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## LINEAR HIGH RESOLUTION FREQUENCY-WAVENUMBER ANALYSIS

by

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#### ABSTRACT

The marginal success of the several high-resolution frequency-wavenumber (f - k) techniques to date is cited from the literature. Their ability to resolve signals from two closely spaced sources is not markedly superior to that of ordinary beamforming. Moreover, such non-linear techniques yield distorted magnitudes and azimuths. The ordinary f - k "spectrum" is shown to be no more than a 1-signal estimator, and the existing high resolution techniques to be but variations of that 1-signal estimator. In this paper the notion of the wavenumber "spectrum" is set aside. Instead, by analogy to the 1-signal estimator (the ordinary f - k "spectrum"), a linear M-signal estimator is developed. The high resolving power of this technique and the fidelity of its estimates is demonstrated theoretically and by computer examples both real and synthetic.

### Introduction

The mathematical development of linear high-resolution frequency-wave number analysis was presented in the Semiannual report to AFOSR for the period ending 1 March 1975. Since that time the software implementation of the theory has been accomplished and the technique has been applied to explosion and earthquake signals recorded at the Large Aperture Seismic Array (LASA) in Montana. The application of the technique to real data is the basis for this report, although for the purposes of completeness and continuity the previously-reported section on mathematical development is also included. The frequency-wavenumber spectrum, which is a multidimensional equivalent of the ordinary frequency spectrum, is used in the sciences for theoretical and experimental analysis of traveling waves. It was introduced formally into seismology by Burg (5) in an application to data analysis. The ordinary unsmoothed three dimensional frequency-wavenumber spectrum of time series data sampled at discrete points in space is given by

$$P(\omega, \overline{k}) = \left| \frac{1}{N} \sum_{n=1}^{N} \left\{ A_n(\omega) \exp[i\alpha_n(\omega)] \right\} \cdot \exp(i\overline{k} \cdot \overline{r_n}) \right|^2$$
(1)

where

n

k

is the index of the spatial sample points.

 $A_n(\omega) \exp[i\alpha_n(\omega)]$ 

is the finite Fourier transform of the nth time series. is the vector wavenumber.

 $\overline{r_n}$  is the vector location of the nth sensor, or sample point.

Each Fourier transform term is equivalent to a sinusoid. For example, the sinusoid for the nth transform at frequency  $\omega$  has amplitude  $A_n(\omega)$  and phase  $\bigotimes_n(\omega)$ (at the center of the time window).

Now  $-\overline{k}\cdot\overline{r_n}$  is the phase delay, between the origin

and  $\overline{r_n}$ , of a plane wave arriving from the azimuth of the vector  $\overline{k}$  and traveling at the phase velocity

$$v = \omega / |\bar{k}|$$

So, multiplication of the transform by the kernel  $\exp(i \vec{k} \cdot \vec{r_n})$  has the effect of <u>advancing</u> the sinusoid by just the amount the wave itself had delayed it. Thus the summation in (1), above, is a beam sum, and the f-k spectrum is just the frequency domain equivalent of ordinary beam steering.

When the traveling-wave delays are exactly compensated for by the beam shifting, i.e., when the true  $\overline{k}$  of the signal is selected, the sinusoids add up constructively with no interference, and the power, P, is maximized. Within certain limits, then, maxima or peaks in f-k space are treated as indications of the presence of traveling plane waves, and the location and size of the maxima are taken as estimates of the speed, bearing, frequency and power of those signals. If more than one signal is present or if there is noise in the data, though exact determinations are no longer possible, the f-k spectrum is still useful for detecting and estimating signals, again within limits.

One of those limitations is imposed by the finite width of the maxima associated with signals. (9) The case is analogous to that of the ordinary frequency spectrum in which components are represented by peaks of finite width.

Plane wave signal peaks in the -k spectrum have a halfpower width of the order

$$\Delta k = \frac{1}{\Delta x}$$

where  $\Delta X$  is the width, or aperture, of the array of spatial sample points. (4) If two signals in the same time window and frequency band are also close enough in phase velocity and azimuth so their wavenumbers, say  $\overline{k}_1$ and  $\overline{k}_2$ , are such:

$$|\overline{k}_1 - \overline{k}_2| < \Delta k,$$

then their maxima in the f-k spectrum are merged and form a single peak. (23) Thus, because the sensor arrays are spatially finite their resolving power is finite. Attempts to increase that power of resolution through data processing technique have required mathematical schemes to reduce the width of the lobe of the signal peak (1-3, 6-15, 17, 19). However, the straight-forward geometric appeal of this approach has proved misleading thus far. In such hybrid spectra signal lobe-widths indeed have been narrowed substantially. Nevertheless, when signal pairs approach each other in the k-plane, resolution still fails as the separation nears  $\Delta k$ , to wit, the lobe half-width for the ordinary f-k spectrum. (2, 11, 13, 15, 20). Observations of other investigators on the shortcomings of various high-resolution frequency-wavenumber techniques are cited below.

Lintz (15) finds that the high-resolution f-k spectral technique of Haney (14) does not significantly improve the capability of a seismic array to detect multiple time-overlapping events from different azimuths.

Galat and Sax (13) experimentally find the high-resolution f-k spectrum of Haney (14), and that of Capon (8), (9), no better at resolving two simultaneously arriving waves than the ordinary f-k spectrum. McCowan and Lintz (17) call attention to an unrecoverable distortion of the true amplitude spectrum in Haney's technique, and the marked disadvantage of spurious peaks under certain conditions which they regard as the inevitable result of using a high-gain procedure.

Seligson (20) describes conditions under which Capon's high-resolution technique displays less "angular resolution" than ordinary beamforming. McDonough (18) concludes that variations in amplitude from sensor to sensor may be expected to produce anomolous behavior in Capon's processor. Of course, just such variation in amplitude from sensor to sensor will result precisely because of the presence of two or more signals. McDonough offers arguments to show that ordinary beamforming is less susceptible to instability resulting from

small signal modeling errors than all other array processors.

Haney, too, notes that in the processor he describes (14) variation in amplitude from sensor-to-sensor could distort the spectrum beyond recognition. He remedies this difficulty by forcing the same amplitude upon each input channel, thus destroying the very amplitude information that would be indicative of the presence of two or more signals.

Woods and Lintz (23) conclude that given favorable conditions, the resolving power of the maximum-likelihood f-k spectrum can be effectively infinite, but, disappointingly, offer computer examples on synthetic data in which the input signal pairs are well spaced to begin with (they are separated by a distance of 0.9 of the main-lobe half-width). Cox (11) also offers theory suggesting that given arbitrarily high signal-to-noise ratios arbitrarily fine resolution should be possible, but he does not offer a method.

It may be argued that the limited resolving power of the several high-resolution techniques results from the wavenumber spectrum being in reality a 1-signal estimator. Indeed the ordinary f-k "spectrum" is a least squares estimator for fitting data to a single plane wave, as shown further on. In routine automated processing of the LASA LP data Mack and Smart (24) found the ordinary spectrum useful for estimating only one signal at a time. Estimates of a possible second

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signal were made by recomputing the wavenumber spectrum after the first (and larger) estimate had been subtracted from the data. They call this process stripping; it is useful, of course, only for estimating signals separated by about the reciprocal of the array diameter or more. At that, such estimates of a pair of signals are not optimum, but first order approximations.

Properly, the f-k spectrum is defined only for signals of infinite spatial extent traversing infinitely large arrays. The effect of a signal of wavenumber k is then confined to the point k in the spectrum. Approximations to this definition are useful if the dimensions of signals and arrays are sufficiently large. Failing that, the "spectrum" reduces to a l-signal estimator as noted. While the high-resolution techniques do attempt to extend the effective array diameter, they all test the wavenumber space with a l-signal probe, as in the ordinary f-k spectrum.

It is proposed here to set aside the notion of a spectrum. Rather we will extend the 1-signal estimator to an M-signal estimator thus to permit the simultaneous removal of the effects of one signal from the estimate of another and so achieve true high-resolution. At the same time, use of simple beamforming (in the k-plane) to estimate each of the M signals will preserve the stability and estimate fidel-

ity of the ordinary f-k spectrum.

