.00
0014     RETURN
0015   END IF
0016   C NJ HAS ALREADY HAD 1 SUBTRACTED FOR RAPID CHI-SQUARE EVALUATION...
0017   IF(NJ.NE.-1 .AND. NJ.NE.LL-1) THEN
0018     WE MUST CONVOLVE THE NONCENTRAL CHI-SQUARE DENSITIES
0019     YY=Y
0020     CALL ADQUA2(O.,Y,VALUE,DGXVI,PUIG,1.D-10,WORK,STACK,
0021               $ SAVE,75,KODE)
0022     IF(KODE.NE.O) THEN
0023       WRITE(5,1) KODE
0024       FORMAT(' PUI ADQUA2 ERROR: KODE=',12)
0025       STOP 'FATAL ERROR'
0026     END IF
0027     PUI=VALUE/BIGK
0028   ELSE
0029     WE ONLY HAVE ONE NONCENTRAL CHI-SQUARE DENSITY WITH 2*LL
0030     DEGREES OF FREEDOM
0031     IF(NJ.EQ.-1) THEN
0032       ALL HOPS UNJAMMED
0033       CALL CHISQE(Y,LL-1,ANJ,F,KODE)
0034     ELSE IF(NJ.EQ.LL-1) THEN
0035       ALL HOPS JAMMED
0036       CALL CHISQE(Y/BIGK,LL-1,AJ,F,KODE)
0037     END IF
0038     F=F/BIGK
0039     END IF
0040     IF(KODE.NE.O) THEN
0041       WRITE(5,11) KODE
0042       FORMAT(' BESSEL FUNCTION ERROR CODE = ',11)
0043       STOP 'FATAL ERROR'
0044     END IF
0045     PUI=F
0046     END IF
0047     RETURN
0048     END

```

```

0001 DOUBLE PRECISION FUNCTION FL(ALPHA,L)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /PARMS/ BIGK, LL
0004 IF(L.NE.O.AND.L.NE.LL) THEN
0005   R=(BIGK-1.)/BIGK
0006   BIGLOG=DLG(BIGK)
0007   ARG=L*BIGLOG+ALPHA/2.
0008   START=DEXP(-ARG)
0009   FL=L-1
0010   DO 100 N=0,100
0011     EN=N
0012     IF(N.EQ.0) THEN
0013       PART=1.00
0014       TERM=DIEXP(ALPHA/2.00,START,LL-1)
0015       SUM=TERM
0016     ELSE
0017       PART=PART*R*(EM+FL1)/EM
0018       TERM=PART*DIEXP(ALPHA/2.00,START,LL+N-1)
0019       DUMMY=SUM+TERM
0020       SUM=DUMMY
0021     END IF
0022     IF(DABS(TERM).LE.1.D-11*DABS(SUM)) GOTO 125
0023     CONTINUE
0024     STOP 'FL SUM DID NOT CONVERGE'
0025     ELSE
0026     IF(L.EQ.0) THEN
0027       A=ALPHA/2.00
0028     ELSE IF(L.EQ.LL) THEN
0029       A=ALPHA/(2.00*BIGK)
0030     END IF
0031     START=DEXP(-A)
0032     SUM=DIEXP(A,START,LL-1)
0033     END IF
0034     FL=L-1.00-SUM
0035     RETURN
0036     END
    
```

```

0001 DOUBLE PRECISION FUNCTION PUIG(X)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /FULCOM/ YY
0004 C AJ = NONCENTRAL PARAMETER FOR JAMMED HOPS
0005 C ANJ = NONCENTRAL PARAMETER FOR NON-JAMMED HOPS
0006 C NJ = NUMBER OF JAMMED HOPS
0007 C MNJ = NUMBER OF NON-JAMMED HOPS
0008 COMMON /PUIPAR/ AJ, ANJ, NJ, MNJ
0009 CALL CHISQ(X/BIGK, NJ, AJ, F1, KODE)
0010 IF(KODE.NE.0) THEN
0011   WRITE(5,1) KODE
0012   FORMAT(' BESSEL FUNCTION ERROR CODE: ',I2)
0013   STOP 'FATAL IN JAMMED HOP DENSITY'
0014 END IF
0015 CALL CHISQ(YY-X,MNJ,ANJ,F2,KODE)
0016 IF(KODE.NE.0) THEN
0017   WRITE(5,1) KODE
0018   STOP 'FATAL IN UNJAMMED HOP DENSITY'
0019 END IF
0020 PUIG=F1+F2
0021 RETURN
0022 END
    
```

/F77/TR:BLOCKS/MR

```

0001 DOUBLE PRECISION FUNCTION DIEXP(X,START,IUP)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 TERM=START
0004 DIEXP=TERM
0005 IF(IUP.EQ.0) RETURN
0006 DO 100 I=1,IUP
0007 F=1
0008 TERM=TERM*X/F
0009 DIEXP=DIEXP+TERM
0010 CONTINUE
0011 RETURN
0012 END
100

```

```

0001 SUBROUTINE CHISOE(X,N,A,DEN,KODE)
0002 C
0003 C NON-CENTRAL CHI-SQUARE DENSITY FOR EVEN DEGREES OF FREEDOM
0004 C DEGREES OF FREEDOM (M) IS M=2*N+2
0005 C
0006 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0007 B=DSORT(X*A)
0008 CALL DXBESI(B,N,BESSEL,KODE)
0009 IF(KODE.NE.0) RETURN
0010 R=X/A
0011 IF(R.NE.0.DO) THEN
0012 POWER=R**(M/2.DO)
0013 ELSE
0014 POWER=1.DO
0015 END IF
0016 DEN=0.500*POWER*DEXP(B-0.500*(X+A))*BESSEL
0017 RETURN
0018 END

```



```

0001 FUNCTION IHASH(BETA, ISIZE)
0002 C AD HOC HASHING FUNCTION
0003 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0004 IF(BETA.LT.1.D0) THEN
0005   IF(BETA.GT.0.5D0) THEN
0006     B=1.D0/(1.01D0-BETA)
0007   ELSE
0008     B=50.D0+1.5D0/(BETA+0.011D0)
0009   END IF
0010 ELSE
0011   B=100000.D0*(BETA-DINT(BETA*1000.D0)/1000.D0)+187.D0
0012 END IF
0013 B=B*23.D0
0014 SIZE=ISIZE
0015 I=DMOD(B+0.5D0, SIZE)+0.5D0
0016 IHASH=I+1
0017 RETURN
0018 END
  
```

```

0001 SUBROUTINE DG16(A,B,F,ANSWER)
0002 C 16-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
0003 C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
0004 C R. H. FRENCH, 28 FEBRUARY 1986
0005
0006 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0007 DIMENSION X(B), W(B)
0008 DATA X/ 0.095012509837637440185D0,
0009 $ 0.281603550779258913230D0,
0010 $ 0.45801677765727386342D0,
0011 $ 0.617876244402643748447D0,
0012 $ 0.755404408355003033895D0,
0013 $ 0.865631202387831743880D0,
0014 $ 0.944575023073232576078D0,
0015 $ 0.989400934991649932596D0 /
0016 DATA W/ 0.189450610455068496285D0,
0017 $ 0.182603415044923588867D0,
0018 $ 0.169156519395002538189D0,
0019 $ 0.149595988816576732081D0,
0020 $ 0.12462897125533872052D0,
0021 $ 0.095158511682492784810D0,
0022 $ 0.062253523938647892863D0,
0023 $ 0.027152459411754094852D0 /
0024 ANSWER=0.D0
0025 BMA02=(B-A)/2.D0
0026 BPA02=(B+A)/2.D0
0027 DO 10 I=1,8
0028 C=X(I)*BMA02
0029 Y1=BPA02+C
0030 Y2=BPA02-C
0031 ANSWER=ANSWER+H(I)*(F(Y1)+F(Y2))
0032 CONTINUE
0033 RETURN
0034 END
  
```

0001 SUBROUTINE DLGSI5(F,A,RESULT)
 C DOUBLE PRECISION 15-POINT LAGUERRE INTEGRATION
 C FOR THE SHIFTED SEMI-INFINITE INTERVAL:
 C INFINITY

$$\frac{F(X)}{X} \quad \frac{F(X)}{X^2} \quad \dots \quad \frac{F(X)}{X^N}$$

$$F(X) \quad DX = RESULT$$

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
 0003 DIMENSION X(15), W(15)
 0004 DATA X/ 0.09330781201700,
 0.49269174030200,
 1.21559541207100,
 2.26994952620400,
 3.66762272175100,
 5.42533662741400,
 7.56591622661300,
 10.12022856801900,
 13.13028248217600,
 16.65440770833000,
 20.77647889944900,
 25.62389422672900,
 31.40751916975400,
 38.53068330648600,
 48.02608557268600 /

DATA W/ 0.23957817031100,
 0.56010084279300,
 0.88700826291900,
 1.2236644021500,
 1.5744487216300,
 1.9447519765300,
 2.3415020566400,
 2.7740419268300,
 3.2556433464000,
 3.8063117142300,
 4.458477538400,
 5.2700177844300,
 6.3595634697300,
 8.0317876321200,
 11.527772100900 /

RESULT=0.00
 DO 10 J=1,15
 RESULT=RESULT+W(J)*F(A+X(J))
 CONTINUE
 RETURN
 END

0001 SUBROUTINE DEXVI(A,B,F,ANSWER)
 C 16-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
 C C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
 C R. H. FRENCH, 28 FEBRUARY 1966

0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
 0003 DIMENSION X(8), W(8)
 0004 DATA X/ 0.09501250983763744018500,
 0.28160355077925691323000,
 0.4580167776572738634200,
 0.61787624440264374844700,
 0.75540440835500303389500,
 0.865631202387831174388000,
 0.9457502307323257607800,
 0.98940093499164993259600 /

DATA W/ 0.18945061045506849628500,
 0.18260341504492358886700,
 0.16915651939500253818900,
 0.14959598881657673208100,
 0.1246289712553387205200,
 0.09515851168249278481000,
 0.0622532338864789286300,
 0.02715245941175409485200 /

ANSWER=0.00
 BPA02=(B-A)/2.00
 RPA02=(B+A)/2.00
 DO 10 I=1,8
 Y1=BPA02+C
 Y2=RPA02-C
 ANSWER=ANSWER+W(I)*(F(Y1)+F(Y2))
 CONTINUE
 ANSWER=ANSWER*BMA02
 RETURN
 END

10

```

0001 SUBROUTINE PSEL2(JSUB,LLL,MM,RHOW1,RHOT1,LUCK,PROB)
C RANDOM WFSX/FH IN PARTIAL BAND TONE JAMMING,
C GIVEN A JAMMING EVENT
C
C JSUB - JAMMING EVENT VECTOR
C LL - NUMBER OF HOPS/SYMBOL
C MM - ALPHABET SIZE
C PROB - RESULTING CONDITIONAL ERROR PROBABILITY
C
C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
C INTEGER RO,RL
C DIMENSION N(0:4),CO(0:12),CL(0:12),
C COEFF(3),COEFL(3),COEFP(3),
C INDEX(9),INDUP(9)
C $
C $
C SYMBOLIC NAMES FOR INDEX-IDENTIFICATION SUBSCRIPTS
C (LOOP NUMBER MODULO 3)
C PARAMETER(LOOPR=1, LOOPK=2, LOOPP=0)
C COMMON /PARMS/ BIGK, LL
C COMMON /RRRCON/ RHOT,RHOW
C
C COMPUTE PARAMETERS
C
C RHOW=RHOW1
C RHOT=RHOT1
C LL=LLL
C BIGK=LUCK
C LMI=LL-1
C FKJLI = DXI(BIGK-1.DO,LMI)
C
C COMPUTE POWERS OF FL
C DO 10 I=0,LL
C N(I)=0
C 10 CONTINUE
C DO 12 I=2,MM
C N(JSUB(I))=N(JSUB(I))+1
C 12 CONTINUE
C
C OVER-ALL SUMMATION INITIALIZATION
C
C SUM=0.DO
C
C --- START LOOP ON RO ---
C
C DO 9000 RO=0,N(0)
C FRO=RO
C IF(RO.EQ.0) THEN
C COEFO=1.DO
C ELSE
C COEFO=-COEFO*((N(0)-FRO+1.DO)/FRO)
C END IF
    
```

```

0029 C --- START LOOP ON KO ---
C KOMAX=RO*LMI
C
C PRE-COMPUTE THE J.C.P. MILLER COEFFICIENTS OVER KO:
C
C CALL JCPMF(CO,KOMAX,RO,LMI)
C DO 8000 KO=0,KOMAX
C
C --- START LOOP ON RL ---
C
C DO 7000 RL=0,M(LL)
C FRL=RL
C IF(RL.EQ.0) THEN
C COEFL=1.DO
C ELSE
C COEFL=-COEFL*((M(LL)-FRL+1.DO)/FRL)
C END IF
C
C --- START LOOP ON KL ---
C KLMAX=RL*LMI
C
C PRE-COMPUTE THE J.C.P. MILLER COEFFICIENTS OVER KL:
C
C CALL JCPMF(CL,KLMAX,RL,LMI)
C DO 6000 KL=0,KLMAX
C IF(KL.EQ.0) THEN
C POWKL=1.DO
C ELSE
C POWKL=POWKL/BIGK
C END IF
C COEFL=CL(KL)*POWKL
C
C --- START THE VARIABLE-LEVEL NESTED LOOPS
C
C LOOPS=3*LMI
C IF(LOOPS.EQ.0) GOTO 555
C SET UP INDEX OF OUTERMOST LOOP
C INDEX(1)=0
C SET UP THE NON-VARYING UPPER LIMITS
C DO 22 I=1,LMI
C INDUP(3*I-2)=N(I)
C 22 CONTINUE
C
C MARK OUTER-MOST LOOP AS JUST ITERATED
C
C NSUB=1
C
C PERFORM INITIALIZATION CODE FOR LOOPS-FROM JUST-ITERATED LOOP
C INWARD
    
```

```

0002 C
0003 C
0004 C
0005 C
0006 C
0007 C
0008 C
0009 C
0010 C
0011 C
0012 C
0013 C
0014 C
0015 C
0016 C
0017 C
0018 C
0019 C
0020 C
0021 C
0022 C
0023 C
0024 C
0025 C
0026 C
0027 C
0028 C
    
```

```

0079 555 AO=RO+RL/BIGK
0080 180=KO+KL
0081 IF(LOOPS.GT.O) THEN
0082 DO 60 I=1,LMI
0083 AO=AO+INDEX(3*I-1)+(INDEX(3*I-2)-INDEX(3*I-1))/BIGK
0084 180=180+INDEX(3*I)
0085 CONTINUE
0086 END IF
0087 PART=GRAL(AO,180,JSUB(1))
C --- ADD A TERM TO THE OVERALL SUM ---
C
0088 PROD=COEF*CO(KO)*COEF*COEFK
0089 IF(LOOPS.GT.O) THEN
0090 DO 70 I=1,LMI
0091 PROD=PROD*COEF(I)*COEFK(I)*COEF(I)
0092 CONTINUE
0093 END IF
0094 PROD=PROD*PART
0095 SUM=SUM+PROD
C --- ITERATE THE LOOPS ---
C
0096 IF(LOOPS.GT.O) THEN
0097 DO 80 NSUB=LOOPS,1,-1
0098 INDEX(NSUB)=INDEX(NSUB)+1
0099 IF(INDEX(NSUB).LE.INDUP(NSUB)) THEN
0100 GOTO 1000
0101 END IF
0102 CONTINUE
C --- OUTERMOST VARIABLE-NEST LOOP REACHED ITS UPPER LIMIT, ---
C --- DONE WITH THE VARIABLE-NEST LOOPS ---
C
0103 END IF
C --- ITERATE THE NORMAL DO LOOPS ---
C
6000 CONTINUE
7000 CONTINUE
8000 CONTINUE
9000 CONTINUE
PROB=1.DO-SUM
RETURN
END
    
```

```

C C FIRST, RESET INDICES OF ALL LOOPS WITHIN ONE JUST ITERATED
C
1000 DO 25 I=NSUB+1,LOOPS
0055 INDEX(I)=0
0056 25 CONTINUE
C
C SECOND, PERFORM FRONT-OF-LOOP CODE FOR EACH LOOP FROM ITERATED
C LOOP ON INWARD
C
DO 30 J=NSUB,LOOPS
0058 IRKP=MOD(1,3)
0059 MAP 1,2,3 INTO 1, 4,5,6 INTO 2; ETC. FOR NESTING LEVEL
0060 NEST=(1+2)/3
0061 IF(IRKP.EQ.LOOPR) THEN
C WE ARE ITERATING AN R-LOOP
0062 IF(INDEX(I).EQ.O) THEN
C FIRST TIME THROUGH:
0063 COEF(NEST)=1.DO
0064 ELSE
C NOT FIRST TIME THROUGH:
0065 UPDATE THE COEFFICIENT
0066 COEF(NEST)=-(COEF(NEST)/FKL1)*
((N(NEST)-INDEX(I)+1.DO)/INDEX(I))
END IF
C SET UP VARIABLE UPPER LIMITS FOR THE ASSOCIATED
C K LOOP
0067 INDUP(I+1)=INDEX(I)
0068 ELSE IF(IRKP.EQ.LOOPK) THEN
C WE ARE ITERATING A K LOOP
0069 IF(INDEX(I).EQ.O) THEN
C FIRST TIME THROUGH:
0070 INITIALIZE THE COEFFICIENT
0071 COEF(NEST)=1.DO
ELSE
C NOT FIRST TIME THROUGH:
0072 UPDATE THE COEFFICIENT
0073 COEF(NEST)=COEF(NEST)*
((INDEX(I-1)-INDEX(I)+1.DO)/INDEX(I))
END IF
C SET UP THE UPPER LIMIT FOR THE ASSOCIATED P LOOP
0074 INDUP(I+1)=ICAPP(LL,NEST,INDEX(I-1),INDEX(I))
0075 ELSE IF(IRKP.EQ.LOOPP) THEN
C WE ARE ITERATING A P LOOP
0076 COEF(NEST)=DEE(LL,NEST,INDEX(I-2),INDEX(I-1),INDEX(I),BIGK)
0077 END IF
0078 30 CONTINUE
C --- DO THE INTEGRAL ---
C
    
```

```

0001 C THE FUNCTION D(P) FOR PBNJ/RHFSK
0002 C
0003 C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0004 DIMENSION G(0:8)
0005 IF(LL.EQ.2) THEN
0006 DEE=DXI(BIGK,IR-K)
0007 IF(MOD(K,2).EQ.1) DEE=-DEE
0008 END IF
0009 ELSE IF(LL.EQ.3) THEN
0010 IF(NEST.EQ.1) THEN
0011 DEE=DXI(BIGK*BIGK,IR-K)*DBINCO(K,IP)*
0012 $ DXI(2.DO*BIGK-1.DO,K-IP)*DXI(BIGK-1.DO,IP)
0013 ELSE IF(NEST.EQ.2) THEN
0014 DEE=DBINCO(IR-K,IP)*DXI(BIGK*(BIGK-2.DO),IR-K-IP)*
0015 $ DXI(BIGK-1.DO,IP)
0016 END IF
0017 ELSE IF(LL.EQ.4) THEN
0018 IF(NEST.EQ.1) THEN
0019 DENOM=3.DO*BIGK*BIGK-3.DO*BIGK+1.DO
0020 COEF1=2.DO*BIGK*BIGK-3.DO*BIGK+1.DO)/DENOM
0021 G(0)=1.DO
0022 G(1)=K*COEF1
0023 BK1=BIGK-1.DO
0024 COEF2=0.500*BK1*BK1/DENOM
0025 DO 41 N=2,IP
0026 G(N)=((K+1-N)*COEF1*G(N-1) + (2*N+2-N)*COEF2*G(N-2))/N
0027 CONTINUE
0028 DEE=DXI(BIGK,3*(IR-K))*DXI(DENOM,K)*G(IP)
0029 IF(MOD(K,2).EQ.1) DEE=-DEE
0030 ELSE IF(NEST.EQ.2) THEN
0031 IUP=MINO(IP,K)
0032 ILOM=MAXO(0,IP-IR+K)
0033 BK3=BIGK-3.DO
0034 TK1=3.DO*BIGK-1.DO
0035 BASE=BK3*BIGK/TK1
0036 SUMQ=0.DO
0037 DO 42 IO=ILOM,IUP
0038 $ SUMQ=SUMQ+DBINCO(IR-K,IP-IO)*DBINCO(K,IO)
0039 $ *DXI(BK3,IR-K-IP)*DXI(TK1,K)*DXI(BASE,IO)
0040 CONTINUE
0041 DEE=SUMQ*DXI(BIGK,IR+IR-K-IP)*DXI(BIGK-1.DO,IP)
0042 ELSE IF(NEST.EQ.3) THEN
0043 G(0)=1.DO
0044 DENOM=DXI(BIGK,3)-3.DO*BIGK*BIGK+3.DO*BIGK
0045 COEF1=(BIGK*BIGK-3.DO*BIGK+2.DO)/DENOM
0046 G(1)=(IR-K)*COEF1
0047 BK1=BIGK-1.DO
0048 COEF2=0.500*BK1*BK1/(BIGK*DENOM)

```

```

0047 DO 43 N=2,IP
0048 $ G(N)=((IR-K+1-N)*COEF1*G(N-1)+
0049 $ (2*(IR-K+1-N)*COEF2*G(N-2)))/N
0050 CONTINUE
0051 DEE=DXI(DENOM,IR-K)*G(IP)
0052 IF(MOD(K,2).EQ.1) DEE=-DEE
0053 END IF
0054 RETURN
0055 END

```

```

0001      FUNCTION ICAPP(LL,NEST,IR,K)
0002      C COMPUTE UPPER SUMMATION LIMIT PL
0003      C
0004      IF(LL.LE.2) THEN
0005          ICAPP=0
0006      ELSE IF(LL.EQ.3) THEN
0007          IF(NEST.EQ.1) THEN
0008              ICAPP=K
0009          ELSE IF(NEST.EQ.2) THEN
0010              ICAPP=IR-K
0011          END IF
0012      ELSE IF(LL.EQ.4) THEN
0013          IF(NEST.EQ.1) THEN
0014              ICAPP=2*K
0015          ELSE IF(NEST.EQ.2) THEN
0016              ICAPP=IR
0017          ELSE IF(NEST.EQ.3) THEN
0018              ICAPP=2*(IR-K)
0019          END IF
0020      END IF
0021      RETURN
0022      END
    
```

```

0001      DOUBLE PRECISION FUNCTION GRAL(AO,IBO,LI)
0002      C THE INNER INTEGRAL FOR CONDITIONAL ERROR PROBABILITY
0003      C
0004      IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0005      COMMON /RRRCOM/ RHOT,RHOW
0006      COMMON /PARMS/ BIGK, LL
0007      A1=1.DO+AO
0008      IF(LL.GT.0 .AND. LI.LT.LI) THEN
0009          BK1=BIGK-1.DO
0010          BASE=BK1/(1.DO+BIGK*AO)
0011          ARG1=(LI-LL)*RHOW/A1
0012          ARG2=-LI*RHOT*BIGK/(1.DO+BIGK*AO)
0013          POWER=1.DO
0014          SUM=0.DO
0015          DO 100 IQ=0,IBO
0016              IF(IQ.GT.0) POWER=POWER*BASE
0017              PH=DSQRT(POWER)
0018              SUP=GMLGPM(IQ+LI-1,IBO-IQ,ARG1,PH)*GM_GPM(LI-1,IQ,ARG2,PH)+SUM
0019              CONTINUE
0020              F=1.DO
0021              DO 110 I=1,IBO
0022                  F=F*(1/A1)
0023              CONTINUE
0024              F=(F/DXI(A1,LI-LL))/DXI(1.DO+BIGK*AO,LI)
0025              GRAL=SUM*F*DEXP((ARG1+ARG2)*AO)
0026              ELSE IF(LI.EQ.0) THEN
0027                  ARG1=-LL*RHOW/A1
0028                  START=DEXP(ARG1*AO-LL*DL0G(A1))
0029                  DO 200 I=1,IBO
0030                      START=START*(1/A1)
0031                  CONTINUE
0032                  GRAL=GMLGPM(LL-1,IBO,ARG1,START)
0033              ELSE IF(LI.EQ.LI) THEN
0034                  A2=1.DO+BIGK*AO
0035                  ARG1=-LL*RHOT/A2
0036                  BASE=BIGK/A2
0037                  START=DEXP(ARG1*BIGK*AO-LL*DL0G(A2))
0038                  DO 300 J=1,IBO
0039                      START=START*(1*BASE)
0040                  CONTINUE
0041                  GRAL=GMLGPM(LL-1,IBO,ARG1,START)
0042              END IF
0043      END
    
```

0001 C DOUBLE PRECISION FUNCTION GNLGPM(ALFA,N,X,PREMUL)
 C GENERALIZED LAGUERRE POLYNOMIALS WITH PREMULTIPLICATION BY
 C A WEIGHTING FACTOR (TO DEFER THE INEVITABLE OVERFLOWS)
 C

```

0002     IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003     TERM=OBINCO(N+IALFA,N)*PREMUL
0004     SUM=TERM
0005     IF(N.EQ.0) GOTO 200
0006     DO 100 M=1,N
0007     TERM=-TERM*((X/M)*((N-M+1.DO)/(IALFA+M)))
0008     SUM=SUM+TERM
0009     CONTINUE
0010     GNLGPM=SUM
0011     RETURN
0012     END
    
```

0001 C SUBROUTINE JCPMF(C,KMAX,R,LMI)
 C COMPUTE J.C.P. MILLER COEFFICIENTS DIVIDED BY K!
 C

```

0002     IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003     INTEGER R
0004     DIMENSION C(0:KMAX)
0005     C(0)=1.DO
0006     IF(LMI.EQ.0) RETURN
0007     DO 100 K=1,KMAX
0008     SUM=0.DO
0009     MUP=MIND(K,LMI)
0010     BC=1.DO
0011     DO 90 N=1,MUP
0012     BC=BC*((K-M+1.DO)/N)
0013     SUM=SUM+BC*((R+1.DO)*N-K)*C(K-M)
0014     CONTINUE
0015     C(K)=SUM/K
0016     CONTINUE
0017     FAC=1.DO
0018     DO 110 K=1,KMAX
0019     FAC=FAC*K
0020     C(K)=C(K)/FAC
0021     CONTINUE
0022     RETURN
0023     END
    
```

```

0001 SUBROUTINE MLINIT(LMAT, LLOW, LMAXC, LMAXR)
C
C THIS SUBROUTINE INITIALIZES A "MATRIX DO-LOOP" STRUCTURE
C DEFINED BY THE FOLLOWING PSEUDO-FORTRAN CODE:
C DO 100 LMAT(1,1)=LLOW(1,1), LUP(1,1), LINC(1,1)
C
C
C DO 100 LMAT(LMAXC,1)=LLOW(LMAXC,1), LUP(LMAXC,1), LINC(LMAXC,1)
C DO 100 LMAT(1,2)=LLOW(1,2), LUP(1,2), LINC(1,2)
C
C
C DO 100 LMAT(LMAXC,2)=LLOW(LMAXC,2), LUP(LMAXC,2), LINC(LMAXC,2)
C DO 100 LMAT(1,LMAXR)=LLOW(1,LMAXR), LUP(1,LMAXR), LINC(1,LMAXR)
C
C
C DO 100 LMAT(LMAXC,LMAXR)=LLOW(LMAXC,LMAXR), LUP(LMAXC,LMAXR),
C LINC(LMAXC,LMAXR)
C
C (STATEMENTS IN RANGE OF LOOP)
C
C 100 CONTINUE
C
C THE COMPANION ROUTINE MLITER HANDLES THE LOOP CONTROL AT THE
C CONTINUE STATEMENT IN THE ABOVE STRUCTURE
C
C USAGE:
C LOGICAL*1 GO
C DIMENSION LMAT(LMAXC,LMAXR), LLOW(LMAXC,LMAXR), LUP(LMAXC,LMAXR)
C DIMENSION LINC(LMAXC,LMAXR)
C (INITIALIZE MATRIX LLOW TO STARTING VALUES OF THE NESTED LOOPS)
C (INITIALIZE MATRIX LUP TO STOPPING VALUES OF THE NESTED LOOPS)
C (INITIALIZE MATRIX LINC TO INCREMENTS OF THE LOOPS)
C CALL MLINIT(LMAT, LLOW, LMAXC, LMAXR)
C 100 CONTINUE
C
C ( STATEMENTS IN RANGE OF LOOPS )
C
C CALL MLITER(LMAT, LLOW, LUP, LINC, LMAXC, LMAXR, GO)
C IF(GO)GOTO 100
C
C WHERE
C LMAT = ARRAY FOR STORAGE OF LOOP INDICES. LMAT(1,1) IS THE
C OUTER-MOST LOOP; LMAT(LMAXC,LMAXR), THE INNER-MOST LOOP.
C LLOW = ARRAY FOR STORAGE OF LOOP STARTING VALUES, IN SAME
C SEQUENCE AS LMAT
C LUP = ARRAY FOR STORAGE OF LOOP FINAL VALUES, IN SAME
C SEQUENCE AS LMAT
C LINC = ARRAY FOR STORAGE OF LOOP INCREMENTS, IN SAME
C SEQUENCE AS LMAT
C LMAX = NUMBER OF LOOPS NESTED
C GO = LOGICAL VARIABLE, .TRUE. IF JUMP BACK TO BEGINNING OF
C STATEMENTS IN THE RANGE OF THE LOOP SHOULD OCCUR,
C .FALSE. OTHERWISE (I.E. OUTER-MOST LOOP TERMINATED)
    
```

```

0001
C
C PROGRAMMER: ROBERT H. FRENCH DATE: 10 MARCH 1986
C
C DIMENSION LMAT(LMAXC,LMAXR), LLOW(LMAXC,LMAXR)
C DO 1 N=1,LMAXR
C DO 1 M=1,LMAXC
C LMAT(M,N)=LLOW(M,N)
C CONTINUE
C RETURN
C END
    
```

```

0001 SUBROUTINE MLITER(LMAT, LLOW, LUP, LINC, LMAXC, LMAXR, GO)
C
C LOOP ITERATION LOGIC FOR A "MATRIX DO-LOOP"
C
C SEE DETAILED COMMENTS IN SUBROUTINE MLINIT FOR USAGE AND
C PARAMETER DEFINITIONS
C
C PROGRAMMER: ROBERT H. FRENCH DATE: 10 MARCH 1986
C
C LOGICAL*1 GO
C DIMENSION LMAT(LMAXC,LMAXR), LLOW(LMAXC,LMAXR), LUP(LMAXC,LMAXR)
C DIMENSION LINC(LMAXC,LMAXR)
C GO=.TRUE.
C DO 100 NDX=1,LMAXR
C NSUB=LMAXR+1-NDX
C DO 100 NDX=1,LMAXC
C MSUB=LMAXC+1-NDX
C LMAT(MSUB,NSUB)=LMAT(MSUB,NSUB)+LINC(MSUB,NSUB)
C IF((LINC(MSUB,NSUB).GE.0.AND.LMAT(MSUB,NSUB).LE.LUP(MSUB,NSUB))
C .OR.
C (LINC(MSUB,NSUB).LT.0.AND.LMAT(MSUB,NSUB).GE.LUP(MSUB,NSUB)))
C RETURN
C LMAT(MSUB,NSUB)=LLOW(MSUB,NSUB)
C CONTINUE
C GO=.FALSE.
C RETURN
C END
    
```

APPENDIX G
COMPUTER PROGRAM FOR
PLOTTING GRAPHICAL RESULTS
FOR THE
ANY-CHANNEL-JAMMED ADAPTIVE GAIN CONTROL RECEIVER

The following pages contain the source code listing for the FORTRAN-77 program used to produce the plotted graphical results for the any-channel-jammed adaptive gain control receiver for FH/RMFSK. The program makes use of the Hewlett-Packard Industry Standard Plotting Package (ISPP) to drive an HP-7470A plotter.

With minor modifications in annotations and file names, this program will serve to plot results for all other receivers. For brevity, the other versions of the plotting program are not included in this report.

```

0001      PROGRAM RINHOPP
C THIS PROGRAM PLOTS THE ERROR PROBABILITY FOR RANDOM M-ARY
C FSK/FH WITH MULTIPLE HOPS/BIT IN THE PRESENCE OF PARTIAL-BAND
C NOISE JAMMING
C
C ANALYSIS: L. E. MILLER, R. H. FRENCH
C PROGRAM: R. H. FRENCH
C
C
C V 3.0.0 - PLOTS
C
C
C      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
C      PARAMETER (NJ=126)
C      PARAMETER (NJZ=NJ+2)
C      REAL*4 PRLOG(NJZ), DBSJR(NJZ), RTEMP
C      CHARACTER*13 FNAME
C COMMON /INPUTS/ DEBNOL(5),LLIST(4),NSLOTS,
C      GAMLST(31),K,MH
C COMMON /SIZE/ NO,NL,NG
C COMMON /SIZE/ NO,NL,NG
C CALL JSLSGO
C CALL GET
C PRLOG(NJ+1)=.5
C PRLOG(NJ+2)=0.625
C DBSJR(NJ+1)=0.
C DBSJR(NJ+2)=-10.
C CALL SETUP(KODE)
C IF(KODE.NE.0) STOP 1
C DO 900 IL=1,NL
C I.L=LLIST(IL)
C DO 600 IO=1,NO
C IOOUT=DEBNOL(IO)
C
C DRAW THE BOX
C
C CALL SLOW
C CALL PLOT(1.25,6.0,-3)
C CALL LOGAX1(0.,0.,'PROBABILITY OF BIT ERROR',.8,.0,-5.,5)
C      LEN('PROBABILITY OF BIT ERROR'),.8,.0,-5.,5)
C CALL AXISI(0.,0.,'BIT ENERGY TO JAMMING DENSITY RATIO (dB)',
C      -LEN('BIT ENERGY TO JAMMING DENSITY RATIO (dB)'),
C      5.,270.,0.,10.)
C
C COMPLETE THE BOX
C CALL AXISI(8.,0.,-1.0,5.,270.,0.,10.)
C CALL LOGAX1(0.,-5.,1.0,8.,0.,-5.,5)
C ANNOTATE M, L, EB/MG
C CALL SYMBOL(7.5,-2.5,0.14,'M=',270.,2)
C RTEMP=MH
C CALL NUMBER(999.,999.,0.14,RTEMP,270.,-1)
C CALL SYMBOL(7.25,-2.5,0.14,'E',270.,1)
C CALL SYMBOL(7.215,999.,0.09,'b',270.,1)
C CALL SYMBOL(7.25,999.,0.14,'/M',270.,2)
C CALL SYMBOL(7.215,999.,0.09,'0',270.,1)
C CALL SYMBOL(7.25,999.,0.14,'s',270.,1)
C RTEMP=DEBNOL(10)
C CALL NUMBER(999.,999.,0.14,RTEMP,270.,6)
C CALL SYMBOL(999.,999.,0.14,'dB',270.,3)
C RTEMP=NSLOTS
C CALL NUMBER(7.0,-2.5,0.14,RTEMP,270.,-1)
C CALL SYMBOL(999.,999.,0.14,'SLOTS',270.,6)
C CALL SYMBOL(6.75,-2.5,0.14,'ACJ-AGC',270.,7)
C CALL PERUP
C DO 700 IG=1,NG
C GAMMA=GAMLST(IG)
C IOOUT=GAMMA*1000.00+0.500
C
C OPEN DATA FILE
C
C
C      WRITE(FNAME,730) MH,LL,IOOUT,IGOUT
C      FORMAT('A',11,11,12.2,14.4,'.DAT')
C OPEN(UNIT=4,FILE=FNAME,STATUS='OLD',FORM='UNFORMATTED',
C      ERR=740)
C GOTO 750
C 740 WRITE(5,741) FNAME
C 741 FORMAT(1X,A13)
C STOP 'FILE NOT FOUND'
C READ(4) MHIM,LLIN,DEBNOL,ISLOTS,GAMMAI,
C      (DBSJR(IN),IN=1,NJ)
C      (PRLOG(IN),IN=1,NJ)
C IF(MHIM.NE.MH.OR.LLIN.NE.LL.OR.DEBNOL.NE.DEBNOL(10)
C      .OR.ISLOTS.NE.NSLOTS.OR.GAMMAI.NE.GAMMA) THEN
C WRITE(5,741) FNAME
C STOP 'FILE CONTENTS ERROR'
C END IF
C
C INTERPOLATE TO EDGE OF THE GRAPH
C
C DO 790 I=1,NJ
C NPTS=1
C IF(PRLOG(I).GE.-5.) GOTO 790
C DY=PRLOG(I)-PRLOG(I-1)
C DX=DBSJR(I)-DBSJR(I-1)
C PART=(-5.)-PRLOG(I-1)
C SLOPE=DX/DY
C DBSJR(I)=DBSJR(I-1)+SLOPE*PART
C PRLOG(I)=-5.
C GOTO 791
C 790 CONTINUE
    
```

```

0002      COMMON /SIZE/ NO,NL,NG
0003      CALL JSLSGO
0004      CALL GET
0005      PRLOG(NJ+1)=.5
0006      PRLOG(NJ+2)=0.625
0007      DBSJR(NJ+1)=0.
0008      DBSJR(NJ+2)=-10.
0009      CALL SETUP(KODE)
0010      IF(KODE.NE.0) STOP 1
0011      DO 900 IL=1,NL
0012      I.L=LLIST(IL)
0013      DO 600 IO=1,NO
0014      IOOUT=DEBNOL(IO)
0015
0016      C DRAW THE BOX
0017
0018      C CALL SLOW
0019      C CALL PLOT(1.25,6.0,-3)
0020      C CALL LOGAX1(0.,0.,'PROBABILITY OF BIT ERROR',.8,.0,-5.,5)
0021      C      LEN('PROBABILITY OF BIT ERROR'),.8,.0,-5.,5)
0022      C CALL AXISI(0.,0.,'BIT ENERGY TO JAMMING DENSITY RATIO (dB)',
0023      C      -LEN('BIT ENERGY TO JAMMING DENSITY RATIO (dB)'),
0024      C      5.,270.,0.,10.)
0025
0026      C COMPLETE THE BOX
0027      C CALL AXISI(8.,0.,-1.0,5.,270.,0.,10.)
0028      C CALL LOGAX1(0.,-5.,1.0,8.,0.,-5.,5)
0029      C ANNOTATE M, L, EB/MG
0030      C CALL SYMBOL(7.5,-2.5,0.14,'M=',270.,2)
0031      C RTEMP=MH
0032      C CALL NUMBER(999.,999.,0.14,RTEMP,270.,-1)
0033      C CALL SYMBOL(7.25,-2.5,0.14,'E',270.,1)
0034      C CALL SYMBOL(7.215,999.,0.09,'b',270.,1)
0035      C CALL SYMBOL(7.25,999.,0.14,'/M',270.,2)
0036      C CALL SYMBOL(7.215,999.,0.09,'0',270.,1)
0037      C CALL SYMBOL(7.25,999.,0.14,'s',270.,1)
0038      C RTEMP=DEBNOL(10)
0039      C CALL NUMBER(999.,999.,0.14,RTEMP,270.,6)
0040      C CALL SYMBOL(999.,999.,0.14,'dB',270.,3)
0041      C RTEMP=NSLOTS
0042      C CALL NUMBER(7.0,-2.5,0.14,RTEMP,270.,-1)
0043      C CALL SYMBOL(999.,999.,0.14,'SLOTS',270.,6)
0044      C CALL SYMBOL(6.75,-2.5,0.14,'ACJ-AGC',270.,7)
0045      C CALL PERUP
0046      C DO 700 IG=1,NG
0047      C GAMMA=GAMLST(IG)
0048      C IOOUT=GAMMA*1000.00+0.500
    
```



```
0044 READ(5,3)LLIST(IN)
0045 IF(LLIST(IN).EQ.0)LLIST(IN)=4
0046 CONTINUE
0047 WRITE(5,23)
0048 FORMAT(' HOPPING SLOTS? [2400]: ', $)
0049 READ(5,24)NSLOTS
0050 FORMAT(I5)
0051 IF(NSLOTS.EQ.0)NSLOTS=2400
0052 WRITE(5,26)
0053 FORMAT(' HOW MANY GAMMA? [10]: ', $)
0054 READ(5,3)NG
0055 IF(NG.EQ.0)NG=10
0056 DO 31 IN=1,NG
0057 WRITE(5,29)IN,DG(IN)
0058 FORMAT(' GAMMA(',I2,') [',I1P8.1,']: ', $)
0059 READ(5,30)GAMLST(IN)
0060 FORMAT(D15.8)
0061 IF(GAMLST(IN).EQ.0.00)GAMLST(IN)=DG(IN)
0062 CONTINUE
0063 RETURN
0064 END
```

APPENDIX H
COMPUTER PROGRAM FOR
CLIPPER RECEIVER WITH M=2 AND L=2

The following pages contain the source code listing for the FORTRAN-77 program used for numerical computations of the performance of the clipper receiver for the case of M=2 and L=2 with the jamming fraction $\gamma=q/N$ as a parameter.


```

0001 C SUBROUTINE GET(MJ,START,DBINC)
0002 C INTERACTIVE INPUT OF PARAMETERS FOR RUN
0003 C
0004 C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0005 C CHARACTER*9 FIELD,BLANK9
0006 C COMMON /INPUTS/ DEBNOL(3),NSLOTS,GAMMST(10),K,M
0007 C COMMON /SIZE/ NO,NG
0008 C DEFAULT LISTS TEMPORARILY NEEDED ARE IN SHARED STORAGE WITH
0009 C THE LARGE CONVOLUTION WORKING ARRAYS
0010 C COMMON /SHARE/ DG(10),DSNR(3,4)
0011 C DATA BLANK9/
0012 WRITE(5,33)
0013 FORMAT(' BITS/SYMBOL (K) [2]: ',S)
0014 READ(5,3)K
0015 IF(K.EQ.0)K=2
0016 MM=2**K
0017 WRITE(5,2)
0018 FORMAT(' HOW MANY EB/MO? [1]: ',S)
0019 READ(5,3)NO
0020 IF(NO.EQ.0)NO=1
0021 DO 7 IN=1,NO
0022 DO=DSNR(IN,K)
0023 WRITE(5,5)IN,DO
0024 FORMAT(' EB/MO(' ,I2,') [ ,F9.6,']: ',S)
0025 READ(5,6)FIELD
0026 FORMAT(A9)
0027 IF(FIELD.EQ.BLANK9) THEN
0028 DEBNOL(IN)=DO
0029 ELSE
0030 DECODE(9,61,FIELD)DEBNOL(IN)
0031 END IF
0032 FORMAT(F9.6)
0033 CONTINUE
0034 C AUTOMATICALLY TAKE 2400 SLOTS
0035 NSLOTS=2400
0036 WRITE(5,26)
0037 FORMAT(' HOW MANY GAMMA? [10]: ',S)
0038 READ(5,3)NG
0039 IF(NG.EQ.0)NG=10
0040 DO 31 IN=1,NG
0041 WRITE(5,29)IN,DG(IN)
0042 FORMAT(' GAMMA(' ,I2,') [ ,1P8.1,']: ',S)
0043 READ(5,30)GAMMST(IN)
0044 FORMAT(D15.8)
0045 IF(GAMMST(IN).EQ.0.D0)GAMMST(IN)=DG(IN)
0046 CONTINUE
0047 WRITE(5,39)
0048 FORMAT(' HOW MANY EB/MJ? [51]: ',S)
0049 READ(5,34,ERR=38) NJ
0050 FORMAT(I3)

```

```

0073 OPEN(UNIT=4, FILE=FNAME, STATUS='NEW', FORM='UNFORMATTED')
0074 WRITE(4) MM,DEBNOL(10),NSLOTS,GAMMA,TAU,TAU2,PEOO
0075 CLOSE(UNIT=4)
0076 WRITE(GNAME,735) MM,IGOUT
0077 FORMAT('EV',I1,'2',I4,'.DAT')
0078 OPEN(UNIT=3, FILE=GNAME, STATUS='OLD', FORM='UNFORMATTED',
0079 READONLY,ERR=770)
0080 WRITE(5,3939)
0081 FORMAT(' READING EVENT FILE')
0082 READ(3) D,IDSUB,MUSED,GOOD
0083 CLOSE(UNIT=3)
0084 GOTO 777
0085 C MUST CREATE EVENT PROBABILITIES
0086 CONTINUE
0087 WRITE(5,3938)
0088 FORMAT(' CREATING EVENT FILE')
0089 CALL GENPTE(MM,NO,NSLOTS,GOOD,A,TASUB,C,ICSUB,
0090 D,IDSUB,MUSED)
0091 OPEN(UNIT=3, FILE=GNAME, STATUS='NEW', FORM='UNFORMATTED')
0092 WRITE(3) D,IDSUB,MUSED,GOOD
0093 IF(.NOT.GOOD) GOTO 700
0094 DO 600 IJ=JJ,NJ
0095 WRITE(5,601) IJ
0096 FORMAT(' IJ=',I3)
0097 DEBNJ=START+(IJ-1)*DBINC
0098 DEBJR(IJ)=DEBNJ
0099 R=10.D0*(DEBNJ/10.D0)
0100 RHOTS=GAMMA*R*EBNO/(GAMMA*R*EBNO)
0101 RHOT=RHOTS/2.D0
0102 RHOT2=RHOT*2.D0
0103 RHOT4=4.D0*RHOT
0104 RHOT8=RHOT*8
0105 CALL PSUBE(MM,PESYM,D,IDSUB,MUSED,PRERR,IPSUB,PEOO)
0106 PE=WORBIT*PESYM
0107 WRITE(6,666) DEBJR(IJ),PE
0108 FORMAT('IX',F7.3,5X,1P12.5)
0109 PRLOG(IJ)=DL0G10(PE)
0110 OPEN(UNIT=4, FILE=FNAME, STATUS='OLD', ACCESS='APPEND',
0111 FORM='UNFORMATTED')
0112 WRITE(4) DEBJR(IJ), PRLOG(IJ)
0113 CLOSE(UNIT=4)
0114 CONTINUE
0115 OPEN(UNIT=4, FILE=FNAME, STATUS='NEW', FORM='UNFORMATTED')
0116 WRITE(4) MM,2,DEBNOL(10),NSLOTS,GAMMA,DEBJR,PRLOG
0117 CLOSE(UNIT=4)
0118 CONTINUE
0119 CONTINUE
0120 STOP 0
0121 END

```

```

0047 IF(NJ.EQ.0) NJ=51
0048 IF(NJ.LT.0 .OR. NJ.GT.51) GOTO 32
0049 WRITE(5,41)
0050 FORMAT(' STARTING VALUE FOR EB/NJ (DB) (0.1: ',5)
0051 READ(5,42,ERR=40) START
0052 FORMAT(F6.3)
0053 IF(NJ.EQ.1) RETURN
0054 WRITE(5,36)
0055 FORMAT(' DB INCREMENT FOR EB/NJ (1.0): ',5)
0056 READ(5,37,ERR=35) DBINC
0057 FORMAT(F6.3)
0058 IF(DBINC.EQ.0.0) DBINC=1.0
0059 RETURN
0060 END
  
```

```

0001 SUBROUTINE PSUBE(M,PE,D,IDSUB,MUSED,PRERR,IPSUB,PEOO)
0002 C COMPUTE UNCONDITIONAL ERROR PROBABILITY
0003 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0004 INTEGER JAM(4),LUP(4),JSUB(4)
0005 LOGICAL*1 GO,MONI,STORE
0006 VIRTUAL PRERR(625),IPSUB(625)
0007 COMMON /SHARE2/ LOM(4),LINC(4)
0008 COMMON /DENPAR/ BIGK, AAB, BAB, L.JAM,
0009 .TAU, TALK, TALK2
0010 COMMON /PARDEW/ RHOM, RHOT, RHOM2, RHOT2, RHOM4, RHOT4,
0011 RHOMB, RHOTB, RHOMH, RHOTH
0012 DATA STORE/.TRUE./
0013 PE=0.00
0014 NPS=0
0015 DO 10 I=1,M
0016 LUP(I)=2
0017 JSUB(I)=0
0018 CONTINUE
0019 CALL LOGN(M,LOM,LUP,JSUB,ISUB)
0020 CALL PUTIN(PEO,PRERR,IPSUB,MPS,625,ISUB,KODE,STORE)
0021 IF(KODE.NE.0) STOP 'PRERR FULL'
0022 JAMI=-1
0023 C START VECTOR-INDEXED LOOP ON JAMMING EVENTS
0024 CALL VLINT(JAM,LOM,M)
0025 CONTINUE
0026 IF(JAMI.NE.JAM(1)) THEN
0027 JAMI=JAM(1)
0028 END IF
0029 CALL EVENT(M,JAM,PIE,D,IDSUB,MUSED)
0030 C BYPASS CONDITIONAL ERROR PROBABILITY CALCULATION IF EVENT
0031 PROBABILITY IS ZERO. THIS SAVES MUCH TIME.
0032 IF(PIE.EQ.0.0)GOTO 101
0033 C SINCE JAMMING PROBABILITIES DEPEND ONLY ON NO. OF CHANNELS
0034 C JAMMED AND NOT THE ARRANGEMENT OF THE CHANNELS, WE CAN SORT
0035 C THE NON-SIGNAL CHANNELS INTO ASCENDING NUMBERS OF HOPS JAMMED.
0036 DO 111 I=1,M
0037 JSUB(I)=JAM(I)
0038 CONTINUE
0039 IF(M.EQ.2) GOTO 199
0040 DO 110 I=2,M-1
0041 DO 120 J=I+1,M
0042 IF(JSUB(J).LT.JSUB(I)) THEN
0043 JTEMP=JSUB(I)
0044 JSUB(I)=JSUB(J)
0045 JSUB(J)=JTEMP
0046 END IF
0047 CONTINUE
0048 120 CONTINUE
0049 110 CONTINUE
0050 199 CONTINUE
  
```

```

0001 $ SUBROUTINE GENPIE(MM,NO,NSLOTS,GOOD,A,IASUB,C,ICSUB,
      D,IDSUB,MUSED)
C SUBROUTINE TO GENERATE EVENT PROBABILITIES
C
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 LOGICAL*1 GO,GOZ,STORE,MOME,GOOD
0004 DIMENSION LUP2(4),LUP3(4)
0005 DIMENSION LUPA(4)
0006 DIMENSION LUPD(4)
0007 DIMENSION LUP1(4)
0008 VIRTUAL A(100),IASUB(100),C(625),ICSUB(625),
      D(625),IDSUB(625)
0009 DIMENSION I(4),II(4),III(4)
C SHARED STORAGE FOR COMMONLY NEEDED CONSTANT ARRAYS
0010 COMMON /SHARE2/ LOW(4),LINC(4)
C SHARED STORAGE FOR: (1) INPUT DEFAULT LISTS,
C (2) CONDITIONAL PROB GEN., AND
C (3) EVENT PROB. GEN. THESE ARE NON-OVERLAPPING USAGES.
0011 COMMON /SHARE/ LUP2,LUP3,LUPD,L1,MUSEA,MUSEC,IERR,
      ISUB,ISUB1,ISUB2,AIN,I,II,III,MM,ADUT,
      CIN,CONUT,DOUT,DJN
      $
0012 DATA I100/100/
0013 DATA LUPA/4*1/
0014 DATA LUP1/4*1/
0015 STORE=.FALSE. => DOWN-T STORE ZERO ELEMENTS OF SPARSE ARRAY.
0016 GOOD=.TRUE.
0017 IF(NO.LE.0) THEN
0018   GOOD=.FALSE.
0019   RETURN
0020 END IF
0021 DO 80 LI=1,MM
0022   LUPD(LI)*2
0023 CONTINUE
C JAMMING PATTERN W/NON-ZERO PROBABILITY ON PER-HOP BASIS
0024 MUSEA=0
C INITIALIZE VECTOR-INDEX LOOP
0025 CALL VLINIT(I,LOW,MM)
0026 CONTINUE
0027 CALL LOGN(MM,LOW,LUPA,1,ISUB)
0028 CALL PRINHOP(I,MM,NO,NSLOTS,AIN)
0029 CALL PUTIM(AIN,A,IASUB,MUSEA,I100,ISUB,IERR,STORE)
0030 IF(IERR.NE.0) STOP 3
C ITERATE VECTOR-INDEX LOOP
0031 CALL VLITER(I,LOW,LUP1,LINC,MM,GO)
0032 IF(GO) GOTO 90
C COMPUTATION STARTS HERE. FIRST COPY A INTO D.
C SINCE ARRAYS ARE A(0:1,0:1,...,0:1) AND D(0:L,0:L,...,0:L)
C THE COPYING MUST BE DONE ON BASIS OF EQUIVALENT LJNEAR
C SUBSCRIPTS RATHER THAN A SIMPLE MOVE OPERATION.
      MUSED=0
0033

```

```

C STORE AS ELEMENT OF A SPARSE ARRAY TO SAVE MEMORY
C EVEN THOUGH WE STORE ZEROS, THE SORTING OF SUBSCRIPTS
C CUTS OUT MANY ELEMENTS.
0002 CALL LOGN(M,LOW,LUP,JSUB,ISUB)
0003 CALL TRY TO FIND CONDITIONAL ERROR PROBABILITY IN STORED ARRAY
0004 CALL LOOKUP(PROB,PRERR,IPSUB,MPS,625,ISUB,STORE,MOME)
0005 C IF IT IS NOT THERE, WE MUST COMPUTE IT
      IF(MOME) THEN
0006   CALL PSEL(JSUB,M,PROB)
0007   ... AND SAVE IT FOR POSSIBLE FUTURE RE-USE
0008   CALL PUTIM(PROB,PRERR,IPSUB,MPS,625,ISUB,KODE,STORE)
      IF(KODE.NE.0) STOP 2
0009 END IF
C SUM UP UNCONDITIONAL ERROR PROBABILITY
0010 PE=PE+PIE*PROB
C ITERATE THE VECTOR-INDEX LOOP
0011 CALL VLITER(JAM,LOW,LUP,LINC,M,GO)
0012 IF(GO) GOTO 100
0013 RETURN
0014 END

```

```

0001 SUBROUTINE EVENT(M,JAM,PIE,D,IDSUB,MUSED)
C SUBROUTINE TO LOOK UP EVENT PROBABILITY FROM STORED ARRAY
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 LOGICAL*1 STORE,MOME
0004 DIMENSION JAM(4),LUP(4)
0005 VIRTUAL D(625),IDSUB(625)
0006 COMMON /SHARE2/ LOW(4),LINC(4)
0007 DATA STORE/.FALSE./
C SET UP ARRAY DESCRIPTION D(0:LL,...,0:LL) WITH M DIMENSIONS
0008 DO 1 I=1,M
0009   LUP(I)*2
0010 CONTINUE
C COMPUTE LINEAR EQUIVALENT SUBSCRIPT FOR JAMMING EVENT
0011 CALL LOGN(M,LOW,LUP,JAM,ISUB)
C LOOK UP THE VALUE, GET 0.DO IF NOT THERE
0012 CALL LOOKUP(PIE,D,IDSUB,MUSED,625,ISUB,STORE,MOME)
0013 RETURN
0014 END

```

```

0001 C SUBROUTINE PUTIN(CIN,C,ICSUB,MUSE,NMAX,K,IERR,STORE)
C THIS SUBROUTINE INSERTS AN ELEMENT INTO A SPARSE ARRAY FOR
C WHICH ONLY THE NON-ZERO ELEMENTS ARE KEPT IN STORAGE IF
C THE SWITCH STORE IS .TRUE.
C
C THE DOUBLE PRECISION VALUE CIN IS STORED AS C(K) WHERE
C THE AUXILIARY ARRAY ICSUB IS USED TO KEEP TRACK OF THE
C SUBSCRIPTS OF THE CORRESPONDING ENTRIES IN C.
C
C USAGE:
C LOGICAL*1 STORE
C DOUBLE PRECISION C,CIN
C VIRTUAL ICSUB(NMAX),C(NMAX)
C CALL PUTIN(CIN,C,ICSUB,MUSE,NMAX,K,IERR,STORE)
C
C WHERE
C CIN = VALUE OF ELEMENT TO STORE
C C = ARRAY IN WHICH NON-ZERO VALUES ARE ACTUALLY STORED
C ICSUB = AUXILIARY ARRAY FOR ACTUAL SUBSCRIPT VALUES
C MUSE = NUMBER OF ELEMENTS OF C CURRENTLY OCCUPIED
C NMAX = SIZE OF ARRAY C
C IERR = ERROR RETURN CODE, 0 IF NO ERROR OR 1 IF THERE IS
C NO ROOM AVAILABLE IN C
C STORE = .TRUE. TO STORE ZEROES EXPLICITLY, ELSE .FALSE.
C NOTE: IF CIN=0 AND THE SUBSCRIPT K IS FOUND IN ICSUB, THEN
C THE ELEMENT IS DELETED BY SHIFTING DOWNWARD ALL
C FOLLOWING ELEMENTS OF THE ARRAY
C
C PROGRAMMER: ROBERT H. FRENCH DATE: 11 JANUARY 1984
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 VIRTUAL ICSUB(NMAX),C(NMAX)
0004 LOGICAL*1 STORE
0005 IERR=0
0006 IF(STORE) GOTO 5
0007 IF(CIN.EQ.0.DO) GOTO 30
0008 IF(MUSE.EQ.0)GOTO 20
0009 DO 10 I=1,MUSE
0010 IF(ICSUB(I).NE.K) GOTO 10
0011 C(I)=CIN
0012 RETURN
0013 CONTINUE
0014 IF(MUSE.LT.NMAX) GOTO 20
0015 IERR=1
0016 RETURN
0017 MUSE=MUSE+1
0018 ICSUB(MUSE)=K
0019 C(MUSE)=CIN
0020 RETURN

```

```

0034 C INITIALIZE VECTOR-INDEX LOOP
0035 CALL VLINIT(I,LOW,MM)
0036 CONTINUE
0037 CALL LOCK(MM,LOW,LUPA,I,ISUB1)
0038 CALL LOCK(MM,LOW,IUPD,I,ISUB2)
0039 CALL LOOKUP(AOUT,A,IASUB,MUSEA,I100,ISUB1,STORE,NOME)
0040 CALL PUTIN(AOUT,D,IDSUB,MUSED,625,ISUB2,IERR,STORE)
0041 C ITERATE VECTOR-INDEX LOOP
0042 CALL VLITER(I,LOW,LUP1,LINC,MM,GO)
0043 IF(GO)GOTO 99
0044 DO 9998 LI=1,1
0045 C ... L-1 CONVOLUTIONS ARE NEEDED ...
0046 C SET UP VECTOR-LOOP UPPER LIMITS FOR THIS CONVOLUTION
0047 DO 125 NN=1,MM
0048 LUP2(MM)=LI
0049 LUP3(MM)=LI+1
0050 MUSEC=0
0051 CALL VLINIT(I,LOW,MM)
0052 CONTINUE
0053 CALL VLINIT(II,LOW,MM)
0054 C LOOK UP ELEMENTS AND PERFORM ONE TERM OF THE CONVOLUTION
0055 CALL LOCK(MM,LOW,IUPD,I,ISUB1)
0056 CALL LOCK(MM,LOW,IUPD,II,ISUB2)
0057 DO 21 NN=1,MM
0058 III(MM)=I(MM)+II(MM)
0059 CONTINUE
0060 CALL LOCK(MM,LOW,IUPD,III,ISUB3)
0061 CALL LOOKUP(AOUT,A,IASUB,MUSEA,I100,ISUB1,STORE,NOME)
0062 CALL LOOKUP(DOUT,D,IDSUB,MUSED,625,ISUB2,STORE,NOME)
0063 CALL LOOKUP(COUT,C,ICSUB,MUSEC,625,ISUB3,STORE,NOME)
0064 CIN=COUT+ADOUT*DOUT
0065 CALL PUTIN(CIN,C,ICSUB,MUSEC,625,ISUB3,IERR,STORE)
0066 IF(IERR.NE.0) STOP 4
0067 CALL VLITER(II,LOW,LUP2,LINC,MM,GO)
0068 IF(GO) GOTO 97
0069 CALL VLITER(I,LOW,LUP1,LINC,MM,GO)
0070 IF(GO) GOTO 98
0071 C COPY C TO D IN SORTED ORDER FOR NEXT ITERATION
0072 MUSED=0
0073 CALL VLINIT(II,LOW,MM)
0074 CONTINUE
0075 CALL LOCK(MM,LOW,IUPD,II,ISUB)
0076 CALL LOOKUP(COUT,C,ICSUB,MUSEC,625,ISUB,STORE,NOME)
0077 DIN=COUT
0078 CALL PUTIN(DIN,D,IDSUB,MUSED,625,ISUB,IERR,STORE)
0079 IF(IERR.NE.0) STOP 5
0080 CALL VLITER(II,LOW,LUP3,LINC,MM,GO)
0081 IF(GO) GOTO 96
0082 CONTINUE
0083 RETURN
0084 END
9998

```

```

0001 SUBROUTINE LOOKUP(COUT,C,ICSUB,N,M,MMAX,K,STORE,NONE)
C THIS SUBROUTINE RETRIEVES AN ELEMENT OF A SPARSE ARRAY WHICH
C HAS BEEN STORED COMPACTLY BY STORING ONLY NON-ZERO ELEMENTS.
C THE ARRAY IS DOUBLE PRECISION.
C USAGE:
C VIRTUAL ICSUB(MMAX), C(MMAX)
C LOGICAL*1 STORE, NONE
C DOUBLE PRECISION COUT
C CALL LOOKUP(COUT,C,ICSUB,N,M,MMAX,K,STORE,NONE)
C WHERE
C COUT = VALUE OF C(K) (OUTPUT FROM SUBROUTINE)
C C = ARRAY USED TO STORE NON-ZERO ELEMENTS
C ICSUB = AUXILIARY ARRAY TO STOP ACTUAL SUBSCRIPTS
C N = NUMBER OF ELEMENTS OF C CURRENTLY IN USE
C MMAX = SIZE OF C
C K = SUBSCRIPT OF SPARSE ARRAY TO LOOK UP
C STORE = .TRUE. IF ZEROS STORED EXPLICITLY, ELSE .FALSE.
C NONE = .FALSE. IF ZEROS NOT STORED OR ZEROS STORED AND
C ELEMENT IS FOUND IN THE STORED ARRAY
C .TRUE. IF ZEROS ARE STORED AND THE ELEMENT IS
C NOT FOUND (OUTPUT QUANTITY)
C
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
C

```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 VIRTUAL ICSUB(MMAX),C(MMAX)
0004 LOGICAL*1 STORE, NONE
0005 NONE=.FALSE.
0006 DO 10 I=1,N
0007 IF(ICSUB(I).NE.K)GOTO 10
0008 COUT=C(I)
0009 RETURN
0010 CONTINUE
0011 IF(STORE) THEN
0012 NONE=.TRUE.
0013 ELSE
0014 COUT=0.
0015 END IF
0016 RETURN
0017 END

```

```

0021 DO 40 I=1,MUSE
0022 J=I
0023 IF(ICSUB(I).EQ.K) GOTO 50
0024 CONTINUE
0025 RETURN
C REMOVE THE ZEROED ELEMENT AND BUMP COUNT OF ENTRIES USED
C
0026 DO 60 I=J,MUSE-1
0027 ICSUB(I)=ICSUB(I+1)
0028 C(I)=C(I+1)
0029 CONTINUE
0030 MUSE=MUSE-1
0031 RETURN
0032 END

```

```

0001 SUBROUTINE VLINIT(LVEC,LLOW,LMAX)
C THIS SUBROUTINE INITIALIZES A "VECTOR DO-LOOP" STRUCTURE
C DEFINED BY THE FOLLOWING PSEUDO-FORTRAN CODE:
C DO 100 LVEC(1)=LLOW(1),LUP(1),LINC(1)
C DO 100 LVEC(2)=LLOW(2),LUP(2),LINC(2)
C :
C :
C DO 100 LVEC(LMAX)=LLOW(LMAX),LUP(LMAX),LINC(LMAX)
C :
C :
C (STATEMENTS IN RANGE OF LOOP)
C 100 CONTINUE
C THE COMPANION ROUTINE VLITER HANDLES THE LOOP CONTROL AT THE
C CONTINUE STATEMENT IN THE ABOVE STRUCTURE
C USAGE:
C LOGICAL*1 GO
C DIMENSION LVEC(LMAX),LLOW(LMAX),LUP(LMAX),LINC(LMAX)
C (INITIALIZE ARRAY LLOW TO STARTING VALUES OF THE NESTED LOOPS)
C (INITIALIZE ARRAY LUP TO STOPPING VALUES OF THE NESTED LOOPS)
C (INITIALIZE ARRAY LINC TO INCREMENTS OF THE LOOPS)
C CALL VLINIT(LVEC,LLOW,LMAX)
C 100 CONTINUE
C :
C : (STATEMENTS IN RANGE OF LOOPS)
C :
C CALL VLITER(LVEC,LLOW,LUP,LINC,LMAX,GO)
C IF(60)GOTO 100
C WHERE
C LVEC = ARRAY FOR STORAGE OF LOOP INDICES. LVEC(1) IS THE
C OUTER-MOST LOOP; LVEC(LMAX) IS THE INNER-MOST LOOP.
C LLOW = ARRAY FOR STORAGE OF LOOP STARTING VALUES, IN SAME
C SEQUENCE AS LVEC
C LUP = ARRAY FOR STORAGE OF LOOP FINAL VALUES, IN SAME
C SEQUENCE AS LVEC
C LINC = ARRAY FOR STORAGE OF LOOP INCREMENTS, IN SAME
C SEQUENCE AS LVEC
C LMAX = NUMBER OF LOOPS NESTED
C GO = LOGICAL VARIABLE. .TRUE. IF JUMP BACK TO BEGINNING OF
C STATEMENTS IN THE RANGE OF THE LOOP SHOULD OCCUR,
C .FALSE. OTHERWISE (I.E. OUTER-MOST LOOP TERMINATED)
C PROGRAMMER: ROBERT H. FRENCH DATE: 11 JANUARY 1984
C
0002 DIMENSION LVEC(LMAX),LLOW(LMAX)
0003 DO 1 N=1,LMAX
0004 LVEC(N)=LLOW(N)
0005 CONTINUE
0006 RETURN
0007 END
  
```

```

0001 SUBROUTINE LCON(NDIM,ILOW,IUP,ISUB,LINER)
C THIS SUBROUTINE COMPUTES THE EQUIVALENT LINEAR SUBSCRIPT FOR
C A MULTIDIMENSIONAL ARRAY OF NDIM DIMENSIONS
C IF THE ARRAY A IS DEFINED AS
C DIMENSION A(ILOW(1):IUP(1),...,ILOW(NDIM):IUP(NDIM))
C AND ISUB(1),...,ISUB(NDIM) IS A SET OF SUBSCRIPTS FOR A,
C THEN THIS SUBROUTINE RETURNS IN LINER THE OFFSET FROM THE
C ORIGIN OF A TO THE ELEMENT A(ISUB(1),...,ISUB(NDIM)), ASSUMING
C THE FIRST SUBSCRIPT VARIES MOST RAPIDLY.
C USAGE:
C DIMENSION ILOW(NDIM),IUP(NDIM),ISUB(NDIM)
C DATA ILOW/lower limits of defined subscripts of array/
C DATA IUP/upper limits of defined subscripts of array/
C ...SET ISUB TO DESIRED SEQUENCE OF SUBSCRIPTS...
C CALL LCON(NDIM,ILOW,IUP,ISUB,LINER)
C WHERE
C NDIM = NUMBER OF DIMENSIONS THE ARRAY HAS
C ILOW = ARRAY OF LOWER SUBSCRIPT ROUNDS
C IUP = ARRAY OF UPPER SUBSCRIPT ROUNDS
C ISUB = ARRAY CONTAINING SUBSCRIPT FOR WHICH LOCATION IS
C TO BE COMPUTED
C LINER = RETURNED VALUE OF OFFSET INTO ARRAY IN MEMORY
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
C
0002 DIMENSION ILOW(NDIM),IUP(NDIM),ISUB(NDIM)
0003 LINER=0
0004 DO 10 I=1,NDIM-1
0005 J=NDIM-I+1
0006 LINER=(LINER+(ISUB(J)-ILOW(J)))*(IUP(J-1)-ILOW(J-1)+1)
0007 CONTINUE
0008 LINER=LINER+ISUB(1)-ILOW(1)
0009 RETURN
0010 END
  
```

```

0001 SUBROUTINE VLITER(LVEC, LLOW, LUP, LINC, LMAX, GO)
C LOOP ITERATION LOGIC FOR A "VECTOR DO-LOOP"
C SEE DETAILED COMMENTS IN SUBROUTINE VLIMIT FOR USAGE AND
C PARAMETER DEFINITIONS
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
C
LOGICAL *I GO
DIMENSION LVEC(LMAX), LLOW(LMAX), LUP(LMAX), LINC(LMAX)
GO = .TRUE.
DO 100 NDX=1, LMAX
  NSUB=LMAX-I-NDX
  LVEC(NSUB)=LVEC(NSUB)+LINC(NSUB)
  IF((LINC(NSUB).GE.0.AND.LVEC(NSUB).LE.LUP(NSUB))
    .OR.(LINC(NSUB).LT.0.AND.LVEC(NSUB).GE.LUP(NSUB))) RETURN
  LVEC(NSUB)=LLOW(NSUB)
100 CONTINUE
GO = .FALSE.
RETURN
END

```

```

0001 SUBROUTINE PRIHOP(I, KM, KQ, KN, AIN)
C THIS SUBROUTINE COMPUTES THE EVENT PROBABILITY FOR ALL
C POSSIBLE JAMMING PATTERNS WITH NON-ZERO PROBABILITY FOR
C L=1 HOP/SYMBOL FOR RMFSK/FH IN PBNJ
C
IMPLICIT DOUBLE PRECISION (A-H, O-Z)
DIMENSION I(4)
AIN=0.00
KJAM=0
DO 1 K=1, KM
  KJAM=KJAM+I(K)
1 CONTINUE
C IF THIS IS AN IMPOSSIBLE CASE, RETURN WITH RESULT = 0.0
IF(KJAM.GT.MIND(KQ, KN)) RETURN
KPMAX=KJAM-1
JPMAX=KM-1
JMAX=MAX0(KPMAX, LPMAX, JPMAX)
PROD=1.00
Q=KQ
DIFF=KN-KQ
DO 100 LOOP=0, JMAX
  F=LOOP
  IF(LOOP.LE.KPMAX) PROD=PROD*(Q-F)
  IF(LOOP.LE.JPMAX) PROD=PROD*(ER-F)
  IF(LOOP.LE.LPMAX) PROD=PROD*(DIFF=Q-F)
100 CONTINUE
AIN=PROD
RETURN
END

```



```

0035 PCOMT=0.00
0036 DO 13 ISECT=1,2
0037 XL=(ISECT-1)*TAU
0038 XU=ISECT*TAU
0039 CALL ADQUAD(XL,XU,CHUNK,DG16,PGRAND,1.0-8,WORK,
$ STACK,HEAP,30,KODE)
0040 CALL TEST2(KODE,10)
0041 PCOMT=PCOMT+CHUNK
0042 13 CONTINUE
C
C DO THE TIE PART OF THE DENSITY
C
C CALL TIES(JSUB,MM,PTIE)
C
C PUT THEM TOGETHER
C
0044 PROB=1.00-PCOMT-PTIE
0045 RETURN
0046 END

```

```

0001 DOUBLE PRECISION FUNCTION PGRAND(BETA)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 INTEGER NCHAN(0:3)
0004 COMMON /SCJAM/ JAMSC
0005 COMMON /JMCWT/ NCHAN
0006 COMMON /DENPAR/ BIGK, AAB, BAB, LJAM,
$ TAU, TAU2, TALK, TALK2
0007 COMMON /PARDEM/ RHOM, RHOT, RHOM2, RHOT2, RHOM4, RHOT4,
$ RHOM8, RHOT8, RHOMH, RHOTH
0008 COMMON /QUES/ Q0, Q1
0009 PROD=1.00
0010 DO 10 I=0,2
0011 IF(NCHAN(I).NE.0) THEN
0012 LJAM=I
0013 X=GL(BETA)
0014 PROD=PROD*DX1(X,NCHAN(I))
0015 END IF
0016 10 CONTINUE
0017 LJAM=JAMSC
0018 Y=PZ1(BETA)
0019 PGRAND=Y*PROD
0020 RETURN
0021 END

```

```

0001 SUBROUTINE ADQUAD(XL,XU,Y,QR,F,TOL,WORK,STACK,HEAP,N,KODE)
C
C ADAPTIVE QUADRATURE ALGORITHM
C XL - LOWER LIMIT OF INTEGRAL (IM)
C XU - UPPER LIMIT OF INTEGRAL (IM)
C Y - VALUE OF INTEGRAL (OUT)
C QR - NAME OF A QUADRATURE RULE SUBROUTINE (IM)
C WITH CALLING SEQUENCE
C CALL QR(XL,XU,F,Y)
C F - NAME OF FUNCTION TO BE INTEGRATED (IM)
C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IM)
C WORK - WORK ARRAY OF SIZE N (IM)
C STACK - SECOND WORK ARRAY OF SIZE M, MUST NOT BE
C SAME ARRAY AS WORK (IM)
C HEAP - THIRD WORK ARRAY OF SIZE N, DISTINCT FROM BOTH WORK AND STACK
C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (IM)
C KODE - ERROR INDICATOR (OUT)
C 0 -- NO ERROR
C 1 -- WORK ARRAYS TOO SMALL
C 2 -- EPS DIVIDED TO ZERO, EITHER ASKING FOR TOO
C TIGHT A TOLERANCE OR ROUND-OFF PREVENTS
C ATTAINING REQUIRED ACCURACY
C R. H. FRENCH, 14 AUGUST 1984
C

```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 EXTERNAL F
0004 DIMENSION WORK(N),STACK(N),HEAP(N)
0005 KODE=0
0006 Y=0.00
0007 WORK(1)=XU
0008 CALL QR(XL,XU,F,T)
0009 HEAP(1)=T
0010 A=XL
0011 NPPTS=1
0012 EPS=TOL
0013 STACK(1)=EPS
0014 B=WORK(NPPTS)
0015 XM=(A+B)*0.500
0016 CALL QR(A,XM,F,P1)
0017 CALL QR(XM,B,F,P2)
0018 IF(DABS(T-P1-P2).LE.EPS) GOTO 20
C SPLIT IT
0019 NPPTS=NPPTS+1
0020 IF(NPPTS.GT.N) THEN
0021 KODE=1
0022 RETURN
0023 END IF
0024 WORK(NPPTS)=XM
0025 HEAP(NPPTS)=P2
0026 T=P1
0027 EPS=EPS/2.00

```

```

0001 SUBROUTINE ADQQA2(XL,XU,Y,QR,F,TOL,WORK,STACK,HEAP,N,KODE)
C ADAPTIVE QUADRATURE ALGORITHM
C XL - LOWER LIMIT OF INTEGRAL (IN)
C XU - UPPER LIMIT OF INTEGRAL (IN)
C Y - VALUE OF INTEGRAL (OUT)
C QR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
C WITH CALLING SEQUENCE
C CALL QR(XL,XU,F,Y)
C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
C WORK - WORK ARRAY OF SIZE N (IN)
C STACK - SECOND WORK ARRAY OF SIZE N, MUST NOT BE
C SAME ARRAY AS WORK (IN)
C HEAP - THIRD WORK ARRAY OF SIZE N, DISTINCT FROM BOTH WORK AND STACK
C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (IN)
C KODE - ERROR INDICATOR (OUT)
C 0 -- NO ERROR
C 1 -- WORK ARRAYS TOO SMALL
C 2 -- EPS DIVIDED TO ZERO, EITHER ASKING FOR TOO
C TIGHT A TOLERANCE OR ROUND-OFF PREVENTS
C ATTAINING REQUIRED ACCURACY
C R. H. FRENCH, 14 AUGUST 1984
C
  
```

```

0028 IF(EPS.EQ.0.001) THEN
0029 KODE=2
0030 RETURN
0031 END IF
0032 STACK(NPTS)=EPS
0033 GOTO 10
C FINISHED A PIECE
20 Y=Y+P1+P2
0034 EPS=STACK(NPTS)
0035 T=HEAP(NPTS)
0036 NPTS=NPTS-1
0037 A=B
0038 IF(NPTS.EQ.0) RETURN
0039 GOTO 10
0040 END
0041
  
```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 EXTERNAL F
0004 DIMENSION WORK(N),STACK(N),HEAP(N)
0005 KODE=0
0006 Y=0.00
0007 WORK(1)=XU
0008 CALL QR(XL,XU,F,T)
0009 HEAP(1)=T
0010 A=XL
0011 NPTS=1
0012 EPS=TOL
0013 STACK(1)=EPS
0014 B=WORK(NPTS)
0015 XM=(A+B)*0.500
0016 CALL QR(A,XM,F,P1)
0017 CALL QR(XM,B,F,P2)
0018 IF(DABS(T-P1-P2).LE.EPS) GOTO 20
C SPLIT IT
0019 NPTS=NPTS+1
0020 IF(NPTS.GT.N) THEN
0021 KODE=1
0022 RETURN
0023 END IF
0024 WORK(NPTS)=XM
0025 HEAP(NPTS)=P2
0026 T=P1
0027 EPS=EPS/2.00
  
```

```

0001 SUBROUTINE D616(A,B,F,ANSWER)
      C
      C 16-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
      C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
      C R. H. FRENCH, 28 FEBRUARY 1986
      C
      IMPLICIT DOUBLE PRECISION (A-H,O-Z)
      COMMON /WTS/ X(8),M(8)
      ANSWER=0.DO
      BPA02=(B-A)/2.DO
      BPA02=(B+A)/2.DO
      DO 10 I=1,8
      C=X(I)*BPA02
      Y1=BPA02+C
      Y2=BPA02-C
      ANSWER=ANSWER+M(I)*(F(Y1))+F(Y2))
      CONTINUE
      ANSWER=ANSWER*BPA02
      RETURN
      END
  
```

```

0028 IF(EPS.EQ.0.DO) THEN
0029   KODE=2
0030   RETURN
0031   END IF
0032   STACK(MPTS)=EPS
0033   GOTD 10
      C FINISHED A PIECE
0034   Z0
0035   Y=Y+P1+P2
0036   EPS=STACK(MPTS)
0037   T=HEAP(MPTS)
0038   MPTS=MPTS-1
0039   A=B
0040   IF(MPTS.EQ.0) RETURN
0041   GOTD 10
      END
  
```

```

0001 SUBROUTINE D6XV1(A,B,F,ANSWER)
      C
      C 16-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
      C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
      C R. H. FRENCH, 28 FEBRUARY 1986
      C
      IMPLICIT DOUBLE PRECISION (A-H,O-Z)
      COMMON /WTS/ X(8),M(8)
      ANSWER=0.DO
      BPA02=(B-A)/2.DO
      BPA02=(B+A)/2.DO
      DO 10 I=1,8
      C=X(I)*BPA02
      Y1=BPA02+C
      Y2=BPA02-C
      ANSWER=ANSWER+M(I)*(F(Y1))+F(Y2))
      CONTINUE
      ANSWER=ANSWER*BPA02
      RETURN
      END
  
```

```

0001 C DOUBLE PRECISION FUNCTION PZ1(Y)
      C SIGNAL CHANNEL P.D.F. WITH CHANGE OF VARIABLE Y=AX
      C
      C IMPLICIT DOUBLE PRECISION (A-H,O-Z)
      C DIMENSION WORK(30), STACK(30), HEAP(30)
      C LOGICAL*1 REG1, REG2
      C EXTERNAL DGXVI, F20, F21, F22
      C COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      C COMMON /DEMPAR/ BIGK, AAB, BAB, LJAM,
      C TAU, TAU2, TALK, TALK2
      C COMMON /PARDEN/ RHOM, RHOT, RHOM2, RHOT2, RHOMA, RHOTA,
      C RHOMB, RHOT8, RHOMH, RHOTH
      C COMMON /QUES/ QO, Q1
      C COMMON /XCOM/ XCOM
      C COMMON /OUTER/ XXX, XXXK
      C XXX=Y
      C IF(LJAM.GE.1) THEN
      C XXX=Y/BIGK
      C YK=XXXK
      C YTK=(Y-TAU)/BIGK
      C YTK2=(Y-TAU2)/BIGK
      C END IF
      C REG1=Y.GE.0.DO .AND. Y.LT.TAU
      C REG2=Y.GE.TAU .AND. Y.LT.TAU2
  