In the following discussion a 1-signal least squares estimator is developed and is identified with the ordinary f-k spectrum. Analogy to the 1-signal estimator is used to develop an M-signal estimator. Conventional Frequency-Wavenumber Analysis

In the conventional frequency-wavenumber spectrum (ordinary or high-resolution) a single plane wave is hypothesized at each frequency. That model is then tested over the wavenumber space of interest. One attempts to minimize the error

$$E = \sum_{n=1}^{N} \left| U_n - A e^{i \vec{k} \cdot \vec{r}_n} \right|^2$$

by varying A and  $\overline{k}$  where

k

Un are the complex Fourier series terms
 (for the given frequency)
 n is the sensor, or channel, index
 N is the total number of sensors
 A are the location vectors of the
 sensors
 A is the complex Fourier series term
 for the hypothesized plane wave
 (at the given frequency)

is the wavenumber of the hypothetical plane wave (at that given frequency)

$$Ae^{i\vec{k}\cdot\vec{r_n}}$$
,  $n=1,\ldots,N$  is the

model, i.e., the hypothesized plane wave.

Note that also one can write  $\boldsymbol{\epsilon}$  as

$$\epsilon = \sum_{n=1}^{N} \left| U_n e^{-i\vec{k}\cdot\vec{r_n}} - A \right|^2$$

since

$$e^{-ik\cdot \vec{r_n}} = 1.$$

For a given  $\overline{k}, \epsilon$  is minimized by setting A to

$$A = \frac{1}{N} \sum_{n=1}^{N} U_n e^{-i\vec{k}\cdot\vec{r_n}}$$

which is shown by the following:  $a_{n} + ic_{n} \equiv U_{n} e^{-i\vec{k}\cdot\vec{r}_{n}}$ Let  $a + ic \equiv A$ and

Then

$$\begin{aligned} \varepsilon &= \sum_{n=1}^{N} \left| \left( \partial_n - \partial \right) + i \left( C_n - C \right) \right|^2 \\ &= \sum_{n=1}^{N} \left( \partial_n - \partial \right)^2 + \left( C_n - C \right)^2 \end{aligned}$$

Take partial derivatives:

$$\frac{\partial E}{\partial a} = -2 \sum_{n=1}^{N} (a_n - a); \quad \frac{\partial E}{\partial c} = -2 \sum_{n=1}^{N} (c_n - c)$$

$$\frac{\partial^2 e}{\partial a^2} = \frac{\partial^2 e}{\partial c^2} = 2N$$

Setting  $\frac{\partial \mathcal{E}}{\partial \partial} = \frac{\partial \mathcal{E}}{\partial \mathcal{C}} = 0$ ,

$$a = \frac{1}{N} \sum_{n=1}^{N} a_n \qquad C = \frac{1}{N} \sum_{n=1}^{N} C_n$$

and

$$a + ic = A = \frac{1}{N} \sum_{n=1}^{N} (a_n + ic) = \frac{1}{N} \sum_{n=1}^{N} U_n e^{-i\overline{k}\cdot\overline{n}}$$

This expression can be separated into 2 parts, thus:

$$\begin{split} & \in = \sum_{n=1}^{N} (a_n - a)^2 + (c_n - c)^2 \\ &= \sum_{n=1}^{N} a_n^2 - 2a_n a + a^2 + c_n^2 - 2c_n c + c^2 \\ &= \sum_{n=1}^{N} (a_n^2 + c_n^2) - 2a \cdot a_N + a^2 N - 2c \cdot c_N + c^2 N \\ &= \sum_{n=1}^{N} |a_n + ic_n|^2 - \sqrt{N} \left| \sum_{n=1}^{N} a_n + ic_n \right|^2 \end{split}$$

Thus,

$$\epsilon = \sum_{n=1}^{N} \left| U_n \right|^2 - \frac{1}{N} \left| \sum_{n=1}^{N} U_n e^{-i\vec{k}\cdot\vec{r}_n} \right|^2$$

The second term is the ordinary frequency-wavenumber spectrum  $1 \sqrt{1 - 12}$ 

$$P(f, \vec{k}) = \frac{1}{N} \left| \sum_{n=1}^{N} U_{j}(f) e^{-i\vec{k}\cdot\vec{n}} \right|^{n}$$

So,

$$\epsilon = \sum_{n=1}^{N} \left| U_n \right|^2 - P(\vec{k})$$

Since **&** is a squared modulus

€≥0

and

$$\sum_{n=1}^{N} \left| U_{n} \right|^{2} \geq 0$$

since it is a sum of squared modulii.

Similarly

Since

$$\begin{split} \sum_{n=1}^{N} \left| U_{n} \right|^{2} &- P(\overline{k}) \geqslant 0 \\ \sum_{n=1}^{N} \left| U_{n} \right|^{2} \geqslant P(\overline{k}) \end{split}$$

So to minimize  $\boldsymbol{\epsilon}$  one must maximize  $P(\boldsymbol{k})$ .

$$\epsilon(\vec{k}) = \sum_{n=1}^{N} \left| U_n e^{-i\vec{k}\cdot\vec{n}} - A \right|^2$$

becomes exactly zero when

$$U_{n} = e^{i \vec{k} \cdot \vec{r_{n}}} \cdot \frac{1}{N} \sum_{j=1}^{N} U_{j} e^{-i \vec{k} \cdot \vec{r_{j}}} = A e^{i \vec{k} \cdot \vec{r_{n}}}$$

$$n = 1, \dots, N$$

that is, when the data describe a single plane wave exactly. The smaller  $\epsilon(\vec{k})_{\min}$  is, in a given situation, the more likely is the hypothetical plane wave  $\Delta e^{i\vec{k}\cdot\vec{r}}$ 

because the smaller  $\epsilon(\vec{k})$  is, the larger the F-statistic is for the hypothesis. The F-statistic is given by

 $F = (N-1) \cdot P(\vec{k}) \in \vec{k}$ 

This single plane wave model is often applied in attempts to analyze a 2-signal case (or a possible 2-signal case). In such an analysis each signal is treated as if it existed by itself, the presence of the other being ignored with consequent distortion of estimates by mutual interference. This interference can be serious, and if the two signals are not separated in k -space by at least the halfwidth of the main lobe of the array response, they are likely to appear as but one signal, their main lobes having coalesced. Attempts to improve the performance of the single wave hypothesis (in application to the two signal case) have been made in which the main lobe of the array response has been slenderized mathematically by alternative methods of estimation of the wavenumber spectrum. The object has been to reduce the main-lobe half-width and so resolve signal pairs which otherwise have coalesced main-lobes indistinguishable from a signal case. These results have been marginal. In the various high-resolution techniques the influence of the one signal on the analysis of the other has been ignored.

Analysis of possible 2-signal cases calls for a 2-signal model, in particular when the 2-signals are known (or suspected) to be so close together as to have their main lobes merged.

As the 1-signal model serves for both the 0- and the 1-signal case, so one might expect a 2-signal model to be effective in all three cases: 0, 1, or 2-signals.

Multiple Signal Frequency-Wavenumber Analysis

By analogy to the 1-signal model, one would expect to solve a 2-signal model by minimizing the error

$$\boldsymbol{\epsilon} = \sum_{n=1}^{N} \left| \boldsymbol{U}_n - \boldsymbol{A} \boldsymbol{e}^{i \vec{k} \cdot \vec{r_n}} - \boldsymbol{B} \boldsymbol{e}^{i \vec{k} \cdot \vec{r_n}} \right|^2$$

varying A, k, B, and k, where

B

k

is the complex Fourier series term for the second hypothesized plane wave (at the same given frequency)

is the wavenumber of the hypothetical plane wave (at that same given frequency)

There are now two signals to solve for:

Aeikir and Beikir

Let

1

then

$$= \sum_{n=1}^{N} \left| T_n \right|^2 = \sum_{n=1}^{N} T_n^* T_n$$

Again, let

$$A = a + ic$$
,  $A^* = a - ic$ 

Taking first partial derivatives while noting that

$$\frac{\partial A}{\partial a} = \frac{\partial A^*}{\partial a} = /$$
 and  $\frac{\partial A}{\partial c} = -\frac{\partial A^*}{\partial c} = i$ 

$$\frac{\partial e}{\partial a} = \sum_{n=1}^{N} T_n^* (-e^{i\vec{k}\cdot\vec{r_n}}) + T_n (-e^{-i\vec{k}\cdot\vec{r_n}})$$

and

$$\frac{\partial e}{\partial c} = i \sum_{n=1}^{N} T_n^* (-e^{i \vec{k} \cdot \vec{r_n}}) + T_n (e^{-i \vec{k} \cdot \vec{r_n}})$$

Setting

$$\frac{\partial e}{\partial a} = \frac{\partial e}{\partial c} = 0$$

as in the 1-signal case,

$$\frac{\partial \epsilon}{\partial a} + i \frac{\partial \epsilon}{\partial c} = -2 \sum_{n=1}^{N} T_n e^{-i k \cdot r_n} = 0$$