```

```

0021 C TWO HOPS PER SYMBOL
      C
      C GOTO (2100, 2200, 2300), LJAM+1
      C
      C NO OPS JAMMED
  
```

```

0022 IF(REG1) THEN
0023 BARG1=DSORT(RHOMB*Y)
0024 I01=1
0025 CALL DXBT(2100)
0026 PZ1=0.500*DSORT(Y/RHOMH)*DEXP(BARG1-Y-RHOM2)*B1
0027 ELSE IF(REG2) THEN
      C BARG1=DSORT(RHOMA*(Y-TAU))
      C I01=0
0028 CALL DXBT(2101)
0029 PAR1=2.00*DEXP(BARG1-Y+TAU-RHOM)*B1
0030 XCOM=Y+RHOM2
0031 CALL ADQUA2(Y-TAU,TAU,ANSWER,DGXVI,F20,I.D-9,WORK,STACK,
      C HEAP,30,K00)
0032 CALL TEST2(KOD,2100)
0033 PZ1=PART+ANSWER
      C ELSE
0034 PZ1=0.00
0035 END IF
0036 GOTO 9000
  
```

```

0001 C SUBROUTINE TEST(ID)
      C TEST RETURN CODE FROM BESSEL FUNCTION
      C
      C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      C COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      C IF(KODE.EQ.0) RETURN
      C WRITE(5,1) KODE, ID
      C 1 FORMAT(' BESSEL FUNCTION CODE = ',I2,' FROM CALL NUMBER ',I5)
      C STOP 'FATAL ERROR'
      C END
  
```

```

0001 C SUBROUTINE TEST2(KODE, ID)
      C TEST RETURN CODE FROM ADQUAD/ADQUA2
      C
      C IF(KODE.EQ.0) RETURN
      C WRITE(5,1) KODE, ID
      C 1 FORMAT(' ADAPTIVE INTEGRATOR CODE = ',I2,' FROM CALL NUMBER ',I5)
      C STOP 'FATAL ERROR'
      C END
  
```



```

0001      DOUBLE PRECISION FUNCTION F21(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=2, LJAM=1
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM,
      TAU, TAU2, TALK, TALK2
      COMMON /PARDEM/ RHOM, RHOT, RHOM2, RHOT2, RHOM4, RHOT4,
      RHOM8, RHOT8, RHOMH, RHOTH
      COMMON /QUES/ Q0, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXK
      BARG1=DSQRT(RHOM4*U)
      BARG2=DSQRT(RHOT4*(XXX-U/BIGK))
      I01=0
      I02=0
      CALL BPROD(2210)
      F21=DEXP(BARG1+BARG2-XCOM-U+U/BIGK)*B1*B2
      RETURN
      END
  
```

```

0001      SUBROUTINE BPROD(IDENT)
C
C COMPUTE TWO BESSEL FUNCTIONS, ARGUMENTS AND RESULTS IN COMMON
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      CALL DXBT(IDENT)
      CALL DXBST(BARG2,I02,B2,KODE)
      CALL TEST(IDENT*1)
      RETURN
      END
  
```

```

0033      TWO HOPS JAMMED
C
C
C 2300      IF(REG1) THEN
0034          GL=1.00-(1.00+YK)*DEXP(-YK)
0035          ELSE IF(REG2) THEN
0036              GL=1.00-(1.00+TALK2-YK)*DEXP(-YK)
0037          ELSE IF(Y.GE.TAU2) THEN
0038              GL=1.00
0039          ELSE
0040              GL=0.00
0041          END IF
0042          GOTO 9000
0043          CONTINUE
0044          RETURN
0045          END
  
```

```

0001      DOUBLE PRECISION FUNCTION F20(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=2, LJAM=0
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM,
      TAU, TAU2, TALK, TALK2
      COMMON /PARDEM/ RHOM, RHOT, RHOM2, RHOT2, RHOM4, RHOT4,
      RHOM8, RHOT8, RHOMH, RHOTH
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /QUES/ Q0, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXK
      BARG1=DSQRT(RHOM4*U)
      BARG2=DSQRT(RHOM4*(XXX-U))
      I01=0
      I02=0
      CALL BPROD(2210)
      F20=DEXP(BARG1+BARG2-XCOM)*B1*B2
      RETURN
      END
  
```

```

0001 SUBROUTINE TIES(JSUB,MM,PTIE)
C COMPUTE PROBABILITY OF CORRECT DECISION GIVEN THAT
C SEVERAL SATURATED CHANNELS ARE TIED
C
      $ IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      DIMENSION JSUB(MM), LLOW(7), LINC(7), LUP(7), NU(7),
      LOGICAL *I GO
      COMMON /DENPAR/ BIEK, AAB, BAB, LJAM,
      TAU, TAU2, TALK, TALK2
      COMMON /PARDEN/ RHOM, RHOT, RHOM2, RHOT2, RHOM4, RHOT4,
      RHOM8, RHOT8, RHOMH, RHOTH
      COMMON /QUES/ QO, QI
      C NUMBER OF NON-SIGNAL CHANNELS
      MMU=MM-1
      PTIE=0.DO
      CUEO=DEXP(-TAU)
      CUEI=DEXP(-TALK)
      PII=DXI(QO,2-JSUB(1))*DXI(QI,JSUB(1))
      DO 10 I=2,MM
      P2LM(I)=DXI(CUEO,2-JSUB(I))*DXI(CUEI,JSUB(I))
      CONTINUE
      C SET UP VECTOR LOOP PARAMETERS
      DO 20 I=1,MM-1
      LLOW(I)=0
      LINC(I)=1
      LUP(I)=1
      CONTINUE
      PTIE=0.DO
      C START LOOP ON THE TIE EVENTS
      CALL VLIMIT(MU,LLOW,MM-1)
      MUSUM=0
      DO 40 I=1,MM-1
      MUSUM=MUSUM+MU(I)
      CONTINUE
      FRAC=1.DO/(1.DO+MUSUM)
      PROD=1.DO
      DO 50 M=2,MM
      IF(MU(M-1).EQ.1) THEN
      PROD=PROD*P2LM(M)
      ELSE
      PROD=PROD*(1.DO-P2LM(M))
      END IF
      CONTINUE
      PTIE=FRAC*PROD*PTIE
      CALL VLITER(MU,LLOW,LUP,LINC,MM-1,GO)
      IF(GO) GOTO 30
      PTIE=PTIE*PII
      RETURN
      END
0001
0002
0003
0004
0005
0006
0007
0008
0009
0010
0011
0012
0013
0014
0015
0016
0017
0018
0019
0020
0021
0022
0023
0024
0025
0026
0027
0028
0029
0030
0031
0032
0033
0034
0035
0036
0037
0038
0039
0040
0041
  
```

```

0001 DOUBLE PRECISION FUNCTION F22(I)
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=2, LJAM=2
C
      $ IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /DENPAR/ BIEK, AAB, BAB, LJAM,
      TAU, TAU2, TALK, TALK2
      COMMON /PARDEN/ RHOM, RHOT, RHOM2, RHOT2, RHOM4, RHOT4,
      RHOM8, RHOT8, RHOMH, RHOTH
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /QUES/ QO, QI
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      BARG1=DSQRT(RHOT4+U/BIEK)
      BARG2=DSQRT(RHOT4+(XXXX-U/BIEK))
      I01=0
      I02=0
      CALL BPROD(2310)
      F22=DEXP(BARG1+BARG2-XCOM)*B1*B2
      RETURN
      END
0001
0002
0003
0004
0005
0006
0007
0008
0009
0010
0011
0012
0013
0014
0015
0016
  
```

```

0001 SUBROUTINE DXBT(ID)
C CALL DXBES1 AND TEST RETURN CODE
C
      $ IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      CALL DXBES1(BARG1,I01,B1,KODE)
      CALL TEST(ID)
      RETURN
      END
0001
0002
0003
0004
0005
0006
0007
  
```

```

0001 SUBROUTINE SETTAU(MH,PEOO)
C SEARCH FOR OPTIMUM THRESHOLD IN ABSENCE OF JAMMING
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 EXTERNAL PUNJAM
0004 COMMON /DENPAR/ BIGK, AAB, BAB, LJAM,
    $ TAU, TAU2, TALK, TALK2
0005 COMMON /PARDEN/ RHOM, RHOT, RHOM2, RHOT2, RHOM4, RHOT4,
    $ RHOM8, RHOT8, RHOMH, RHOTH
0006 COMMON /QUES/ QO, QI
0007 LJAM=0
C GUESS BASED ON QUADRATIC CURVE FIT
C
0008 IF(MH.EQ.2) THEN
0009 GUESS=0.92500*4-8.47500*2+32.4500
0010 ELSE IF(MH.EQ.4) THEN
0011 GUESS=1.0500*4-9.3500*2+34.500
0012 ELSE IF(MH.EQ.8) THEN
0013 GUESS=1.100*4-9.900*2+36.300
0014 ELSE
0015 GUESS=15.00
0016 END IF
0017 CALL MINSER(PUNJAM,PEMIN,TAUOPT,1.00,GUESS,0.00)
    $
0018 TAU=TAUOPT
0019 TAU2=TAU+TAU
0020 PEOO=PEMIN
0021 RETURN
0022 END
  
```

```

0001 DOUBLE PRECISION FUNCTION PUNJAM(ETA)
C FUNCTION FOR UNJAMMED P(E) FOR OPT. THRESHOLD SEARCH
C
C NOTE: WHEN JAMMING EVENT IS (0,0,...,0), THE VARIABLES
C BIGK, AAB, BAB, TALK, AND TALK2 ARE NOT
C USED IN THE COMPUTATIONS, AND HENCE DO NOT NEED
C TO BE SET UP BEFORE CALLING PSEL FROM THIS FUNCTION
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 DIMENSION NOJAM(4)
0004 COMMON /INPUTS/ DEBIND(3), NSLOTS, GAMLST(10), K, MH
0005 COMMON /DENPAR/ BIGK, AAB, BAB, LJAM,
    $ TAU, TAU2, TALK, TALK2
0006 DATA NOJAM/0,0,0,0/
0007 LJAM=0
0008 TAU=ETA
0009 TAU2=TAU+ETA
0010 CALL PSEL(NOJAM,MH,P)
0011 PUNJAM=P
0012 RETURN
0013 END
  
```

```
0001 SUBROUTINE MINSER(F,FMIN,XMIN,STEP,GUESS,BLIM,ULIM,TOL)
C SEARCH FOR MINIMUM OF F(X) OVER THE INTERVAL BLIM <= X <= ULIM
C TROUBLE MAY OCCUR IF F(X) HAS MULTIPLE LOCAL MINIMA WITHIN THE
C SEARCH INTERVAL OR IF THE FUNCTION IS VERY STEEP AND STEP IS
C TOO BIG.
C F = NAME OF FUNCTION TO BE MINIMIZED
C FMIN = MINIMUM VALUE OF F(X) OVER INTERVAL
C XMIN = VALUE OF X FOR WHICH FMIN OCCURS
C STEP = INITIAL STEP SIZE FOR SEARCH
C GUESS = INITIAL GUESS AT XMIN, BLIM <= GUESS <= ULIM
C BLIM = LOWER LIMIT OF SEARCH INTERVAL
C ULIM = UPPER LIMIT OF SEARCH INTERVAL
C TOL = TOLERANCE ON XMIN; SEARCH STOPS WHEN DX < TOL
C NOTE: F MUST BE A DOUBLE PRECISION FUNCTION OF ONE
C DOUBLE PRECISION ARGUMENT. ANY PARAMETERS CAN BE PASSED
C FROM THE CALLER VIA A COMMON BLOCK.
C PROGRAMMER: ROBERT H. FRENCH DATE: 17 MARCH 1986
C
```

```
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 X=GUESS
0004 SLMAX=DABS(X-BLIM)
0005 SUMAX=DABS(ULIM-X)
0006 DX=DMINI(STEP,SLMAX,SUMAX)
0007 TEST=TOL
0008 FO=F(X)
0009 F1=F(X+DX)
0010 ARE WE GOING IN THE RIGHT DIRECTION?
0011 IF(F1.LE.FO) GOTO 100
0012 DX=-DX
0013 F1=F(X+DX)
0014 IF(F1.LE.FO) GOTO 100
0015 DX=DX/10.DO
0016 IF(DABS(DX).GE.TEST) GOTO 10
0017 XMIN=X
0018 FMIN=FO
0019 RETURN
0020
```

```
0021 C ALL OK
0022 C PAST MIN?
0023 IF(F2.GE.F1) GOTO 200
0024 F0=F1
0025 F1=F2
0026 X=X+DX
0027 GOTO 100
0028 C MIN MAY BE AT AN ENDPOINT. CUT STEP SIZE AND TRY AGAIN
0029 C IF INCREMENT NOT TOO SMALL.
0030 IF(DABS(DX).LE.TEST) GOTO 120
0031 X=X+DX
0032 F0=F1
0033 DX=DX/10.00
0034 F1=F(X+DX)
0035 GOTO 100
0036 C MIN MUST BE AT THE ENDPOINT (OR WITHIN MINIMUM DX THEREOF)
0037 IF(X2.LE.BLIM) GOTO 122
0038 C MIN AT X=ULIM
0039 XMIN=ULIM
0040 FMIN=F(XMIN)
0041 RETURN
0042 C MIN AT BLIM
0043 XMIN=BLIM
0044 GOTO 121
0045 C HAVE PASSED MIN. IS IT LOCATED CLOSELY ENOUGH YET?
0046 IF(DABS(DX).LE.TEST) GOTO 300
0047 C ... NO, CUT STEP SIZE AND TRY AGAIN
0048 GOTO 115
0049 C DOWNE!
0050 C SINCE FO >= F1 & F2 >= F1 AND ABS(DX) < MIN. DX, CALL F1 THE MIN.
0051 FMIN=F1
0052 XMIN=X+DX
0053 RETURN
0054 END
```

```
0055 C NOW GOING RIGHT DIRECTION. KEEP GOING UNTIL PAST MINIMUM
0056 C BY ONE STEP.
0057 X2=X+DX+DX
0058 IF(X2.GT.ULIM .OR. X2.LT.BLIM) GOTO 110
```


APPENDIX I
COMPUTER PROGRAM FOR
CLIPPER RECEIVER FOR M=4 AND L=2

The following pages contain the source code listing for the FORTRAN-77 program used for numerical computations of the performance of the clipper receiver for the case of M=4 and L=2 with a numerical search for the worst-case jamming fraction. By increasing the array A to 256 elements and the arrays C and D to 6561 elements each, and changing the array size parameters to calls to PUTIN and LOOKUP, the program may also be used for the case M=8, L=2.

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0001 PROGRAM CLPRAN
C THIS PROGRAM COMPUTES THE ERROR PROBABILITY FOR RANDOM 4-ARY
C FSK/FH WITH 2 HOPS/BIT IN THE PRESENCE OF PARTIAL-BAND
C NOISE JAMMING BY NUMERICAL INTEGRATION FOR THE CLIPPER RECEIVER
C
C ANALYSIS: L. E. MILLER, R. H. FRENCH
C PROGRAM: R. H. FRENCH
C
C V 2.1.0 - COMPUTATIONS ONLY
C
C
C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
PARAMETER (LJ=51)
CHARACTER*1 YES, NO, REPLY, BLANK
CHARACTER*13 FNAME, GAMMA
LOGICAL DOTAU, TEST
REAL*4 PLOG(LJ), DBSJR(LJ), QOPT(LJ)
VIRTUAL A(100), IASUB(100), G(625), ICSUB(625)
VIRTUAL B(625), IDSUB(625), PRERR(625), IPSUB(625)
VIRTUAL PESAN(2400)
C COMMON /INPUTS/ PASSES PARAMETERS OF THE RUN
COMMON /INPUTS/ DEBNOL(3), NSLOTS,K,MM
C COMMON /SIZE/ PASSES NUMBERS OF PARAMETERS
COMMON /SIZE/ NO
COMMON /DENPAR/ BIGK, AAB, BAB, L,JAM,
TAU, TAU2, TALK, TALK2
COMMON /PARGEN/ RHOM, RIOT
DATA YES, NO, BLANK /'Y', 'N', ' ' /
CALL ERSET(29,TRUE,.,FALSE,.,TRUE,.,FALSE,.,15)
CALL GET(NJ,START,DBINC)
SLOTS=NSLOTS
WORBIT=0.5DO*MM/(MM-1.DO)
DO 800 I=1,NO
DOTAU=.TRUE.
EBNO=10.00**((DEBNOL(10)/10).DO)
RHOM=K*EBNO/2.DO
IOUTPUT=DEBNOL(10)
C OPEN DATA FILE
C
C
IGOUT=GAMMA*1000.DO+0.500
WRITE(FNAME,730) MM,IOUTPUT
FORMAT('GJ',I1,'.I2'.I2,2, '.DAT')
730 WRITE(6,776) MM,DEBNOL(10)
776 FORMAT('ICLIPPER RECEIVER. OPTIMUM GAMMA RESULTS:/
, M=',L,5X, 'L=2',5X, 'EB/NO=',F8.4//', EB/NJ (dB)',
, 5X, 'p(e)',15X, 'Qopt')
$
$ WRITE(5,733) FNAME
733 FORMAT(' WORKING ON ',A13)
OPEN(UNIT=4,FILE=FNAME,STATUS='OLD',FORM='UNFORMATTED',
ERR=750)
$

```

```

C HAVE AN EXISTING FILE, READ TO SEE HOW FAR WE GOT BEFORE
C
1300 READ(4) MMIN, EBNOIN, NSLIN, TAU, TAU2, PEOD
C WE HAVE READ A VALUE OF TAU, SO WE WON'T NEED TO RECOMPUTE
C IT UNTIL EITHER EB/NO, MM, OR LL CHANGES
C
DOTAU=.FALSE.
JJ=0
JJ=JJ+1
740 READ(A,END=742) DBSJR(JJ), PRLOG(JJ), QOPT(JJ)
GOTO 740
742 CLOSE(UNIT=4)
GOTO 755
C NO EXISTING FILE, THIS IS THE FIRST TIME:
C CREATE FILE HEADER RECORD
C
750 JJ=1
... WE CAN'T READ TAU FROM A FILE, SO IF THE FIRST TIME
THROUGH THE LOOP ON EB/NO WE MUST COMPUTE IT. BUT IF
EB/NO HASN'T CHANGED, WE DON'T NEED TO RECOMPUTE IT SINCE
THE THRESHOLD IS NOT A FUNCTION OF GAMMA NOR OF EB/NJ.
IF(DOTAU) THEN
WRITE(5,757)
FORMAT(' SETTING THRESHOLD')
CALL SETTAU(MM,PEOD)
DOTAU=.FALSE.
WRITE(5,1991) MM, TAU
WRITE(6,1991) MM, TAU
FORMAT(' M=',12, ' L=2 OPT THRES = ',1PD15.8)
END IF
OPEN(UNIT=4,FILE=FNAME,STATUS='NEW',FORM='UNFORMATTED')
WRITE(4) MM,DEBNOL(10),NSLOTS,TAU,TAU2,PEOD
CLOSE(UNIT=4)
DO 600 IJ=JJ,NJ
IF(IJ.GE.3) THEN
IQQ=QOPT(IJ-1)-QOPT(IJ-2)+0.500
IF(IQQ.EQ.0) IQQ=1
Q=QOPT(IJ-1)
IQ=Q
ELSE
IQQ=1
Q=1.DO
END IF
DO=IQQ
C GIVE PROGRESS MESSAGE TO TI:
601 WRITE(5,601) IJ
FORMAT(' IJ=',13)
DEBNJ=START+(IJ-1)*DBINC

```

```

0070 DBSJR(IJ)=DEBNJ
0071 DO 602 IJ=1,2400
0072 PESAV(IJ)=0.00
0073 R=10.00**((DEBNJ/10.00)
C PRIME THE ALGORITHM WITH DUMMY OLD VALUES OF P(E)
0074 P1=0.00
0075 P2=0.00
0076 GAMMA=0/SLOTS
0077 WRITE(GNAME,735) MM,IQ
0078 FORMAT('EQ',11,'2',14.4,'.DAT')
0079 OPEN(UNIT=3,FILE=GNAME,STATUS='OLD',FORM='UNFORMATTED',
$ READONLY,ERR=770)
0080 WRITE(5,3939)
0081 FORMAT(' READING EVENT FILE')
0082 READ(3) D,IDSUB,NUSED,GOOD
0083 CLOSE(UNIT=3)
0084 GOTO 777
C IF FILE FOR EVENT PROBABILITIES DOES NOT EXIST, CALCULATE THEM
C AND CREATE A FILE.
0085 770 CONTINUE
0086 WRITE(5,3938)
0087 FORMAT(' CREATING EVENT FILE')
0088 CALL GENPTE(MM,IQ,NSLOTS,GOOD,A,IASUB,C,ICSUB,
D,IDSUB,NUSED)
0089 OPEN(UNIT=3,FILE=GNAME,STATUS='NEW',FORM='UNFORMATTED')
0090 WRITE(3) D,IDSUB,NUSED,GOOD
0091 CLOSE(UNIT=3)
0092 IF(.NOT.GOOD) GOTO 700
0093 RHOTS=GAMMA*R*EBNO/(GAMMA*R+EBNO)
0094 RHOT=K*RHOTS/2.00
C EVALUATE THE PROBABILITY
0095 CALL PSUBE(MM,PESYM,D,IDSUB,NUSED,PRERR,IPSUB,PEOO,
$ PESAV,IQ)
0096 P3=PESYM
0097 IF(P3.GT.P2 .AND. IQ.LT.NSLOTS) THEN
C KEEP ON GOING, WE ARE NOT PAST THE MAXIMUM
P1=P2
P2=P3
IQ=MINO(IQ+IQ,NSLOTS)
Q=DMINI(Q+IQ,SLOTS)
GOTO 709
ELSE
PMAX=DMAX1(P1,P2,P3)
EPS=0.00100*PMAX
TEST=(DABS(P1-P2)).LE.EPS .AND. DABS(P1-P3).LE.EPS .AND.
DABS(P2-P3).LE.EPS)
IF( TEST .OR. IDO.EQ.1
$ IF( (.NOT.TEST) .AND. IQ.EQ.NSLOTS)) THEN
C WE ARE DONE WHEN ALL 3 ARE CLOSE TOGETHER OR WHEN DO=1
C OR WHEN WE REACHED FULL-BAND JAMMING AND P(E) IS STILL
C INCREASING
PORT=PMAX
0108

```

```

0109 IF(P2.GT.P3) THEN
C THE OPTIMUM MUST BE THE MIDDLE POINT OF THE 3
QOPT(IJ)=Q-DQ
0110 QOPT(IJ)=Q-DQ
0111 IF(QOPT(IJ).EQ.0.00) QOPT(IJ)=1.00
C PREVENT ROUND-OFF FROM MAKING QOPT VS. EB/NJ NON-MONOTONIC
0112 IF(IJ.GT.1) THEN
IF(QOPT(IJ).LT.QOPT(IJ-1)) QOPT(IJ)=QOPT(IJ-1)
END IF
0113 ELSE
0114 THE OPTIMUM IS FULL-BAND JAMMING
QOPT(IJ)=NSLOTS
END IF
0115 GOTO 665
ELSE
0116 NOT LOCATED SUFFICIENTLY ACCURATELY, CUT DQ AND TRY AGAIN
Q=Q-DQ-DQ
0117 I=I-IDQ-IDQ
0118 D=D-IDQ
0119 P2=P1
0120 P1=0.00
0121 Q=Q+DQ
0122 IQ=IQ+IDQ
0123 GOTO 709
END IF
0124 END IF
0125 PE=WORBIT*POPT
0126 WRITE(6,666) DBSJR(IJ),PE,QOPT(IJ)
0127 FORMAT(1X,F7.3,5X,IPD12.5,5X,IPD12.5)
0128 PRLOG(IJ)=DLOG10(PE)
0129 OPEN(UNIT=4,FILE=FNAME,STATUS='OLD',ACCESS='APPEND',
$ FORM='UNFORMATTED')
0130 WRITE(4) DBSJR(IJ), PRLOG(IJ), QOPT(IJ)
0131 CLOSE(UNIT=4)
0132 CONTINUE
0133 OPEN(UNIT=4,FILE=FNAME,STATUS='NEW',FORM='UNFORMATTED')
0134 WRITE(4) MM,2,DEBNOL(10),NSLOTS,DBSJR,PRLOG,QOPT
0135 CLOSE(UNIT=4)
0136 WRITE(6,776) MM,DEBNOL(10)
0137 DO 689 IJ=1,NJ
0138 WRITE(6,666) DBSJR(IJ),IQ,**PRLOG(IJ),QOPT(IJ)
0139 CONTINUE
0140 WRITE(6,688) TAU
0141 FORMAT(////,OPTIMUM THRESHOLD FOR ABOVE IS ETA/SIGMA**2 =',
$ F7.3)
0142 CONTINUE
0143 700 CONTINUE
0144 800 CONTINUE
0145 900 CONTINUE
0146 STOP 0
0147 END

```

```

0001 SUBROUTINE GET(N), START, DBINC)
C
C INTERACTIVE INPUT OF PARAMETERS FOR RUN
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 CHARACTER*9 FIELD, BLANK9
0004 COMMON /INPUTS/ DEBNOL(3), NSLOTS, K, MM
0005 COMMON /SIZE/ NO
0006 COMMON /SHARE/ DSNR(3,4)
0007 DATA BLANK9/'
0008 WRITE(5,33)
0009 FORMAT(' BITS/SYMBOL (K) [2]: ', $)
0010 READ(5,3)K
0011 IF(K.EQ.0)K=2
0012 MM=2**K
0013 WRITE(5,2)
0014 FORMAT(' HOW MANY EB/NO? [1]: ', $)
0015 READ(5,3)NO
0016 FORMAT(12)
0017 IF(NO.EQ.0)NO=1
0018 DO 7 IN=1,NO
0019 DO=DSNR(IN,K)
0020 WRITE(5,5)IN,DO
0021 FORMAT(' EB/NO(' ,I2,') [' ,F9.6,'] : ', $)
0022 READ(5,6)FIELD
0023 FORMAT(A9)
0024 IF(FIELD.EQ.BLANK9) THEN
0025 DEBNOL(IN)=DO
0026 ELSE
0027 DECODE(9,61,FIELD)DEBNOL(IN)
0028 FORMAT(F9.6)
0029 END IF
0030 CONTINUE
0031 NSLOTS=2400
0032 WRITE(5,39)
0033 FORMAT(' HOW MANY EB/NO? [51]: ', $)
0034 READ(5,34,ERR=38) NJ
0035 FORMAT(13)
0036 IF(NJ.EQ.0) NJ=51
0037 IF(NJ.LT.0 .OR. NJ.GT.51) GOTO 32
0038 WRITE(5,41)
0039 FORMAT(' STARTING VALUE FOR EB/NJ (08) [50.]: ', $)
0040 READ(5,42,ERR=40) START
0041 IF(START.EQ.0.00) START=50.00
0042 FORMAT(F6.3)
0043 IF(NJ.EQ.1) RETURN
0044 WRITE(5,36)
0045 FORMAT(' DB INCREMENT FOR EB/NJ [-1.07]: ', $)
0046 READ(5,37,ERR=35) DBINC
0047 FORMAT(F6.3)
0048 IF(DBINC.EQ.0.00) DBINC=-1.00
0049 RETURN
0050 END

```

```

0001 SUBROUTINE PSIBE(M,PE,D,IDSUB,MUSED,PRERR,IPSUB,PEOD,
C PESAV,IQ)
C
C COMPUTE UNCONDITIONAL ERROR PROBABILITY
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 INTEGER JAM(4),LUP(4),JSUB(4)
0004 LOGICAL*1 GO,NOME,STORE
0005 VIRTUAL PRERR(625),IPSUBI(625)
0006 VIRTUAL DI(625),IDSUB(625)
0007 VIRTUAL PESAV(2400)
0008 COMMON /SHARE2/ LOM(4),LINC(4)
0009 COMMON /DEMPAR/ B1GK, AAB, BAB, LJAM,
C TAU, TAUZ, TALK, TALKZ
COMMON /PARDEN/ RHM, RHOT
DATA STORE/.TRUE./
0010 IF(PESAV(IQ).NE.0.00) THEN
0011 PE=PESAV(IQ)
0012 RETURN
0013 END IF
0014 PE=0.00
0015 NPS=0
0016 DO 10 I=1,M
0017 LUP(I)=2
0018 JSUB(I)=0
0019 CONTINUE
0020 IO
0021 C THE ALL-ZERO JAMMING EVENT P(E) IS AVAILABLE FROM THE SEARCH FOR
C THE OPTIMUM THRESHOLD, SO PUT IT INTO THE ARRAY OF SAVED VALUES
C START VECTOR-INDEXED LOOP ON JAMMING EVENTS
CALL VINIT(JAM,LOM,M)
0022 CONTINUE
0023 IF(JAMI.NE.JAM(1)) THEN
0024 CALL PUTIN(PEOD,PRERR,IPSUB,NPS,625,ISUB,STORE)
0025 IF(KODE.NE.0) STOP 'PRERR FULL'
JAMI=1
0026 C START VECTOR-INDEXED LOOP ON JAMMING EVENTS
CALL VINIT(JAM,LOM,M)
0027 CONTINUE
0028 IF(JAMI.NE.JAM(1)) THEN
0029 CALL PUTIN(PEOD,PRERR,IPSUB,NPS,625,ISUB,STORE)
0030 IF(KODE.NE.0) STOP 'PRERR FULL'
JAMI=1
0031 END IF
0032 CALL EVENT(M,JAM,PIE,D,IDSUB,MUSED)
C BYPASS CONDITIONAL ERROR PROBABILITY CALCULATION IF EVENT
C PROBABILITY IS ZERO. THIS SAVES MUCH TIME.
IF(PIE.EQ.0.00)GOTO 101
C SINCE JAMMING PROBABILITIES DEPEND ONLY ON NO. OF CHANNELS
C HAVING JAM(I) HOPS JAMMED AND NOT THE ARRANGEMENT OF THE
C CHANNELS, WE CAN SORT THE NON-SIGNAL CHANNELS INTO ASCENDING
C NUMBERS OF HOPS JAMMED. THIS REDUCES NUMBER OF DISTINCT
C CONDITIONAL ERROR PROBABILITIES WHICH MUST BE SAVED TO AVOID
C RECOMPUTING THEM UNNECESSARILY.
DO 111 I=1,M
0033 JSUB(I)=JAM(I)
0034 CONTINUE
0035 111

```

```

0036 IF(M.EQ.2) GOTO 199
0037 DO 110 I=2,M-1
0038 DO 120 J=I+1,M
0039 IF(JSUB(J).LT.JSUB(I)) THEN
0040 JTMP=JSUB(I)
0041 JSUB(I)=JSUB(J)
0042 JSUB(J)=JTMP
0043 END IF
0044 CONTINUE
0045 110 CONTINUE
0046 199 CONTINUE
C STORE AS ELEMENT OF A SPARSE ARRAY TO SAVE MEMORY
C EVEN THOUGH WE STORE ZEROS, THE SORTING OF SUBSCRIPTS
C CUTS OUT MANY ELEMENTS.
0047 CALL LOCK(M,LOW,LUP,JSUB,ISUB)
0048 C TRY TO FIND CONDITIONAL ERROR PROBABILITY IN STORED ARRAY
0049 CALL LOOKUP(PROB,PRERR,IPSUB,MPS,625,ISUB,STORE,HOME)
0050 C IF IT IS NOT THERE, WE MUST COMPUTE IT
0051 IF(HOME) THEN
0052 CALL PSEL(JSUB,M,PROB)
0053 C ... AND SAVE IT FOR POSSIBLE FUTURE RE-USE
0054 CALL PUTIN(PROB,PRERR,IPSUB,MPS,625,ISUB,KODE,STORE)
0055 IF(KODE.NE.0) STOP 2
0056 END IF
0057 C SUM UP UNCONDITIONAL ERROR PROBABILITY
0058 PE=PE+IE*PROB
C ITERATE THE VECTOR-INDEX LOOP
101 CALL VITER(JAM,LOW,LUP,LINC,M,60)
0056 IF(60) GOTO 100
0057 RETURN
0058 END

```

```

0001 C SUBROUTINE EVENT(M,JAM,PIE,D,IDSUB,MUSED)
C SUBROUTINE TO LOOK UP EVENT PROBABILITY FROM STORED ARRAY
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 LOGICAL*1 STORE,HOME
0004 DIMENSION JAM(4),LUP(4)
0005 VIRTUAL D(625),IDSUB(625)
0006 COMMON /SHARE2/ LOW(4),LINC(4)
0007 DATA STORE/,FALSE./
C SET UP ARRAY DESCRIPTION D(0:LL,....,0:LL) WITH M DIMENSIONS
0008 DO I I=1,M
0009 LUP(I)=2
0010 CONTINUE
1 C COMPUTE LINEAR EQUIVALENT SUBSCRIPT FOR JAMMING EVENT
0011 CALL LOCK(M,LOW,LUP,JAM,ISUB)
C LOOK UP THE VALUE, GET 0.0 IF NOT THERE
0012 CALL LOOKUP(PIE,D,IDSUB,MUSED,625,ISUB,STORE,HOME)
0013 RETURN
0014 END

```

```

0001      $ SUBROUTINE GEMPIE(NM,NO,MSLOTS,GOOD,A,IASUB,C,ICSUB,
          $ D,IOSUB,MUSED)
          C SUBROUTINE TO GENERATE EVENT PROBABILITIES
          C
          C IMPLICIT DOUBLE PRECISION (A-H,O-Z)
          LOGICAL*1 GO,GOZ,STORE,MOME,GOOD
          DIMENSION LUP2(4),LUP3(4)
          DIMENSION IUPA(4)
          DIMENSION IUPD(4)
          DIMENSION LUP1(4)
          VIRTUAL A(100),IASUB(100),C(625),ICSUB(625),
          $ D(625),IOSUB(625)
          DIMENSION I(4),II(4),III(4)
          C SHARED STORAGE FOR COMMONLY NEEDED CONSTANT ARRAYS
          COMMON /SHAREZ/ LON(4),LINC(4)
          C SHARED STORAGE FOR:
          C (1) INPUT DEFAULT LISTS,
          C (2) CONDITIONAL PROB GEN., AND
          C (3) EVENT PROB. GEN. THESE ARE NON-OVERLAPPING USAGES.
          COMMON /SHARE/ LUP2,LUP3,LUPD,L1,MUSEA,MUSEC,IERR,
          $ ISUB,ISUB1,ISUB2,AIN,I,II,III,NK,AGUT,
          $ CIN,COU,DOU,DIH
          $
          $ DATA I100/100/
          DATA IUPA/4*1/
          DATA LUP1/4*1/
          C STORE=.FALSE. => DON'T STORE ZERO ELEMENTS OF SPARSE ARRAY.
          STORE=.FALSE.
          GOOD=.TRUE.
          IF(.NOT.LE.O) THEN
            GOOD=.FALSE.
            RETURN
          END IF
          DO 80 LI=1,NM
            IUPD(LI)=2
          80 CONTINUE
          C JAMMING PATTERN #/NON-ZERO PROBABILITY ON PER-HOP BASIS
          MUSEA=0
          C INITIALIZE VECTOR-INDEX LOOP
          CALL VLINIT(I,LOW,NM)
          90 CONTINUE
          CALL LON(MH,LOW,IUPA,I,ISUB)
          CALL PRHOP(I,MH,NO,MSLOTS,AIN)
          CALL PUTIN(AIN,A,IASUB,MUSEA,1100,ISUB,IERR,STORE)
          IF(IERR.NE.O) STOP 3
          C ITERATE VECTOR-INDEX LOOP
          CALL VLITER(I,LOW,LUP1,LINC,NM,GO)
          IF(GO) GOTO 90
          C COMPETITION STARTS HERE. FIRST COPY A INTO D.
          C SINCE ARRAYS ARE A(O:1,O:1,...,O:1) AND D(O:L,O:L,...,O:L)
          C THE COPYING MUST BE DONE ON BASIS OF EQUIVALENT LINEAR
          C SUBSCRIPTS RATHER THAN A SIMPLE MOVE OPERATION.
  
```

```

0033      MUSED=0
          C INITIALIZE VECTOR-INDEX LOOP
          CALL VLINIT(I,LOW,NM)
          99 CONTINUE
          CALL LON(MH,LOW,IUPA,I,ISUB1)
          CALL LON(MH,LOW,LUPD,I,ISUB2)
          CALL LOOKUP(AGUT,A,IASUB,MUSEA,1100,ISUB1,STORE,MOME)
          CALL PUTIN(AGUT,D,IOSUB,MUSED,625,ISUB2,IERR,STORE)
          CALL VLITER(I,LOW,LUP1,LINC,NM,GO)
          IF(GO) GOTO 99
          C ... L-1 CONVOLUTIONS ARE NEEDED ...
          DO 9998 LI=1,I
          C SET UP VECTOR-LOOP UPPER LIMITS FOR THIS CONVOLUTION
          9 DO 125 NM=2,NM
          DO LUP2(NM)=L1
          LUP3(NM)=L1+1
          MUSED=0
          CALL VLINIT(I,LOW,NM)
          98 CONTINUE
          CALL VLINIT(II,LOW,NM)
          97 CONTINUE
          C LOOK UP ELEMENTS AND PERFORM ONE TERM OF THE CONVOLUTION
          CALL LON(MH,LOW,IUPA,I,ISUB1)
          CALL LON(MH,LOW,LUPD,II,ISUB2)
          DO 21 NM=1,NM
            III(NM)=I(NM)+II(NM)
          CONTINUE
          CALL LON(MH,LOW,LUPD,III,ISUB3)
          CALL LOOKUP(AGUT,A,IASUB,MUSEA,1100,ISUB1,STORE,MOME)
          CALL LOOKUP(AGUT,D,IOSUB,MUSED,625,ISUB2,STORE,MOME)
          CALL LOOKUP(AGUT,C,ICSUB,MUSEC,625,ISUB3,STORE,MOME)
          CIN=COU+AGUT*DOU
          CALL PUTIN(CIN,C,ICSUB,MUSEC,625,ISUB3,IERR,STORE)
          IF(IERR.NE.O) STOP 4
          CALL VLITER(II,LOW,LUP2,LINC,NM,GO2)
          IF(GO2) GOTO 97
          CALL VLITER(I,LOW,LUP1,LINC,NM,GO)
          IF(GO) GOTO 98
          C COPY C TO D IN SORTED ORDER FOR NEXT ITERATION
          MUSED=0
          CALL VLINIT(II,LOW,NM)
          96 CONTINUE
          CALL LON(MH,LOW,LUPD,II,ISUB)
          CALL LOOKUP(AGUT,C,ICSUB,MUSEC,625,ISUB,STORE,MOME)
          DIN=COU
          CALL PUTIN(DIN,D,IOSUB,MUSED,625,ISUB,IERR,STORE)
          IF(IERR.NE.O) STOP 5
          CALL VLITER(II,LOW,LUP3,LINC,NM,GO)
          IF(GO) GOTO 96
          9998 CONTINUE
          RETURN
          END
  
```

```

0001 SUBROUTINE PUTIN(CIN,C,ICSUB,MUSE,NMAX,K,IERR,STORE)
C THIS SUBROUTINE INSERTS AN ELEMENT INTO A SPARSE ARRAY FOR
C WHICH ONLY THE NON-ZERO ELEMENTS ARE KEPT IN STORAGE IF
C THE SWITCH STORE IS .TRUE.
C
C THE DOUBLE PRECISION VALUE CIN IS STORED AS C(K) WHERE
C THE AUXILIARY ARRAY ICSUB IS USED TO KEEP TRACK OF THE
C SUBSCRIPTS OF THE CORRESPONDING ENTRIES IN C.
C
C USAGE:
C LOGICAL*1 STORE
C DOUBLE PRECISION C,CIN
C VIRTUAL ICSUB(NMAX),C(NMAX)
C CALL PUTIN(CIN,C,ICSUB,MUSE,NMAX,K,IERR,STORE)
C
C WHERE
C CIN = VALUE OF ELEMENT TO STORE
C C = ARRAY IN WHICH NON-ZERO VALUES ARE ACTUALLY STORED
C ICSUB = AUXILIARY ARRAY FOR ACTUAL SUBSCRIPT VALUES
C MUSE = NUMBER OF ELEMENTS OF C CURRENTLY OCCUPIED
C NMAX = SIZE OF ARRAY C
C IERR = ERROR RETURN CODE, 0 IF NO ERROR OR 1 IF THERE IS
C NO ROOM AVAILABLE IN C
C STORE = .TRUE. TO STORE ZEROS EXPLICITLY, ELSE .FALSE.
C NOTE: IF CIN=0 AND THE SUBSCRIPT K IS FOUND IN ICSUB, THEN
C THE ELEMENT IS DELETED BY SHIFTING DOWNWARD ALL
C FOLLOWING ELEMENTS OF THE ARRAY
C
C PROGRAMMER: ROBERT H. FRENCH DATE: 11 JANUARY 1984
  
```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 VIRTUAL ICSUB(NMAX),C(NMAX)
0004 LOGICAL*1 STORE
0005 IERR=0
0006 IF(STORE) GOTO 5
0007 IF(CIN.EQ.0.DO) GOTO 30
0008 IF(MUSE.EQ.0)GOTO 20
0009 DO 10 I=1,MUSE
0010 IF(ICSUB(I).NE.K) GOTO 10
0011 C(I)=CIN
0012 RETURN
0013 CONTINUE
0014 IF(MUSE.LT.NMAX) GOTO 20
0015 IERR=1
0016 RETURN
0017 MUSE=MUSE+1
0018 ICSUB(MUSE)=K
0019 C(MUSE)=CIN
0020 RETURN
0021 DO 40 I=1,MUSE
0022 J=I
0023 IF(ICSUB(I).EQ.K) GOTO 50
0024 CONTINUE
  
```

```

0025 RETURN
C REMOVE THE ZEROED ELEMENT AND BUMP COUNT OF ENTRIES USED
C
C 50 DO 60 I=J,MUSE-1
0026 ICSUB(I)=ICSUB(I+1)
0027 C(I)=C(I+1)
0028 CONTINUE
0029 MUSE=MUSE-1
0030 RETURN
0031 END
0032
  
```

```

0001 SUBROUTINE LOOKUP(COUT,C,ICSUB,N,NMAX,K,STORE,MOME)
C THIS SUBROUTINE RETRIEVES AN ELEMENT OF A SPARSE ARRAY WHICH
C HAS BEEN STORED COMPACTLY BY STORING ONLY NON-ZERO ELEMENTS.
C
C USAGE:
C VIRTUAL ICSUB(NMAX),C(NMAX)
C LOGICAL*1 STORE,MOME
C DOUBLE PRECISION COUT
C CALL LOOKUP(COUT,C,ICSUB,N,NMAX,K,STORE,MOME)
C
C WHERE
C COUT = VALUE OF C(K) (OUTPUT FROM SUBROUTINE)
C C = ARRAY USED TO STORE NON-ZERO ELEMENTS
C ICSUB = AUXILIARY ARRAY TO STORE ACTUAL SUBSCRIPTS
C N = NUMBER OF ELEMENTS OF C CURRENTLY IN USE
C NMAX = SIZE OF C
C K = SUBSCRIPT OF SPARSE ARRAY TO LOOK UP
C STORE = .TRUE. IF ZEROS STORED EXPLICITLY, ELSE .FALSE.
C MOME = .FALSE. IF ZEROS NOT STORED
C .TRUE. IF ZEROS ARE STORED
C
C PROGRAMMER: ROBERT H. FRENCH DATE: 11 JANUARY 1984
  
```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 VIRTUAL ICSUB(NMAX),C(NMAX)
0004 LOGICAL*1 STORE,MOME
0005 MOME=.FALSE.
0006 DO 10 I=1,N
0007 IF(ICSUB(I).NE.K)GOTO 10
0008 COUT=C(I)
0009 RETURN
0010 CONTINUE
0011 IF(STORE) THEN
0012 MOME=.TRUE.
0013 ELSE
0014 COUT=0.
0015 END IF
0016 RETURN
0017 END
  
```

```

0001 SUBROUTINE LLOC(NDIM, ILOW, IUP, ISUB, LLINEAR)
C THIS SUBROUTINE COMPUTES THE EQUIVALENT LINEAR SUBSCRIPT FOR
C A MULTIDIMENSIONAL ARRAY OF NDIM DIMENSIONS
C IF THE ARRAY A IS DEFINED AS
C DIMENSION A(ILOW(1):IUP(1),...,ILOW(NDIM):IUP(NDIM))
C AND ISUB(1),...,ISUB(NDIM) IS A SET OF SUBSCRIPTS FOR A,
C THEN THIS SUBROUTINE RETURNS IN LINEAR THE OFFSET FROM THE
C ORIGIN OF A TO THE ELEMENT A(ISUB(1),...,ISUB(NDIM)), ASSUMING
C THE FIRST SUBSCRIPT VARIES MOST RAPIDLY.
C USAGE:
C DIMENSION ILOW(NDIM),IUP(NDIM),ISUB(NDIM)
C DATA ILOW/lower limits of defined subscripts of array/
C DATA IUP/upper limits of defined subscripts of array/
C ...SET ISUB TO DESIRED SEQUENCE OF SUBSCRIPTS....
C CALL LLOC(NDIM, ILOW, IUP, ISUB, LLINEAR)
C WHERE
C NDIM = NUMBER OF DIMENSIONS THE ARRAY HAS
C ILOW = ARRAY OF LOWER SUBSCRIPT BOUNDS
C IUP = ARRAY OF UPPER SUBSCRIPT BOUNDS
C ISUB = ARRAY CONTAINING SUBSCRIPT FOR WHICH LOCATION IS
C TO BE COMPUTED
C LLINEAR = RETURNED VALUE OF OFFSET INTO ARRAY IN MEMORY
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
C
C DIMENSION ILOW(NDIM),IUP(NDIM),ISUB(NDIM)
C LLINEAR=0
C DO 10 I=1,NDIM-1
C J=NDIM-I+1
C LLINEAR=(LLINEAR+(ISUB(J)-ILOW(J)))*(IUP(J-1)-ILOW(J-1)+1)
C CONTINUE
C LLINEAR=LLINEAR+ISUB(1)-ILOW(1)
C RETURN
C END
10

```

```

0001 SUBROUTINE VLINIT(LVEC,LLOW,LMAX)
C THIS SUBROUTINE INITIALIZES A "VECTOR DO-LOOP" STRUCTURE
C DEFINED BY THE FOLLOWING PSEUDO-FORTRAN CODE:
C DO 100 LVEC(1)=LLOW(1),LUP(1),LINC(1)
C DO 100 LVEC(2)=LLOW(2),LUP(2),LINC(2)
C
C DO 100 LVEC(LMAX)=LLOW(LMAX),LUP(LMAX),LINC(LMAX)
C
C (STATEMENTS IN RANGE OF LOOP)
C 100 CONTINUE
C THE COMPANION ROUTINE VLITER HANDLES THE LOOP CONTROL AT THE
C CONTINUE STATEMENT IN THE ABOVE STRUCTURE
C USAGE:
C LOGICAL*1 GO
C DIMENSION LVEC(LMAX),LLOW(LMAX),LUP(LMAX),LINC(LMAX)
C (INITIALIZE ARRAY LLOW TO STARTING VALUES OF THE NESTED LOOPS)
C (INITIALIZE ARRAY LUP TO STOPPING VALUES OF THE NESTED LOOPS)
C (INITIALIZE ARRAY LINC TO INCREMENTS OF THE LOOPS)
C CALL VLINIT(LVEC,LLOW,LMAX)
C 100 CONTINUE
C
C ( STATEMENTS IN RANGE OF LOOPS)
C
C CALL VLITER(LVEC,LLOW,LUP,LINC,LMAX,GO)
C IF(GO)GOTO 100
C WHERE
C LVEC = ARRAY FOR STORAGE OF LOOP INDICES. LVEC(1) IS THE
C OUTER-MOST LOOP; LVEC(LMAX), THE INNER-MOST LOOP.
C LLOW = ARRAY FOR STORAGE OF LOOP STARTING VALUES, IN SAME
C SEQUENCE AS LVEC
C LUP = ARRAY FOR STORAGE OF LOOP FINAL VALUES, IN SAME
C SEQUENCE AS LVEC
C LINC = ARRAY FOR STORAGE OF LOOP INCREMENTS, IN SAME
C SEQUENCE AS LVEC
C LMAX = NUMBER OF LOOPS NESTED
C GO = LOGICAL VARIABLE, TRUE IF JUMP BACK TO BEGINNING OF
C STATEMENTS IN THE RANGE OF THE LOOP SHOULD OCCUR,
C .FALSE. OTHERWISE (I.E. OUTER-MOST LOOP TERMINATED)
C PROGRAMMER: ROBERT H. FRENCH DATE: 11 JANUARY 1984
C
C DIMENSION LVEC(LMAX),LLOW(LMAX)
C DO 1 N=1,LMAX
C LVEC(N)=LLOW(N)
C CONTINUE
C RETURN
C END

```

```

0001 SUBROUTINE VLITER(LVEC,LLON,LUP,LINC,LMAX,GO)
C LOOP ITERATION LOGIC FOR A "VECTOR DO-LOOP"
C SEE DETAILED COMMENTS IN SUBROUTINE VLIMIT FOR USAGE AND
C PARAMETER DEFINITIONS
C
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
C
LOGICAL*1 GO
DIMENSION LVEC(LMAX),LLOW(LMAX),LUP(LMAX),LINC(LMAX)
GO=.TRUE.
DO 100 I=1,LMAX
  NSUB=LMAX+1-NDX
  LVEC(NSUB)=LVEC(NSUB)+LINC(NSUB)
  IF((LINC(NSUB).GE.0.AND.LVEC(NSUB).LE.LUP(NSUB))
    $ .OR.(LINC(NSUB).LT.0.AND.LVEC(NSUB).GE.LUP(NSUB))) RETURN
  CONTINUE
100 GO=.FALSE.
RETURN
END

```

```

0001 SUBROUTINE PR1HOP(I,KM,KO,KH,AJM)
C THIS SUBROUTINE COMPUTES THE EVENT PROBABILITY FOR ALL
C POSSIBLE JAMMING PATTERNS WITH NON-ZERO PROBABILITY FOR
C L=1 HOP/SYMBOL FOR RHF5K/FH IN PBWJ
C
C IMPLICIT DOUBLE PRECISION (A-H,O-Z)
DIMENSION I(4)
AJM=0.DO
KJAM=0
DO 1 K=1,KH
  KJAM=KJAM+I(K)
  CONTINUE
1
C IF THIS IS AN IMPOSSIBLE CASE, RETURN WITH RESULT = 0.0
IF(KJAM.GT.MIND(KO,KH)) RETURN
KPMAX=KJAM-1
LPMAX=KH-KJAM-1
JPMAX=KH-1
IMAX=MAX0(KPMAX,LPMAX,JPMAX)
PROD=1.DO
O=KO
DO 100 LOOP=0,IMAX
  DIFF=KN-KO
  EN=KN
  F=LOOP
  IF(LOOP.LE.KPMAX) PROD=PROD*(O-F)
  IF(LOOP.LE.JPMAX) PROD=PROD/(EN-F)
  IF(LOOP.LE.LPMAX) PROD=PROD*(DIFF+O-F)
  CONTINUE
100 AIM=PROD
RETURN
END

```

```

0001 SUBROUTINE FSEL(JSUB,MM,PROB)
C
C RANDON MFSK/FH IN PARTIAL BAND NOISE JAMMING.
C GIVEN A JAMMING EVENT, WITH CLIPPER RECEIVER
C
C JSUB - JAMMING EVENT VECTOR
C MM - ALPHABET SIZE
C PROB - RESULTING CONDITIONAL ERROR PROBABILITY
C

```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 DIMENSION JSUB(MM), WORK(30), STACK(30), HEAP(30)
0004 EXTERNAL D616, PERAND
0005 INTEGER NCHAN(0:3)
0006 COMMON /JAMCHT/ NCHAN
0007 COMMON /DEMPAR/ B1GK, AAB, BAB, L,JAM,
      TAU, TAU2, TAU3, TAU4, TAU5
0008 COMMON /PARDEN/ RHOM, RHOT
0009 COMMON /QUES/ Q0, Q1
0010 COMMON /SCJAM/ JAMSC
0011 KSUB=0
0012 DO 6 I=1,MM
0013 KSUB=KSUB+JSUB(I)
0014 6 CONTINUE
C SET UP VALUES WHICH WILL REMAIN IF THIS IS THE NOTHING-JAMMED CASE
0015 Q0=Q(2.00*DSORT(0.500*RHOM),DSORT(2.00*TAU))
0016 Q1=1.00
0017 TAU=0.00
C IF ANYTHING IS JAMMED, SET UP JAMMING-RELATED QUANTITIES
0018 IF(KSUB.NE.0) THEN
0019 B1GK=RHOM/RHOT
0020 BK1=B1GK-1.00
0021 AAB=B1GK/BK1
0022 BAB=1.00/BK1
0023 TAU=TAU/B1GK
0024 TAU2=TAU2/B1GK
0025 Q1=Q(2.00*DSORT(0.500*RHOT),DSORT(2.00*TAU))
0026 END IF
C
C COUNT NUMBER OF NONSIGNAL CHANNELS WITH LM HOPS JAMMED
C
DO 10 I=0,2
NCHAN(I)=0
CONTINUE
10 DO 11 J=2,MM
KSUB=JSUB(I)
NCHAN(KSUB)=NCHAN(KSUB)+1
CONTINUE
11 JAMSC=JSUB(I)
C
C DO THE CONTINUOUS PART OF THE DENSITY IN SECTIONS
C
PCORT=0.00
0035

```

```

0001 BLOCK DATA
C INITIALIZE SHARED CONSTANTS
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 COMMON /SHARE/ DSNR(3,4)
0004 COMMON /SHARE2/ LOM(4),LINC(4)
0005 COMMON /MTS/ X(8),M(8)
C
C WEIGHTS AND ABSISSAS FOR 16-POINT GAUSSIAN QUADRATURE
C
DATA X/ 0.09501250983763744018500,
$ 0.28160355077925891323000,
$ 0.4580167765722738634200,
$ 0.6178762440264374844700,
$ 0.75540440835500303389500,
$ 0.86563120238783174388000,
$ 0.94457502307323257607800,
$ 0.98940093499164993259600 /
DATA M/ 0.18945061045506849628500,
$ 0.18260341504492358886700,
$ 0.16915651939500253818900,
$ 0.14959598881657673206100,
$ 0.1246289712553387205200,
$ 0.09515851168249278481000,
$ 0.06225352393864789286300,
$ 0.02715245941175409485200 /
C DEFAULT LISTS FOR INTERACTIVE PARAMETER INPUTS
C ARE SHARED WITH LARGE WORKING STORAGE ARRAYS SINCE THEY
C MAY BE DESTROYED ONCE THE INPUT PARAMETERS ARE SET UP
DATA DSNR /13.3524700, 12.313300, 10.9444300,
$ 10.60657200, 9.628400, 8.3524800,
$ 9.0940100, 8.169600, 6.97199500,
$ 8.0793500, 7.199600, 6.06964600/
C FREQUENTLY NEEDED CONSTANT ARRAYS AND SCALARS
DATA LOM/4*0/,LINC/4*1/
END
0009
0010

```



```

0028 IF(EPS.EQ.0.00) THEN
0029   KODE=2
0030   RETURN
0031 END IF
0032 STACK(MPTS)=EPS
0033 GOTO 10
C FINISHED A PIECE
0034 20
0035 Y=Y+PI*P2
0036 EPS=STACK(MPTS)
0037 T=HEAP(MPTS)
0038 A=B
0039 IF(MPTS.EQ.0) RETURN
0040 GOTO 10
0041 END
  
```

```

0001 SUBROUTINE ADQUA2(XL,XU,Y,QR,F,TOL,WORK,STACK,HEAP,N,KODE)
C ADAPTIVE QUADRATURE ALGORITHM
C XL - LOWER LIMIT OF INTEGRAL (IN)
C XU - UPPER LIMIT OF INTEGRAL (IN)
C Y - VALUE OF INTEGRAL (OUT)
C QR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
C WITH CALLING SEQUENCE
C CALL QR(XL,XU,F,Y)
C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
C WORK - WORK ARRAY OF SIZE N (IN)
C STACK - SECOND WORK ARRAY OF SIZE N, MUST NOT BE
C HEAP - THIRD WORK ARRAY, SIZE N, DISTINCT FROM WORK AND STACK
C SAME ARRAY AS WORK (IN)
C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (IN)
C KODE - ERROR INDICATOR (OUT)
C 0 -- NO ERROR
C 1 -- WORK ARRAYS TOO SMALL
C 2 -- EPS DIVIDED TO ZERO, EITHER ASKING FOR TOO
C TIGHT A TOLERANCE OR ROUND-OFF PREVENTS
C ATTAINING REQUIRED ACCURACY
C R. H. FRENCH, 14 AUGUST 1984
C
  
```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 EXTERNAL F
0004 DIMENSION WORK(N),STACK(N),HEAP(N)
0005 KODE=0
0006 Y=0.00
0007 WORK(1)=XU
0008 CALL QR(XL,XU,F,T)
0009 HEAP(1)=T
0010 A=XL
0011 MPTS=1
0012 EPS=TOL
0013 STACK(1)=EPS
0014 B=WORK(MPTS)
0015 XM=(A+B)*0.500
0016 CALL QR(A,XM,F,P1)
0017 CALL QR(XM,B,F,P2)
0018 IF(DABS(T-P1-P2).LE.EPS) GOTO 20
C SPLIT IT
MPTS=MPTS+1
IF(MPTS.GT.N) THEN
KODE=1
RETURN
END IF
WORK(MPTS)=XM
HEAP(MPTS)=P2
T=P1
EPS=EPS/2.00
  
```

```

0028 IF(EPS.EQ.0.DO) THEN
0029   KODE=2
0030   RETURN
0031 END IF
0032 STACK(NPTS)=EPS
0033 GOTO 10

C FINISHED A PIECE
0034 Z0
0035 Y=Y+PI*P2
0036 EPS=STACK(NPTS)
0037 T=HEAP(NPTS)
0038 NPTS=NPTS-1
0039 A=B
0040 IF(NPTS.EQ.0) RETURN
0041 END
  
```

```

0001 SUBROUTINE DG16(A,B,F,ANSWER)
C
C 16-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
C R. H. FRENCH, 28 FEBRUARY 1986
C
      IMPLICIT DOUBLE PRECISION (A-H,O-Z)
      COMMON /NTS/ X(8),M(8)
      ANSWER=0.00
      BMAO2=(B-A)/2.DO
      BPAO2=(B+A)/2.DO
      DO 10 I=1,8
      C=X(I)*BMAO2
      Y1=BPAO2+C
      Y2=BPAO2-C
      ANSWER=ANSWER+W(I)*(F(Y1)+F(Y2))
0011 CONTINUE
0012 ANSWER=ANSWER*BMAO2
0013 RETURN
0014 END
0015
  
```

```

0001 SUBROUTINE DGXVI(A,B,F,ANSWER)
C
C 16-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
C R. H. FRENCH, 28 FEBRUARY 1986
C
      IMPLICIT DOUBLE PRECISION (A-H,O-Z)
      COMMON /NTS/ X(8),M(8)
      ANSWER=0.00
      BMAO2=(B-A)/2.DO
      BPAO2=(B+A)/2.DO
      DO 10 I=1,8
      C=X(I)*BMAO2
      Y1=BPAO2+C
      Y2=BPAO2-C
      ANSWER=ANSWER+W(I)*(F(Y1)+F(Y2))
0011 CONTINUE
0012 ANSWER=ANSWER*BMAO2
0013 RETURN
0014 END
0015
  
```

```
0001 SUBROUTINE TEST(ID)
      C TEST RETURN CODE FROM BESSEL FUNCTION
      C
      C TEST RETURN CODE FROM ADQUAD/ADQUAZ
      C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KCODE, I01, I02
      IF(KCODE.EQ.0) RETURN
      WRITE(5,1) KCODE, ID
      1 FORMAT(' BESSEL FUNCTION CODE = ',I2,
             ' FROM CALL NUMBER ',I5)
      $ STOP 'FATAL ERROR'
      END
```

```
0002 SUBROUTINE TEST2(KODE, ID)
      C TEST RETURN CODE FROM ADQUAD/ADQUAZ
      C
      IF(KODE.EQ.0) RETURN
      WRITE(5,1) KODE, ID
      1 FORMAT(' ADAPTIVE INTEGRATOR CODE = ',I2,
             ' FROM CALL NUMBER ',I5)
      $ STOP 'FATAL ERROR'
      END
```

```

0001      DOUBLE PRECISION FUNCTION PZ1(Y)
          C
          C   SIGNAL CHANNEL P.D.F. WITH CHANGE OF VARIABLE Y=X
          C
          C   IMPLICIT DOUBLE PRECISION (A-H,O-Z)
          C   DIMENSION WORK(30), STACK(30), HEAP(30)
          C   LOGICAL *1 REG1, REG2
          C   EXTERNAL DGXV1, F20, F21, F22
          C   COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
          C   COMMON /DENPAR/ BTGK, AAB, BAB, LJAM,
          C   TAU, TAU2, TALK, TALK2
          C   COMMON /PARDEN/ RHOM, RHOT
          C   COMMON /QUES/ QO, Q1
          C   COMMON /XCOM/ XCOM
          C   COMMON /OUTER/ XXX, XXXX
          C   XXX=Y
          C   IF (LJAM.GE.1) THEN
          C     XXXX=Y/BTGK
          C     YK=XXXX
          C     YTK=(Y-TAU)/BTGK
          C     YTK2=(Y-TAU2)/BTGK
          C   END IF
          C   REG1=Y.GE.0.DO .AND. Y.LT.TAU
          C   REG2=Y.GE.TAU .AND. Y.LT.TAU2
          C
          C   TWO HOPS PER SYMBOL
          C
          C   GOTO (2100, 2200, 2300), LJAM+1
          C
          C   NO HOPS JAMMED
          C
          C   IF (REG1) THEN
          C     BARG1=DSQRT(8.DO*RHOT*YK)
          C     I01=1
          C     CALL DXBT(2100)
          C     PZ1=0.500*DSQRT(2.DO*Y/RHOM)*DEXP(BARG1-Y-2.DO*RHOM)*B1
          C   ELSE IF (REG2) THEN
          C     BARG1=DSQRT(4.DO*RHOM*(Y-TAU))
          C     I01=0
          C     CALL DXBT(2101)
          C     PART=2.DO*Q1*DEXP(BARG1-Y+TAU-RHOM)*B1
          C     XCOM=Y+2.DO*RHOM
          C     CALL ADQUAZ(Y-TAU,TAU,ANSWER,DGXV1,F20,1.D-9,WORK,STACK,
          C       HEAP,30,K00)
          C     CALL TEST2(K00,2100)
          C     PZ1=PART+ANSWER
          C   ELSE
          C     PZ1=0.DO
          C   END IF
          C   CONTINUE
          C   RETURN
          C   END
  
```

```

0001      ONE HOP JAMMED
          C
          C   IF (REG1) THEN
          C     XCOM=YK+RHOM+RHOT
          C     CALL ADQUAZ(0.DO,Y,ANSWER,DGXV1,F21,1.D-9,WORK,STACK,
          C       HEAP,30,K00)
          C     CALL TEST2(K00,2200)
          C     PZ1=ANSWER/BTGK
          C   ELSE IF (REG2) THEN
          C     BARG1=DSQRT(4.DO*RHOT*YTK)
          C     I01=0
          C     CALL DXBT(2201)
          C     PART=QO*DEXP(BARG1-YK+TALK-RHOT)*B1/BTGK
          C     BARG1=DSQRT(4.DO*RHOM*(Y-TAU))
          C     I01=0
          C     CALL DXBT(2202)
          C     PART=PART+Q1*DEXP(BARG1-Y+TAU-RHOM)*B1
          C     XCOM=YK+RHOM+RHOT
          C     CALL ADQUAZ(Y-TAU,TAU,ANSWER,DGXV1,F21,1.D-9,WORK,STACK,
          C       HEAP,30,K00)
          C     CALL TEST2(K00,2202)
          C     PZ1=PART+ANSWER/BTGK
          C   ELSE
          C     PZ1=0.DO
          C   END IF
          C   GOTO 9000
          C
          C   TWO HOPS JAMMED
          C
          C   IF (REG1) THEN
          C     BARG1=DSQRT(8.DO*RHOT*YK)
          C     I01=1
          C     CALL DXBT(2300)
          C     PZ1=0.500*DSQRT(2.DO*Y/RHOM)*DEXP(BARG1-YK-2.DO*RHOT)
          C       *B1/BTGK
          C   ELSE IF (REG2) THEN
          C     BARG1=DSQRT(4.DO*RHOT*YTK)
          C     I01=0
          C     CALL DXBT(2301)
          C     PART=2.DO*Q1*DEXP(BARG1-YTK-RHOT)*B1/BTGK
          C     XCOM=YK+2.DO*RHOT
          C     CALL ADQUAZ(Y-TAU,TAU,ANSWER,DGXV1,F22,1.D-9,WORK,STACK,
          C       HEAP,30,K00)
          C     CALL TEST2(K00,2300)
          C     PZ1=PART+ANSWER/(BTGK*BTGK)
          C   ELSE
          C     PZ1=0.DO
          C   END IF
          C   CONTINUE
          C   RETURN
          C   END
  
```

```

0001 DOUBLE PRECISION FUNCTION GL(Y)
C
C NON-SIGNAL CHANNEL CUMULATIVE DISTRIBUTION FUNCTION
C WITH CHANGE OF VARIABLE Y=AX
C
      IMPLICIT DOUBLE PRECISION (A-H,O-Z)
      LOGICAL*1 REG1, REG2, REG3
      COMMON /DENPAR/ BIG1, AAB, BAB, LJAM,
      TAU, TAU2, TAU3, TAU4, TAU5
      REG1=Y.GE.O.DO .AND. Y.LT.TAU
      REG2=Y.GE.TAU .AND. Y.LT.TAU2
      IF(LJAM.GT.O) THEN
        YK=Y/BIG1
      ELSE
        YK=Y
      END IF
      YTK=(Y-TAU)/BIG1
      YTK2=(Y-TAU2)/BIG1
      END IF
C
C TWO HOPS PER SYMBOL
C
      GOTO (2100, 2200, 2300), LJAM+1
C
C NO HOPS JAMMED
C
      IF(REG1) THEN
        GL=1.DO-(1.DO+Y)*DEXP(-Y)
      ELSE IF(REG2) THEN
        GL=1.DO-(1.DO+TAU2-Y)*DEXP(-Y)
      ELSE IF(Y.GE.TAU2) THEN
        GL=1.DO
      ELSE
        GL=0.DO
      END IF
      GOTO 9000
C
C ONE HOP JAMMED
C
      IF(REG1) THEN
        GL=1.DO-AAB*DEXP(-YK)+BAB*DEXP(-Y)
      ELSE IF(REG2) THEN
        GL=1.DO-AAB*DEXP(TAU-TAUK-Y)+BAB*DEXP(TAUK-TAU-YK)
      ELSE
        GL=1.DO
      END IF
      GOTO 9000
C
C TWO HOPS JAMMED
C
      IF(REG1) THEN
        GL=1.DO-(1.DO+YK)*DEXP(-YK)
      ELSE
        GL=1.DO
      END IF
      GOTO 9000
C
      END FUNCTION GL

```

```

0001 DOUBLE PRECISION FUNCTION F20(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=2, LJAM=0
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /DENPAR/ BIG1, AAB, BAB, LJAM,
      TAU, TAU2, TAU3, TAU4, TAU5
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /QUES/ QO, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      BARG1=DSORT(4.DO+RHOM*U)
      BARG2=DSORT(4.DO+RHOM*(XXX-U))
      I01=0
      I02=0
      CALL BPROO(2110)
      F20=DEXP(BARG1+BARG2-XCOM)*B1*B2
      RETURN
      END

```

```

0001 DOUBLE PRECISION FUNCTION F20(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=2, LJAM=0
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /DENPAR/ BIG1, AAB, BAB, LJAM,
      TAU, TAU2, TAU3, TAU4, TAU5
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /QUES/ QO, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      BARG1=DSORT(4.DO+RHOM*U)
      BARG2=DSORT(4.DO+RHOM*(XXX-U))
      I01=0
      I02=0
      CALL BPROO(2110)
      F20=DEXP(BARG1+BARG2-XCOM)*B1*B2
      RETURN
      END

```

```

0001 DOUBLE PRECISION FUNCTION F20(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=2, LJAM=0
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /DENPAR/ BIG1, AAB, BAB, LJAM,
      TAU, TAU2, TAU3, TAU4, TAU5
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /QUES/ QO, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      BARG1=DSORT(4.DO+RHOM*U)
      BARG2=DSORT(4.DO+RHOM*(XXX-U))
      I01=0
      I02=0
      CALL BPROO(2110)
      F20=DEXP(BARG1+BARG2-XCOM)*B1*B2
      RETURN
      END

```

```

0001 DOUBLE PRECISION FUNCTION F20(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=2, LJAM=0
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /DENPAR/ BIG1, AAB, BAB, LJAM,
      TAU, TAU2, TAU3, TAU4, TAU5
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /QUES/ QO, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      BARG1=DSORT(4.DO+RHOM*U)
      BARG2=DSORT(4.DO+RHOM*(XXX-U))
      I01=0
      I02=0
      CALL BPROO(2110)
      F20=DEXP(BARG1+BARG2-XCOM)*B1*B2
      RETURN
      END

```

```

0001      DOUBLE PRECISION FUNCTION F22(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=2, LJAM=2
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM,
      &          TAU, TAU2, TALK, TALK2
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /QUES/ QO, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      BARG1=DSORT(4.DO+RHOT*U/BIGK)
      BARG2=DSORT(4.DO+RHOT*(XXXX-U/BIGK))
      I01=0
      I02=0
      CALL BPROD(2310)
      F22=DEXP(BARG1+BARG2-XCOM)*B1*B2
      RETURN
      END
  
```

```

0001      SUBROUTINE DXBT(ID)
C
C CALL DXBESI AND TEST RETURN CODE
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      CALL DXBESI(BARG1,I01,B1,KODE)
      CALL TEST(ID)
      RETURN
      END
  
```

```

0001      DOUBLE PRECISION FUNCTION F21(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=2, LJAM=1
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM,
      &          TAU, TAU2, TALK, TALK2
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /QUES/ QO, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      BARG1=DSORT(4.DO+RHOM*U)
      BARG2=DSORT(4.DO+RHOT*(XXXX-U/BIGK))
      I01=0
      I02=0
      CALL BPROD(2210)
      F21=DEXP(BARG1+BARG2-XCOM-U/U/BIGK)*B1*B2
      RETURN
      END
  
```

```

0001      SUBROUTINE BPROD(IDENT)
C
C COMPUTE TWO BESSEL FUNCTIONS, ARGUMENTS AND RESULTS IN COMMON
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      CALL DXBT(IDENT)
      CALL DXBESI(BARG2,I02,B2,KODE)
      CALL TEST(IDENT+1)
      RETURN
      END
  
```

0001 C FUNCTION FOR UNJAMMED P(E) FOR OPT. THRESHOLD SEARCH

0002 C NOTE: WHEN JAMMING EVENT IS (0,0,....,0), THE VARIABLES

0003 C BIGK, AAB, BAB, TALK, AND TALK2 ARE NOT

0004 C USED IN THE COMPUTATIONS.

0005 C IMPLICIT DOUBLE PRECISION(A-H,O-Z)

0006 C DIMENSION NOJAM(4)

0007 C COMMON /DENPAR/ BIGK, AAB, BAB, LJAM,

0008 C COMMON /PARDEN/ RHOM, RHOT

0009 C COMMON /QUES/ Q0, Q1

0010 C MRU=MM-1

0011 C CUEO=DEXP(-TAU)

0012 C CUE1=DEXP(-TAUK)

0013 C P1L=DXI(Q0,2-JSUB(1))*DXI(Q1,JSUB(1))

0014 C DO 10 I=2,MM

0015 C P2LM(I)=DXI(CUEO,2-JSUB(I))*DXI(CUE1,JSUB(I))

0016 C 10 CONTINUE

0017 C C SET UP VECTOR LOOP PARAMETERS

0018 C DO 20 I=1,MM-1

0019 C LLOW(I)=0

0020 C LINC(I)=1

0021 C LUP(I)=1

0022 C CONTINUE

0023 C PTIE=0.00

0024 C CALL VLIMIT(MU,LLON,MM-1)

0025 C MUSUM=0

0026 C DO 40 I=1,MM-1

0027 C MUSUM=MUSUM+MU(I)

0028 C CONTINUE

0029 C FRAC=1.00/(1.00+MUSUM)

0030 C PROD=1.00

0031 C DO 50 M=2,MM

0032 C IF(MU(M-1).EQ.1) THEN

0033 C PROD=PROD*P2LM(M)

0034 C ELSE

0035 C PROD=PROD*(1.00-P2LM(M))

0036 C END IF

0037 C PTIE=FRAC*PROD*PTIE

0038 C CALL VLIMIT(MU,LLON,LUP,LINC,MM-1,60)

0039 C IF(GO) GO TO 30

0040 C PTIE=PTIE*P1L

0041 C RETURN

0042 C END

0001 C DOUBLE PRECISION FUNCTION PUNJAM(ETA)

0002 C FUNCTION FOR UNJAMMED P(E) FOR OPT. THRESHOLD SEARCH

0003 C NOTE: WHEN JAMMING EVENT IS (0,0,....,0), THE VARIABLES

0004 C BIGK, AAB, BAB, TALK, AND TALK2 ARE NOT

0005 C USED IN THE COMPUTATIONS.

0006 C IMPLICIT DOUBLE PRECISION(A-H,O-Z)

0007 C DIMENSION NOJAM(4)

0008 C COMMON /DENPAR/ BIGK, AAB, BAB, LJAM,

0009 C COMMON /PARDEN/ RHOM, RHOT

0010 C COMMON /QUES/ Q0, Q1

0011 C MRU=MM-1

0012 C CUEO=DEXP(-TAU)

0013 C CUE1=DEXP(-TAUK)

0014 C P1L=DXI(Q0,2-JSUB(1))*DXI(Q1,JSUB(1))

0015 C DO 10 I=2,MM

0016 C P2LM(I)=DXI(CUEO,2-JSUB(I))*DXI(CUE1,JSUB(I))

0017 C 10 CONTINUE

0018 C C SET UP VECTOR LOOP PARAMETERS

0019 C DO 20 I=1,MM-1

0020 C LLOW(I)=0

0021 C LINC(I)=1

0022 C LUP(I)=1

0023 C CONTINUE

0024 C PTIE=0.00

0025 C CALL VLIMIT(MU,LLON,MM-1)

0026 C MUSUM=0

0027 C DO 40 I=1,MM-1

0028 C MUSUM=MUSUM+MU(I)

0029 C CONTINUE

0030 C FRAC=1.00/(1.00+MUSUM)

0031 C PROD=1.00

0032 C DO 50 M=2,MM

0033 C IF(MU(M-1).EQ.1) THEN

0034 C PROD=PROD*P2LM(M)

0035 C ELSE

0036 C PROD=PROD*(1.00-P2LM(M))

0037 C END IF

0038 C PTIE=FRAC*PROD*PTIE

0039 C CALL VLIMIT(MU,LLON,LUP,LINC,MM-1,60)

0040 C IF(GO) GO TO 30

0041 C PTIE=PTIE*P1L

0042 C RETURN

0043 C END

0001 C SUBROUTINE SETTAU(MM,PEOO)

0002 C SEARCH FOR OPTIMUM THRESHOLD IN ABSENCE OF JAMMING

0003 C EXTERNAL PUNJAM

0004 C COMMON /DENPAR/ BIGK, AAB, BAB, LJAM,

0005 C COMMON /PARDEN/ RHOM, RHOT

0006 C COMMON /QUES/ Q0, Q1

0007 C LJAM=0

0008 C GUESS BASED ON QUADRATIC CURVE FIT

0009 C IF(MM.EQ.2) THEN

0010 C GUESS=0.92500*4-8.47500*2+32.4500

0011 C ELSE IF(MM.EQ.4) THEN

0012 C GUESS=1.05000*4-9.35000*2+34.500

0013 C ELSE IF(MM.EQ.8) THEN

0014 C GUESS=1.10000*4-9.90000*2+36.300

0015 C ELSE

0016 C GUESS=15.00

0017 C END IF

0018 C CALL MINSER(PUNJAM,PEMIN,TAUKOPT,1.00,GUESS,0.00,

0019 C 50.00,0.0100)

0020 C TAU=TAUKOPT

0021 C TAU2=TAU+TAU

0022 C PE00=PEMIN

0023 C RETURN

0024 C END

0001 C SEARCH FOR OPTIMUM THRESHOLD IN ABSENCE OF JAMMING

0002 C EXTERNAL PUNJAM

0003 C COMMON /DENPAR/ BIGK, AAB, BAB, LJAM,

0004 C COMMON /PARDEN/ RHOM, RHOT

0005 C COMMON /QUES/ Q0, Q1

0006 C LJAM=0

0007 C GUESS BASED ON QUADRATIC CURVE FIT

0008 C IF(MM.EQ.2) THEN

0009 C GUESS=0.92500*4-8.47500*2+32.4500

0010 C ELSE IF(MM.EQ.4) THEN

0011 C GUESS=1.05000*4-9.35000*2+34.500

0012 C ELSE IF(MM.EQ.8) THEN

0013 C GUESS=1.10000*4-9.90000*2+36.300

0014 C ELSE

0015 C GUESS=15.00

0016 C END IF

0017 C CALL MINSER(PUNJAM,PEMIN,TAUKOPT,1.00,GUESS,0.00,

0018 C 50.00,0.0100)

0019 C TAU=TAUKOPT

0020 C TAU2=TAU+TAU

0021 C PE00=PEMIN

0022 C RETURN

0023 C END

```

0001 C SUBROUTINE MINSER(F,FMIN,XMIN,STEP,GUESS,BLIM,ULIM,TOL)
C SEARCH FOR MINIMUM OF F(X) OVER THE INTERVAL BLIM <= X <= ULIM
C TROUBLE MAY OCCUR IF F(X) HAS MULTIPLE LOCAL MINIMA WITHIN THE
C SEARCH INTERVAL OR IF THE FUNCTION IS VERY STEEP AND STEP IS
C TOO BIG.
C
C F = NAME OF FUNCTION TO BE MINIMIZED
C FMIN = MINIMUM VALUE OF F(X) OVER INTERVAL
C XMIN = VALUE OF X FOR WHICH FMIN OCCURS
C STEP = INITIAL STEP SIZE FOR SEARCH
C GUESS = INITIAL GUESS AT XMIN, BLIM <= GUESS <= ULIM
C BLIM = LOWER LIMIT OF SEARCH INTERVAL
C ULIM = UPPER LIMIT OF SEARCH INTERVAL
C TOL = TOLERANCE ON XMIN; SEARCH STOPS WHEN DX < TOL
C
C NOTE: F MUST BE A DOUBLE PRECISION FUNCTION OF ONE DOUBLE PRECISION
C ARGUMENT. ANY PARAMETERS CAN BE PASSED FROM THE CALLER VIA
C A COMMON BLOCK.
C
C PROGRAMMER: ROBERT H. FRENCH DATE: 17 MARCH 1986
C
C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
X=GUESS
SLMAX=DABS(X-BLIM)
SUMAX=DABS(ULIM-X)
DX=DMIN1(STEP,SLMAX,SUMAX)
TEST=TOL
10 FO=F(X)
F1=F(X+DX)
C ARE WE GOING IN THE RIGHT DIRECTION?
IF(F1.LE.FO) GOTO 100
C ... NO, SWITCH DIRECTION
DX=-DX
F1=F(X+DX)
IF(F1.LE.FO) GOTO 100
C ELSE WE MUST BE CLOSE TO A MIN. AT X=GUESS, SO CUT
C STEP SIZE AND TRY AGAIN
DX=DX/10.DO
IF(DABS(DX).GE.TEST) GOTO 10
C CLOSE ENOUGH AT GUESS
12 XMIN=X
FMIN=FO
RETURN
C
C NOW GOING RIGHT DIRECTION.
C KEEP GOING UNTIL PAST MINIMUM BY ONE STEP.
C
100 X2=X+DX+DX
C HAVE WE REACHED END POINT?
IF(X2.GT.ULIM .OR. X2.LT.BLIM) GOTO 110

```

```

0021 C ALL OK
105 F2=F(X2)
C PAST MIN?
IF(F2.GE.F1) GOTO 200
C ...NO, STEP AGAIN
FO=F1
F1=F2
X=X+DX
GOTO 100
C MIN MAY BE AT AN ENDPOINT. CUT STEP SIZE AND TRY AGAIN
C IF INCREMENT NOT TOO SMALL.
110 IF(DABS(DX).LE.TEST) GOTO 120
X=X+DX
FO=F1
DX=DX/10.DO
F1=F(X+DX)
GOTO 100
C MIN MUST BE AT THE ENDPOINT (OR WITHIN MINIMUM DX THEREOF)
120 IF(X2.LE.BLIM) GOTO 122
C MIN AT X=ULIM
XMIN=ULIM
FMIN=F(XMIN)
RETURN
C MIN AT BLIM
122 XMIN=BLIM
GOTO 121
C HAVE PASSED MIN. IS IT LOCATED CLOSELY ENOUGH YET?
200 IF(DABS(DX).LE.TEST) GOTO 300
C ... NO, CUT STEP SIZE AND TRY AGAIN
GOTO 115
C DONE!
C SINCE FO >= F1 & F2 >= F1 AND ABS(DX)<MIN. DX, CALL F1 THE MIN.
300 FMIN=F1
XMIN=X+DX
RETURN
END
0041
0042
0043
0044

```


APPENDIX J
COMPUTER PROGRAM FOR
CLIPPER RECEIVER WITH L=3 HOPS/SYMBOL

The following pages contain the source code listing for the FORTRAN-77 program used for numerical computations of the performance of the clipper receiver when L=3 hops/symbol, using a numerical search for the worst-case jamming fractions. If $M > 4$ the sizes of the arrays used in computing event probabilities must be increased and the corresponding array-size parameters in calls to PUTIN and LOOKUP changed accordingly.