Therefore,

$$A = \frac{1}{N} \sum_{n=1}^{N} (U_n - B e^{i \vec{k} \cdot \vec{r_n}}) \cdot e^{-i \vec{k} \cdot \vec{r_n}}$$

Analogously

$$B = \frac{1}{N} \sum_{n=1}^{N} (U_n - A e^{i \vec{k} \cdot \vec{r_n}}) \cdot e^{-i \vec{k} \cdot \vec{r_n}}$$

In this form A and B are optimized, that is, they produce the minimum value of  $\leq$  for any arbitrary pair of k and k. Adopting the notation:

$$P = \frac{1}{N} \sum_{n=1}^{N} U_n e^{-i \vec{k} \cdot \vec{r}_n}$$

$$Q = \frac{1}{N} \sum_{n=1}^{N} U_n e^{-i \vec{k} \cdot \vec{r}_n}$$

$$E = \frac{1}{N} \sum_{n=1}^{N} e^{i (\vec{k} - \vec{k}) \cdot \vec{r}_n}$$

One may write simply:

 $A = P - B \cdot E \quad \text{and} \quad B = Q - A \cdot E^{*}$ Rearranging to solve A and B simultaneously  $P = A + B \cdot E$   $Q = A \cdot E^{*} + B$   $A = \frac{\begin{vmatrix} P & E \\ Q & 1 \end{vmatrix}}{\begin{vmatrix} I & E \\ E^{*} & 1 \end{vmatrix}}, \qquad B = \frac{\begin{vmatrix} I & P \\ E^{*} & Q \end{vmatrix}}{\begin{vmatrix} I & E \\ E^{*} & 1 \end{vmatrix}}$   $A = (P - Q E)/(I - E^{*}E)$ 

 $B = (Q - PE^{*})/(I - E^{*}E)$ 

Written out at length,

$$A = \frac{\frac{1}{N}\sum_{n=1}^{N}U_{n}e^{-i\vec{k}\cdot\vec{r_{n}}} - \frac{1}{N^{2}}\sum_{n=1}^{N}U_{n}e^{-i\vec{k}\cdot\vec{r_{n}}}\sum_{n=1}^{N}e^{i(\vec{k}-\vec{k})\cdot\vec{r_{n}}}}{|-\frac{1}{N}\sum_{n=1}^{N}e^{i(\vec{k}-\vec{k})\cdot\vec{r_{n}}}\prod_{j=1}^{N}e^{-i(\vec{k}-\vec{k})\cdot\vec{r_{j}}}}$$

and B is similar in form.

Introducing a factor of 
$$\frac{1}{N}$$
 into  $\epsilon = \frac{1}{N} \sum_{n=1}^{N} T_n * T_n$ 

$$= \frac{1}{N} \sum_{n=1}^{N} \left( U_{n}^{*} - A^{*} e^{-i \vec{k} \cdot \vec{r}_{n}} - B^{*} e^{-i \vec{k} \cdot \vec{r}_{n}} \right) \times \left( U_{n} - A e^{i \vec{k} \cdot \vec{r}_{n}} - B e^{i \vec{k} \cdot \vec{r}_{n}} \right)$$

$$\begin{aligned} \epsilon &= \frac{1}{N} \sum_{\lambda=1}^{N} U_{n}^{*} U_{n} \\ &- (A^{*} \frac{1}{N} \sum_{\lambda=1}^{N} U_{n} e^{-i\vec{k}\cdot\vec{r}_{n}} + A \frac{1}{N} \sum_{\lambda=1}^{N} U_{n}^{*} e^{i\vec{k}\cdot\vec{r}_{n}}) \\ &- (B^{*} \frac{1}{N} \sum_{\lambda=1}^{N} U_{n} e^{-i\vec{k}\cdot\vec{r}_{n}} + B \frac{1}{N} \sum_{\lambda=1}^{N} U_{n}^{*} e^{i\vec{k}\cdot\vec{r}_{n}}) \\ &+ (A^{*}A + B^{*}B) \\ &+ (A^{*}B \frac{1}{N} \sum_{\lambda=1}^{N} e^{i(\vec{k}\cdot\vec{k})\cdot\vec{r}_{n}} + AB^{**} \frac{1}{N} \sum_{\lambda=1}^{N} e^{-i(\vec{k}\cdot\vec{k})\cdot\vec{r}_{n}}) \\ &\epsilon &= \frac{1}{N} \sum_{\lambda=1}^{N} U_{n}^{*} U_{n} - (A^{*}P + AP^{*}) - (B^{*}Q + BQ^{*}) \end{aligned}$$

$$\mathcal{E} = \frac{1}{N} \sum_{n=1}^{N} U_n^* U_n - (A^* P + A P^*) - (B^* Q + B Q^*)$$
  
+ (A^\* A + B^\* B) + (A^\* B E + A B^\* E^\*)

Rearranging the terms in  $oldsymbol{\epsilon}$  ,

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$$= \frac{1}{N} \sum_{n=1}^{N} U_{n}^{*} U_{n} - (A^{*}P + AP^{*}) - (B^{*}Q + BQ^{*})$$
  
+ A^{\*}(A + BE) + B^{\*}(AE^{\*} + B)

and recalling that

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$$P = A + BE \quad and \quad Q = AE^* + B$$

$$\epsilon = \frac{1}{N} \sum_{n=1}^{N} U_n^* U_n - (AP^* + BQ^*)$$

Further, substituting

0

A = P - BE and  $B = Q - AE^*$ 

$$\epsilon = \frac{1}{N} \sum_{n=1}^{N} |U_n|^2 - (\frac{P^*P + Q^*Q - PQ^*E^* - P^*QE}{1 - E^*E})$$

$$\frac{-\frac{1}{N}\sum_{j=1}^{N}\left|e^{i\vec{k}\cdot\vec{r_{j}}}\right|^{2}}{|-|\frac{1}{N}\sum_{j=1}^{N}e^{i(\vec{k}\cdot\vec{r_{j}})}-e^{i\vec{k}\cdot\vec{r_{j}}}|^{2}}$$

The identity of these last 2 equations may be demonstrated by noting that the numerator (above) equals

$$\frac{1}{N}\sum_{n=1}^{N}\left|e^{i\vec{k}\cdot\vec{r_{n}}}P - e^{i\vec{k}\cdot\vec{r_{n}}}Q\right|^{2}$$

$$= \frac{1}{N}\sum_{n=1}^{N}\left(e^{-i\vec{k}\cdot\vec{r_{n}}}P^{*} - e^{-i\vec{k}\cdot\vec{r_{n}}}Q^{*}\right)\left(e^{i\vec{k}\cdot\vec{r_{n}}}P - e^{i\vec{k}\cdot\vec{r_{n}}}Q\right)$$

$$= \frac{1}{N}\sum_{n=1}^{N}\left\{P^{*}P + Q^{*}Q - PQ^{*}e^{-i(\vec{k}-\vec{k})\cdot\vec{r_{n}}} - P^{*}Qe^{i(\vec{k}-\vec{k})\cdot\vec{r_{n}}}\right\}$$

$$= P^{*}P + Q^{*}Q - PQ^{*}e^{-i(\vec{k}-\vec{k})\cdot\vec{r_{n}}} - P^{*}Qe^{i(\vec{k}-\vec{k})\cdot\vec{r_{n}}}$$

Since  $\leftarrow$  is a sum of squares, by definition it must be non-negative everywhere. Therefore the second of the 2 terms in  $\leftarrow$ , above, must always be

$$\leq \frac{1}{N} \sum_{n=1}^{N} \left| U_n \right|^2$$

Thus to minimize 🧲 , one must maximize

$$\frac{\frac{1}{N}\sum_{n=1}^{N}\left|e^{i\vec{k}\cdot\vec{r_{n}}}\cdot\frac{1}{N}\sum_{j=1}^{N}U_{j}e^{-i\vec{k}\cdot\vec{y}}-e^{i\vec{k}\cdot\vec{r_{n}}}\cdot\frac{1}{N}\sum_{j=1}^{N}U_{j}e^{-i\vec{k}\cdot\vec{r_{j}}}\right|^{2}}{\left|-\left|\frac{1}{N}\sum_{n=1}^{N}e^{i(\vec{k}-\vec{k})\cdot\vec{r_{n}}}\right|^{2}}\right|^{2}}$$

This is the 2-signal test, analogous to the ordinary frequencywavenumber spectrum, which is the 1-signal test. However, it is more convenient to retain the form

This 2-signal f-k "spectrum" then is computed from 3 beams (as the ordinary f-k spectrum is computed from 1 beam). The beams are

$$P = \frac{1}{N} \sum_{n=1}^{N} U_n e^{-ik \cdot m}$$

the mean of the data transforms that have been beamed to  $\mathbf{k}$  (one of the two wavenumber variables),

$$Q = \frac{1}{N} \sum_{n=1}^{N} U_{n} e^{-i\vec{k}\cdot\vec{m}}$$

the mean of the data transforms after beaming to  $\frac{1}{4}$  (the other wavenumber variable),

$$E = \frac{1}{N} \sum_{n=1}^{N} e^{i(\vec{k} - \vec{k}) \cdot \vec{k}}$$

which is the (complex) array response

This 2-signal test is solved as is the ordinary 4-k spectrum, numerically, by searching the wavenumber space of

interest. Now, however, there are 4 dimensions to search, over which to test the error criterion.

It is instructive to submit a known pair of pure, noiseless signals to the 2-signal test to illustrate the function of the elements of the expression:

Let

$$U_{n} = Fe^{i\vec{k}\cdot\vec{r}_{n}} + Ge^{i\vec{k}\cdot\vec{r}_{n}}, n=1,...,N$$

Beaming them exactly to  $\mathbf{k}$  and  $\mathbf{k}$  (since these are known in this special case),

$$P = \frac{1}{N} \sum_{n=1}^{N} (F e^{i\vec{k}\cdot\vec{r_n}} + G e^{i\vec{k}\cdot\vec{r_n}}) \cdot e^{-i\vec{k}\cdot\vec{r_n}}$$
$$= \frac{1}{N} \sum_{n=1}^{N} F + G e^{i(\vec{k}-\vec{k})\cdot\vec{r_n}}$$
$$= F + G \cdot \frac{1}{N} \sum_{n=1}^{N} e^{i(\vec{k}-\vec{k})\cdot\vec{r_n}} = F + G E$$

anđ

$$Q = \frac{1}{N} \sum_{n=1}^{N} (Fe^{i\vec{k}\cdot\vec{n}} + Ge^{i\vec{k}\cdot\vec{n}}) \cdot e^{-i\vec{k}\cdot\vec{n}}$$
$$= FE^* + G$$

Then

$$A = (P + QE)/(I - E^{*}E)$$
  
= (F + GE - (FE^{\*}+G)E)/(I - E^{\*}E)  
= (F + GE - FE^{\*}E - GE)/(I - E^{\*}E)  
= F(I - E^{\*}E)/(I - E^{\*}E) = F

$$B = G(I - E^*E) / (I - E^*E) = G$$

Thus

$$\begin{aligned} \varepsilon &= \frac{1}{N} \sum_{n=1}^{N} \left| U_n - A e^{i\vec{k}\cdot\vec{r_n}} - B e^{i\vec{k}\cdot\vec{r_n}} \right|^2 \\ &= \frac{1}{N} \sum_{n=1}^{N} \left| F e^{i\vec{k}\cdot\vec{r_n}} + G e^{i\vec{k}\cdot\vec{r_n}} - F e^{i\vec{k}\cdot\vec{r_n}} - G e^{i\vec{k}\cdot\vec{r_n}} \right|^2 \end{aligned}$$

 $\epsilon = 0$ 

This little exercise clarifies a bit the function of the array response,  $E_{\rm ,}$  in the signal models  $A_{\rm and}\ B_{\rm c}$  .

The development of the 2-signal test, of course, suggests the derivation of a 3-signal test, by analogy: First, the form of the test would be, analogously,

$$\epsilon = \frac{1}{N} \sum_{n=1}^{N} \left| U_n - A e^{i\vec{k}\cdot\vec{m}} - B e^{i\vec{k}\cdot\vec{m}} - C e^{i\vec{k}\cdot\vec{m}} \right|^2$$

Introducing the notation

.

$$R = \frac{1}{N} \sum_{n=1}^{N} U_n e^{-i\vec{k}\cdot\vec{r_n}} \qquad E_n = \frac{1}{N} \sum_{n=1}^{N} e^{i(\vec{k}-\vec{k})\cdot\vec{r_n}}$$
$$E_n = \frac{1}{N} \sum_{n=1}^{N} e^{i(\vec{k}-\vec{k})\cdot\vec{r_n}} \qquad E_n = \frac{1}{N} \sum_{n=1}^{N} e^{i(\vec{k}-\vec{k})\cdot\vec{r_n}}$$

and expanding  $\epsilon$  :

$$= \frac{1}{N} \sum_{n=1}^{N} \left( U_n^* - A^* e^{-i\vec{k}\cdot\vec{r_n}} - B^* e^{-i\vec{k}\cdot\vec{r_n}} - C^* e^{-i\vec{k}\cdot\vec{r_n}} \right)$$

$$\times \left( U_n - A e^{i\vec{k}\cdot\vec{r_n}} - B e^{i\vec{k}\cdot\vec{r_n}} - C e^{i\vec{k}\cdot\vec{r_n}} \right)$$

$$\begin{split} \boldsymbol{\epsilon} &= \frac{1}{N} \sum_{n=1}^{N} U_{n}^{*} U_{n} \\ &- (A^{*} \frac{1}{N} \sum_{n=1}^{N} U_{n} e^{i\vec{k}\cdot\vec{r}_{n}} + A \frac{1}{N} \sum_{n=1}^{N} U_{n}^{*} e^{i\vec{k}\cdot\vec{r}_{n}}) \\ &- (B^{*} \frac{1}{N} \sum_{n=1}^{N} U_{n} e^{-i\vec{k}\cdot\vec{r}_{n}} + B \frac{1}{N} \sum_{n=1}^{N} U_{n}^{*} e^{i\vec{k}\cdot\vec{r}_{n}}) \\ &- (C^{*} \frac{1}{N} \sum_{n=1}^{N} U_{n} e^{-i\vec{k}\cdot\vec{r}_{n}} + C \frac{1}{N} \sum_{n=1}^{N} U_{n}^{*} e^{i\vec{k}\cdot\vec{r}_{n}}) \\ &+ (A^{*}A + B^{*}B + C^{*}C) \\ &+ (A^{*}B \frac{1}{N} \sum_{n=1}^{N} e^{i(\vec{k}\cdot\vec{k})\cdot\vec{r}_{n}} + AB^{*} \frac{1}{N} \sum_{n=1}^{N} e^{-i(\vec{k}\cdot\vec{k})\cdot\vec{r}_{n}}) \\ &+ (B^{*}C \frac{1}{N} \sum_{n=1}^{N} e^{i(\vec{k}\cdot\vec{k})\cdot\vec{r}_{n}} + BC^{*} \frac{1}{N} \sum_{n=1}^{N} e^{-i(\vec{k}\cdot\vec{k})\cdot\vec{r}_{n}}) \\ &+ (A^{*}C \frac{1}{N} \sum_{n=1}^{N} e^{i(\vec{k}\cdot\vec{k})\cdot\vec{r}_{n}} + AC^{*} \frac{1}{N} \sum_{n=1}^{N} e^{-i(\vec{k}\cdot\vec{k})\cdot\vec{r}_{n}}) \\ &+ (A^{*}C \frac{1}{N} \sum_{n=1}^{N} e^{i(\vec{k}\cdot\vec{k})\cdot\vec{r}_{n}} + AC^{*} \frac{1}{N} \sum_{n=1}^{N} e^{-i(\vec{k}\cdot\vec{k})\cdot\vec{r}_{n}}) \\ &+ (A^{*}A + B^{*}B + C^{*}C) \\ &+ (A^{*}A + B^{*}B + C^{*}C) \end{aligned}$$