```

0001      PROGRAM CLIP3R
C THIS PROGRAM COMPUTES THE ERROR PROBABILITY FOR RANDOM M-ARY
C FSK/FH WITH 3 HOPS/BIT IN THE PRESENCE OF PARTIAL-BAND
C NOISE JAMMING BY NUMERICAL INTEGRATION FOR THE CLIPPER RECEIVER
C AND BETWEEN-CONDITIONAL-PROB. RESTART CAPABILITY
C
C ANALYSIS: L. E. MILLER, R. H. FRENCH
C PROGRAM: R. H. FRENCH
C
C V 3.1.0 - COMPUTATIONS ONLY
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      CHARACTER (LJ=11)
      CHARACTER*1 YES, NO, REPLY, BLANK
      CHARACTER*13 FNAME, GAMME
      LOGICAL DOTAU, TEST
      LOGICAL*1 GOOD, RESTR1
      DIMENSION POFQ(50), IOL(50)
      REAL*4 PLOG(LJ), DBSJR(LJ), QOPT(LJ)
      VIRTUAL A(100), IASUB(100), C(625), ICSUB(625)
      VIRTUAL D(625), IDSUB(625), PRIERR(625), IPSUB(625)
C COMMON /INPUTS/ PASSES PARAMETERS OF THE RUN
      COMMON /INPUTS/ DEBNOL(3), NSLOTS,K,MM
C COMMON /SIZE/ PASSES NUMBERS OF PARAMETERS
      COMMON /SIZE/ NO
      COMMON /ENPAR/ BIGK, AAB, BAB, LJM, AAB2, BAD2, AABAB,
      TAU, TAU2, TAU3, TALK, TALK2, TALK3
C COMMON /PARDEN/ RHOX, RHOY
C COMMON /RSPAR/ JAM(4), JAMI, MPS, IJ, RESTR1
      DATA YES, NO, BLANK /'Y', 'N', ' ' /
      CALL JSLGO
      CALL ERSET(29, .TRUE., .FALSE., .TRUE., .FALSE., .15)
      SLOTS=NSLOTS
      WORBIT=0.500*MM/(MM-1.00)
      DO 800 IO=1, NO
      DOTAU=.TRUE.
      EBNO=10.00** (DEBNOL(IO)/10.00)
      RHOX=K*EBNO/3.00
      IODUT=DEBNOL(IO)
C OPEN DATA FILE
C
      WRITE(FNAME,730) MM, IODUT
      FORMAT('COJ', I1, '3', I2, 2, '.DAT')
      WRITE(6,776) MM, DEBNOL(IO)
      FORMAT('CLIPPER RECEIVER, OPTIMUM GAMMA RESULTS /
      ' M= ', I2, 'S= ', I3, 'X= ', EB/NO= ', FR. A//', EB/NO (dB)',
      ' 5X', 'P(e)'.15X, 'Qopt')
      WRITE(5,733) FNAME
      FORMAT(' WORKING ON ', A13)
      OPEN(UNIT=4, FILE=FNAME, STATUS='OLD', FORM='UNFORMATTED',
      ERR=750)
  
```

```

0003      C HAVE AN EXISTING FILE, READ TO SEE HOW FAR WE GOT BEFORE
0035      READ(4) MMJM, EBNOMN, NSLTN, TAU, TAU2, TAU3, PEOD
C WE HAVE READ A VALUE OF TAU, SO WE NOW'NT NEED TO RECOMPUTE IT UNTIL
C EITHER EB/NO, MM, OR LL CHANGES
C
      DOTAU=.FALSE.
      JJ=0
      740      JJ=JJ+1
      READ(4, END=742) DBSJR(JJ), PLOG(JJ), QOPT(JJ)
      GOTO 740
      742      CLOSE(UNIT=4)
      $      OPEN(UNIT=4, FILE='RESUME.DAT', STATUS='UNKNOWN',
      READ(4) JAM, JAMI, PRERR, MPS, IPSUB, JJJM
      CLOSE(UNIT=4)
      RESTR1=JJ.EQ.JJJM
      GOTO 755
C NO RESUME DATA FILE, SO WE MUST BE STARTING
C A POINT FROM THE BEGINNING
      756      RESTR1=.FALSE.
      GOTO 755
C NO EXISTING FILE, THIS IS THE FIRST TIME:
C CREATE FILE HEADER RECORD
      750      JJ=1
      C ... WE CAN'T READ TAU FROM A FILE, SO IF THE FIRST TIME
      C THROUGH THE LOOP ON EB/NO WE MUST COMPUTE IT. BUT IF
      C EB/NO HASN'T CHANGED, WE DON'T NEED TO RECOMPUTE IT SINCE
      C THE THRESHOLD IS NOT A FUNCTION OF GAMMA NOR OF EB/NO.
      RESTR1=.FALSE.
      IF(DOTAU) THEN
      WRITE(5,757)
      FORMAT(' SETTING THRESHOLD')
      CALL SETTAU(MM, PEOD)
      DOTAU=.FALSE.
      WRITE(5,1991) MM, TAU
      WRITE(6,1991) MM, TAU
      FORMAT(' M= ', I2, ' L=3 OPT THRES = ', JPD15.B)
      END IF
      FNAME(UNIT=4, FILE=FNAME, STATUS='NEW', FORM='UNFORMATTED')
      WRITE(4) MM, DEBNOL(IO), NSLOTS, TAU, TAU2, TAU3, PEOD
      CLOSE(UNIT=4)
      DO 600 IJ=JJ, MJ
      IF(IJ.GE.3) THEN
      IDO=QOPT(IJ-1)-QOPT(IJ-2)+0.500
      IF(IDO.EQ.0) IDO=1
      755      0066
      0065
      0064
      0063
      0062
      0061
      0060
      0059
      0058
      0057
      0056
      0055
      0054
      0053
      0052
      0051
      0050
      0049
      0048
      0047
      0046
      0045
      0044
      0043
      0042
      0041
      0040
      0039
      0038
      0037
      0036
  
```

```

0067 Q=QOPT(IJ-1)
0068 IQ=Q
0069 ELSE
0070 IQQ=1
0071 Q=1.00
0072 IQ=1
0073 END IF
0074 DQ=IDQ

C GIVE PROGRESS MESSAGE TO T:
0075 WRITE(5,601) IJ
0076 FORMAT(' IJ=',I3)
0077 DEBNJ=START+(IJ-1)*OBINC
0078 DBSJR(IJ)=DEBNJ
0079 R=10.00**((DEBNJ)/10.00)
0080 DO 602 IQS=1.50
0081 POFQ(IQS)=0.00
0082 IQL(IQS)=0
0083 CONTINUE
0084 IQS=0

C PRIME THE ALGORITHM WITH DUMMY OLD VALUES OF P(E)
0085 P1=0.00
0086 P2=0.00
0087 GAMMA=Q/SLOTS
0088 WRITE(GNAME,735) MM,IQ
0089 FORMAT('EQ',I1,'3',I4,'.DAT')
0090 OPEN(UNIT=3,FILE=GNAME,STATUS='OLD',FORM='UNFORMATTED',
      $ READONLY,ERR=770)
0091 WRITE(5,3939)
0092 FORMAT(' READING EVENT FILE')
0093 READ(3) D,IDSUB,MUSED,GOOD
0094 CLOSE(UNIT=3)
0095 GOTO 777

C IF FILE FOR EVENT PROBABILITIES DOES NOT EXIST, CALCULATE THEM
      C AND CREATE A FILE.
0096 CONTINUE
0097 WRITE(5,3938)
0098 FORMAT(' CREATING EVENT FILE')
0099 CALL GENP(E,MM,IQ,MSLOTS,GOOD,A,IASUB,
      $ C,ICSUB,D,IDSUB,MUSED)
0100 OPEN(UNIT=3,FILE=GNAME,STATUS='NEW',FORM='UNFORMATTED')
0101 WRITE(3) D,IDSUB,MUSED)
0102 CLOSE(UNIT=3)
0103 IF(.NOT.GOOD) GOTO 700
0104 RHOTS=GAMMA*R*EBNO/(GAMMA*R+EBNO)
0105 RHOT=K*RHOTS/3.00

C EVALUATE THE PROBABILITY
0106 DO 780 NOS=1,IQS
0107 IF(IQL(NOS).EQ.10) THEN
0108 PESYM=POFQ(NOS)
0109 GOTO 781
0110 END IF
0111 CONTINUE
780

```

```

C NOT IN STORED LIST, COMPUTE IT
0112 CALL PSUBE(MM,PESYM,D,IDSUB,MUSED,PRERR,IPSUB,PEDD)
0113 IQS=IQS+1
0114 IF(IQS.GT.50) THEN
0115 IQS=50
0116 ELSE
0117 POFQ(IQS)=PESYM
0118 IQL(IQS)=IQ
0119 END IF
0120 P3=PESYM
0121 IF(P3.GT.P2 .AND. IO.LT.MSLOTS) THEN
      C KEEP ON GOING, WE ARE NOT PAST THE MAXIMUM
      P1=P2
0122 P2=P3
0123 IQ=MIMO(IQ+IDQ,MSLOTS)
0124 Q=DMINI(Q+DQ,SLOTS)
0125 GOTO 709
0126 ELSE
0127 PMAX=DMAX1(P1,P2,P3)
0128 EPS=0.00100*PMAX
0129 TEST=(DABS(P1-P2)).LE.EPS .AND. DABS(P1-P3).LE.EPS .AND.
      $ DABS(P2-P3).LE.EPS)
0130 IF( TEST .OR. IOQ.EQ.1
      $ .OR. (.NOT.TEST) .AND. IO.EQ.MSLOTS) THEN
      C WE ARE DONE WHEN ALL 3 ARE CLOSE TOGETHER OR WHEN DO=1
      C OR WHEN WE REACHED FULL-BAND JAMMING AND P(E) IS STILL
      C INCREASING
      POPT=PMAX
0131 IF(P2.GT.P3) THEN
      C THE OPTIMUM MUST BE THE MIDDLE POINT OF THE 3
      QOPT(IJ)=Q-DQ
0132 IF(QOPT(IJ).EQ.0.00) QOPT(IJ)=1.00
0133 IF(IJ.GT.1) THEN
      C PREVENT ROUND-OFF FROM MAKING QOPT VS. EB/NI NON-MONOTONIC
      IF(QOPT(IJ).LT.QOPT(IJ-1)) QOPT(IJ)=QOPT(IJ-1)
      END IF
      ELSE
      C THE OPTIMUM IS FULL-BAND JAMMING
      QOPT(IJ)=MSLOTS
0134 END IF
0135 GOTO 665
0136 ELSE
      C NOT LOCATED SUFFICIENTLY ACCURATELY, CUT DQ AND TRY AGAIN
      Q=Q-DQ-DQ
0137 IQ=IQ-IDQ-IDQ
0138 IDQ=IDQ/2
0139 DQ=IDQ
0140 P2=P1
0141 P1=0.00
0142 Q=Q+DQ
0143 IO=IO+IDQ

```

```

0001 SUBROUTINE GET(NJ,START,DBINC)
0002 IMPLICIT DOUBLE PRECISION FOR REAL
0003 CHARACTER*9 FIELD,BLANK9
0004 COMMON /INPUTS/ DEBNOL(3),MSLOTS,K,MM
0005 COMMON /SIZE/ NO
0006 C DEFAULT LISTS TEMPORARILY NEEDED ARE IN SHARED STORAGE WITH
0007 C THE LARGE CONVOLUTION WORKING ARRAYS
0008 COMMON /SHARE/ DSNR(3,4)
0009 DATA BLANK9/'
0010 WRITE(5,33)
0011 FORMAT(' BITS/SYMBOL (K) [2]: ',5)
0012 READ(5,3)K
0013 IF(K.EQ.0)K=2
0014 MM=2**K
0015 WRITE(5,2)
0016 FORMAT(' HOW MANY EB/MO? [1]: ',5)
0017 READ(5,3)NO
0018 IF(NO.EQ.0)NO=1
0019 DO 7 IN=1,NO
0020 DO=DSNR(IN,K)
0021 WRITE(5,5)IN,DO
0022 FORMAT(' EB/MO('',12,'') ['',F9.6,'']: ',5)
0023 READ(5,6)FIELD
0024 FORMAT(A9)
0025 IF(FIELD.EQ.BLANK9) THEN
0026 DEBNOL(IN)=DO
0027 ELSE
0028 DECODE(9,6i,FIELD;DEBNOL(IN)
0029 FORMAT(F9.6)
0030 END IF
0031 CONTINUE
0032 MSLOTS=2400
0033 WRITE(5,39)
0034 FORMAT(' HOW MANY EB/NJ? [11]: ',5)
0035 READ(5,34,ERR=38) NJ
0036 FORMAT(I3)
0037 IF(NJ.EQ.0) NJ=11
0038 IF(NJ.LT.0 .OR. NJ.GT.11) GOTO 32
0039 WRITE(5,41)
0040 FORMAT(' STARTING VALUE FOR EB/NJ (DB) [50.7]: ',5)
0041 READ(5,42,ERR=40) START
0042 IF(START.EQ.0.00) START=50.00
0043 FORMAT(F6.3)
0044 IF(NJ.EQ.1) RETURN
0045 WRITE(5,36)
0046 FORMAT(' DB INCREMENT FOR EB/NJ [-5.0]: ',5)
0047 READ(5,37,ERR=35) DBINC
0048 IF(DBINC.EQ.0.00) DBINC=-5.00
0049 RETURN
0050 END
  
```

```

0152 GOTO 709
0153 END IF
0154 END IF
0155 PE=NOBRT+POPT
0156 WRITE(6,666) DBSJR(IJ),PE,QOPT(IJ)
0157 FORMAT(1X,F7.3,5X,1PD12.5,5X,1PD12.5)
0158 PRLOG(IJ)=DK.0610(PE)
0159 OPEN(UNIT=4,FILE=FNAME,STATUS='OLD',ACCESS='APPEND',
0160 FORM='UNFORMATTED')
0161 WRITE(4) DBSJR(IJ), PRLOG(IJ), QOPT(IJ)
0162 CLOSE(UNIT=4)
0163 CONTINUE
0164 OPEN(UNIT=4,FILE=FNAME,STATUS='NEW',FORM='UNFORMATTED')
0165 WRITE(4) MM,3,DEBNOL(10),MSLOTS,DBSJR,PRLOG,QOPT
0166 CLOSE(UNIT=4)
0167 WRITE(6,776) MM,DEBNOL(10)
0168 DO 689 IJ=1,NJ
0169 WRITE(6,666) DBSJR(IJ),10,**PRLOG(IJ),QOPT(IJ)
0170 CONTINUE
0171 WRITE(6,688) TAU
0172 FORMAT(////', OPTIMUM THRESHOLD FOR ABOVE IS ETA/SIGMA**2 = ',
0173 F7.3)
0174 CONTINUE
0175 CONTINUE
0176 CONTINUE
0177 STOP 0
0178 END
  
```

```
0001 SUBROUTINE PSUBE(M,PE,D,IDSUB,NUSED,PRERR,IPSUB,PEOO)
C COMPUTE UNCONDITIONAL ERROR PROBABILITY
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 INTEGER LUP(4),JSUB(4)
0004 LOGICAL *I GO,MORE,STORE,RESTR
C WE DO WANT TO STORE ZERO ELEMENTS OF THE DENSITY FUNCTION,
C SINCE IT SAVES TIME TO AVOID REPEATING THE UNDERFLOWS
VIRTRAL PRERR(E25),IPSUB(625)
VIRTUAL O(625),IDSUB(625)
COMMON /RSPAR/ JAM(4),JAM1,MPS,IJ,RESTR7
COMMON /SHAREZ/ LOM(4),LINC(4)
COMMON /DENPAR/ BIG, AAB, BAB, LJAM, AAB2, BAB2, AABBBAB,
$ TAU, TAU2, TAU3, TAU4, TAU5, TAU6, TAU7, TAU8, TAU9, TAU10, TAU11, TAU12, TAU13
COMMON /PARDEN/ RHOM, RHOT
DATA STORE/.TRUE./
PE=0.00
DO 10 I=1,M
LUP(I)=3
JSUB(I)=0
CONTINUE
10 IF(.NOT.RESTR) THEN
MPS=0
C THE ALL-ZERO JAMMING EVENT P(E) IS AVAILABLE FROM THE
C SEARCH FOR THE OPTIMUM THRESHOLD, SO PUT IT INTO THE
C ARRAY OF SAVED VALUES
CALL LOCN(M,LOM,LUP,JSUB,ISUB)
CALL PUTIN(PEO,PRERR,IPSUB,MPS,625,ISUB,KODE,STORE)
IF(KODE.NE.O) STOP 'PRERR FULL'
JAM1=-1
ELSE
C RESTART BYPASSES VECTOR LOOP INITIALIZATION CALL
GOTO 100
END IF
END IF
C START VECTOR-INDEXED LOOP ON JAMMING EVENTS
CALL VINIT(JAM,LOM,M)
CONTINUE
100 IF(JAM1.NE.JAM(1)) THEN
C UPDATE TEST VALUE FOR NEXT TIME, AND ...
JAM1=JAM(1)
END IF
CALL EVENT(M,JAM,PE,O,IDSUB,NUSED)
C BYPASS CONDITIONAL ERROR PROBABILITY CALCULATION IF EVENT
C PROBABILITY IS ZERO. THIS SAVES MUCH TIME.
IF(PIE.EQ.O.DO)GOTO 101
C SINCE JAMMING PROBABILITIES DEPEND ONLY ON NO. OF CHANNELS
C HAVING JAM(1) HOPS JAMMED AND NOT THE ARRANGEMENT OF THE
C CHANNELS, WE CAN SORT THE NON-SIGNAL CHANNELS INTO ASCENDING
C NUMBERS OF HOPS JAMMED. THIS REDUCES NUMBER OF DISTINCT
C CONDITIONAL ERROR PROBABILITIES WHICH MUST BE SAVED TO AVOID
C RECOMPUTING THEM UNNECESSARILY.
```

```
0033 DO 111 I=1,M
0034 JSUB(I)=JAM(I)
0035 CONTINUE
111 IF(M.EQ.2) GOTO 199
0036 DO 110 I=2,M-1
0037 DO 120 J=I+1,M
0038 IF(JSUB(J).LT.JSUB(I)) THEN
0039 JTEMP=JSUB(I)
0040 JSUB(I)=JSUB(J)
0041 JSUB(J)=JTEMP
0042 END IF
0043 CONTINUE
0044 CONTINUE
0045 110 CONTINUE
0046 199 CONTINUE
C STORE AS ELEMENT OF A SPARSE ARRAY TO SAVE MEMORY
C EVEN THOUGH WE STORE ZEROS, THE SORTING OF SUBSCRIPTS
C CUTS OUT MANY ELEMENTS.
CALL LOCN(M,LOM,LUP,JSUB,ISUB)
C TRY TO FIND CONDITIONAL ERROR PROBABILITY IN STORED ARRAY
CALL LOOKUP(PROB,PRERR,IPSUB,MPS,625,ISUB,STORE,MNME)
C IF IT IS NOT THERE, WE MUST COMPUTE IT
IF(NONE) THEN
CALL PSEL(JSUB,M,PROB)
C ... AND SAVE IT FOR POSSIBLE FUTURE RE-USE
CALL PUTIN(PROB,PRERR,IPSUB,MPS,625,ISUB,KODE,STORE)
IF(KODE.NE.O) STOP 2
END IF
C SUM UP UNCONDITIONAL ERROR PROBABILITY
PE=PE+PIE*PROB
C SAVE WHAT WE HAVE SO FAR FOR RESTART CAPABILITY
C
0055 OPEN(UNIT=4,FILE='RESUME.DAT',STATUS='UNKNOWN',
$ FORM='UNFORMATTED')
0056 WRITE(4) JAM,JAM1,PRERR,MPS,IPSUB,IJ
0057 CLOSE(UNIT=4)
C ONCE WE HAVE UPDATED RESTART INFORMATION, THE RESTART
C FLAG BECOMES FALSE.
C
0058 RESTR=.FALSE.
C ITERATE THE VECTOR-INDEX LOOP
101 CALL VITER(JAM,LOM,LUP,LINC,M,GO)
0060 IF(GO) GOTO 100
0061 RETURN
0062 END
```

```

0001 C SUBROUTINE EVENT(M,JAM,PIE,D,IDSUB,NUSED)
C SUBROUTINE TO LOOK UP EVENT PROBABILITY FROM STORED ARRAY
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 LOGICAL *I STORE,NOME
0004 DIMENSION JAM(4),LUP(4)
0005 VIRTUAL D(625),IDSUB(625)
0006 COMMON /SHAREZ/ LOW(4),LINC(4)
0007 DATA STOREZ/FALSE./
C SET UP ARRAY DESCRIPTION D(O:LL,...,O:LL) WITH M DIMENSIONS
0008 DO I=1,M
0009 LUP(I)=3
0010 CONTINUE
C COMPUTE LINEAR EQUIVALENT SUBSCRIPT FOR JAMMING EVENT
0011 CALL LOCN(M,LOW,LUP,JAM,ISUB)
C LOOK UP THE VALUE, GET O.DO IF NOT THERE
0012 CALL LOOKUP(PIE,D,IDSUB,NUSED,625,ISUB,STORE,NOME)
0013 RETURN
0014 END
    
```

```

0001 $ SUBROUTINE GENPIE(MM,MQ,MSLOTS,GOOD,A,TASUB,C,ICSUB,
C SUBROUTINE TO GENERATE EVENT PROBABILITIES
C
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 LOGICAL *I GO,GOZ,STORE,NOME,GOOD
0004 DIMENSION LUP2(4),LUP3(4)
0005 DIMENSION IUPA(4)
0006 DIMENSION IUPD(4)
0007 DIMENSION LUP1(4)
0008 VIRTUAL A(100),TASUB(100),C(625),ICSUB(625),
$ D(625),IDSUB(625)
0009 DIMENSION I(4),II(4),III(4)
C SHARED STORAGE FOR COMMONLY NEEDED CONSTANT ARRAYS
0010 COMMON /SHAREZ/ LOW(4),LINC(4)
C SHARED STORAGE FOR: (1) INPUT DEFAULT LISTS, (2) CONDITIONAL PROB GEN.,
C AND (3) EVENT PROB. GEN. THESE ARE NON-OVERLAPPING USAGES.
0011 COMMON /SHARE/ LUP2,LUP3,IUPD,LI,MUSEA,MUSEC,ERR,
$ ISUB,ISUB1,ISUB2,AIN,I,II,III,MN,ADUT,CIN,COU,
$ DOUT,DIN
    
```

```

0012 DATA I100/I00/
0013 DATA IUPA/A=1/
0014 DATA LUP1/A=1/
0015 STORE=.FALSE. => DON'T STORE ZERO ELEMENTS OF SPARSE ARRAY.
0016 GOOD=.TRUE.
0017 IF(MQ.LE.O) THEN
0018 GOOD=.FALSE.
0019 RETURN
0020 END IF
0021 DO 80 LI=1,MM
0022 IUPD(LI)=3
0023 CONTINUE
C JAMMING PATTERN W/NON-ZERO PROBABILITY ON PER-HOP BASIS
0024 MUSEA=0
C INITIALIZE VECTOR-INDEX LOOP
0025 CALL VLINIT(I,LOW,LUP1,LINC,MM,GO)
0026 CONTINUE
90 CONTINUE
0027 CALL LOCN(MM,LOW,IUPA,I,ISUB)
0028 CALL PRIHOP(I,MM,MQ,MSLOTS,AIN)
0029 CALL PUTIN(AIN,A,TASUB,MUSEA,I100,ISUB,IEERR,STORE)
0030 IF(IEERR.NE.O)STOP 3
C ITERATE VECTOR-INDEX LOOP
0031 CALL VLITER(I,LOW,LUP1,LINC,MM,GO)
0032 IF(GO) GOTO 90
C COMPUTATION STARTS HERE. FIRST COPY A INTO D.
C SINCE ARRAYS ARE AID:1,0:1,...,0:1) AND D(O:L,O:L,...,0:L)
C THE COPYING MUST BE DONE ON BASIS OF EQUIVALENT LINEAR
C SUBSCRIPTS RATHER THAN A SIMPLE MOVE OPERATION.
0033 NUSED=0
    
```

```

0001 SUBROUTINE PUTIN(CIM,C,ICSUB,MUSE,NMAX,K,IERR,STORE)
C THIS SUBROUTINE INSERTS AN ELEMENT INTO A SPARSE ARRAY FOR
C WHICH ONLY THE NON-ZERO ELEMENTS ARE KEPT IN STORAGE IF
C THE SWITCH STORE IS .TRUE.
C THE DOUBLE PRECISION VALUE CIM IS STORED AS C(K) WHERE
C THE AUXILIARY ARRAY ICSUB IS USED TO KEEP TRACK OF THE
C SUBSCRIPTS OF THE CORRESPONDING ENTRIES IN C.
C USAGE: LOGICAL=I STORE
C DOUBLE PRECISION C,CIM
C VIRTUAL ICSUB(NMAX),C(NMAX)
C CALL PUTIN(CIM,C,ICSUB,MUSE,NMAX,K,IERR,STORE)
C WHERE
C CIM = VALUE OF ELEMENT TO STORE
C C = ARRAY IN WHICH NON-ZERO VALUES ARE ACTUALLY STORED
C ICSUB = AUXILIARY ARRAY FOR ACTUAL SUBSCRIPT VALUES
C MUSE = NUMBER OF ELEMENTS OF C CURRENTLY OCCUPIED
C NMAX = SIZE OF ARRAY C
C IERR = ERROR RETURN CODE, 0 IF NO ERROR OR 1 IF THERE IS
C NO ROOM AVAILABLE IN C
C STORE = .TRUE. TO STORE ZEROS EXPLICITLY, ELSE .FALSE.
C NOTE: IF CIM=0 AND THE SUBSCRIPT K IS FOUND IN ICSUB, THEN
C THE ELEMENT IS DELETED BY SHIFTING DOWNWARD ALL
C FOLLOWING ELEMENTS OF THE ARRAY
C PROGRAMMER: ROBERT H. FRENCH DATE: 11 JANUARY 1984
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 VIRTUAL ICSUB(NMAX),C(NMAX)
0004 LOGICAL=I STORE
0005 IERR=0
0006 IF(STORE) GOTO 5
0007 IF(CIM.EQ.O.DO) GOTO 30
0008 IF(MUSE.EQ.O)GOTO 20
0009 DO 10 I=1,MUSE
0010 IF(ICSUB(I).NE.K) GOTO 10
0011 C(I)=CIM
0012 RETURN
0013 CONTINUE
0014 IF(MUSE.LT.NMAX) GOTO 20
0015 IERR=1
0016 RETURN
0017 MUSE=MUSE+1
0018 ICSUB(MUSE)=K
0019 C(MUSE)=CIM
0020 RETURN
0021 DO 40 I=1,MUSE
0022 J=I
0023 IF(ICSUB(I).EQ.K) GOTO 50
0024 CONTINUE

```

```

0034 C INITIALIZE VECTOR-INDEX LOOP
0035 CALL VLIMIT(I,LOW,MM)
0036 CONTINUE
0037 CALL LOCN(MM,LOW,IUPA,1,ISUB1)
0038 CALL LOCN(MM,LOW,IUPD,1,ISUB2)
0039 CALL LOOKUP(AOUT,A,IASUB,MUSEA,I100,ISUB1,STORE,MOME)
0040 CALL PUTIN(AOUT,D,IDSUB,MUSED,625,ISUB2,IERR,STORE)
0041 CALL VLIMIT(I,LOW,IUP1,LINC,MM,60)
0042 IF(60)GOTO 99
0043 C ... L-1 CONVOLUTIONS ARE NEEDED ...
0044 DO 998 L1=1,2
0045 DO 125 NN=1,MM
0046 LUP2(NN)=L1
0047 LUP3(NN)=L1+1
0048 MUSEC=0
0049 CALL VLIMIT(I,LOW,MM)
0050 CONTINUE
0051 CALL VLIMIT(II,LOW,MM)
0052 CONTINUE
0053 CALL LOCN(MM,LOW,IUPA,1,ISUB1)
0054 CALL LOCN(MM,LOW,IUPD,1,ISUB2)
0055 DO 21 NN=1,MM
0056 III(NN)=I(NN)+II(NN)
0057 CONTINUE
0058 CALL LOCN(MM,LOW,IUPD,1,ISUB3)
0059 CALL LOOKUP(AOUT,A,IASUB,MUSEA,I100,ISUB1,STORE,MOME)
0060 CALL LOOKUP(DOUT,D,IDSUB,MUSED,625,ISUB2,STORE,MOME)
0061 CALL LOOKUP(COUT,C,ICSUB,MUSEC,625,ISUB3,STORE,MOME)
0062 CIM=COUT+AOUT*DOUT
0063 CALL PUTIN(CIM,C,ICSUB,MUSEC,625,ISUB3,IERR,STORE)
0064 IF(IERR.NE.O) STOP 4
0065 C ITERATE VECTOR-LOOP FOR ARRAY D
0066 CALL VLIMIT(II,LOW,IUP2,LINC,MM,602)
0067 IF(602) GOTO 97
0068 C ITERATE VECTOR-LOOP FOR ARRAY A
0069 CALL VLIMIT(I,LOW,IUP1,LINC,MM,60)
0070 IF(60) GOTO 98
0071 MUSED=0
0072 CALL VLIMIT(II,LOW,MM)
0073 CONTINUE
0074 CALL LOCN(MM,LOW,IUPD,1,ISUB)
0075 CALL LOOKUP(COUT,C,ICSUB,MUSEC,625,ISUB,STORE,MOME)
0076 DIM=COUT
0077 CALL PUTIN(DIM,D,IDSUB,MUSED,625,ISUB,IERR,STORE)
0078 IF(IERR.NE.O) STOP 5
0079 CALL VLIMIT(II,LOW,IUP3,LINC,MM,60)
0080 IF(60) GOTO 96
0081 CONTINUE
0082 RETURN
0083 END
9988
0079

```

```

0025 RETURN
C
C REMOVE THE ZEROED ELEMENT AND BUMP COUNT OF ENTRIES USED
C
0026 50 DO 60 I=J,NUSE-1
0027 ICSUB(I)=ICSUB(I+1)
0028 C(I)=C(I+1)
0029 CONTINUE
0030 NUSE=NUSE-1
0031 RETURN
0032 END

```

```

0001 SUBROUTINE LOOKUP(COUT,C,ICSUB,N,M,MAX,K,STORE,NOME)
C
C THIS SUBROUTINE RETRIEVES AN ELEMENT OF A SPARSE ARRAY WHICH
C HAS BEEN STORED COMPACTLY BY STORING ONLY NON-ZERO ELEMENTS.
C
C THE ARRAY IS DOUBLE PRECISION.
C
C USAGE:
C VIRTUAL ICSUB(MMAX), C(MMAX)
C LOGICAL *I STORE, NOME
C DOUBLE PRECISION COUT
C CALL LOOKUP(COUT,C,ICSUB,N,M,MAX,K,STORE,NOME)
C WHERE
C COUT = VALUE OF C(K) (OUTPUT FROM SUBROUTINE)
C C = ARRAY USED TO STORE NON-ZERO ELEMENTS
C ICSUB = AUXILIARY ARRAY TO STORE ACTUAL SUBSCRIPTS
C N = NUMBER OF ELEMENTS OF C CURRENTLY IN USE
C MMAX = SIZE OF C
C K = SUBSCRIPT OF SPARSE ARRAY TO LOOK UP
C STORE = .TRUE. IF ZEROES STORED EXPLICITLY, ELSE .FALSE.
C NOME = .FALSE. IF ZEROES NOT STORED OR ZEROES STORED AND
C ELEMENT IS FOUND IN THE STORED ARRAY
C .TRUE. IF ZEROES ARE STORED AND THE ELEMENT IS
C NOT FOUND (OUTPUT QUANTITY)
C
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
C

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```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 VIRTUAL ICSUB(MMAX),C(MMAX)
0004 LOGICAL *I STORE, NOME
0005 NOME=.FALSE.
0006 DO 10 I=1,N
0007 IF(ICSUB(I).NE.K)GOTO 10
0008 COUT=C(I)
0009 RETURN
0010 CONTINUE
0011 IF(STORE) THEN
0012 NOME=.TRUE.
0013 ELSE
0014 COUT=0.
0015 END IF
0016 RETURN
0017 END

```

```

0001 SUBROUTINE LOGCN(NDIM, ILOW, IUP, ISUB, LLINEAR)
C THIS SUBROUTINE COMPUTES THE EQUIVALENT LINEAR SUBSCRIPT FOR
C A MULTIDIMENSIONAL ARRAY OF NDIM DIMENSIONS
C
C IF THE ARRAY A IS DEFINED AS
C DIMENSION A(ILOW(1):IUP(1),...,ILOW(NDIM):IUP(NDIM))
C AND ISUB(1),...,ISUB(NDIM) IS A SET OF SUBSCRIPTS FOR A,
C THEN THIS SUBROUTINE RETURNS IN LINEAR THE OFFSET FROM THE
C ORIGIN OF A TO THE ELEMENT A(ISUB(1),...,ISUB(NDIM)), ASSUMING
C THE FIRST SUBSCRIPT VARIES MOST RAPIDLY.
C
C USAGE:
C DIMENSION ILOW(NDIM),IUP(NDIM),ISUB(NDIM)
C DATA ILOW/lower limits of defined subscripts of array/
C DATA IUP/upper limits of defined subscripts of array/
C ...SET ISUB TO DESIRED SEQUENCE OF SUBSCRIPTS...
C CALL LOGCN(NDIM,ILOW,IUP,ISUB,LLINEAR)
C
C WHERE
C NDIM = NUMBER OF DIMENSIONS THE ARRAY HAS
C ILOW = ARRAY OF LOWER SUBSCRIPT BOUNDS
C IUP = ARRAY OF UPPER SUBSCRIPT BOUNDS
C ISUB = ARRAY CONTAINING SUBSCRIPT FOR WHICH LOCATION IS
C TO BE COMPUTED
C LLINEAR = RETURNED VALUE OF OFFSET INTO ARRAY IN MEMORY
C
C
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
  
```

```

0002 DIMENSION ILOW(NDIM),IUP(NDIM),ISUB(NDIM)
0003 LLINEAR=0
0004 DO 10 I=1,NDIM-1
0005 J=NDIM-I+1
0006 LLINEAR=(LLINEAR+(ISUB(J)-ILOW(J)))*(IUP(J-1)-ILOW(J-1)+1)
0007 CONTINUE
0008 LLINEAR=LLINEAR+ISUB(1)-ILOW(1)
0009 RETURN
0010 END
  
```

```

0001 SUBROUTINE VLINIT(LVEC,LLOW,LMAX)
C THIS SUBROUTINE INITIALIZES A "VECTOR DO-LOOP" STRUCTURE
C DEFINED BY THE FOLLOWING PSEUDO-FORTRAN CODE:
C DO 100 LVEC(1)=LLOW(1),LUP(1),LINC(1)
C DO 100 LVEC(2)=LLOW(2),LUP(2),LINC(2)
C
C
C DO 100 LVEC(LMAX)=LLOW(LMAX),LUP(LMAX),LINC(LMAX)
C
C (STATEMENTS IN RANGE OF LOOP)
C
C 100 CONTINUE
C
C THE COMPANION ROUTINE VLITER HANDLES THE LOOP CONTROL AT THE
C CONTINUE STATEMENT IN THE ABOVE STRUCTURE
C
C USAGE:
C LOGICAL*1 GO
C DIMENSION LVEC(LMAX),LLOW(LMAX),LUP(LMAX),LINC(LMAX)
C (INITIALIZE ARRAY LLOW TO STARTING VALUES OF THE NESTED LOOPS)
C (INITIALIZE ARRAY LUP TO STOPPING VALUES OF THE NESTED LOOPS)
C (INITIALIZE ARRAY LINC TO INCREMENTS OF THE LOOPS)
C CALL VLINIT(LVEC,LLOW,LMAX)
C 100 CONTINUE
C
C : (STATEMENTS IN RANGE OF LOOPS)
C
C CALL VLITER(LVEC,LLOW,LUP,LINC,LMAX,GO)
C IF(GO)GOTO 100
C
C WHERE
C LVEC = ARRAY FOR STORAGE OF LOOP INDICES. LVEC(1) IS THE
C OUTER-MOST LOOP; LVEC(LMAX), THE INNER-MOST LOOP.
C LLOW = ARRAY FOR STORAGE OF LOOP STARTING VALUES, IN SAME
C SEQUENCE AS LVEC
C LUP = ARRAY FOR STORAGE OF LOOP FINAL VALUES, IN SAME
C SEQUENCE AS LVEC
C LINC = ARRAY FOR STORAGE OF LOOP INCREMENTS, IN SAME
C SEQUENCE AS LVEC
C LMAX = NUMBER OF LOOPS NESTED
C GO = LOGICAL VARIABLE, .TRUE. IF JUMP BACK TO BEGINNING OF
C STATEMENTS IN THE RANGE OF THE LOOP SHOULD OCCUR,
C .FALSE. OTHERWISE (I.E. OUTER-MOST LOOP TERMINATED)
C
C PROGRAMMER: ROBERT H. FRENCH DATE: 11 JANUARY 1984
C
0002 DIMENSION LVEC(LMAX),LLOW(LMAX)
0003 DO 1 N=1,LMAX
0004 LVEC(N)=LLOW(N)
0005 CONTINUE
0006 RETURN
0007 END
  
```