 $+(A^*BE_1 + AB^*E_1^*) + (B^*CE_3 + BC^*E_3^*) + (A^*CE_1 + AC^*E_2^*)$ 

Now noting that in the 2-signal test

$$P = A + BE$$
 and  $Q = AE^* + B$ 

so that

$$A = \frac{\begin{vmatrix} P & E \\ Q & 1 \end{vmatrix}}{\begin{vmatrix} I & E \\ E^* & 1 \end{vmatrix}} \qquad B = \frac{\begin{vmatrix} I & P \\ E^* & Q \end{vmatrix}}{\begin{vmatrix} I & E \\ E^* & 1 \end{vmatrix}}$$

one recognizes that, in the 3-signal test,

$$P = A + BE_1 + CE_2$$
  
 $Q = AE_1^* + B + CE_3$   
 $R = AE_2^* + BE_3^* + C$ 

and, defining

$$den \equiv \begin{vmatrix} I & E_{i} & E_{2} \\ E_{i}^{*} & I & E_{3} \\ E_{2}^{*} & E_{3}^{*} & I \end{vmatrix}$$

$$A = \begin{vmatrix} P & E_{i} & E_{3} \\ P & E_{3}^{*} & E_{3}^{*} \end{vmatrix} \cdot den^{-1} \quad etc., or$$

$$A = \begin{bmatrix} P(1 - E_{3}^{*}E_{3}) + Q(E_{3}^{*}E_{2} - E_{i}) + R(E_{i}E_{3} - E_{2}) \end{bmatrix} / den$$

$$B = \begin{bmatrix} P(E_{2}^{*}E_{3} - E_{i}^{*}) + Q(1 - E_{2}^{*}E_{2}) + R(E_{i}^{*}E_{2} - E_{3}) \end{bmatrix} / den$$

$$C = \begin{bmatrix} P(E_{i}^{*}E_{3}^{*} - E_{2}^{*}) + Q(E_{i}^{*}E_{2}^{*} - E_{3}^{*}) + R(1 - E_{i}^{*}E_{2}) \end{bmatrix} / den$$

Now rearranging  $\epsilon$  ,

$$= \frac{1}{N} \sum_{n=1}^{N} U_{n}^{*} U_{n} - (A^{*}P + B^{*}Q + C^{*}R) - (AP^{*} + BQ^{*} + CR^{*})$$

 $+A^{*}(A+BE_{1}+CE_{3})+B^{*}(AE_{1}^{*}+B+CE_{3})+C^{*}(AE_{2}^{*}+BE_{3}^{*}+C)$ 

and substituting P ,  $Q\,,$  and R

$$\leq = \frac{1}{N} \sum_{n=1}^{N} U_{n}^{*} U_{n} - (AP^{*} + BQ^{*} + CR^{*})$$

To minimize  $\boldsymbol{\leftarrow}$  , then, one will maximize

AP\* + BQ\* + CR\*

the 3-signal test, or 3-signal analog to the conventional, 1-signal frequency-wavenumber spectrum. The function is composed of 6 beams: P, Q, and R, the 3 beams of the data,  $U_{A}$ , and  $E_{1}$ ,  $E_{2}$ , and  $E_{3}$ , the 3 beams of the array response.

Remembering the 1-signal test (conventional f-k

spectrum),

$$\frac{1}{N} \in = \frac{1}{N} \sum_{n=1}^{N} \left| U_n - A e^{i \vec{k} \cdot \vec{r_n}} \right|^2$$
$$= \frac{1}{N} \sum_{n=1}^{N} \left( U_n^* - A^* e^{-i \vec{k} \cdot \vec{r_n}} \right) \left( U_n - A e^{i \vec{k} \cdot \vec{r_n}} \right)$$

we may rewrite it as

$$\frac{1}{N} \in = \frac{1}{N} \sum_{n=1}^{N} U_n^* U_n - A^* P - A P^* - A^* A$$

(since  $P = \frac{1}{N} \sum_{n=1}^{N} U_n e^{i \vec{k} \cdot \vec{r_n}}$ ), or

$$\frac{1}{N} = \frac{1}{N} \sum_{n=1}^{N} U_{n}^{*} U_{n} - A^{*}A - AP^{*} + A^{*}A$$
$$= \frac{1}{N} \sum_{n=1}^{N} U_{n}^{*} U_{n} - AP^{*}$$

Thus

$$A P^* = \frac{1}{N} \left( \frac{1}{N} \left| \sum_{n=1}^{N} U_n e^{i \vec{k} \cdot \vec{r}_n} \right|^2 \right)$$

is the expression one must maximize in order to minimize the error. So the f-k spectrum (for the 1-signal, conventional, case) is

and

$$AP^*$$
,  
 $AP^* + BQ^*$  is the 2-signal test,  
 $AP^* + BQ^* + CR^*$  is the 3-signal test.

and

In the 1-signal test  $A = \frac{P}{I}$ 

For the 2-signal test

$$A = \frac{\begin{vmatrix} P & E_{i} \\ Q & 1 \end{vmatrix}}{\begin{vmatrix} I & E_{i} \\ E_{i}^{*} & I \end{vmatrix}}, \qquad B = \frac{\begin{vmatrix} I & P \\ E_{i}^{*} & Q \end{vmatrix}}{\begin{vmatrix} I & E_{i} \\ E_{i}^{*} & I \end{vmatrix}}$$

For the 3-signal test |DEE|

$$A = \frac{|P E_{1} E_{2}|}{|R E_{3}^{*}|}, etc.$$

$$A = \frac{|R E_{3}^{*}|}{|E E_{1}^{*}|}, etc.$$

This formalism makes evident the relationship between the successive tests. Thus one may extropolate and directly write the expression for the M-signal test in simple, terse form. For example, the 4-signal test is

 $AP^{*} + BQ^{*} + CR^{*} + DS^{*}$ in which S, the sum of the data beamed to yet a 4th point  $\mathbf{k}$ , is introduced into the sequence P, Q, and R; and in which

$$A = \frac{|P E_{1} E_{2} E_{4}|}{|Q | E_{3} E_{3}|}$$

$$A = \frac{|P E_{1} E_{3} E_{3}|}{|S E_{3} E_{3}^{*}|}$$

$$= \frac{|P E_{1} E_{3} E_{3}|}{|S E_{3}^{*}|}$$

and  $E_1$  is the array response at (k-k),  $E_5$ , that at (k-k), etc.

Note that the four-signal test is computed from 10 beams; 4 beams of the input,  $U_{\Lambda}$ , and 6 of the array response. In general, the M-signal test requires M beams of input data  $(U_{\Lambda})$ , and M(M-I)/2 beams on the array response, for a total of M(M+I)/2 beams to compute the least-squares error at any point in the 2M-dimensional space. But the beams on the array response are computed from the same complex triqonometric terms that are required for the M beams of the input data. So the M-signal test requires evaluation of 2MN sine and cosine terms to compute the error at any point ( N is the number of sensors in the array). Thus the number of trigonometric terms requiring computation increases linearly with Μ.

It must be noted that a multiple signal test is not everywhere well-behaved, but has a singularity. For example, in the case of the 2-signal test, if

so that



and

is undefined. The value it will take on at k = kÆ depends on the direction from which Though this can, of course, be shown analytically, it is a bit tedious for repetition here. The contoured map of an example (figure 1 ) displays this characteristic graphically. The contoured function is the 2-signal test

AP\*+ BQ\*

with k held fixed as k varies over the plane. Note that the contour lines all run together at k = k. The may range arbitrarily close to k but must not take on that value exactly. The data in this figure consist of 2 closely spaced signals. The fixed vector, k, was set at the peak of their merged main lobes.

One might dismiss this singularity from practical consideration since signals of identical speed and bearing are indistinguishable by array methods. The test for 2 signals at the same wavenumber location is thus unnecessary anyway. But if the 2-signal test, say, is applied to data composed of only 1 signal, must not both the probe vectors approach the same point, i.e., the wavenumber location of the input signal, in order to merge and reduce the function to the 1-signal test? We have seen that when the data,  $U_{\Lambda}$ , consist of the same number of signals as that for which one is testing, the test performs as expected: the error <u>is</u> minimized at the wavenumber location of those input signals, and the signals <u>are</u> recovered undistorted. Suppose, though, that the 2-signal test, say, is applied to data consisting of just plane wave.

Ta = U. - Aeik. - Beik.

in the error expression

$$\boldsymbol{\epsilon} = \frac{1}{N} \sum_{n=1}^{N} \left| \boldsymbol{T}_{n} \right|^{2}$$

We have to maximize

 $P = F \frac{1}{N} \sum_{n=1}^{N} e^{i(\vec{k} - \vec{k}) \cdot \vec{k}}, \text{ and } Q = F \frac{1}{N} \sum_{n=1}^{N} e^{i(\vec{k} - \vec{k}) \cdot \vec{k}}$ If  $\vec{k}$  goes to  $\vec{k}$ , then

$$P = F$$
,  $Q = FE^*$ 

and

becomes

$$A = (F - F E^*E) / (I - E^*E) = F$$

and

and

When k goes to k the error is minimized, the signal, F, is recovered undistorted, and the hypothesized second signal vanishes. This solution is invariant though

 $\vec{k}$  be permitted to range over the entire k-plane, excepting the point  $\vec{k}$ . Thus the 2-signal test does not reduce to the ordinary f - k spectrum in the presence of a single plane wave, and  $\vec{k}$  is not required to go to  $\vec{k}$  nor would the gradient of  $\boldsymbol{\epsilon}$  with respect to  $\vec{k}$ lead to  $\vec{k}$  (if one were using a steepest descent technique to minimize  $\boldsymbol{\epsilon}$ ).