```

0001 SUBROUTINE PRJHOP(I,KM,KQ,KM,AIN)
C
C THIS SUBROUTINE COMPUTES THE EVENT PROBABILITY FOR ALL
C POSSIBLE JAMMING PATTERNS WITH NON-ZERO PROBABILITY FOR
C L=1 HOP/SYMBOL FOR RHFSK/FH IN PBAJ
C
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 DIMENSION I(4)
0004 AIN=0.DO
0005 KJAM=0
0006 DO 1 K=1,KM
0007 KJAM=KJAM+I(K)
0008 1 CONTINUE
C IF THIS IS AN IMPOSSIBLE CASE, RETURN WITH RESULT = 0.0
  IF(KJAM.GT.MIND(KQ,KM)) RETURN
  KPMAX=KJAM-1
  LPMAX=KM-KJAM-1
  JPMAX=KM-1
  IMAX=MAXO(KPMAX,LPMAX,JPMAX)
  PROD=1.DO
  Q=KQ
  DIFFNO=KN-KQ
  EN=KN
  DO 100 LOOP=0,IMAX
    F=LOOP
    IF(LOOP.LE.KPMAX) PROD=PROD*(Q-F)
    IF(LOOP.LE.JPMAX) PROD=PROD/(EN-F)
    IF(LOOP.LE.LPMAX) PROD=PROD*(DIFFNO-F)
  100 CONTINUE
  AIN=PROD
  RETURN
  END
  
```

```

0001 SUBROUTINE VLITER(LVEC,LLOW,LUP,LINC,LMAX,GO)
C
C LOOP ITERATION LOGIC FOR A *VECTOR DO-LOOP*
C
C SEE DETAILED COMMENTS IN SUBROUTINE VLIMIT FOR USAGE AND
C PARAMETER DEFINITIONS
C
C PROGRAMMER: ROBERT H. FRENCH
C DATE: 11 JANUARY 1984
C
0002 LOGICAL*1 GO
0003 DIMENSION LVEC(LMAX),LLOW(LMAX),LUP(LMAX),LINC(LMAX)
0004 GO=.TRUE.
0005 DO 100 I=1,LMAX
0006 NSUB=LMAX+1-IDX
0007 LVEC(NSUB)=LVEC(NSUB)+LINC(NSUB)
0008 IF(LINC(NSUB).GE.0.AND.LVEC(NSUB).LE.LUP(NSUB)) RETURN
  $ .OR.(LINC(NSUB).LT.0.AND.LVEC(NSUB).GE.LUP(NSUB)) RETURN
0009 LVEC(NSUB)=LLOW(NSUB)
0010 CONTINUE
0011 GO=.FALSE.
0012 RETURN
0013 END
  
```

```
0001 SUBROUTINE ADQUAD(XL,XU,Y,QR,F,TOL,WORK,STACK,HEAP,N,KODE)
C ADAPTIVE QUADRATURE ALGORITHM
C XL - LOWER LIMIT OF INTEGRAL (IN)
C XU - UPPER LIMIT OF INTEGRAL (IN)
C Y - VALUE OF INTEGRAL (OUT)
C QR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
C WITH CALLING SEQUENCE
C CALL QR(XL,XU,F,Y)
C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
C WORK - WORK ARRAY OF SIZE N (IN)
C STACK - SECOND WORK ARRAY OF SIZE N, MUST NOT BE
C HEAP - THIRD WORK ARRAY, SIZE N, DISTINCT FROM WORK AND STACK
C SAME ARRAY AS WORK (IN)
C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (:N)
C KODE - ERROR INDICATOR (OUT)
C 0 -- NO ERROR
C 1 -- WORK ARRAYS TOO SMALL
C 2 -- EPS DIVIDED TO ZERO, EITHER ASKING FOR TOO
C TIGHT A TOLERANCE OR ROUND-OFF PREVENTS
C ATTAINING REQUIRED ACCURACY
C R. H. FRENCH, 14 AUGUST 1984
C
```

```
0002 BLOCK DATA
C INITIALIZE SHARED CONSTANTS
IMPLICIT DOUBLE PRECISION (A-H,O-Z)
COMMON /SHARE/ DSHR(3,4)
COMMON /SHARE/ LOM(4),LINC(4)
COMMON /M5/ X(5),M(5)

0006 C WEIGHTS AND ABSISSAS FOR 10-POINT GAUSSIAN QUADRATURE
C
DATA X/ 0.1488743398163100,
$ 0.43339539412924700,
$ 0.67940956829902400,
$ 0.8650633666898500,
$ 0.97390652851717200 /
DATA W/ 0.29552422471475300,
$ 0.26926671930999600,
$ 0.21908636251598200,
$ 0.14945134915058100,
$ 0.06667134430868800 /

0007 C DEFAULT LISTS FOR INTERACTIVE PARAMETER INPUTS
C ARE SHARED WITH LARGE WORKING STORAGE ARRAYS. SINCE THEY
C MAY BE DESTROYED ONCE THE INPUT PARAMETERS ARE SET UP
DATA DSHR /13.3524700, 12.313300, 10.9444300,
$ 10.60657200, 9.628400, 8.3524500,
$ 9.0940100, 8.169000, 6.97199500,
$ 8.0783500, 7.199600, 6.06964600/

0008 C FREQUENTLY NEEDED CONSTANT ARRAYS AND SCALARS
DATA LOM/4*0/,LINC/4*1/
END
```

```
0009 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0010 EXTERNAL F
0011 DIMENSION WORK(N),STACK(N),HEAP(N)
0012 KODE=0
0013 Y=0.DO
0014 WORK(1)=XU
0015 CALL QR(XL,XU,F,T)
0016 HEAP(1)=T
0017 A=XL
0018 MPTS=1
0019 EPS=TOL
0020 STACK(1)=EPS
0021 B=WORK(MPTS)
0022 XM=(A+B)*0.500
0023 CALL QR(A,XM,F,P1)
0024 CALL QR(XM,B,F,P2)
0025 IF(DABS(T-P1-P2).LE.EPS) GOTD 20
0026 C SPLIT IT
0027 MPTS=MPTS+1
0028 IF(MPTS.GT.N) THEN
0029 KODE=1
0030 RETURN
0031 END IF
0032 WORK(MPTS)=XM
0033 HEAP(MPTS)=P2
0034 T=P1
0035 EPS=EPS/2.DO
```

```

0028 IF(EPS.EQ.0.00) THEN
0029   KODE=2
0030   RETURN
0031 END IF
0032 STACK(NPTS)=EPS
0033 GOTO 10
C FINISHED A PIECE
0034 20
0035 Y=Y+P1+P2
0036 EPS=STACK(NPTS)
0037 T=HEAP(NPTS)
0038 NPTS=NPTS-1
0039 A=8
0039 IF(NPTS.EQ.0) RETURN
0040 GOTO 10
0041 END

```

```

0001 SUBROUTINE ADQUA2(XL,XU,Y,QR,F,TOL,WORK,STACK,HEAP,N,KODE)
C ADAPTIVE QUADRATURE ALGORITHM
C XL - LOWER LIMIT OF INTEGRAL (IN)
C XU - UPPER LIMIT OF INTEGRAL (IN)
C Y - VALUE OF INTEGRAL (OUT)
C QR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
C WITH CALLING SEQUENCE
C CALL QR(XL,XU,F,Y)
C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
C WORK - WORK ARRAY OF SIZE N (IN)
C STACK- SECOND WORK ARRAY OF SIZE N, MUST NOT BE
C HEAP- THIRD WORK ARRAY, SIZE N, DISTINCT FROM WORK AND STACK
C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (IN)
C KODE - ERROR INDICATOR (OUT)
C 0 -- NO ERROR
C 1 -- WORK ARRAYS TOO SMALL
C 2 -- EPS DIVIDED TO ZERO, EITHER ASKING FOR TOO
C TIGHT A TOLERANCE OR ROUND-OFF PREVENTS
C ATTAINING REQUIRED ACCURACY
C R. H. FRENCH, 14 AUGUST 1984

```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 EXTERNAL F
0004 DIMENSION WORK(N),STACK(N),HEAP(N)
0005 KODE=0
0006 Y=0.00
0007 WORK(1)=XU
0008 CALL QR(XL,XU,F,T)
0009 HEAP(1)=T
0010 A=XL
0011 NPTS=1
0012 EPS=TOL
0013 STACK(1)=EPS
0014 B=WORK(NPTS)
0015 XM=(A+B)*0.500
0016 CALL QR(A,XM,F,P1)
0017 CALL QR(XM,B,F,P2)
0018 IF(DABS(T-P1-P2).LE.EPS) GOTO 20
C SPLIT IT
0019 NPTS=NPTS+1
0020 IF(NPTS.GT.N) THEN
0021   KODE=1
0022   RETURN
0023 END IF
0024 WORK(NPTS)=XM
0025 HEAP(NPTS)=P2
0026 T=P1
0027 EPS=EPS/2.00

```

```

0001 SUBROUTINE D610(A,B,F,ANSWER)
C 10-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
C
C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
C R. H. FRENCH, 28 FEBRUARY 1986
C
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 COMMON /WTS/ X(5),M(5)
0004 ANSWER=0.D0
0005 BMA02=(B-A)/2.D0
0006 BPA02=(B+A)/2.D0
0007 DO 10 J=1,5
0008 C=X(1)*BMA02
0009 Y1=BPA02+C
0010 Y2=BPA02-C
0011 ANSWER=ANSWER+W(1)*(F(Y1)+F(Y2))
0012 CONTINUE
0013 ANSWER=ANSWER*BMA02
0014 RETURN
0015 END

```

```

0028 IF(EPS.EQ.0.D0) THEN
0029   KODE=2
0030   RETURN
0031 END IF
0032 STACK(NPTS)=EPS
0033 GOTO 10
C FINISHED A PIECE
0034 20 Y=Y+D1+P2
0035 EPS=STACK(NPTS)
0036 T=HEAP(NPTS)
0037 NPTS=NPTS-1
0038 A=B
0039 IF(NPTS.EQ.0) RETURN
0040 GOTO 10
0041 END

```

```

0028 IF(EPS.EQ.0.DO) THEN
0029   KODE=2
0030   RETURN
0031 END IF
0032 STACK(NPTS)=EPS
0033 GOTO 10
      C FINISHED A PIECE
0034   Y=Y+P1+P2
0035   EPS=STACK(NPTS)
0036   T=HEAP(NPTS)
0037   NPTS=NPTS-1
0038   A=B
0039 IF(NPTS.EQ.0) RETURN
0040 GOTO 10
0041 END

```

```

0001 SUBROUTINE ADQUA3(XL,XU,Y,QR,F,TOL,WORK,STACK,HEAP,N,KODE)
      C ADAPTIVE QUADRATURE ALGORITHM
      C XL - LOWER LIMIT OF INTEGRAL (IN)
      C XU - UPPER LIMIT OF INTEGRAL (IN)
      C Y - VALUE OF INTEGRAL (OUT)
      C QR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
      C WITH CALLING SEQUENCE
      C CALL QR(XL,XU,F,Y)
      C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
      C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
      C WORK - WORK ARRAY OF SIZE N (IN)
      C STACK - SECOND WORK ARRAY OF SIZE N, MUST NOT BE
      C SAME ARRAY AS WORK (IN)
      C HEAP - THIRD WORK ARRAY, SIZE N, DISTINCT FROM WORK AND STACK
      C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (IN)
      C KODE - ERROR INDICATOR (OUT)
      C 0 -- NO ERROR
      C 1 -- WORK ARRAYS TOO SMALL
      C 2 -- EPS DIVIDED TO ZERO, EITHER ASKING FOR TOO
      C TIGHT A TOLERANCE OR ROUND-OFF PREVENTS
      C ATTAINING REQUIRED ACCURACY
      C R. H. FRENCH, 14 AUGUST 1984

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0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 EXTERNAL F
0004 DIMENSION WORK(N),STACK(N),HEAP(N)
0005 KODE=0
0006 Y=0.DO
0007 WORK(1)=XU
0008 CALL QR(XL,XU,F,T)
0009 HEAP(1)=T
0010 A=XL
0011 NPTS=1
0012 EPS=TOL
0013 STACK(1)=EPS
0014 B=WORK(NPTS)
0015 XM=(A+B)*0.5D0
0016 CALL QR(A,XM,F,P1)
0017 CALL QR(XM,B,F,P2)
      C *** MAKE IT A RELATIVE TEST FOR THIS INTEGRAL ***
0018 IF(DABS(T-P1-P2).LE.DABS(T*EPS)) GOTO 20
      C SPLIT IT
0019 NPTS=NPTS+1
0020 IF(NPTS.GT.N) THEN
0021   KODE=1
0022   RETURN
0023 END IF
0024 WORK(NPTS)=XM
0025 HEAP(NPTS)=P2
0026 T=P1
0027 EPS=EPS/2.DO

```

0001 SUBROUTINE DGX(A,B,F,ANSWER)
C 10-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
C
C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
C R. H. FRENCH, 28 FEBRUARY 1986
C

0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 COMMON /MTS/ X(5),M(5)
0004 ANSWER=0.00
0005 BMA02=(B-A)/2.00
0006 BPA02=(B+A)/2.00
0007 DO 10 I=1,5
0008 C=X(I)*BMA02
0009 Y1=BPA02+C
0010 Y2=BPA02-C
0011 ANSWER=ANSWER+M(I)*(F(Y1)+F(Y2))
0012 CONTINUE
0013 ANSWER=ANSWER*BMA02
0014 RETURN
0015 END

0001 SUBROUTINE DGTEM(A,B,F,ANSWER)
C 10-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
C
C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
C R. H. FRENCH, 28 FEBRUARY 1986
C

0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 COMMON /MTS/ X(5),M(5)
0004 ANSWER=0.00
0005 BMA02=(B-A)/2.00
0006 BPA02=(B+A)/2.00
0007 DO 10 I=1,5
0008 C=X(I)*BMA02
0009 Y1=BPA02+C
0010 Y2=BPA02-C
0011 ANSWER=ANSWER+M(I)*(F(Y1)+F(Y2))
0012 CONTINUE
0013 ANSWER=ANSWER*BMA02
0014 RETURN
0015 END

```

0001 C SUBROUTINE MINSER(F,FMIN,XMIN,STEP,GUESS,BLIM,ULIM,TOL)
C SEARCH FOR MINIMUM OF F(X) OVER THE INTERVAL BLIM <= X <= ULIM
C TROUBLE MAY OCCUR IF F(X) HAS MULTIPLE LOCAL MINIMA WITHIN THE
C SEARCH INTERVAL OR IF THE FUNCTION IS VERY STEEP AND STEP IS
C TOO BIG.
C
C F = NAME OF FUNCTION TO BE MINIMIZED
C FMIN = MINIMUM VALUE OF F(X) OVER INTERVAL
C XMIN = VALUE OF X FOR WHICH FMIN OCCURS
C STEP = INITIAL STEP SIZE FOR SEARCH
C GUESS = INITIAL GUESS AT XMIN, BLIM <= GUESS <= ULIM
C BLIM = LOWER LIMIT OF SEARCH INTERVAL
C ULIM = UPPER LIMIT OF SEARCH INTERVAL
C TOL = TOLERANCE ON XMIN; SEARCH STOPS WHEN DX < TOL
C
C NOTE: F MUST BE A DOUBLE PRECISION FUNCTION OF ONE DOUBLE PRECISION
C ARGUMENT. ANY PARAMETERS CAN BE PASSED FROM THE CALLER VIA
C A COMMON BLOCK.
  
```

```

C PROGRAMMER: ROBERT H. FRENCH DATE: 17 MARCH 1986
C
0002 C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 X=GUESS
0004 SUMAX=DABS(X-BLIM)
0005 SUMAX=DABS(ULIM-X)
0006 DX=DMINI(STEP,SUMAX,SUMAX)
0007 TEST=TOL
0008 F0=F(X)
0009 F1=F(X+DX)
0010 C ARE WE GOING IN THE RIGHT DIRECTION?
0011 IF(F1.LE.F0) GOTO 100
0012 C ... NO, SWITCH DIRECTION
0013 DX=-DX
0014 F1=F(X+DX)
0015 IF(F1.LE.F0) GOTO 100
0016 C ELSE WE MUST BE CLOSE TO A MIN. AT X=GUESS, SO CUT
0017 C STEP SIZE AND TRY AGAIN
0018 DX=DX/10.DO
0019 IF(DABS(DX).GE.TEST) GOTO 10
0020 C CLOSE ENOUGH AT GUESS
0021 I2 XMIN=X
0022 FMIN=F0
0023 RETURN
0024 C NOW GOING RIGHT DIRECTION.
0025 C KEEP GOING UNTIL PAST MINIMUM BY ONE STEP.
0026 I100 X2=X+DX+DX
0027 C HAVE WE REACHED END POINT?
0028 IF(X2.GT.ULIM .OR. X2.LT.BLIM) GOTO 110
  
```

```

0021 C ALL OK F2=F(X2)
0022 C PAST MIN? IF(F2.GE.F1) GOTO 200
0023 C ... NO, STEP AGAIN F0=F1
0024 F1=F2
0025 X=X+DX
0026 GOTO 100
C MIN MAY BE AT AN ENDPOINT. CUT STEP SIZE AND TRY AGAIN
C IF INCREMENT NOT TOO SMALL.
0027 I110 IF(DABS(DX).LE.TEST) GOTO 120
0028 X=X+DX
0029 F0=F1
0030 DX=DX/10.DO
0031 F1=F(X+DX)
0032 GOTO 100
C MIN MUST BE AT THE ENDPOINT (OR WITHIN MINIMUM DX THEREOF)
0033 I120 IF(X2.LE.BLIM) GOTO 122
C MIN AT X=ULIM
0034 XMIN=ULIM
0035 FMIN=F(XMIN)
0036 RETURN
C MIN AT BLIM
0037 I122 XMIN=BLIM
0038 GOTO 121
C HAVE PASSED MIN. IS IT LOCATED CLOSELY ENOUGH YET?
0039 I200 IF(DABS(DX).LE.TEST) GOTO 300
C ... NO, CUT STEP SIZE AND TRY AGAIN
0040 GOTO 115
C DONE!
0041 C SINCE F0 >= F1 & F2 >= F1 AND ABS(DX) < MIN. DX, CALL F1 THE MIN.
0042 FMIN=F1
0043 RETURN
0044 END
  
```

```

0001 SUBROUTINE PSEL(JSUB,MM,PROB)
C RANDOM MFSK/FH IN PARTIAL BAND NOISE JAMMING,
C GIVEN A JAMMING EVENT, WITH CLIPPER RECEIVER
C
C JSUB - JAMMING EVENT VECTOR
C LL - NUMBER OF HOPS/SYMBOL
C MM - ALPHABET SIZE
C PROB - RESULTING CONDITIONAL ERROR PROBABILITY
C
C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
C DIMENSION JSUB(MM), WORK(30), STACK(30), HEAP(30)
C EXTERNAL DG10, PGRAND
C INTEGER NCHAN(0:3)
C COMMON /JAMCNT/ NCHAN
C COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
C TAUK1, TAUK2, TAUK3
C COMMON /PARDEN/ RHOM, RHOT
C COMMON /QUES/ Q0, Q1
C COMMON /SCJAM/ JAMSC
C KSUB=0
C DO 6 I=1,MM
C KSUB=KSUB+JSUB(I)
C CONTINUE
6
C SET UP VALUES WHICH WILL REMAIN IF THIS IS THE NOTHING-JAMMED CASE
Q0=Q(2.00*DSQRT(0.500*RHOM),DSQRT(2.00*TAU))
Q1=1.00
TAUK=0.00
C IF ANYTHING IS JAMMED, SET UP JAMMING-RELATED QUANTITIES
IF(KSUB.NE.0) THEN
BIGK=RHOM/RHOT
BK1=BIGK-1.00
AAB=BIGK/BK1
AAB2=AAB*AAB
BAB=1.00/BK1
BAB2=BAB*BAB
PABAB=AAB*BAB
TAUK=TAU/BIGK
TAUK2=TAU2/BIGK
TAUK3=TAU3/BIGK
Q1=Q(2.00*DSQRT(0.500*RHOT),DSQRT(2.00*TAUK))
END IF
C
C COUNT NUMBER OF NONSIGNAL CHANNELS WITH LM HOPS JAMMED
C
DO 10 I=0,3
NCHAN(I)=0
CONTINUE
10
DO 11 I=2,MM
KSUB=JSUB(I)
NCHAN(KSUB)=NCHAN(KSUB)+1
CONTINUE
11
JAMSC=JSUB(1)
0008
0009
0010
0011
0012
0013
0014
0015
0016
0017
0018
0019
0020
0021
0022
0023
0024
0025
0026
0027
0028
0029
0030
0031
0032
0033
0034
0035
0036
0037
0038

```

```

0001 DOUBLE PRECISION FUNCTION PGRAND(BETA)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 INTEGER NCHAN(0:3)
0004 COMMON /SCJAM/ JAMSC
0005 COMMON /JAMCNT/ NCHAN
0006 COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AAB8AB,
      TAU, TAU2, TAU3, TAU4, TAU5, TAU6, TAU7, TAU8, TAU9
0007 COMMON /PARDEN/ RHOM, RHOT
0008 COMMON /QUES/ QO, Q1
0009 PROD=1.00
0010 DO 10 I=0,3
0011 IF(NCHAN(I).NE.0) THEN
0012 LJAM=I
0013 X=GL(BETA)
0014 PROD=PROD*DXI(X,NCHAN(I))
0015 END IF
0016 CONTINUE
0017 LJAM=JAMSC
0018 Y=PZ1(BETA)
0019 PGRAND=Y*PROD
0020 RETURN
0021 END
  
```

```

0001 SUBROUTINE TEST(ID)
      C TEST RETURN CODE FROM BESSEL FUNCTION
      C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
0004 IF(KODE.EQ.0) RETURN
0005 WRITE(5,1) KODE, ID
0006 FORMAT(' BESSEL FUNCTION CODE = ',12,' FROM CALL NUMBER ',15)
0007 STOP 'FATAL ERROR'
0008 END
  
```

```

0001 SUBROUTINE TEST2(KODE, ID)
      C TEST RETURN CODE FROM ADQUAD/ADQUAZ/ADQUA3
      C
0002 IF(KODE.EQ.0) RETURN
0003 WRITE(5,1) KODE, ID
0004 FORMAT(' ADAPTIVE INTEGRATOR CODE = ',12,' FROM CALL NUMBER ',15)
0005 STOP 'FATAL ERROR'
0006 END
  
```

```

0001 DOUBLE PRECISION FUNCTION PZ1(Y)
      C
0002 SIGNAL CHANNEL P.D.F. WITH CHANGE OF VARIABLE Y=X
      C
0003 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0004 DIMENSION WORK1(30), STACK1(30), HEAP1(30)
0005 LOGICAL *I REG1, REG2, REG3
0006 EXTERNAL DGX, F30A1, F30A2, F30B1,
      F30B2, F31A, F31B1, F31B2, F31B3, F31C2,
      F32A, F32B1, F32B2, F32B3, F32C2,
      F33A1, F33A2, F33B1, F33B2
0007 COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
0008 COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AAB8AB,
      TAU, TAU2, TAU3, TAU4, TAU5, TAU6, TAU7, TAU8, TAU9
0009 COMMON /PARDEN/ RHOM, RHOT
0010 COMMON /QUES/ QO, Q1
0011 COMMON /XCOM/ XCOM
0012 COMMON /OUTER/ XXX, XXXX
0013 XXX=Y
0014 IF(LJAM.GE.1) THEN
0015 XXX=X/Y/BIGK
0016 YTK=(Y-TAU)/BIGK
0017 YTK2=(Y-TAU2)/BIGK
0018 END IF
0019 REG1=Y.GE.0.D0 .AND. Y.LT.TAU
0020 REG2=Y.GE.TAU .AND. Y.LT.TAU2
0021 REG3=Y.GE.TAU2 .AND. Y.LT.TAU3
  
```

```

0022 GOTO (3100, 3200, 3300, 3400), LJAM+1
      C
0023 IF(REG1) THEN
0024 BARG1=DSORT(12.00*RHOM*Y)
0025 I01=2
0026 CALL DXBT(3100)
0027 PZ1={Y/(3.00*RHOM)}*DEXP(BARG1-Y-3.00*RHOM)*B1
0028 ELSE IF(REG2) THEN
0029 BARG1=DSORT(8.00*RHOM*(Y-TAU))
0030 I01=1
0031 CALL DXBT(3101)
0032 PART=1.500*00*DSORT(2.00*(Y-TAU)/RHOM)
      *DEXP(BARG1-Y+TAU-2.00*RHOM)*B1
0033 XCOM=Y+3.00*RHOM
0034 CALL ADQUAZ(Y-TAU, TAU, ANSWER, DGX, F30A1, 1, D-9, WORK1, STACK1,
      HEAP1, 30, KOD)
0035 CALL TEST2(KOD, 3100)
0036 PART=PART+ANSWER/2.00
  
```

```

0022 GOTO (3100, 3200, 3300, 3400), LJAM+1
      C
0023 IF(REG1) THEN
0024 BARG1=DSORT(12.00*RHOM*Y)
0025 I01=2
0026 CALL DXBT(3100)
0027 PZ1={Y/(3.00*RHOM)}*DEXP(BARG1-Y-3.00*RHOM)*B1
0028 ELSE IF(REG2) THEN
0029 BARG1=DSORT(8.00*RHOM*(Y-TAU))
0030 I01=1
0031 CALL DXBT(3101)
0032 PART=1.500*00*DSORT(2.00*(Y-TAU)/RHOM)
      *DEXP(BARG1-Y+TAU-2.00*RHOM)*B1
0033 XCOM=Y+3.00*RHOM
0034 CALL ADQUAZ(Y-TAU, TAU, ANSWER, DGX, F30A1, 1, D-9, WORK1, STACK1,
      HEAP1, 30, KOD)
0035 CALL TEST2(KOD, 3100)
0036 PART=PART+ANSWER/2.00
  
```

```
0037 $ CALL ADQUA2(TAU,Y,ANSWER,DGX,F30A2,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
0038 $ CALL TEST2(K00,3101)  
0039 PZ1=PART+ANSWER  
0040 ELSE IF(REG3) THEN  
0041 BARG1=DSQRT(4.00*RHOW*(Y-TAU2))  
0042 I01=0  
0043 CALL DXBT(3103)  
0044 PART=3.00*Q0*DEXP(BARG1-Y+TAU2-RHOW)*B1  
0045 XCOM=Y-TAU-2.00*RHOW  
0046 CALL ADQUA2(Y-TAU,TAU2,ANSWER,DGX,F30B1,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
0047 $ CALL TEST2(K00,3103)  
0048 PART=PART+3.00*Q0*ANSWER  
0049 XCOM=Y+3.00*RHOW  
0050 CALL ADQUA2(Y-TAU,TAU2,ANSWER,DGX,F30B2,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
0051 $ CALL TEST2(K00,3104)  
0052 PZ1=PART+ANSWER  
0053 ELSE  
0054 PZ1=0.00  
0055 END IF  
0056 GOTO 9000  
C  
C  
C 3200 IF(REG1) THEN  
0057 XCOM=YK+2.00*RHOW-RHOT  
0058 CALL ADQUA2(0.00,Y,ANSWER,DGX,F31A,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
0059 $ CALL TEST2(K00,3200)  
0060 PZ1=ANSWER*0.500/BIGK  
0061 ELSE IF(REG2) THEN  
0062 BARG1=DSQRT(8.00*RHOW*(Y-TAU))  
0063 I01=1  
0064 PART=0.500*Q1*DEXP(BARG1-Y+TAU-2.00*RHOW)*B1  
0065 XCOM=YK+2.00*RHOW-RHOT  
0066 CALL ADQUA2(Y-TAU,TAU,ANSWER,DGX,F31B1,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
0067 $ CALL TEST2(K00,3201)  
0068 PART=PART+0.500*ANSWER/BIGK  
0069 XCOM=YK+RHOW-RHOT  
0070 CALL ADQUA2(0.00,Y-TAU,ANSWER,DGX,F31B2,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
0071 $ CALL TEST2(K00,3202)  
0072 PART=PART+2.00*Q0*ANSWER/BIGK  
0073 XCOM=Y+2.00*RHOW-RHOT  
0074 CALL ADQUA2(TAU,Y,ANSWER,DGX,F31B3,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
0075 $ CALL TEST2(K00,3203)  
0076 PZ1=PART+ANSWER/BIGK  
0077  
0078
```

```
0079 ELSE IF(REG3) THEN  
0080 BARG1=DSQRT(4.00*RHOT*(YK-TAU2))  
0081 I01=0  
0082 CALL DXBT(3204)  
0083 PART=Q0*Q0*DEXP(BARG1-YK+TAUK2-RHOT)*B1/BIGK  
0084 BARG1=DSQRT(4.00*RHOW*(Y-TAU2))  
0085 I01=0  
0086 CALL DXBT(3205)  
0087 PART=PART+1.00*Q0*Q1*DEXP(BARG1-Y+TAU2-RHOW)*B1  
0088 XCOM=Y-TAU-2.00*RHOW  
0089 CALL ADQUA2(Y-TAU,TAU2,ANSWER,DGX,F30B1,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
$  
0090 CALL TEST2(K00,3204)  
0091 PART=PART+Q1*ANSWER  
0092 XCOM=YK-TAU-RHOW-RHOT  
0093 CALL ADQUA2(Y-TAU,TAU2,ANSWER,DGX,F31C2,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
$  
0094 CALL TEST2(K00,3205)  
0095 PART=PART+2.00*Q0*ANSWER/BIGK  
0096 XCOM=YK+2.00*RHOW-RHOT  
0097 CALL ADQUA2(Y-TAU,TAU2,ANSWER,DGX,F31B3,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
$  
0098 CALL TEST2(K00,3206)  
0099 PZ1=PART+ANSWER/BIGK  
0100 ELSE  
0101 PZ1=0.00  
0102 END IF  
0103 GOTO 9000  
C  
C  
C TWO HOPS JAMMED  
C 3300 IF(REG1) THEN  
0104 XCOM=Y+2.00*RHOT-RHOW  
0105 CALL ADQUA2(0.00,Y,ANSWER,DGX,F32A,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
0106 $ CALL TEST2(K00,3301)  
0107 PZ1=0.500*ANSWER/BIGK  
0108 ELSE IF(REG2) THEN  
0109 BARG1=DSQRT(8.00*RHOT*YTK)  
0110 I01=1  
0111 CALL DXBT(3302)  
0112 PART=0.500*Q0*DSQRT(2.00*(Y-TAU)/RHOW)  
0113 *DEXP(BARG1-YK+TAUK-2.00*RHOT)*B1/BIGK  
0114 XCOM=Y+2.00*RHOT-RHOW  
0115 CALL ADQUA2(Y-TAU,TAU,ANSWER,DGX,F32B1,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
$  
0116 CALL TEST2(K00,3302)  
0117 PART=PART+ANSWER*0.500/BIGK  
0118 XCOM=Y-TAU-RHOW-RHOT  
0119 CALL ADQUA2(0.00,Y-TAU,ANSWER,DGX,F32B2,1.0-9,WORK1,STACK1,  
HEAP1,30,K00)  
$  
0120 CALL TEST2(K00,3303)
```



```
0033 ELSE IF (REG3) THEN  
0034 YMT2=Y-TAU2  
0035 GL=1.DO-DEXP(TAU-TAUK-Y)*(AAB*(1.DO+Y)-AABBAB-AAB*YMT2)  
$ -BAB2*DEXP(TAUK2-TAU2-YK)  
0036 ELSE IF (Y.GE.TAU3) THEN  
0037 GL=1.DO  
0038 ELSE  
0039 GL=0.DO  
0040 END IF  
0041 GOTO 9000  
C  
C TMO HOPS JAMMED  
C  
3300 IF (REG1) THEN  
0042 GL=1.DO-BAB2*DEXP(-Y)-AAB*DEXP(-YK)  
0043 * (1.DO-BAB*YK)  
$  
0044 ELSE IF (REG2) THEN  
0045 GL=1.DO-DEXP(TAUK-YK)  
$ *(DEXP(-Y)*(BAB2-BAB*YTK)  
$ +DEXP(-YK)*(AAB2-BAB*(Y-TAU2)))  
$ +2.DO+AABBAB*DEXP(TAU-TAUK-Y)  
0046 ELSE IF (REG3) THEN  
0047 GL=1.DO-DEXP(TAUK-TAU-YK)*(BAB*YTK2  
$ -BAB*(1.DO+YK)-AABBAB)  
$ -AAB2*DEXP(TAU2-TAUK2-Y)  
0048 ELSE IF (Y.GE.TAU3) THEN  
0049 GL=1.DO  
0050 ELSE  
0051 GL=0.DO  
0052 END IF  
0053 GOTO 9000  
C  
C THREE HOPS JAMMED  
C  
3400 IF (Y.GE.0.DO .AND. Y.LT.TAU) THEN  
0054 GL=1.DO-DEXP(-YK)*(1.DO+YK+0.500*YK*YK)  
0055 ELSE IF (Y.GE.TAU .AND. Y.LT.TAU2) THEN  
0056 GL=1.DO-DEXP(-YK)*(1.DO+TAUK+0.500*TAUK*TAUK  
$ +(1.DO+TAUK)*YTK-YTK*YTK)  
0057 ELSE IF (Y.GE.TAU2 .AND. Y.LT.TAU3) THEN  
0058 GL=1.DO-DEXP(-YK)*(1.DO+2.DO*TAUK  
$ +0.500*TAUK*TAUK-(2.DO+TAUK)*YTK2+0.500*YTK2*YTK2)  
0059 ELSE IF (Y.GE.TAU3) THEN  
0060 GL=1.DO  
0061 ELSE  
0062 GL=0.DO  
0063 END IF  
0064 CONTINUE  
0065 RETURN  
0066  
0067
```

```
0001 DOUBLE PRECISION FUNCTION GL(Y)  
C  
C NON-SIGNAL CHANNEL CUMULATIVE DISTRIBUTION FUNCTION  
C WITH CHANGE OF VARIABLE Y=A  
C  
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)  
0003 LOGICAL*1 REG1, REG2, REG3  
0004 COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,  
$ TAU, TAU2, TAU3, TAUK, TAUK2, TAUK3  
0005 REG1=Y.GE.0.DO .AND. Y.LT.TAU  
0006 REG2=Y.GE.TAU .AND. Y.LT.TAU2  
0007 REG3=Y.GE.TAU2 .AND. Y.LT.TAU3  
0008 IF (LJAM.GT.0) THEN  
0009 YK=Y/BIGK  
0010 YTK=(Y-TAU)/BIGK  
0011 YTK2=(Y-TAU2)/BIGK  
0012 END IF  
C  
C THREE HOPS PER SYMBOL  
C  
3000 GOTO (3100, 3200, 3300, 3400), LJAM+1  
C  
C NO HOPS JAMMED  
C  
3100 IF (REG1) THEN  
0014 GL=1.DO-(1.DO+Y+0.500*Y*Y)*DEXP(-Y)  
0015 ELSE IF (REG2) THEN  
0016 YMT=Y-TAU2  
0017 YMT=Y-TAU  
0018 GL=1.DO-(1.DO+TAU+0.500*TAU*TAU  
$ +(1.DO+TAU)*YMT-YMT*YMT)*DEXP(-Y)  
0019 ELSE IF (REG3) THEN  
0020 YMT2=Y-TAU2  
0021 GL=1.DO-(1.DO+TAU+TAU+0.500*TAU*TAU  
$ -(2.DO+TAU)*YMT2+0.500*YMT2*YMT2)*DEXP(-Y)  
0022 ELSE IF (Y.GE.TAU3) THEN  
0023 GL=1.DO  
0024 ELSE  
0025 GL=0.DO  
0026 END IF  
0027 GOTO 9000  
C  
C ONE HOP JAMMED  
C  
3200 IF (REG1) THEN  
0028 GL=1.DO-AAB2*DEXP(-YK)+BAB*(1.DO+AAB*Y)*DEXP(-Y)  
0029 ELSE IF (REG2) THEN  
0030 YMT=Y-TAU  
0031 GL=1.DO-DEXP(-YMT)*(DEXP(-YK)*(AAB*(AAB+YMT))+  
$ DEXP(-Y)*(BAB*(BAB+Y-TAU2)))  
$ +2.DO*AABBAB*DEXP(TAUK-TAU-YK)  
0032
```



```

0001      DOUBLE PRECISION FUNCTION F3081(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=0,
C THIRD REGION, FIRST INTEGRAL
C
0002      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
0004      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      $      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
0005      COMMON /PARDEN/ RHOM, RHOT
0006      COMMON /QUES/ Q0, Q1
0007      COMMON /XCOM/ XCON
0008      COMMON /OUTER/ XXX, XXXX
0009      BARG1=DSORT(4.D0)*HOM*(XXX-U)
0010      BARG2=DSORT(4.D0)*RHOM*(U-TAU)
0011      I01=0
0012      I02=0
0013      CALL BPROD(3120)
0014      F3081=DEXP(BARG1+BARG2-XCON)*B1*B2
0015      RETURN
0016      END

```

```

0001      DOUBLE PRECISION FUNCTION F31A(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=1,
C FIRST REGION
C
0002      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      $      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
0004      COMMON /PARDEN/ RHOM, RHOT
0005      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
0006      COMMON /QUES/ Q0, Q1
0007      COMMON /XCOM/ XCON
0008      COMMON /OUTER/ XXX, XXXX
0009      BARG1=DSORT(4.D0)*RHOT*(XXXX-U/BIGK)
0010      BARG2=DSORT(2.D0*RHOM*U)
0011      I01=0
0012      I02=1
0013      CALL BPROD(3210)
0014      F31A=DEXP(BARG1+BARG2-XCON)*B1*B2*DSORT(2.D0*U/RHOM)
0015      RETURN
0016      END

```

```

0001      DOUBLE PRECISION FUNCTION F3082(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=0,
C SECOND REGION, SECOND INTEGRAL
C
0002      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003      EXTERNAL DGTEN, F308A
0004      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
0005      COMMON /INWORK/ WORK(30), STACK(30), HEAP(30)
0006      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      $      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
0007      COMMON /PARDEN/ RHOM, RHOT
0008      COMMON /QUES/ Q0, Q1
0009      COMMON /XCOM/ XCON
0010      COMMON /OUTER/ XXX, XXXX
0011      COMMON /INNER/ WNN, WNNK
0012      WNN=U
0013      CALL ADQUA3(U-TAU,TAU,ANSWER,DGTEN,F308A,I.D-10,
      $      WORK,STACK,HEAP,30,KOD)
0014      CALL TEST2(KOD,3120)
0015      BARG1=DSORT(4.D0*RHOM*(XXX-U))
0016      I01=0
0017      CALL OXBT(3121)
0018      F3082=DEXP(BARG1-XCON)*B1*ANSWER
0019      RETURN
0020      END

```

```

0001      DOUBLE PRECISION FUNCTION F31B1(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=1,
C SECOND REGION, FIRST INTEGRAL
C
0002      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      $      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
0004      COMMON /PARDEN/ RHOM, RHOT
0005      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
0006      COMMON /QUES/ Q0, Q1
0007      COMMON /XCOM/ XCON
0008      COMMON /OUTER/ XXX, XXXX
0009      BARG1=DSORT(8.D0*RHOM*U)
0010      UK=U/BIGK
0011      BARG2=DSORT(4.D0*RHOM*(XXXX-UK))
0012      I01=1
0013      I02=0
0014      CALL BPROD(3220)
0015      S=DSORT(2.D0*U/RHOM)
0016      F31B1=DEXP(BARG1+BARG2-XCON-U*UK)*B1*B2*S
0017      RETURN
0018      END

```



```
0001 DOUBLE PRECISION FUNCTION F33A2(U)
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=3,
C SECOND INTEGRAL
  IMPLICIT DOUBLE PRECISION(A-H,O-Z)
  EXTERNAL DGTEN, F33ABA
  COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
  COMMON /WORK/ WORK(30), STACK(30), HEAP(30)
  COMMON /DENPAR/ BIGK, AAB, BAB, LAM, AAB2, BAB2, AABRAB,
  & TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
  COMMON /PARDEN/ RHOM, RHOT
  COMMON /QUES/ Q0, Q1
  COMMON /XCOM/ XCOM
  COMMON /OUTER/ XXX, XXXX
  COMMON /INNER/ WNN, WNNK
  UK=U/BIGK
  WNNK=UK
  CALL ADQUA3(U-TAU,TAU,ANSWER,DGTEN,F33ABA,1,D-10,
  & WORK,STACK,HEAP,30,KOD)
  CALL TEST2(KOD,3412)
  BARG1=DSQRT(4.D0*RHOM*(XXXK-UK))
  I01=0
  CALL DXBT(3112)
  F33A2=DEXP(BARG1-XCOM)*B1*ANSWER
  RETURN
  END
```

```
0001 DOUBLE PRECISION FUNCTION F33ABA(U)
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=3,
C SECOND INTEGRAL'S INNER INTEGRAL; ALSO L=3, LJAM=2, SECOND
C REGION, THIRD INTEGRAL, INNER INTEGRAL
  IMPLICIT DOUBLE PRECISION(A-H,O-Z)
  COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
  COMMON /DENPAR/ BIGK, AAB, BAB, LAM, AAB2, BAB2, AABRAB,
  & TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
  COMMON /PARDEN/ RHOM, RHOT
  COMMON /QUES/ Q0, Q1
  COMMON /XCOM/ XCOM
  COMMON /OUTER/ XXX, XXXX
  COMMON /INNER/ WNN, WNNK
  UK=U/BIGK
  BARG1=DSQRT(4.D0*RHOT*(WNNK-UK))
  CALL DBESI(BARG1,0,B1,KODE)
  CALL TEST(13410)
  BARG2=DSQRT(4.D0*RHOT*(WNNK-UK))
  CALL DBESI(BARG2,0,B2,KODE)
  CALL TEST(13411)
  F33ABA=B1*B2
  RETURN
  END
```

```
0001 DOUBLE PRECISION FUNCTION F33B1(U)
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=3,
C THIRD REGION, FIRST INTEGRAL
  IMPLICIT DOUBLE PRECISION(A-H,O-Z)
  COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
  COMMON /DENPAR/ BIGK, AAB, BAB, LAM, AAB2, BAB2, AABRAB,
  & TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
  COMMON /PARDEN/ RHOM, RHOT
  COMMON /QUES/ Q0, Q1
  COMMON /XCOM/ XCOM
  COMMON /OUTER/ XXX, XXXX
  UK=U/BIGK
  BARG1=DSQRT(4.D0*RHOT*(XXXK-UK))
  BARG2=DSQRT(4.D0*RHOT*(UK-TAUK))
  I01=0
  CALL BPROD(3420)
  F33B1=DEXP(BARG1+BARG2-XCOM)*B1*B2
  RETURN
  END
```

```
0001 DOUBLE PRECISION FUNCTION F33B2(U)
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=3,
C SECOND REGION, SECOND INTEGRAL
  IMPLICIT DOUBLE PRECISION(A-H,O-Z)
  EXTERNAL DGTEN, F33ABA
  COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
  COMMON /WORK/ WORK(30), STACK(30), HEAP(30)
  COMMON /DENPAR/ BIGK, AAB, BAB, LAM, AAB2, BAB2, AABRAB,
  & TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
  COMMON /PARDEN/ RHOM, RHOT
  COMMON /QUES/ Q0, Q1
  COMMON /XCOM/ XCOM
  COMMON /OUTER/ XXX, XXXX
  COMMON /INNER/ WNN, WNNK
  UK=U/BIGK
  WNNK=UK
  CALL ADQUA3(U-TAU,TAU,ANSWER,DGTEN,F33ABA,1,D-10,
  & WORK,STACK,HEAP,30,KOD)
  CALL TEST2(KOD,3420)
  BARG1=DSQRT(4.D0*RHOT*(XXXK-UK))
  I01=0
  CALL DXBT(3421)
  F33B2=DEXP(BARG1-XCOM)*B1*ANSWER
  RETURN
  END
```

```
0001      DOUBLE PRECISION FUNCTION F32B2(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=2,
C SECOND REGION, SECOND INTEGRAL
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      $      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU2, TAU3
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /QUES/ Q0, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      UK=U/BIGK
      BARG1=DSQRT(4.DO+RHOM*(XXX-TAU-U))
      I01=0
      BARG2=DSQRT(4.DO+RHOM*(XXX-TAU-U))
      I02=0
      CALL BPROD(3322)
      F32B2=DEXP(BARG1+BARG2-XCOM-UK+U)*B1*B2
      RETURN
      END
```

```
0001      DOUBLE PRECISION FUNCTION F32A(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=2,
C FIRST REGION
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      $      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU2, TAU3
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /QUES/ Q0, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      UK=U/BIGK
      BARG1=DSQRT(4.DO+RHOM*(XXX-U))
      BARG2=DSQRT(2.DO+RHOT*(UK))
      I01=0
      I02=1
      CALL BPROD(3310)
      F32A=DEXP(BARG1+BARG2-XCOM)*B1*B2*DSQRT(2.DO+U/RHOM)
      RETURN
      END
```

```
0001      DOUBLE PRECISION FUNCTION F32B3(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=2,
C SECOND REGION, THIRD INTEGRAL
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      EXTERNAL DGTEN, F33ABA
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /INMORK/ WOK:(30), STACK(30), HEAP(30)
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      $      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU2, TAU3
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /QUES/ Q0, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      COMMON /INNER/ MMM, MMMM
      UK=U/BIGK
      WMMK=UK
      CALL ADQUA3(U-TAU,TAU,ANSWER,DGTEN,F33ABA,I,D-I0,
      $      WOK,STACK,HEAP,30,KOD)
      CALL TEST2(KOD,3322)
      BARG1=DSQRT(4.DO+RHOM*(XXX-U))
      I01=0
      CALL DXBT(3324)
      F32B3=DEXP(BARG1-XCOM-UK+U)*B1*ANSWER
      RETURN
      END
```

```
0001      DOUBLE PRECISION FUNCTION F32B1(U)
C
C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=2,
C SECOND REGION, FIRST INTEGRAL
C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      $      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU2, TAU3
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /QUES/ Q0, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      UK=U/BIGK
      BARG1=DSQRT(8.DO+RHOT*(UK))
      BARG2=DSQRT(4.DO+RHOM*(XXX-U))
      I01=1
      I02=0
      CALL BPROD(3320)
      S=DSQRT(2.DO+U/RHOM)
      F32B1=DEXP(BARG1+BARG2-XCOM-UK+U)*B1*B2*S
      RETURN
      END
```

```

0001 SUBROUTINE TIES(JSUB,MM,PTIE)
      C COMPUTE PROBABILITY OF CORRECT DECISION GIVEN THAT
      C SEVERAL SATURATED CHANNELS ARE TIED
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      DIMENSION JSUB(MM), LLOW(7), LINC(7), LUP(7), MU(7),
      PZLM(2:8)
      LOGICAL I=1 GO
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /QUES/ Q0, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      UK=U/BIGK
      BARG1=DSORT(4.DO*RHOM*(XXX-U))
      I01=0
      BARG2=DSORT(4.DO*RHOT*(UK-TAUK1))
      I02=0
      CALL BPROD(3330)
      F32C2=DEXP(BARG1+BARG2-XCOM-UK+U)*B1*B2
      RETURN
      END
  
```

```

0001 DOUBLE PRECISION FUNCTION F32C2(U)
      C INTEGRAND FUNCTION FOR SIGNAL CHANNEL DENSITY, L=3, LJAM=2,
      C THIRD REGION, SECOND INTEGRAL
      C
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      COMMON /LOCALS/ BARG1, BARG2, B1, B2, KODE, I01, I02
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /QUES/ Q0, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      UK=U/BIGK
      BARG1=DSORT(4.DO*RHOM*(XXX-U))
      I01=0
      BARG2=DSORT(4.DO*RHOT*(UK-TAUK1))
      I02=0
      CALL BPROD(3330)
      F32C2=DEXP(BARG1+BARG2-XCOM-UK+U)*B1*B2
      RETURN
      END
  
```

J-27

```

0001 SUBROUTINE TIES(JSUB,MM,PTIE)
      C COMPUTE PROBABILITY OF CORRECT DECISION GIVEN THAT
      C SEVERAL SATURATED CHANNELS ARE TIED
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      DIMENSION JSUB(MM), LLOW(7), LINC(7), LUP(7), MU(7),
      PZLM(2:8)
      LOGICAL I=1 GO
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /QUES/ Q0, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      UK=U/BIGK
      BARG1=DSORT(4.DO*RHOM*(XXX-U))
      I01=0
      BARG2=DSORT(4.DO*RHOT*(UK-TAUK1))
      I02=0
      CALL BPROD(3330)
      F32C2=DEXP(BARG1+BARG2-XCOM-UK+U)*B1*B2
      RETURN
      END
  
```

```

0001 SUBROUTINE TIES(JSUB,MM,PTIE)
      C COMPUTE PROBABILITY OF CORRECT DECISION GIVEN THAT
      C SEVERAL SATURATED CHANNELS ARE TIED
      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
      DIMENSION JSUB(MM), LLOW(7), LINC(7), LUP(7), MU(7),
      PZLM(2:8)
      LOGICAL I=1 GO
      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
      TAU, TAU2, TAU3, TAU, TAU2, TAU3, TAU, TAU2, TAU3
      COMMON /PARDEN/ RHOM, RHOT
      COMMON /QUES/ Q0, Q1
      COMMON /XCOM/ XCOM
      COMMON /OUTER/ XXX, XXXX
      UK=U/BIGK
      BARG1=DSORT(4.DO*RHOM*(XXX-U))
      I01=0
      BARG2=DSORT(4.DO*RHOT*(UK-TAUK1))
      I02=0
      CALL BPROD(3330)
      F32C2=DEXP(BARG1+BARG2-XCOM-UK+U)*B1*B2
      RETURN
      END
  
```

J-27

```

0001      DOUBLE PRECISION FUNCTION PUNJAM(ETA)
C      C FUNCTION FOR UNJAMMED P(E) FOR OPT. THRESHOLD SEARCH
C
C      NOTE: WHEN JAMMING EVENT IS (0,0,....,0), THE VARIABLES
C      BIGK, AAB, BAB, TAU1, TAU2, TAU3, AND TAU4 ARE NOT
C      USED IN THE COMPUTATIONS, AND HENCE DO NOT NEED
C      TO BE SET UP BEFORE CALLING PSEL FROM THIS FUNCTION
C
0002      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003      DIMENSION NOJAM(4)
0004      COMMON /INPUTS/ DEBINOL(3),NSLOTS,K,MM
0005      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
          $      TAU, TAU2, TAU3, TAU4, TAU1, TAU2, TAU3, TAU4
0006      DATA NOJAM/0,0,0,0/
0007      LJAM=0
0008      TAU=ETA
0009      TAU2=TAU+ETA
0010      TAU3=TAU2+ETA
0011      CALL PSEL(NOJAM,MM,P)
0012      PUNJAM=P
0013      RETURN
0014      END
  
```

```

0001      SUBROUTINE SETTAU(MM,PEOO)
C      C SEARCH FOR OPTIMUM THRESHOLD IN ABSENCE OF JAMMING
C
C      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0002      EXTERNAL PUNJAM
0003      COMMON /DENPAR/ BIGK, AAB, BAB, LJAM, AAB2, BAB2, AABAB,
          $      TAU, TAU2, TAU3, TAU4, TAU1, TAU2, TAU3, TAU4
0004      COMMON /PARDEN/ RHOM, RHOT
0005      COMMON /QUES/ QO, Q1
0006      LJAM=0
0007
C      C GUESS BASED ON PREVIOUS RESULTS
C
0008      GUESS=8.DO
0009      STEP=1.DO
0010      C FOR M=4, L=3 THE OPTIMUM THRESHOLD IS 8.15
          $      CALL MINSER(PUNJAM,PEMIN,TAUOPT,STEP,GUESS,0.DO,
          $      50.DO,0.01DO)
0011      TAU=TAUOPT
0012      TAU2=TAU+TAU
0013      TAU3=TAU2+TAU
0014      PEOO=PEMIN
0015      RETURN
0016      END
  
```

APPENDIX K
COMPUTER PROGRAM FOR
SELF-NORMALIZING RECEIVER WITH M=2, L=2

The following pages contain the source code listing for the FORTRAN-77 program used for numerical computations of the performance of the self-normalizing receiver when M=2 and L=2, with jamming fraction γ as a parameter.

```

0001 PROGRAM SELF22
C SELF NORMALIZING RECEIVER, M=2, L=2
C ANALYSIS: L.E. MILLER
C PROGRAM: R.H. FRENCH
C
C IMPLICIT DOUBLE PRECISION (A-H,O-Z)
C PARAMETER(M=2, L=2, NSLOTS=2400)
C CHARACTER*13 FNAME
C REAL DEBNJ(126), PELOG(126)
C COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGX
C COMMON /PARMS/ MO, NJ, MG, DEBNOL(5), GAMLST(10),
C DEBNJL(126), DJ, DBJO
C CALL ERRSET(29, .TRUE., .FALSE., .TRUE., .FALSE., 15)
C CALL GET
C DO 900 IO=1, MO
C EBNOL=IO, DO=((DEBNOL(IO)/10.00)
C RHOM=EBNOL/2.00
C TORUT=DEBNOL(IO)
C DO 800 IG=1, MG
C GAMMA=GAMLST(IG)
C IO=GAMMA*NSLOTS+0.500
C IGOUT=1000.00*GAMMA+0.500
C
C C PROGRESS FILE
C
C WRITE(FNAME,1) IOOUT,IGOUT
C 1 FORMAT('S22',I2.2,I4.4,'.DAT')
C WRITE(5,2) FNAME
C 2 FORMAT(' WORKING ON FILE ',A13)
C OPEN(UNIT=4, FILE=FNAME, STATUS='OLD', ERR=810,
C FORM='UNFORMATTED', ACCESS='SEQUENTIAL')
C
C HAVE A PROGRESS FILE, READ IT
C
C READ(6) EBNJIN, GAMMIN, DBJOIN, DJJM
C IF(EBNJIN.NE.DEBNOL(10) .OR. GAMMIN.NE.GAMMA .OR.
C DBJOIN.NE.DBJO .OR. DJJM.NE.DJ) STOP 'FILE CORRUPT'
C JJ=0
C 801 JJ=JJ+1
C READ(4, END=802) DEBNJ(JJ), PELOG(JJ)
C GO TO 801
C 802 CLOSE(UNIT=4)
C GO TO 820
C
C NO FILE, MUST CREATE IT
C
C 810 OPEN(UNIT=4, FILE=FNAME, STATUS='NEW', FORM='UNFORMATTED',
C ACCESS='SEQUENTIAL')
C WRITE(4) DEBNOL(10), GAMMA, DBJO, DJ
C CLOSE(UNIT=4)
0034 JJ=1
C KEEP ON GOING
C
0035 DO 700 IJ=JJ, NJ
0036 WRITE(5,821) IJ
0037 FORMAT(' IJ = ', I3)
0038 DEBNJ=DBJO*(IJ-1)*DJ
0039 DEBNJ(IJ)=DEBNJ
0040 EBNJ=10.00*((DEBNJ/10.00)
0041 RHOJ=EBNJ/2.00
0042 RHOJ=GAMMA*RHOJ
0043 RHOJ=RHOJ*(RHOM/(RHOJ*(RHOM)
0044 BIGX=RHOM/RHOT
0045 CALL PSURE(10, PE)
0046 PELOG(IJ)=DLOG10(PE)
0047 OPEN(UNIT=4, FILE=FNAME, STATUS='OLD', ACCESS='APPEND',
C FORM='UNFORMATTED')
C WRITE(4) DEBNJ(IJ), PELOG(IJ)
0048 CLOSE(UNIT=4)
0049 CONTINUE
0050 OPEN(UNIT=4, FILE=FNAME, STATUS='NEW', ACCESS='SEQUENTIAL',
C FORM='UNFORMATTED')
0051 WRITE(4) M, L, DEBNOL(10), GAMMA, NSLOTS, NJ, DEBNJ, PELOG
C CLOSE(UNIT=4)
0052 WRITE(6,710) M, L, DEBNOL(10), GAMMA, NSLOTS,
C (DEBNJ(IJ), IO, DO=((PELOG(IJ), IJ=1, NJ)
0053 CLOSE(UNIT=4)
0054 WRITE(5,710) M, L, DEBNOL(10), GAMMA, NSLOTS,
C (DEBNJ(IJ), IO, DO=((PELOG(IJ), IJ=1, NJ)
0055 $ 5X, 'EB/NO = ', F8.5, ' dB', 5X, 'GAMMA= ', I10.3, 5X, 'NSLOTS= ', I4,
C ' EB/NJ (dB)', 8X, 'P(E) / <N> (4X, OPF4.1, 8X, I10.3 /))
C CONTINUE
0056 CONTINUE
0057 CONTINUE
0058 STOP 'PLEASE PURGE S22*.DAT'
0059 END

```

```

0001 SUBROUTINE GET
0002 C
0003 C INTERACTIVE RUN PARAMETER INPUTS
0004 C
0005 C
0006 C
0007 C
0008 C
0009 C
0010 C
0011 C
0012 C
0013 C
0014 C
0015 C
0016 C
0017 C
0018 C
0019 C
0020 C
0021 C
0022 C
0023 C
0024 C
0025 C
0026 C
0027 C
0028 C
0029 C
0030 C
0031 C
0032 C
0033 C
0034 C
0035 C
0036 C
0037 C
0038 C
0039 C
0040 C
0041 C
0042 C
0043 C
0044 C
0045 C

IMPLICIT DOUBLE PRECISION(A-H,O-Z)
DIMENSION DGAM(10), DOLST(5)
CHARACTER*8 REPLY, BLANKS
COMMON /PARMS/ NO, NJ, NG, DEBINOL(5), GAMLST(10),
DEBINJL(126), DJ, DBJO
$ DATA DGAM/1.0-3, 2.0-3, 5.0-3,
1.0-2, 2.0-2, 5.0-2,
1.0-1, 2.0-1, 5.0-1, 1.00/
$ DATA DOLST/ 13.3524700, 12.313300, 10.9444300, 14.8925300,
16.02713500/
$ DATA BLANKS/ ' ' /
1 WRITE(5,2)
2 FORMAT(' HOW MANY EB/MO? [11] ', $)
3 READ(5,3,ERR=1) NO
4 FORMAT(I1)
5 IF(NO.EQ.0) NO=1
6 IF(NO.LT.0 .OR. NO.GT.5) GOTO 1
7 DO 8 IN=1,NO
8 WRITE(5,5) IN, DOLST(IN)
9 FORMAT(3X, 'EB/MO(' , I1, ') [' , F8.5, ' dB]: ', $)
10 READ(5,6,ERR=4) REPLY
11 FORMAT(A8)
12 IF(REPLY.EQ.BLANKS) THEN
13 DEBINOL(IN)=DOLST(IN)
14 ELSE
15 READ(REPLY,7,ERR=4) DEBINOL(IN)
16 FORMAT(F8.5)
17 END IF
18 CONTINUE
19 WRITE(5,10)
20 FORMAT(' HOW MANY GAMMA? [10] ', $)
21 READ(5,11,ERR=9) NG
22 FORMAT(I2)
23 IF(NG.EQ.0) NG=10
24 IF(NG.LT.0 .OR. NG.GT.10) GOTO 9
25 DO 15 IN=1,NG
26 WRITE(5,13) IN, DGAM(IN)
27 FORMAT(3X, 'GAMMA(' , I2, ') [' , F5.3, ']: ', $)
28 READ(5,14,ERR=12) GAMLST(IN)
29 FORMAT(F8.6)
30 IF(GAMLST(IN).EQ.0.00) GAMLST(IN)=DGAM(IN)
31 IF(GAMLST(IN).LE.0.00 .OR. GAMLST(IN).GT.1.00) GOTO 12
32 CONTINUE
33 WRITE(5,17)
34 FORMAT(' HOW MANY EB/NJ? [126] ', $)
35 READ(5,18,ERR=16) NJ
36 FORMAT(I3)
37 IF(NJ.EQ.0) NJ=126
  
```

```

0046 IF(NJ.LT.0 .OR. NJ.GT.126) GOTO 16
0047 WRITE(5,20)
0048 FORMAT(' STARTING VALUE FOR EB/NJ [0 dB]: ', $)
0049 READ(5,21,ERR=19) DBJO
0050 FORMAT(F5.2)
0051 DJ=0.00
0052 IF(NJ.GT.1) THEN
0053 WRITE(5,23)
0054 FORMAT(' INCREMENT FOR EB/NJ [0.4 dB]: ', $)
0055 READ(5,24,ERR=22) DJ
0056 FORMAT(F5.0)
0057 IF(DJ.EQ.0.00) DJ=0.400
0058 END IF
0059 RETURN
0060 END
  
```

```

0042 CALL ADQUAD(0.DO,1.DO,PART,D616,F10B,1.D-9,WORK,STACK,HEAP,
$      20,KODE)
0043 IF(KODE.NE.0) STOP 'ADQUAD ERROR SIXTH CALL'
0044 PART=PI1*PI1*PART
0045 PB=PB+PART
0046 CALL ADQUAD(0.DO,1.DO,PART,D616,F10C,1.D-9,WORK,STACK,HEAP,
$      20,KODE)
0047 IF(KODE.NE.0) STOP 'ADQUAD ERROR SEVENTH CALL'
0048 PART=2.DO*PI1*PI2*PART
0049 PB=PB+PART
0050 PART=0.500*DEXP(-RHOT)*(1.DO+RHOT/3.DO)
0051 PART=PI2*PI2*PART
0052 PB=PB+PART
0053 PE=PB
0054 RETURN
0055 END

```

```

0001 DOUBLE PRECISION FUNCTION P101(Z)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0004 D=1.D+*(BIGK-1.DO)*Z
0005 X=BIGK*RHOM*Z/D
0006 P101=BIGK*DEXP(X-RHOM)*(1.DO+X)/(D*D)
0007 RETURN
0008 END

```

```

0001 DOUBLE PRECISION FUNCTION P110(Z)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0004 D=BIGK*(BIGK-1.DO)*Z
0005 X=RHOT*Z/D
0006 P110=BIGK*DEXP(X-RHOT)*(1.DO+X)/(D*D)
0007 RETURN
0008 END

```

```

0001 DOUBLE PRECISION FUNCTION FO1A(V)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0004 FO1A=(1.DO-V)*DEXP(-RHOM*V)*PI01(V)
0005 RETURN
0006 END

```

```

0001 SUBROUTINE PSUBE(IQ,PE)
C COMPUTE ERROR PROBABILITY
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 PARAMETER (SLOTS=2400.DO)
0004 PARAMETER (SLOTS1=2399.DO)
0005 DIMENSION WORK(20), STACK(20), HEAP(20)
0006 EXTERNAL D616, FO1A, FO1B, FO1C, FO1D, F10B, F10C
0007 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
C COMPUTE ELEMENTAL EVENT PROBABILITIES
C

```

```

Q=IQ
PI2=Q*(Q-1.DO)/SLOTPR
PI1=Q*(SLOTS-Q)/SLOTPR
PI0=(SLOTS-Q)*(SLOTS-Q)/SLOTPR
PART=PI0*PI0*0.500*DEXP(-RHOM)*(1.DO+RHOM/3.DO)
PB=PART
CALL ADQUAD(0.DO,1.DO,PART,D616,FO1A,1.D-9,WORK,STACK,HEAP,
$      20,KODE)
IF(KODE.NE.0) STOP 'ADQUAD ERROR FIRST CALL'
PART=2.DO*PI0*PI1*PART
PB=PB+PART
CALL ADQUAD(0.DO,1.DO,PART,D616,FO1B,1.D-9,WORK,STACK,HEAP,
$      20,KODE)
IF(KODE.NE.0) STOP 'ADQUAD ERROR SECOND CALL'
PART=PI1*PI1*PART
PB=PB+PART
CALL ADQUAD(0.DO,1.DO,PART,D616,FO1A,1.D-9,WORK,STACK,HEAP,
$      20,KODE)
IF(KODE.NE.0) STOP 'ADQUAD ERROR THIRD CALL'
PART=2.DO*PI0*PI1*PART
PB=PB+PART
RDIFF=RHOM-RHOT
RSUM=RHOM+RHOT
PART=DEXP(-RHOT)*(RHOM+RDIFF-RSUM)
PART=PART*DEXP(-RHOM)*(RHOT+RDIFF+RSUM)
PART=PART/DXI(RDIFF,3)
PART=2.DO*PI0*PI2*PART
PB=PB+PART
CALL ADQUAD(0.DO,1.DO,PART,D616,FO1C,1.D-9,WORK,STACK,HEAP,
$      20,KODE)
IF(KODE.NE.0) STOP 'ADQUAD ERROR FOURTH CALL'
PART=2.DO*PI1*PI1*PART
PB=PB+PART
CALL ADQUAD(0.DO,1.DO,PART,D616,FO1D,1.D-9,WORK,STACK,HEAP,
$      20,KODE)
IF(KODE.NE.0) STOP 'ADQUAD ERROR FIFTH CALL'
PART=2.DO*PI1*PI2*PART
PB=PB+PART

```

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PDP-11 FORTRAN-77 V4.0-1 09:50:27 17-Jul-86
 SELFL2M2.FTN;13 /F77/TR:BLOCKS/MR

```

0001 DOUBLE PRECISION FUNCTION F01B(V)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0004 OMV=1.00-V
0005 D=V+BIGK*OMV
0006 F01B=PI01(V)*BIGK*DEXP(-V*(RHOM/D))*OMV/D
0007 RETURN
0008 END
  
```

PDP-11 FORTRAN-77 V4.0-1 09:50:29 17-Jul-86
 SELFL2M2.FTN;13 /F77/TR:BLOCKS/MR

```

0001 DOUBLE PRECISION FUNCTION F10A(V)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0004 F10A=(1.00-V)*DEXP(-RHOM*V)*PI10(V)
0005 RETURN
0006 END
  
```

PDP-11 FORTRAN-77 V4.0-1 09:50:31 17-Jul-86
 SELFL2M2.FTN;13 /F77/TR:BLOCKS/MR

```

0001 DOUBLE PRECISION FUNCTION PI10I(V)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0004 OMV=1.00-V
0005 D=BIGK*V+OMV
0006 PI10I=OMV*DEXP(-BIGK*(RHOT*V/D))/D
0007 RETURN
0008 END
  
```

PDP-11 FORTRAN-77 V4.0-1 09:50:32 17-Jul-86
 SELFL2M2.FTN;13 /F77/TR:BLOCKS/MR

```

0001 DOUBLE PRECISION FUNCTION F01C(V)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0004 F01C=PI01(V)*PI10I(V)
0005 RETURN
0006 END
  
```

PDP-11 FORTRAN-77 V4.0-1 09:50:34 17-Jul-86
 SELFL2M2.FTN;13 /F77/TR:BLOCKS/MR

```

0001 DOUBLE PRECISION FUNCTION F01D(V)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0004 F01D=(1.00-V)*DEXP(-RHOT*V)*PI10I(V)
0005 RETURN
0006 END
  
```

PDP-11 FORTRAN-77 V4.0-1 09:50:35 17-Jul-86
 SELFL2M2.FTN;13 /F77/TR:BLOCKS/MR

```

0001 DOUBLE PRECISION FUNCTION F10B(V)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0004 F10B=PI10(V)*PI10I(V)
0005 RETURN
0006 END
  
```

PDP-11 FORTRAN-77 V4.0-1 09:50:37 17-Jul-86
 SELFL2M2.FTN;13 /F77/TR:BLOCKS/MR

```

0001 DOUBLE PRECISION FUNCTION F10C(V)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0004 F10C=(1.00-V)*DEXP(-RHOT*V)*PI10(V)
0005 RETURN
0006 END
  
```

```

0001 SUBROUTINE ADQUAD(XL,XU,Y,OR,F,TOL,WORK,STACK,HEAP,N,KODE)
C ADAPTIVE QUADRATURE ALGORITHM
C XL - LOWER LIMIT OF INTEGRAL (IN)
C XU - UPPER LIMIT OF INTEGRAL (IN)
C Y - VALUE OF INTEGRAL (OUT)
C OR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
C WITH CALLING SEQUENCE
C CALL OR(XL,XU,F,Y)
C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
C WORK - WORK ARRAY OF SIZE N (IN)
C STACK - SECOND WORK ARRAY OF SIZE N, MUST NOT BE
C HEAP - THIRD WORK ARRAY, SIZE N, DISTINCT FROM WORK AND STACK
C SAME ARRAY AS WORK (IN)
C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (IN)
C KODE - ERROR INDICATOR (OUT)
C 0 -- M) ERROR
C 1 -- WORK ARRAYS TOO SMALL
C 2 -- EPS DIVIDED TO ZERO, EITHER ASKING FOR TOO
C TIGHT A TOLERANCE OR ROUND-OFF PREVENTS
C ATTAINING REQUIRED ACCURACY
C R. H. FRENCH, 14 AUGUST 1984
C
C IMPLICIT DOUBLE PRECISION(A-H,O-Z)
EXTERNAL F
DIMENSION WORK(N),STACK(N),HEAP(N)
KODE=0
Y=0.D0
WORK(1)=XU
CALL OR(XL,XU,F,T)
HEAP(1)=T
A=XL
NPTS=1
EPS=TOL
STACK(1)=EPS
B=WORK(NPTS)
XM=(A+B)*0.5D0
CALL OR(A,XM,F,P1)
CALL OR(XM,B,F,P2)
IF(DABS(T-P1-P2).LE.EPS) GOTO 20
C SPLIT IT
NPTS=NPTS*1
IF(NPTS.GT.N) THEN
KODE=1
RETURN
END IF
WORK(NPTS)=XM
HEAP(NPTS)=P2
T=P1
EPS=EPS/2.D0

```

```

0028 IF(EPS.EQ.0.00) THEN
0029 KODE=2
0030 RETURN
0031 END IF
0032 STACK(NPTS)=EPS
0033 GOTO 10
C FINISHED A PIECE
0034 Y=Y+P1+P2
0035 EPS=STACK(NPTS)
0036 T=HEAP(NPTS)
0037 NPTS=NPTS-1
0038 A=B
0039 IF(NPTS.EQ.0) RETURN
0040 GOTO 10
0041 END

```

0001 SUBROUTINE DG16(A,B,F,ANSHER)
 C 16-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
 C REF.: ABRAHAMOITZ & STEGUN, F.O. 25.4.30 AND TABLE 25.4
 C R. H. FRENCH, 28 FEBRUARY 1986
 C IMPLICIT DOUBLE PRECISION (A-H,O-Z)
 C DIMENSION X(8),W(8)
 C WEIGHTS AND ABSCISSAS FOR 16-POINT GAUSSIAN QUADRATURE

0004	DATA X/	0.09501250983763744018500,
		0.28160355077925891323000,
		0.45801677765722738634200,
		0.6178762440264374844700,
		0.75540440835500303389500,
		0.86563120238783174388000,
		0.94457502307323257607800,
		0.98940093499164993259600 /
0005	DATA W/	0.18945061045506849628500,
		0.18260341504492358886700,
		0.16915651939500253818900,
		0.14959598881657673208100,
		0.12462897125553387205200,
		0.09515851168249278481000,
		0.06225352393864789286300,
		0.027152455941175409485200 /

0006 ANSHER=0.D0
 0007 BPA02=(B-A)/2.D0
 0008 BPA02=(B+A)/2.D0
 0009 DO 10 I=1,8
 0010 C=X(I)*BMA02
 0011 Y1=BPA02+C
 0012 Y2=BPA02-C
 0013 ANSHER=ANSHER+H*(F(Y1)+F(Y2))
 0014 CONTINUE
 0015 ANSHER=ANSHER*BMA02
 0016 RETURN
 0017 END

APPENDIX L
COMPUTER PROGRAM FOR
SELF-NORMALIZING RECEIVER WITH $M=2$, $L=3$

The following pages contain the source code listing for the FORTRAN-77 program used for numerical computations of the performance of the self-normalizing receiver when $M=2$ and $L=3$. The program searches numerically for the worst-case jamming fraction.

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```
0001 PROGRAM SELF23
C SELF NORMALIZING RECEIVER, M=2, L=3
C ANALYSIS: L.E. MILLER
C PROGRAM: R.H. FRENCH
C METHOD: SELF-CONVOLUTION OF THE MARGINAL DENSITY
C
C IMPLICIT DOUBLE PRECISION (A-H,O-Z)
PARAMETER(M=2, L=3, NSLOTS=2400, SLOTS=2400.00)
PARAMETER (SLOTPR=5757600.00)
LOGICAL TEST
CHARACTER*13 FNAME
VIRTUAL PEOLD(2400)
COMMON /PIES/ P10, P11, P12
COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK1
COMMON /RONGON/ AOM, AIN, AZN, A3N, A0T, A1T, A2T, A3T,
$ BO, B1, CO, C1
COMMON /PARMS/ NO, NJ, DEBNOL(5),
$ DEBNJ(126), DJ, DRJO
CALL ERRSET(29, .TRUE., .FALSE., .TRUE., .FALSE., 15)
CALL GET
DC 900 I0=1, NO
RHOM=EBNO/L
RHOW=EBNO/L
ADN=1.00+RHOM+RHOW/6.00
AIN=RHOM+RHOW/2.00-1.00
AZN=-RHOM*(1.00+RHOW/2.00)
A3N=-RHOM+RHOW/6.00
I0OUT=DEBNOL(I0)
C PROGRESS FILE
C
C WRITE(FNAME,1) I0OUT
1 FORMAT('S23',I2.2,'GOPT.DAT')
C WRITE(5,2) FNAME
2 FORMAT(' WORKING ON FILE ',A13)
OPEN(UNIT=4, FILE=FNAME, STATUS='OLD', ERR=810,
$ FORM='UNFORMATTED', ACCESS='SEQUENTIAL')
C HAVE A PROGRESS FILE, READ IT
C
READ(4) EBNOIN, DRJOIN, DJIN
IF(EBNOIN.NE.DEBNOL(I0) .OR.
$ DRJOIN.NE.DRJO .OR. DJIN.NE.DJ) STOP 'FILE CORRUPT'
JJ=0
801 JJ=JJ+1
READ(4, END=802) DEBNJ(JJ), PELOG(JJ), COPT(JJ)
GOTO 801
802 CLOSE(UNIT=4)
GOTO 820
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0035
C NO FILE, MUST CREATE IT
C
810 OPEN(UNIT=4, FILE=FNAME, STATUS='NEW', FORM='UNFORMATTED',
$ ACCESS='SEQUENTIAL')
WRITE(4) DEBNOL(I0), DRJO, DJ
CLOSE(UNIT=4)
JJ=1
C KEEP ON GOING
C
820 DO 700 IJ=JJ, NJ
C INVALIDATE SAVED PE ARRAY
DO 737 IJK=1, 2400
PEOLD(IJK)=-1.00
CONTINUE
737 IF(IJ.GE.3) THEN
I0Q=COPT(IJ-1)-COPT(IJ-2)+0.500
IF(I0Q.EQ.0) I0Q=1
Q=COPT(IJ-1)
I0=Q
ELSE
I0Q=1
Q=1.00
I0=1
END IF
EQ=I0Q
WRITE(5,821) IJ
821 FORMAT(' IJ = ',I3)
DEBNJ(IJ)=DRJO+(IJ-1)*DJ
EBRJ=I0.00+(DRBNJ/I0.00)
C PRIME THE ALGORITHM WITH DUMMY OLD VALUES OF P(E)
P1=0.00
P2=0.00
GAMMA=Q/SLOTS
709 WRITE(5,828) I0
828 FORMAT('S',Q',I5)
P10=(SLOTS-Q)*(SLOTS-1.00-Q)/SLOTPR
P11=Q*(SLOTS-Q)/SLOTPR
P12=Q*(Q-1.00)/SLOTPR
RHOJ=EBRJ/L
RHOJ1=GAMMA+RHOJ
RHOT=RHOJ1+RHOM/(RHOJ1+RHOM)
BIGK=RHOM/RHOT
BIGK1=BIGK-1.00
A0T=1.00+RHOT+RHOT+RHOT/6.00
A1T=0.500+RHOT+RHOT-1.00
A2T=-RHOT*(1.00+0.500+RHOT)
A3T=-RHOT+RHOT/6.00
BO=RHOM+RHOM*(RHOM-RHOT)-2.00+RHOW+RHOT
B1=RHOT*(RHOM-RHOT)*(RHOM+RHOW+RHOT-RHOT)
```

```
0120 CLOSE(UNIT=4)  
0121 CONTINUE  
0122 OPEN(UNIT=4, FILE=FNAME, STATUS='NEW', ACCESS='SEQUENTIAL',  
FORM='UNFORMATTED')  
0123 WRITE(4) M, L, DEBNOL(10), NSLOTS, NJ, DEBNJ, PELOG, QOPT  
0124 CLOSE(UNIT=4)  
0125 WRITE(6,710) M, L, DEBNOL(10), NSLOTS,  
(DEBNJ(IJ), IO.DO==PELOG(IJ), QOPT(IJ), IJ=1, NJ)  
0126 FORMAT('1SELF-NORMALIZING RECEIVER FOR M=', I4/  
5X, 'EB/NO = ', F8.5, 'dB', 5X, 'NSLOTS=', I4/  
5X, 'EB/NJ (dB)', 8X, 'P(E)', 8X, 'QOPT' / < NJ > (4X, 'PPF4.1.8X, 1PD10.3,  
3X, 'OPF5.0/'))  
0127 CONTINUE  
0128 CONTINUE  
0129 STOP 'PLEASE PURGE S23*.DAT'  
0130 END
```

```
0079 C0=RHOT*(RHOM-RHOT)+2.DO*(RHOM+RHOT)  
0080 C1=RHOM*(RHOM-RHOT)*(RHOT+RHOT-RHOM*(RHOT-RHOM))  
0081 CALL PSUBE(PE, PECD, IQ)  
0082 P3=PE  
0083 IF(P3.GT.P2 .AND. IQ.LT.NSLOTS) THEN  
P1=P2  
P2=P3  
IQ=MINO(IQ-IDQ, NSLOTS)  
Q=DMINI(Q-IDQ, SLOTS)  
GOTO 709  
ELSE  
PMAX=DMAXI(P1, P2, P3)  
EPS=0.001DO*PMAX  
TEST=(DABS(P1-P2)).LE.EPS .AND. DABS(P1-P3).LE.EPS .AND.  
DABS(P2-P3).LE.EPS  
$ IF( TEST .OR. IDQ.EQ.1  
$ .OR. ((.NOT.TEST) .AND. IQ.EQ.NSLOTS)) THEN  
C WE ARE DONE WHEN ALL 3 ARE CLOSE TOGETHER OR WHEN DQ=1  
C OR WHEN WE REACHED FULL-BAND JAMMING AND P(E) IS STILL  
C INCREASING  
POPT=PMAX  
IF(P2.GT.P3) THEN  
C THE OPTIMUM MUST BE THE MIDDLE POINT OF THE 3  
C QOPT(IJ)=Q-IDQ  
IF(QOPT(IJ).EQ.0.DO) QOPT(IJ)=1.DO  
C PREVENT ROUND-OFF FROM MAKING QOPT VS. EB/NJ NON-MONOTONIC  
IF(IJ.GT.1) THEN  
QOPT(IJ).I.T.QOPT(IJ-1) QOPT(IJ)=QOPT(IJ-1)  
END IF  
ELSE  
C THE OPTIMUM IS FULL-BAND JAMMING  
QOPT(IJ)=NSLOTS  
END IF  
GOTO 665  
ELSE  
C NOT LOCATED SUFFICIENTLY ACCURATELY, CUT DQ AND TRY AGAIN  
Q=Q-IDQ-IDQ  
IQ=IQ-IDQ-IDQ  
IDQ=IDQ/2  
DQ=IDQ  
P2=P1  
P1=0.DO  
Q=Q-IDQ  
IQ=IQ-IDQ  
GOTO 709  
END IF  
END IF  
PELOG(IJ)=DLOG10(POPT)  
0665 OPEN(UNIT=4, FILE=FNAME, STATUS='OLD', ACCESS='APPEND',  
FORM='UNFORMATTED')  
$ WRITE(4) DEBNJ(IJ), PELOG(IJ), QOPT(IJ)  
0119
```