## Numerical Solution of the Multiple Signal Test

One might propose to carry out the numerical solution of a multiple signal test by a straightforward search of the entire wavenumber space of interest, as is done in the computation of the conventional f-k spectrum. But the multiple signal test may be used in more practical fashion, with greater efficiency, as a follow-up to the ordinary f-k spectrum. Since a high-resolution array process by design is intended to separate signals otherwise unresolvable, there is sound justification to limit its use to the vicinity of signals tentatively identified beforehand by less powerful but faster techniques. This is an advantageous circumstance, since an M-signal test is a function of 2M dimensions of wavenumber and would otherwise prove computationally less efficient. Applying the 2 - signal test to the highest peak of an ordinary f - k spectrum, then, one hypothesizes the presence of 2 plane waves which appear as only 1 because of their proximity. By the hypothesis the spectral peak lies within the area of the main lobe of either signal and thus  $\in$  may be minimized directly by

the method of steepest descent. This is the procedure used here.

Since, as has been shown earlier,  
$$\overline{k} = \overline{k}$$

is prohibited, the descent cannot begin from any one single point in the k-plane, as, for example, the peak under consideration. But any pair of points in that vicinity is suitable; all lead to the same solution. A convenient pair are (1) the peak, and (2) the adjacent minimum of  $\boldsymbol{\leftarrow}$  with respect to, say, k when k is fixed at the peak as in the previously discussed figure 1. The gradient of  $\epsilon$  is computed at this pair and  $\boldsymbol{\epsilon}$  itself then recomputed at a new location down the gradient. The length of this first step in the descent is some fraction of the width of the arrayresponse main-lobe, thus chosen to ensure that the process does not jump from the vicinity of the solution into the range of an adjacent relative minimum. The gradient is newly computed at this second location; another somewhat smaller step is taken down the gradient; the gradient is once more computed, now at this third location, and so forth in successively smaller steps until the point is reached in that 4-dimensional space at which the gradient goes to zero.

Some examples with synthetic data for the LASA LP array follow.

Figure 2 represents the array response of LASA. The function is mapped in contours of 3 db intervals down from the peak at the center. At present 7 long-period vertical seismometers remain at LASA: the D-ring, 2 elements of the C-ring, and the center sensor at AO. Thus, the half-width of the main lobe is about 0.016 cycle/km. At 20 seconds period and 3.5 km/sec that half-width intercepts nearly 70 degrees of azimuth.

Since the error expression for the 2-signal test is a function of 4 dimensions of wavenumber and cannot be presented in map form as are ordinary f-k spectral sections, numerical presentations must be resorted to. Figure 3 presents the first and last page of computer print-out of the successive steps in the solution of a single-frequency test case of 2 noiseless signals separated by 1/8 of the array main-lobe half-width. One signal is 2 magnitudes greater than the other (ten times the amplitude) and is 180° out of phase with it at the array center such that they destructively interfere with each other. At the upper right of the first page are the signal descriptions; beneath that are the array coordinates (in km.) and the Fourier transforms of the signals. Thereafter each successive line represents a step in the

optimization. The E format number at the left in each line is the error term. The amplitude, phase and coordinates (kx, ky) at each pair of points are given in the two columns enclosed by vertical lines. The unit vector of the gradient is given by the 4 numbers at the right of the page. The size of the step from the previous point just precedes the unit vector.

The solution is given at the bottom of the second page by the complex transforms, wavenumber locations and the final error value. The solution is both accurate and precise; the high resolution has introduced no distortion such as characterizes the non-linear techniques.

The computer routines of figure 3 that apply the 2signal test were introduced into a general f-k analysis program called FKSCAN which was styled after FKCOMB [Mack and Smart (24)]. To this automated high-resolution processor synthetic time series were submitted for trial analysis. One test consisted of a unit plane-wave from  $356^{\circ}$  at 3.5 km/s to which synthetic random noise was added to make the signal-tonoise power ratio equal to 4. To this combination was added a second plane-wave, 2 magnitudes larger than the first, from  $302^{\circ}$ , also at 3.5 km/s. In the band of interest ( 5-23 seconds period) the 2 signals overlap in wavenumber space. At 23.3 seconds period they are separated by 0.7 the main-lobe half-

width; at 16.0 seconds, by 1.0. Each signal was of 20 seconds period enclosed in 192 second cosine envelope. They arrive at the array center at the same instant. A 256-second time window was applied for the analysis. The resulting bulletin from FKSCAN is given in figure 4, and is self-explanatory. The larger signal is shown arriving from 301° at 3.518 km/s; the smaller from 354° at 3.059 km/s. They differ in apparent magnitude by 2.05 (from the ratio of the power summed over the band).

The last item on the second page gives a measure of the assurance one would have had of the validity of such a detection had it appeared in processing of real data. In routine processing of such 4 minute, 16 second windows each interval yields 2 suits of vectors, or detections. At the rate of 2 suits of random vectors per time interval, so anamolous an angular concentration of vectors would appear only once every 11 days, on the average. [This algorithm, installed in FKSCAN to provide an independent detection statistic separate from the F-statistic, is based on a probability expression developed by the author which he intends subsequently to submit for publication.]

Thus, in this modest test of the 2-signal detector functioning in the presence of random noise, the small, "hidden" signal is recovered as a strong detection.

The array, the relative magnitudes and azimuthal spacing of these 2 test signals, and the frequency band in which the search was made anticipate a test on real LASA data in which similar conditions were expected. The 2 seismic events sought in the real data are recorded in the U.S. Department of the Interior's Earthquake Data Reports 36-74 and 43-74. On 31 May 74 at 0313:11 an earthquake occurred in the vicinity of Unimak Island in the Aleutians. At 0326:57 a second event occurred in eastern Kazakh SSR. The first quake had a body wave magnitude of 4.8 and a surface wave magnitude of 4.6. The second event was recorded as  $M_{\rm p}$  5.9, with  $M_{\rm S}$  measurements unavailable. The Unimak signal was expected at LASA about 0330 from  $302^{\circ}$ , with the Kazakh signal expected about 0406, in the ongoing coda, from 356°. Figure 5 displays the seismograms of this interval for all seven LASA stations. The figure begins at 0329:08 and continues past 0415. The anticipated onset of the Kazakh surface wave is marked by the arrow at 0406:09. The circled numerals at the bottom of the figure number the successive time windows, indicated by arrows, that were submitted to the high-resolution array processor.

Figures 6 through 11 are the resulting bulletins for the 6 time-windows marked on the seismograms. The relatively narrow band from 16-23 seconds period was chosen for this analysis because it was anticipated that the faint signal from

from Kazakh was most likely to appear in these frequencies if at all. As before there are two sets of tentative detections in each time-window, that is, 2 detections per frequency. The suit at the upper left contains, at each frequency, the signal pick of greatest power. One might call these the primary detections. The suit at the upper right contains the smaller signal picks. Before each suit is submitted to azimuthal distribution analysis, the program computes the straight line through the frequency-wavenumber origin (0, 0, 0) which, in the least squares sense, best fits all the vectors in the suit. The back azimuth and phase velocity of that mean are then printed in the bulletin.

The back-azimuth of the mean of the primary signal picks in succeeding time-windows, then, for these 6 intervals reads, in sequence: 328°, 319°, 323°, 353°, 356°, and 318°. In windows 1, 2, and 3 the detector is "triggering" on the ongoing coda from the Unimak earthquake. But in windows 4 and 5 it turns and indicates the back azimuth of the Kazakh site. Then in window 6 it returns to the Unimak coda.

Figures 12 and 17 (one for each of these 6 time-windows) are contoured plots in 3 db intervals, of the conventional wavenumber spectra, integrated over the frequency band 0.043-0.063 Hz (16-23 seconds period) after the secondary detections derived from the high-resolution processor have been filtered

out. These spectra make visible the observations of the previous paragraph.

and the second s

### Conclusions

The high resolving power of the linear multiple-signal analysis and the fidelity of its estimates have been demonstrated by computer examples and by application to real signals.

Computer examples indicate that this technique is capable, in the absence of noise, of exactly recovering the amplitude, phase, and velocity of two simultaneously arriving Rayleigh waves at, for example, LASA, which differ in azimuth by as little as  $8^{\circ}$ , even if one signal is 10 times larger than the other. In the case of the simultaneous arrival of a small signal with S/N of 2 and a signal 100 times larger, with a difference in azimuth between the two of 54°, the magnitude (Ms) of the small signal can be recovered with less than 3% distortion.

The extraction of the Rayleigh wave arriving from a nuclear test in Kazakh from the coda of an Aleutian earthquake demonstrates the practical application of the technique. It should now be possible to utilize long period array data to obtain accurate amplitude and phase information for small events which were previously "hidden" in the coda of much larger events. In addition, the linear multiple signal estimator should make possible the decomposition of large surface waves into primary and multipath components on the basis of differences in arrival azimuth. Better estimates of the true amplitude and phase will result by removal of the multipath effects, and the spectrum and angle of approach of the multipath components will provide information as to the location and nature of the conditions which give rise to multipaths.

# FIGURES

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Figure 1. A contoured map of a 2-signal test of synthetic data (see page 27) with one probe point held fixed while the other ranges over the wavenumber plane. When both probes occupy the same point the function is ambiguous, its value varying with the direction from which one probe approaches the other. Note the intersection of the contours.



Figure 2. The array response of the 7-element long-period vertical seismic array at LASA. The contour interval is in 3 db steps. The scale is 0.01 cycle/km per inch.

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13: Figure

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3. The

successive steps in 2-signal analysis. See p 33.

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PERISO	SECONDS	23-273	19.592	17-067	GREATER THAN	E VECTORS AT THE					EF & IS THAT OF T .8 DEJREES OR LESS	CITY-AZIMUTH LINE	S GREATER TAN CETTV-AZINUTH LINE T 55 -005 CYCLESV	esolution f
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	S AT THE RIGHT HAVING FESTATISTICS SAEA	9F THE 3ROUP 9F \$ VECT3RS JETWEE Less Will 9Ccur at Random BV THE Line Through F-K Sace Vielded AV Les/KK*			
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( : 6 F-STATISTIC THE VOST AND VALOUS CONCENTRATION OF VECTORS IN THE ABOVE LIST OF 5 IS THAT OF THE GROUD OF 5 VECTORS DETWEEV 303 AND 341 Dedrees. The equivalent of 5 be more out of 5 vectors within 37.9 degrees of Less will been at Awodm on the Average if Snee LVERT SUDALED FITS OF VELL 5 AVENUMBER VECTORS TO A CONSTANT VELOCITY-AZEMUTH LINE THROUGH F-K SOACE VIELDED AN APPAGENT PHASE VELOCITY OF 31.442 FITS AT BUY DEGREES AN APPAGENT OF THE FIT OF .0005 CYCLES/K\* HIGH RESOLUTION FREQUENCY-WAVENUMBER ANALYSIS APPLIED TO THE DATA OF THE PRECEDING PHASE THERE ARE 5 VECTORS AT THE LETT, ABOVE, HAVING FASTATISTICS GREATER TANN 10-00 A LEAST SQUARES FIT OF ALL 5 NAVENUMLER VECTORS TO A CONSTANT VELOCITY-AZIMUTH LINE THROUGH FAK SPACE VIELDED AN APPARENT PHASE VELOCITY OF JOONE AT 319 DEJREESN WITH AN ANS EAROR TO THE FIT OF .0005 CYCLESNAM. ·C.C1 THE FULLDAING LIST REPEATSA IN AZIMUTAL BROER, ALL THE ASBVE VECTORS AT THE LEFT HAVING F-STATISTICS GREATER THAN 13+3-FBRTEACH VECTOR THE LAST COLUMN GIVESY IN DEGREESA THE DISTANCE FROM THE PRECEDING VECTOR. ~~ ....... FOR FALL FOLLOWING LIST REPEATON IN AZIMUTAL 04364, ALL THE AGOVE VECTORS AT THE RIGHT HAVING F-STATISTICS GREATER THAN FOR THE VECTOR THE VECTOR THE VECTOR THE VECTOR THE VECTOR. APPARENT -----PONER •19545 •1965 •1965 •3935 -A LEAST FILSE ARE 5 VECTORS AT THE RIGHT, ABOVEL HAVING FESTATISTICS GREATER THAN 10.0. A LEAST SQLARES FIT 3F ALL 5 MAVENUMLER VECTORS TO A CONSTANT VELOCITY-ALIUTA LINE THROUGH F-K SPACE VIELDED VELOCITY OF 4+575 KIVIS AT 342 DEGREES, MITH AN RWS ERROR TO THE FIT 95 +011 CYCLESNAM BACA AZITUTA DEGREES EAST OF EAST OF S12 382. Z APPARENT APPARENTAPPARENT APPARENT APPARENT APPARENT APPARENTAPPARENT APPARENTAPPARENTAPPARENT APPARENTAPPA 3-775-5-775-5-573-5-272-3-272-PERIOD IN IN SECONDS 23-273 219-285 19-285 17-067 17-067 F-STATISTIC 10 + O + O 0 00 + 0 4 00 0 + BULLETIN PRESENTS THE RESULTS OF 23.28.62 DEGREES DECKEES 322.1 37 . 5 P3.69 -1725 -1965 -1976 -1976 DERIES 21.53 1981-010 1981-010 1981-010 BACK ALINUTH ALINUTH IN CEGREES EAST 95 N92TH C81839 -----F-STAT F-STAT 1 + 1 - 1 + 10 5 - 1 - 10 10 + 10 10 + 10 10 + 10 10 + 10 10 1.8 APPAGENT PLASE VELOCITY 3+775 SPEED \* 01 × 0 × 0 \* 0 × 0 × 0 \* 0 × 0 × 0 SPEED THE FOLLOWING H102121 H10-171 PERICO IN SECONDS 23-273 21-233 19-092 17-067 17-067 50 Figure G: 0353:21 to 0357:36 GMT, array. 7. 31MAY74, LASA LP

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HIGH RESOLUTION FREGUENCY-WAVENUMBER ANALYSIS APPLIED TO THE DATA OF INE PRECEDING	PERIED PARENT AZIMUTA PERIED PARENT AZIMUTA PARE IN DEGREG IN VELDETTY EAST OF VERTH PORE OF SECOVOS IN KINS VERTH PORE OF SOF CARE 703, 505 CARE 705 CARE 705 CA	533         51         536         73         51         53         53         53         53         53         55         57         5	AL BADER, ALL THE ABOVE VECTORS AT THE LEFT HAVING F-STATISTICS OPEATER TWAN 10-0- The Distance From the preceding vector.	WE, HAVING F-STATISTICS DREATER THAN 10.0. UNES TO A CONSTANT VELOCITY-AZIMUTH LINE THROUGH F-K SPACE VIELDED AN APPARENT PHASE AN ANS ERROR TO THE FIT OF -014 CYCLESKAM AN ENS ERROR THE FIT OF -014 CYCLESKAM AL BROER THE ABBVE VECTORS AT THE RIGHT HAVING F-STATISTICS GREATER THAN 10.0.
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THE FELLAINS BULLETIN PRESENTS THE RESULTS OF HIGH RESELUTION FREQUENCY-WAVENUMBER ANALYSIS APPLIED TO THE DATA OF THE PRECEDING Tive allos...