```
0001 SUBROUTINE GET
C INTERACTIVE RUN PARAMETER INPUTS
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 DIMENSION DOLST(5)
0004 CHARACTER*8 REPLY, BLANKS
0005 COMMON /PARMS/ NO, NJ, DEBNL(5),
0006 DATA DOLST/ 13.3528700, 12.313300, 10.9444300, 14.8925300,
0007 16.02713500/
0008 DATA BLANKS/' '
0009 WRITE(5,2)
0010 FORMAT(' HOW MANY EB/NO? [1] ', $)
0011 READ(5,3,ERR=1) NO
0012 FORMAT(I1)
0013 IF(NO.EQ.0) NO=1
0014 DO 8 I=1,NO
0015 WRITE(5,5) I, DOLST(I)
0016 FORMAT(3X, 'EB/NO(', I1, ') [', F8.5, ' dB]: ', $)
0017 READ(5,6,ERR=4) REPLY
0018 FORMAT(A8)
0019 IF(REPLY.EQ.BLANKS) THEN
0020 DEBNL(I)=DOLST(I)
0021 ELSE
0022 READ(REPLY,7,ERR=4) DEBNL(I)
0023 FORMAT(F8.5)
0024 END IF
0025 CONTINUE
0026 WRITE(5,17)
0027 FORMAT(' HOW MANY EB/NJ? [5] ', $)
0028 READ(5,18,ERR=16) NJ
0029 FORMAT(I3)
0030 IF(NJ.EQ.0) NJ=51
0031 IF(NJ.LT.0 .OR. NJ.GT.126) GOTO 16
0032 WRITE(5,20)
0033 FORMAT(' STARTING VALUE FOR EB/NJ [50 dB]: ', $)
0034 READ(5,21,ERR=19) DBJ0
0035 FORMAT(F5.2)
0036 IF(DBJ0.EQ.0.00) DBJ0=50.00
0037 DJ=0.00
0038 IF(NJ.GT.1) THEN
0039 WRITE(5,23)
0040 FORMAT(' INCREMENT FOR EB/NJ [-1 dB]: ', $)
0041 READ(5,24,ERR=22) DJ
0042 FORMAT(F5.0)
0043 IF(DJ.EQ.0.00) DJ=-1.00
0044 END IF
0045 RETURN
0046 END
```

```
0001 SUBROUTINE PSUBE(PE,PEOLD,IQ)
C COMPUTE ERROR PROBABILITY
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 VIRTUAL PEOLD(2400)
0004 DIMENSION WORK(20), STACK(20), HEAP(20)
0005 EXTERNAL DG10A, PDF
0006 IF(PEOLD(IQ).NE.-1.D0) THEN
C WE HAVE A SAVED VALUE FROM A LARGER STEP SIZE
C
0007 PE=PEOLD(IQ)
0008 ELSE
C COMPUTE ELEMENTAL EVENT PROBABILITIES
C
0009 CALL ADQUAI(0.D0,1.500,PE,DG10A,PDF,1.0-3,1.0-1,
0010 WORK,STACK,HEAP,20,KODE)
0011 IF(KODE.NE.0) STOP 'ADQUAI ERROR'
0012 PEOLD(IQ)=PE
0013 END IF
0014 RETURN
0015 END
```

PDP-11 FORTRAN-77 V4.0-1 /F77/MR
SEFL3428.FTH:16

PDP-11 FORTRAN-77 V4.0-1 /F77/MR
SEFL3428.FTH:16

0029 STACK(NPTS)=EPS
0030 GOTO 10
C FINISHED A PIECE
20
0031 V=Y+PI+PZ
0032 EPS=STACK(NPTS)
0033 T=HEAP(M*Y)
0034 NPTS=NPTS-1
0035 A=B
0036 IF(NPTS.EQ.0) RETURN
0037 GOTO 10
0038 END

0001 § SUBROUTINE ADQUAL(XL,XU,Y,OR,F,TN,ABSTOL,
WORK,STACK,HEAP,N,KODE)
C ADAPTIVE QUADRATURE ALGORITHM
C XL - LOWER LIMIT OF INTEGRAL (IN)
C XU - UPPER LIMIT OF INTEGRAL (IN)
C Y - VALUE OF INTEGRAL (OUT)
C OR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
C WITH CALLING SEQUENCE
C CALL OR(XL,XU,F,Y)
C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
C ABSTOL-ABSOLUTE ERROR TOLERANCE (IN)
C WORK - WORK ARRAY OF SIZE M (IN)
C STACK- SECOND WORK ARRAY OF SIZE N. MUST NOT BE
C HEAP- THIRD WORK ARRAY, SIZE M, DISTINCT FROM WORK AND STACK
C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (IN)
C KODE - ERROR INDICATOR (OUT)
C 0 -- NO ERROR
C 1 -- WORK ARRAYS TOO SMALL
C R. H. FRENCH, 14 AUGUST 1984

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 EXTERNAL F
0004 DIMENSION WORK(N),STACK(N),HEAP(N)
0005 KODE=0
0006 Y=0.00
0007 WORK(1)=XU
0008 CALL OR(XL,XU,F,T)
0009 HEAP(1)=T
0010 A=XL
0011 NPTS=1
0012 EPS=TOL
0013 STACK(1)=EPS
0014 B=WORK(NPTS)
0015 XM=(A+B)*0.500
0016 CALL OR(A,XM,F,P1)
0017 CALL OR(XM,B,F,P2)
0018 TEST=DMAXI(EPS,DABS(T),ABSTOL)
0019 IF(DABS(T)-P1-P2).LE.TEST.OR. DABS(T).LE.ABSTOL) GOTO 20
C SPLIT IT
NPTS=NPTS+1
IF(NPTS.GT.N) THEN
KODE=1
RETURN
END IF
WORK(NPTS)=XM
HEAP(NPTS)=P2
T=P1
EPS=DMAXI(EPS/2.00,5.0E-16)

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 EXTERNAL F
0004 DIMENSION WORK(N),STACK(N),HEAP(N)
0005 KODE=0
0006 Y=0.00
0007 WORK(1)=XU
0008 CALL OR(XL,XU,F,T)
0009 HEAP(1)=T
0010 A=XL
0011 NPTS=1
0012 EPS=TOL
0013 STACK(1)=EPS
0014 B=WORK(NPTS)
0015 XM=(A+B)*0.500
0016 CALL OR(A,XM,F,P1)
0017 CALL OR(XM,B,F,P2)
0018 TEST=DMAXI(EPS,DABS(T),ABSTOL)
0019 IF(DABS(T)-P1-P2).LE.TEST.OR. DABS(T).LE.ABSTOL) GOTO 20
C SPLIT IT
NPTS=NPTS+1
IF(NPTS.GT.N) THEN
KODE=1
RETURN
END IF
WORK(NPTS)=XM
HEAP(NPTS)=P2
T=P1
EPS=DMAXI(EPS/2.00,5.0E-16)

```

0001      SUBROUTINE ADQUA2(XL,XU,Y,OR,F,TOL,ABSTOL,
          WORK,STACK,HEAP,N,KODE)
          C
          C ADAPTIVE QUADRATURE ALGORITHM
          C XL - LOWER LIMIT OF INTEGRAL (IN)
          C XU - UPPER LIMIT OF INTEGRAL (IN)
          C Y - VALUE OF INTEGRAL (OUT)
          C OR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
          C      WITH CALLING SEQUENCE
          C      CALL QR(XL,XU,F,Y)
          C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
          C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
          C ABSTOL - ABSOLUTE ERROR TOLERANCE (IN)
          C WORK - WORK ARRAY OF SIZE N (IN)
          C STACK - SECOND WORK ARRAY OF SIZE N, MUST NOT BE
          C HEAP - THIRD WORK ARRAY, SIZE N, DISTING FROM WORK AND STACK
          C      SAME ARRAY AS WORK (IN)
          C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (IN)
          C KODE - ERROR INDICATOR (OUT)
          C      0 -- NO ERROR
          C      1 -- WORK ARRAYS TOO SMALL
          C R. H. FRENCH, 14 AUGUST 1984
  
```

```

0029      STACK(NPTS)=EPS
0030      GOTO 10
          C FINISHED A PIECE
0031      20      Y=Y+P1+P2
0032      EPS=STACK(NPTS)
0033      T=HEAP(NPTS)
0034      NPTS=NPTS-1
0035      A=B
0036      IF(NPTS.EQ.0) RETURN
0037      GOTO 10
0038      END
  
```

```

0002      IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003      EXTERNAL F
0004      DIMENSION WORK(N),STACK(N),HEAP(N)
0005      KODE=0
0006      Y=0.00
0007      WORK(1)=XU
0008      CALL QR(XL,XU,F,T)
0009      HEAP(1)=T
0010      A=XL
0011      NPTS=1
0012      EPS=TOL
0013      STACK(1)=EPS
0014      B=WORK(NPTS)
0015      XM=(A+B)*0.500
0016      CALL QR(A,XM,F,P1)
0017      CALL QR(XM,B,F,P2)
0018      TEST=OMAXI(EPS*DABS(T),ABSTOL)
0019      IF(DABS(T-P1-P2).LE.TEST .OR. DABS(T).LE.ABSTOL) GOTO 20
          C SPLIT IT
0020      NPTS=NPTS+1
0021      IF(NPTS.GT.N) THEN
0022      KODE=1
0023      RETURN
0024      END IF
0025      WORK(NPTS)=XM
0026      HEAP(NPTS)=P2
0027      T=P1
0028      EPS=OMAXI(EPS/2.00,5.0-16)
  
```

```

0001 SUBROUTINE ADQUA3(XL,XU,Y,OR,F,TOL,ABSTOL,
      WORK,STACK,HEAP,N,KODE)
      C
      C ADAPTIVE QUADRATURE ALGORITHM
      C XL - LOWER LIMIT OF INTEGRAL (IN)
      C XU - UPPER LIMIT OF INTEGRAL (IN)
      C Y - VALUE OF INTEGRAL (OUT)
      C OR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
      C WITH CALLING SEQUENCE
      C CALL OR(XL,XU,F,Y)
      C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
      C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
      C ABSTOL - ABSOLUTE ERROR TOLERANCE (IN)
      C WORK - WORK ARRAY OF SIZE N (IN)
      C STACK - SECOND WORK ARRAY, SIZE N, DISTINCT FROM WORK AND STACK
      C HEAP - THIRD WORK ARRAY, SIZE N, DISTINCT FROM WORK AND STACK
      C SAME ARRAY AS WORK (IN)
      C N - SIZE OF WORK AND STACK; MAX. NO. OF BISECTIONS (IN)
      C KODE - ERROR INDICATOR (OUT)
      C 0 -- NO ERROR
      C 1 -- WORK ARRAYS TOO SMALL
      C R. H. FRENCH, 14 AUGUST 1984
      C
  
```

```

0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 EXTERNAL F
0004 DIMENSION WORK(N),STACK(N),HEAP(N)
0005 KODE=0
0006 Y=0.DO
0007 WORK(1)=XU
0008 CALL OR(XL,XU,F,T)
0009 HEAP(1)=T
0010 A=XL
0011 NPTS=1
0012 EPS=TOL
0013 STACK(1)=EPS
0014 B=WORK(NPTS)
0015 XM=(A+B)*0.500
0016 CALL OR(A,XM,F,P1)
0017 CALL OR(XM,B,F,P2)
0018 TEST=DMAX1(EPS*DABS(T),ABSTOL)
0019 IF(DABS(T-P1-P2).LE.TEST .OR. DABS(T).LE.ABSTOL) GOTO 20
      C SPLIT IT
      NPTS=NPTS+1
      IF(NPTS.GT.N) THEN
        KODE=1
        RETURN
      END IF
      WORK(NPTS)=XM
      HEAP(NPTS)=P2
      T=P1
      EPS=DMAX1(EPS/2.DO,5.D-16)
  
```

```

0029 STACK(NPTS)=EPS
0030 GOTO 10
      C FINISHED A PIECE
0031 20 Y=Y+P1+P2
0032 EPS=STACK(NPTS)
0033 T=HEAP(NPTS)
0034 NPTS=NPTS-1
0035 A=B
0036 IF(NPTS.EQ.0) RETURN
0037 GOTO 10
0038 END
  
```

```

0001 BLOCK DATA
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /GOMTS/ X(5),M(5)
      C
      C HEIGHTS AND ABSISSAS FOR 10-POINT GAUSSIAN QUADRATURE
      C
      DATA X/ 0.14887433898163100,
      $ 0.43339539412924700,
      $ 0.67940956829902400,
      $ 0.865063366668898500,
      $ 0.97390652851717200 /
      DATA W/ 0.29552422471475300,
      $ 0.26926671930999600,
      $ 0.21908636251598200,
      $ 0.14945134915058100,
      $ 0.08667134430868800 /
      END
  
```

```
0001 SUBROUTINE D610C(A,B,F,ANSWER)
C
C 10-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
C
C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
C R. H. FRENCH, 30 MAY 1986
C
      IMPLICIT DOUBLE PRECISION (A-H,O-Z)
      COMMON /GQMTS/ X(5),M(5)
      ANSWER=0.DO
      BMAO2=(B-A)/2.DO
      BPAO2=(B+A)/2.DO
      DO 10 I=1,5
      C=X(I)*BMAO2
      Y1=BPAO2+C
      Y2=BPAO2-C
      ANSWER=ANSWER+W(I)*(F(Y1)+F(Y2))
      CONTINUE
      ANSWER=ANSWER*BMAO2
      RETURN
      END
      10
```

```
0001 SUBROUTINE D610A(A,B,F,ANSWER)
C
C 10-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
C
C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
C R. H. FRENCH, 30 MAY 1986
C
      IMPLICIT DOUBLE PRECISION (A-H,O-Z)
      COMMON /GQMTS/ X(5),M(5)
      ANSWER=0.DO
      BMAO2=(B-A)/2.DO
      BPAO2=(B+A)/2.DO
      DO 10 I=1,5
      C=X(I)*BMAO2
      Y1=BPAO2+C
      Y2=BPAO2-C
      ANSWER=ANSWER+W(I)*(F(Y1)+F(Y2))
      CONTINUE
      ANSWER=ANSWER*BMAO2
      RETURN
      END
      10
```

```
0001 DOUBLE PRECISION FUNCTION PDF(Z)
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 PARAMETER (CUSP=1.DO)
0004 EXTERNAL D610B, COMV02
0005 DIMENSION WORK(20), STACK(20), HEAP(20)
0006 COMMON /PASSZ/ ZEE
0007 ZEE=Z
0008 XL=DMAX1(0.DO,Z-1.DO)
0009 XU=DMIN1(2.DO,Z)
0010 IF(XL.GE.CUSP.OR.XU.LE.CUSP) THEN
0011 CALL ADQUA2(XL,XU,PDF,D610B,COMV02,1.D-4,1.D-10,
$ WORK,STACK,HEAP,20,KODE)
$ IF(KODE.NE.0) STOP 'ADQUA2 ERROR'
ELSE
$ CALL ADQUA2(XL,CUSP,PX,D610B,COMV02,1.D-4,1.D-10,
$ WORK,STACK,HEAP,20,KODE)
$ IF(KODE.NE.0) STOP 'ADQUA2-1 ERROR'
$ CALL ADQUA2(CUSP,XU,PY,D610B,COMV02,1.D-4,1.D-10,
$ WORK,STACK,HEAP,20,KODE)
$ IF(KODE.NE.0) STOP 'ADQUA2-2 ERROR'
PDF=PX+PY
END IF
RETURN
END
```

```
0001 SUBROUTINE D610B(A,B,F,ANSWER)
C
C 10-POINT GAUSSIAN INTEGRATION OVER ARBITRARY INTERVAL
C
C REF.: ABRAMOWITZ & STEGUN, EQ. 25.4.30 AND TABLE 25.4
C R. H. FRENCH, 30 MAY 1986
C
      IMPLICIT DOUBLE PRECISION (A-H,O-Z)
      COMMON /GQMTS/ X(5),M(5)
      ANSWER=0.DO
      BMAO2=(B-A)/2.DO
      BPAO2=(B+A)/2.DO
      DO 10 I=1,5
      C=X(I)*BMAO2
      Y1=BPAO2+C
      Y2=BPAO2-C
      ANSWER=ANSWER+W(I)*(F(Y1)+F(Y2))
      CONTINUE
      ANSWER=ANSWER*BMAO2
      RETURN
      END
      10
```


0001 DOUBLE PRECISION FUNCTION F1011(Z)
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 COMMON /PASSM/ DUBEYU
0004 F1011=PI0(Z)*PI1(DUBEYU-Z)
0005 RETURN
0006 END

0001 DOUBLE PRECISION FUNCTION F01(Z)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, B1GK, B1GK1
0004 F01=DEXP(RHOM*Z-RHOT)*((1.00+RHOM*Z)
0005 RETURN
0006 END

0001 DOUBLE PRECISION FUNCTION F01(Z)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, B1GK, B1GK1
0004 RNZ=RHOM*Z
0005 BK1Z=B1GK1*Z
0006 D=1.00+BK1Z
0007 A=B1GK*RNZ/D
0008 P01=B1GK*DEXP(A-RHOM)*((1.00+A)/D)/D
0009 RETURN
0010 END

0001 DOUBLE PRECISION FUNCTION P10(Z)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, B1GK, B1GK1
0004 RTZ=RHOT*Z
0005 BK1Z=B1GK1*Z
0006 D=B1GK-BK1Z
0007 A=RTZ/D
0008 P10=B1GK*DEXP(A-RHOT)*((1.00+A)/D)/D
0009 RETURN
0010 END

0001 DOUBLE PRECISION FUNCTION F010(Z)
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 COMMON /PASSM/ DUBEYU
0004 F010=PO0(Z)*PI0(DUBEYU-Z)
0005 RETURN
0006 END

0001 DOUBLE PRECISION FUNCTION F010(Z)
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 COMMON /PASSM/ DUBEYU
0004 F010=PO1(Z)*PO1(DUBEYU-Z)
0005 RETURN
0006 END

0001 DOUBLE PRECISION FUNCTION F0110(Z)
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 COMMON /PASSM/ DUBEYU
0004 F0110=PO1(Z)*PI0(DUBEYU-Z)
0005 RETURN
0006 END

0001 DOUBLE PRECISION FUNCTION F0111(Z)
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 COMMON /PASSM/ DUBEYU
0004 F0111=PO1(Z)*PI1(DUBEYU-Z)
0005 RETURN
0006 END

0001 DOUBLE PRECISION FUNCTION F1010(Z)
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 COMMON /PASSM/ DUBEYU
0004 F1010=PI0(Z)*PI0(DUBEYU-Z)
0005 RETURN
0006 END

PDP-11 FORTRAN-77 V4.0-1 10:02:46 17-Jul-86 Page 31
SELF1328.FTN:16 /F77/WR

0001 DOUBLE PRECISION FUNCTION P11(Z)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK, BIGK1
0004 P11=DEXP(RHOT*Z-RHOT)*(1.DO-RHOT*Z)
0005 RETURN
0006 END

APPENDIX M
COMPUTER PROGRAM FOR
PRACTICAL ADAPTIVE GAIN CONTROL RECEIVER

The following pages contain the source code listing for the FORTRAN-77 program used for numerical computations of the performance of the practical adaptive gain control receiver for FH/RMFSK.

```

0001      PROGRAM PRAC22
C
C PRACTICAL ACJ/AGC RECEIVER, M=2, L=2
C
C ANALYSIS: L.E. MILLER
C PROGRAM: R.H. FRENCH
C
0002      IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003      PARAMETER(M=2, L=2, NSLOTS=2400)
0004      CHARACTER*13 FNAME
0005      REAL DEBNJ(126), PELOG(126)
0006      COMMON /ROSE/ RHOH, RHOT, GAMMA, BIGK
0007      COMMON /PARMS/ MO, NJ, NG, DEBNOL(5), GAMLST(10),
C
C CALL ERRSET;29,.TRUE,..FALSE,..TRUE,..FALSE,..15)
C CALL GET
0008      DO 900 IO=1,MO
0009      EBNO=10.DO**((DEBNOL(IO)/10.DO)
0010      RHOH=EBNO/L
0011      IOOUT=DEBNOL(IO)
0012      DO 800 IG=1,NG
0013      GAMMA=GAMLST(IG)
0014      IO=GAMMA*NSLOTS+0.500
0015      IOOUT=1000.DO*GAMMA+0.500
0016      IOOUT=1000.DO*GAMMA+0.500
0017
C
C PROGRESS FILE
C
0018      WRITE(FNAME,1) IOOUT,IOOUT
0019      FORMAT('822',12.2,14.4,'.DAT')
0020      WRITE(5,2) FNAME
0021      FORMAT(' WORKING ON FILE ',A13)
0022      OPEN(UNIT=4,FILE=FNAME,STATUS='OLD',ERR=810,
C
C FORM='UNFORMATTED',ACCESS='SEQUENTIAL')
C
C HAVE A PROGRESS FILE, READ IT
C
0023      READ(4) EBNOJM, GAMMIN, DBJOJM, DJJM
0024      IF(EBNOJM.NE.DEBNOL(IO).OR. GAMMIN.NE.GAMMA .OR.
C
C DBJOJM.NE.DBJO .OR. DJJM.NE.DJ) STOP 'FILE CORRUPT'
C
C JJ=0
0025      JJ=0
0026      JJ=JJ+1
0027      READ(4,END=802) DEBNJ(JJ), PELOG(JJ)
0028      GOTO 801
0029      CLOSE(UNIT=4)
0030      GOTO 820
C
C NO FILE, MUST CREATE IT
C
0031      OPEN(UNIT=4,FILE=FNAME,STATUS='NEW',FORM='UNFORMATTED',
C
C ACCESS='SEQUENTIAL')
0032      WRITE(4) DEBNOL(IO), GAMMA, DBJO, DJ
0033      CLOSE(UNIT=4)

```

1-2

```

0034      JJ=1
C
C KEEP ON GOING
C
0035      DO 700 IJ=JJ,NJ
0036      WRITE(5,821) IJ
0037      FORMAT(' IJ = ',I3)
0038      DEBNJ=DBJO*(IJ-1)*DJ
0039      DEBNJ(IJ)=DBEKNJ
0040      EBNO=10.DO**((DBEKNJ/10.DO)
0041      RHOJ=EBNJ/L
0042      RHOJ)=GAMMA*RHOJ
0043      RHOI=RHOJI+RHOH/(RHOJI+RHOH)
0044      BIGK=RHOH/RHOT
0045      CALL PSUBE(IQ,PEI)
0046      PELOG(IJ)=DLOG10(PEI)
0047      OPEN(UNIT=4,FILE=FNAME,STATUS='OLD',ACCESS='APPEND',
C
C FORM='UNFORMATTED')
C
C WRITE(4) DEBNJ(IJ), PELOG(IJ)
0048      CLOSE(UNIT=4)
0049      CONTINUE
0050      700
0051      OPEN(UNIT=4,FILE=FNAME,STATUS='NEW',ACCESS='SEQUENTIAL',
C
C FORM='UNFORMATTED')
0052      WRITE(4) M, L, DEBNOL(IO), GAMMA, NSLOTS, NJ,
C
C DEBNJ, PELOG
C
C CLOSE(UNIT=4)
0053      WRITE(6,710) M, L, DEBNOL(IO), GAMMA, NSLOTS,
C
C (DEBNJ(IJ),10.DO**PELOG(IJ),IJ=1,NJ)
0054      FORMAT('IPRACTICAL ACJ-AGC RECEIVER FOR M= ',I1,
C
C ' AND L= ',I1/5X,'EB/NO = ',F8.5,' dB',5X,'GAMMA=',
C
C '1PD10.3,5X,'NSLOTS=',I4/' EB/NO (dB)',8X,'PIE)/'
C
C <NJ>(4X,DPF4.1,8X,1PD10.3/))
0055      800
0056      CONTINUE
0057      900
0058      STOP 'PLEASE PURGE 822*.DAT'
0059      END

```

```

0001 SUBROUTINE GET
C INTERACTIVE RUN PARAMETER INPUTS
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 DIMENSION DGAM(10), DOLST(5)
0004 CHARACTER*8 REPLY, BLANKS
0005 COMMON /PARMS/ NO, NJ, NG, DEBNOL(5), GAMLST(10),
0006 $ DEBNJL(126), DJ, DBJO
0007 $ DATA DGAM/1.0-3, 2.0-3, 5.0-3,
0008 $ 1.0-2, 2.0-2, 5.0-2,
0009 $ 1.0-1, 2.0-1, 5.0-1, 1.00/
0010 $ DATA DOLST/ 13.3524700, 12.313300, 10.9444300,
0011 $ 14.8925300, 16.02713500/
0012 $ DATA BLANKS/ ' '
0013 1 WRITE(5,2)
0014 2 FORMAT(' HOW MANY EB/NO? [1] ', $)
0015 3 READ(5,3,ERR=1) NO
0016 4 FORMAT(I1)
0017 5 IF(NO.EQ.0) NO=1
0018 6 IF(NO.LT.0 .OR. NO.GT.5) GOTO 1
0019 DO 8 IN=1,NO
0020 7 WRITE(5,5) IN, DOLST(IN)
0021 8 FORMAT(3X, 'EB/NO?', I1, ' ', F8.5, ' dB]: ', $)
0022 9 READ(5,6,ERR=4) REPLY
0023 10 FORMAT(A8)
0024 11 IF(REPLY.EQ.BLANKS) THEN
0025 DEBNOL(IN)=DOLST(IN)
0026 ELSE
0027 READ(REPLY,7,ERR=4) DEBNOL(IN)
0028 12 FORMAT(F8.5)
0029 END IF
0030 8 CONTINUE
0031 9 WRITE(5,10)
0032 10 FORMAT(' HOW MANY GAMMA? [10] ', $)
0033 11 READ(5,11,ERR=9) NG
0034 12 FORMAT(I2)
0035 13 IF(NG.EQ.0) NG=10
0036 14 IF(NG.LT.0 .OR. NG.GT.10) GOTO 9
0037 DO 15 IM=1,NG
0038 15 WRITE(5,13) IN, DGAM(IN)
0039 16 FORMAT(3X, GAMMA', I2, ' ', F5.3, ']: ', $)
0040 17 READ(5,14,ERR=12) GAMLST(IN)
0041 18 FORMAT(F8.6)
0042 19 IF(GAMLST(IN).EQ.0.D0) GAMLST(IN)=DGAM(IN)
0043 20 IF(GAMLST(IN).LE.0.D0 .OR. GAMLST(IN).GT.1.00) GOTO 12
0044 CONTINUE
0045 16 WRITE(5,17)
0046 17 FORMAT(' HOW MANY EB/NO? [126] ', $)
0047 18 READ(5,18,ERR=16) NJ
0048 19 FORMAT(I3)
0049 20 IF(NJ.EQ.0) NJ=126
    
```

```

0046 IF(NJ.LT.0 .OR. NJ.GT.126) GOTO 16
0047 19 WRITE(5,20)
0048 20 FORMAT(' STARTING VALUE FOR EB/NJ [0 dB]: ', $)
0049 21 READ(5,21,ERR=19) DBJO
0050 22 FORMAT(F5.2)
0051 DJ=0.D0
0052 23 IF(NJ.GT.1) THEN
0053 24 WRITE(5,23)
0054 25 FORMAT(' INCREMENT FOR EB/NJ [0.4 dB]: ', $)
0055 26 READ(5,24,ERR=22) DJ
0056 27 FORMAT(F5.0)
0057 28 IF(DJ.EQ.0.D0) DJ=0.400
0058 END IF
0059 RETURN
0060 END
    
```

```

0001 SUBROUTINE PSUBE(IQ,PE)
C COMPUTE ERROR PROBABILITY
C
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 PARAMETER (SLOTS=2400.D0)
0004 PARAMETER (SLOTSI=2399.D0)
0005 PARAMETER (SLOTPI=5757600.D0)
0006 DIMENSION WORK(20), STACK(20), HEAP(20)
0007 EXTERNAL DGAU20, PDF
0008 COMMON /ROSE/ RHOM, RHOI, GAMMA, BIGK
0009 COMMON /PIES/ PIO, PII, PIZ
C COMPUTE ELEMENTAL EVENT PROBABILITIES
C
0010 Q=10
0011 PIO=(SLOTS-Q)*(SLOTSI-Q)/SLOTPI
0012 PII=Q*(SLOTS-Q)/SLOTPI
0013 PIZ=Q*(Q-1.D0)/SLOTPI
0014 CALL ADQUAI(0.D0,1.D0,PE,DGAU20,PDF,1.0-4,1.0-15,
0015 $ WORK,STACK,HEAP,20,KODE)
0016 IF(KODE.NE.0) STOP 'ADQUAI ERROR'
0017 G1=6(1.D0)
0018 PE=2.D0*PE+G1*6I
0019 RETURN
0020 END
    
```

```

0001 SUBROUTINE ADQUAI(XL,XU,Y,OR,F,TOL,ABSTOL,
      2 WORK,STACK,HEAP,N,KODE)
      3
      4 C ADAPTIVE QUADRATURE ALGORITHM
      5 C XL - LOWER LIMIT OF INTEGRAL (IN)
      6 C XU - UPPER LIMIT OF INTEGRAL (IN)
      7 C Y - VALUE OF INTEGRAL (OUT)
      8 C OR - NAME OF A QUADRATURE RULE SUBROUTINE (IN)
      9 C WITH CALLING SEQUENCE
     10 CALL QR(XL,XU,F,Y)
     11 C F - NAME OF FUNCTION TO BE INTEGRATED (IN)
     12 C TOL - ERROR TOLERANCE FOR FINAL ANSWER (IN)
     13 C ABS'OL-ABSOLUTE ERROR TOLERANCE (IN)
     14 C WORK - WORK ARRAY OF SIZE N (IN)
     15 C STACK- SECOND WORK ARRAY OF SIZE M, MUST NOT BE
     16 C HEAP- THIRD WORK ARRAY, SIZE N, DISTINCT FROM WORK AND STACK (IN)
     17 C N - SIZE OF WORK AND STACK; MAX. NO. OF RISECTIONS (IN)
     18 C KODE - ERROR INDICATOR (OUT)
     19 C 0 -- NO ERROR
     20 C 1 -- WORK ARRAYS TOO SMALL
     21 C R. H. FRENCH, 14 AUGUST 1984
     22
     23 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
     24 EXTERNAL F
     25 DIMENSION WORK(M),STACK(M),HEAP(M)
     26 KODE=0
     27 Y=0.DO
     28 WORK(1)=XU
     29 CALL QR(XL,XU,F,T)
     30 HEAP(1)=T
     31 A=XL
     32 NPTS=1
     33 EPS=TOL
     34 STACK(1)=EPS
     35 B=WORK(NPTS)
     36 XM=(A+B)*0.500
     37 CALL QR(A,XM,F,P1)
     38 CALL QR(XM,B,F,P2)
     39 TEST=DMAX1(EPS*DABS(T),ABSTOL)
     40 IF(DABS(T-P1-P2).LE.TEST) GOTO 20
     41
     42 C SPLIT IT
     43 NPTS=NPTS+1
     44 IF(NPTS.GT.M) THEN
     45   KODE=1
     46   RETURN
     47 END IF
     48 WORK(NPTS)=XM
     49 HEAP(NPTS)=P2
     50 T=P1
     51 EPS=DMAX1(EPS/2.DO,5.D-16)
     52 STACK(NPTS)=EPS
     53 GOTO 10
  
```

```

0001 C FINISHED A PIECE
      2 Y=Y+P1+P2
      3 EPS=STACK(NPTS)
      4 T=HEAP(NPTS)
      5 NPTS=NPTS-1
      6 A=B
      7 IF(NPTS.EQ.0) RETURN
      8 GOTO 10
      9 END
  
```

```

PDP-11 FORTRAN-77 V4.0-1 /F77//TP:BLOCKS/MR
0001 DOUBLE PRECISION FUNCTION PDF(Z)
0002 IMPLICIT DOUBLE PRECISION (A-H,O-Z)
0003 PDF=F(Z)*G(Z)
0004 RETURN
0005 END
  
```

```

PDP-11 FORTRAN-77 V4.0-1 /F77//TR:BLOCKS/MR
0001 DOUBLE PRECISION FUNCTION F(X)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /PIES/ PIO, PII, P12
0004 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0005 XI=X+1.DO
0006 RNK=RHOM*X
0007 RTX=RHOT*X
0008 PART1=DEXP(-RNK/XI)*(1.DO+RHOM/XI)/(XI*X1)
0009 D=BIGK*X+1.DO
0010 PART2=BIGK*DEXP(-BIGK*RTX/D)*(1.DO+RHOT/D)/(D*D)
0011 D=X+BIGK
0012 PART3=BIGK*DEXP(-RNK/D)*(1.DO+BIGK*RHOM/D)/(D*D)
0013 PART4=DEXP(-RTX/XI)*(1.DO+RHOT/XI)/(XI*X1)
0014 F=PIO*PART1+PII*(PART2+PART3)+P12*PART4
0015 RETURN
0016 END
  
```

```

PDP-11 FORTRAN-77 V4.0-1 /F77//TR:BLOCKS/MR
0001 DOUBLE PRECISION FUNCTION F(X)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /PIES/ PIO, PII, P12
0004 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0005 XI=X+1.DO
0006 RNK=RHOM*X
0007 RTX=RHOT*X
0008 PART1=DEXP(-RNK/XI)*(1.DO+RHOM/XI)/(XI*X1)
0009 D=BIGK*X+1.DO
0010 PART2=BIGK*DEXP(-BIGK*RTX/D)*(1.DO+RHOT/D)/(D*D)
0011 D=X+BIGK
0012 PART3=BIGK*DEXP(-RNK/D)*(1.DO+BIGK*RHOM/D)/(D*D)
0013 PART4=DEXP(-RTX/XI)*(1.DO+RHOT/XI)/(XI*X1)
0014 F=PIO*PART1+PII*(PART2+PART3)+P12*PART4
0015 RETURN
0016 END
  
```

```

PDP-11 FORTRAN-77 V4.0-1 /F77//TR:BLOCKS/MR
0001 DOUBLE PRECISION FUNCTION F(X)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /PIES/ PIO, PII, P12
0004 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0005 XI=X+1.DO
0006 RNK=RHOM*X
0007 RTX=RHOT*X
0008 PART1=DEXP(-RNK/XI)*(1.DO+RHOM/XI)/(XI*X1)
0009 D=BIGK*X+1.DO
0010 PART2=BIGK*DEXP(-BIGK*RTX/D)*(1.DO+RHOT/D)/(D*D)
0011 D=X+BIGK
0012 PART3=BIGK*DEXP(-RNK/D)*(1.DO+BIGK*RHOM/D)/(D*D)
0013 PART4=DEXP(-RTX/XI)*(1.DO+RHOT/XI)/(XI*X1)
0014 F=PIO*PART1+PII*(PART2+PART3)+P12*PART4
0015 RETURN
0016 END
  
```

```
0001 DOUBLE PRECISION FUNCTION G(X)
0002 IMPLICIT DOUBLE PRECISION(A-H,O-Z)
0003 COMMON /PIES/ PIO, PII, PIZ
0004 COMMON /ROSE/ RHOM, RHOT, GAMMA, BIGK
0005 XI=X+1.DO
0006 PART1=DEXP(-RHOM/XI)/XI
0007 PART2=DEXP(-RHOT/XI)/XI
0008 XK1=1.DO+BIGK*X
0009 PART3=BIGK*DEXP(-RHOM/XK1)/XK1
0010 XK=X+BIGK
0011 PART4=DEXP(-BIGK*RHOT/XK)/XK
0012 G=X*(PIO*PART1+PII*(PART4+PART3)+PIZ*PART2)
0013 RETURN
0014 END
```


REFERENCES

1. J.S. Lee, L.E. Miller, R.H. French, Y.K. Kim, and A.P. Kadrichu, "The optimum jamming effects on frequency-hopping M-ary FSK systems under certain ECCM receiver design strategies, Final report, Office of Naval Research Contract N00014-83-C-0312, October 1984. (AD-A147766)
2. Milton Abramowitz and Irene A. Stegun (eds.), Handbook of Mathematical Functions, National Bureau of Standards Applied Mathematics Series 55. Washington, D.C.: Government Printing Office, June 1964, Ninth printing, November 1970.
3. I.S. Gradshteyn and I.M. Ryzhik, Table of Integrals, Series, and Products, Corrected and Enlarged Edition. New York: Academic Press, 1980.
4. Daniel I.A. Cohen, Basic Techniques of Combinatorial Theory. New York: Wiley and Sons, 1978.
5. A.J. Viterbi and I.M. Jacobs, "Advances in Coding and Modulation for Non-Coherent Channels Affected by Fading, Partial Band, and Multiple Access Interference", in Advances in Comm. Syst., A.J. Viterbi (ed.), New York: Academic Press, 1975, pp. 279-308.
6. J.E. Blanchard, "A Slow Frequency Hopping Technique That is Robust to Repeat Jamming," IEEE Military Electronics Conf. Record, October 1982, paper 14.1.
7. D.J. Torrieri, Principles of Military Communications. Dedham, MA: Artech, 1981.
8. H. Hite, et al., "High Frequency Counter-Countermeasures Study", August 5, 1983, Hughes Aircraft Co., Fullerton, CA, under Contract DAAK80-81-C-0119, AD-C032993.
9. D.J. Torrieri, "Information-Bit Error Rate for Block Codes", IEEE Trans. on Comm., Vol. COM-32, April 1984, pp. 474-476.
10. J.S. Lee, et al., "Probability of Error Analyses of a BFSK Frequency Hopping System With Diversity Under Partial-Band Jamming Interference - Part I: Performance of Square-Law Linear Combining Soft Decision Receiver", IEEE Trans. on Commun., Vol. COM-32, June 1984, pp. 645-653.
11. J.S. Lee, et al., "Probability of Error Analyses of a BFSK Frequency Hopping System With Diversity Under Partial-Band Jamming Interference - Part II: Performance of Square-Law Nonlinear Combining Soft Decision Receivers", IEEE Trans. on Commun., Vol. COM-32, December 1984, pp. 1243-1250.
12. K.S. Gong, "Performance of Diversity Combining Techniques for FH/MFSK in Worst Case Partial Band Noise and Multi-Tone Jamming", Proc. 1983 IEEE Military Commun. Conf., pp. 17-21.

J. S. LEE ASSOCIATES, INC.

13. Y.K. Kim, et al., "The Exact Performance Analyses of Two Types of Adaptive Receivers for Multi-Hops Per Symbol FH/MFSK Systems in Partial-Band Noise Jamming and System Thermal Noise", Proc. IEEE Global Telecom. Conf. 1983, pp. 1309-1314.
14. J.S. Lee, et al., "Signal Design and Detection Strategies for LPI Communications in Electronic Warfare Environments", J.S. Lee Associates, Inc., May 1983, Final report on Contract N00014-81-C-0534, AD-B073026L.
15. J.S. Lee, et al., "Error Performance Analyses of Linear and Nonlinear Combining Square-Law Receivers for L-Hops Per Bit FH/BFSK Waveforms in Worst-Case Partial-Band Jamming", Proc. 1983 IEEE Military Comm. Conf., pp. 22-28.
16. J.S. Bird and E.B. Felstead, "Antijam Performance of Fast Frequency-Hopped M-ary NCFSK an Overview," IEEE Journal on Selected Areas in Communications, Vol. SAC-4, March 1986, pp. 216-233.
17. D.J. Torrieri, "Frequency Hopping with Multiple Frequency-Shift Keying and Hard Decisions," IEEE Trans. on Commun., Vol. COM-32, May 1984, pp. 574-582.
18. E.J. Kelly, et al., "The Sensitivity of Radiometric Measurements", J. Soc. Indust. Appl. Math., Vol. 11, pp. 235-257, June 1963.
19. J.S. Bendat and A.G. Piersol, Random Data: Analysis and Measurement Procedures, New York: Wiley-Interscience, 1971.
20. A. Gelb, ed., Applied Optimal Estimation. Cambridge: MIT Press, 1974.
21. L.E. Miller, J.S. Lee and A.P. Kadrichu, "Probability of Error Analyses of a BFSK Frequency-Hopping System with Diversity under Partial-Band Jamming Interference--Part III: Performance of a Square-Law Self-Normalizing Soft Decision Receiver," IEEE Trans. on Commun., Vol. COM-34, No. 7, July 1986, pp. 669-675.