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31-33       324       11       12-55       325       12-55<	<pre>3. 31. 1156 33. 33. 1156 33. 33. 1156 35. 31. 1156 35. 1151 31. 1156 35. 1156 3</pre>	21.333	3.554	297.	.5031 01.	33	23.273	1.531.	201.0	-292E 01.	28
<pre>19-05 0-040 33 100 20 100 100 10-00 3-259 310 10-00 3-259 329 329 10-00 3-259 329 10-00 3-259 329 10-00 3-259 329 329 10-00 3-259 329 329 329 329 329 329 329 329 329 32</pre>	<pre>First ME First ME 100 100 100 100 100 100 100 100 1000 1</pre>		3.6%6.	31.	.1265 03.	372	19.635	4.179.	12	·872E 32.	**
177050     +.027     329     .495     229     .695     329     .655     130     .655     130     .655     .655     .655     .655     .655     .655     .655     .655     .655     .655     .655     .655     .655     .655     .655     .655     .655     .655     .655	THE FALLENT TO THE LET THE LET TO THE TOTAL THE TOTAL TO THE TOTAL TOTAL TO THE TOTAL TO THE TOTAL TOTAL TOTAL TO THE TOTAL TO	44241		188	.55 32.	125	18-286	1.755.	101	·162 3541.	1.00
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AZITUTH SEED F-STAT PERIDD DEGREES 31 3-046 372 21-33 61-8 32 4-027 60 17-07 1-7 33 4-025 120 13-02 1-0 34-7 95 120 13-02 1-0 36-00 13-2 36-00 13-2 19-05 13-1 19-05 13-1 19-05 13-1 19-05 13-1 19-05 13-1 19-05 14-6 AMAN 10-05 THERE ARE 6 VECTORS AT THE RIGHT ABOVE, MAVING F-STATISTICS GREATER THAN 10-05 EAST 50-04 TEST 3F ALL 6 AMENUMER VECTORS TO A CONSTANT ALINE THEOLOGH F-K SPACE YIELDED AN APPAGENT PH	AZIWUL SPEC FESTAT PERIDO JEGEES 31 5006 72 2:33 5:9 32 0.00 120 10 23 0.00 120 10 33 0.00 120 10 36 0.00 120 10 30 0.00 10	EACH VECTO	1 THE LAST	COLUMN 3 IN	EST IN DEGR	EEST THE DISTANCE FROM	THE PARCENT 10 VE				
31       3:445       372       21:33       61:3         32       4:427       60       17:07       17         33       4:454       120       13:07       17         33       3:454       120       13:07       17         34       3:454       120       13:07       10:00         35       3:454       120       13:07       10:01         329       3:457       13:02       13:02       13:02         329       3:454       13:02       13:02       13:02         329       3:455       13:02       13:02       13:02         329       3:454       14       14       14         329       5:030       13:02       13:02       13:02         329       5:0455       31       14       14         329       5:0455       31       44       10:00         5:0470       5:0455       31       5:045       5:045         5:0455       5:0455       5:0455       5:045       5:045         5:0455       5:0455       5:0455       5:045       5:045         5:0470       5:0455       5:0455       5:045       5:045     <	31       3000       372       21.03       10.0         32       3000       10.0       10.0       10.0         320       3000       10.0       10.0       10.0         320       3000       10.0       10.0       10.0         320       3000       10.0       10.0       10.0         320       3000       10.0       10.0       10.0         321       3000       10.0       10.0       10.0         321       3000       10.0       10.0       10.0         321       3000       10.0       10.0       10.0         321       3000       10.0       10.0       10.0         321       3000       10.0       10.0       10.0         320       3000       10.0       10.0       10.0         320       3000       3000       4000       4000         320       3000       3000       4000       4000         3200       3000       3000       3000       3000         3200       3000       3000       3000       3000         3200       3000       3000       3000       3000         3200 <td>AZIMUTH</td> <td>SPEED F-ST</td> <td>AT PERICO</td> <td>CEGREES</td> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td>	AZIMUTH	SPEED F-ST	AT PERICO	CEGREES						
32       4:047       50       15.29       1.0         231       3:55       55       15.29       1.0         325       3:55       55       15.29       1.0         326       3:25       55       15.29       1.0         329       3:25       55       15.00       13:2         329       3:25       55       15:05       13:1         329       3:25       55       15:05       13:2         329       3:25       55       15:1       15:1         329       3:25       55       15:1       15:1         329       3:25       15:1       15:1       15:1         329       5:21       15:1       16:00       13:2         329       5:31       14:1       15:1       16:1         329       5:21       2:1       2:1       2:1       2:1         320       5:21       2:1       2:1       2:1       3:1       3:1         5:51       2:1       2:1       2:1       3:2       2:1       3:1       3:1       2:1         5:51       2:1       2:1       3:1       2:1       3:1       3:1       3	32 ***ET 10 10:00 10:0 20 3:05 10:00 10:0 20 3:05 10:00 10:0 20 3:05 10:00 10:0 20 3:05 10:0 10:0 10:0 10:0 10:0 10:0 10:0 10:0 10:0 10:0 10:0 10:0 10:0 10:0 10:0 10:0 10:0	31	3.646 372	51-33	61.3		•				
237 3:54: 32 53:27 26:2 316 3:74:7 95 19:0 329 3:35 65 19:1 329 3:35 65 16:00 13:2 THERE ARE 6 VECTURE AT THE RIGHT, ABOVE, HAVING F-STATISTICS GREATER THAN 10:0 THERE ARE 6 VECTURE AT THE RIGHT, ABOVE, HAVING F-STATISTICS GREATER THAN 10:0 THERE FIT 3F ALL 6 AVENUTERA VELOCITY-AZIMUTH LINE THROUGH F-K SPACE YIELDED AN APPARENT PA	201       3007       2007       2002       1000         310       3005       10       1000       1000         310       3005       10       1000       1000         310       3005       10       1000       1000         310       3005       10       1000       1000         310       3005       10       000       1000         310       3005       4000       4000       4000         2010       1000       4000       4000       4000         2011       000       000       000       4000         2011       000       000       000       4000         2011       000       000       000       4000         2011       000       000       000       4000         2011       000       000       000       4000         2011       000       000       000       4000         2011       000       000       000       4000         2010       000       000       000       4000         2010       000       000       000       4000         2010       000 <t< td=""><td>32</td><td>4.027 60</td><td>18.29</td><td>1.0</td><td></td><td></td><td></td><td></td><td></td><td></td></t<>	32	4.027 60	18.29	1.0						
THERE ARE & VECTURS AT THE RIGHT, ABOVE, MAVING F-STATISTICS GREATER THAN 10-0. East SQUARES FIL & AVENUTERA VECTORS TO A CONSTANT VELOCITY-AZIMUTH LINE THROUGH F-K SPACE VIELDED AN APARENT PA	THERE ARE 6 VECTORS AT THE RIGHT, ABOVE, MAVING F-STATISTICS GREATER THAN 10-0. THERE ARE 6 VECTORS AT THE RIGHT, ABOVE, MAVING F-STATISTICS GREATER THAN 10-0. EAST GOUARES FIT 3F ALL 6 , AGENUSER VECTORS TO THE FIT 0F .018 CVCLES/KH- EOTIV DF1392 RAVE AT 335 DEDREES WITH AN RYS ERROR TO THE FIT 0F .018 CVCLES/KH- 	262	3.554 30	23.27	264.2						
THERE & VECTURS AT THE RIGHT, ABOVE, HAVING F-STATISTICS GREATER THAN 10-0. East Squares Fit of All & AVENUTER VECTORS TO A CONSTANT VELOCITY-AZIMUTH LINE THABUGH F-K SPACE VIELDED AN APPARENT PA	THERE ARE 6 VECTORS AT THE RIGHT, ABBVE, HAVING F-STATISTICS GREATER THAN 10.00. East 500ARES FIT 3F ALL 6 AAGENUEER VECTORS TO THE FIT 96018 CYCLESVAN. BELTY 97 **3892 KAVS AT 335 DEGREES WITH AN RYS EROR TO THE FIT 96018 CYCLESVAN. BELTY 97 **3892 KAVS AT 335 DEGREES WITH AN RYS EROR TO THE FIT 96018 CYCLESVAN. BELTY 97 **3892 KAVS AT 335 DEGREES WITH AN RYS EROR TO THE FIT 96018 CYCLESVAN. ALL 45 THE AT 335 DEGREES WITH AN RYS EROR TO THE FIT 96018 CYCLESVAN. ALL 45 THE ALL 55 THE ALL AN RYS EROR TO THE FIT 96018 CYCLESVAN. ALL THE FOLLSTING LIST GEPEATS. IN AZIMUTHAL 33DER ALL THE AUGVE VECTORS AT THE RIGHT HAVING F-STATISTICS STATER THAN A FACH VECTOR THE LAST CELLON JIVES, IN DEGREES. THE DISTANCE FROM THE PRECEDING VECTOR.	- 626	5.235 B	16.00	13.2						
THERE ARE 6 VECTORS AT THE RIGHT, ABOVE, HAVING F-STATISTICS GREATER THAN 10-0. Least squares fit of anounder vectors to a constant velocity-azimuth line though F-K space vielded an apparent pr cecity DF 1000-2017 at 335 degrees, with an RMS eargh to the Fit of -018 cycles/KH-	THERE ARE 6 VECTORS AT THE RIGHT, ABOVE, MAVING F-STATISTICS GREATER THAN 10.0. LEAST GOUARES FIT JF ALL 6 AAVENUMERA VECTORS TO A CONSTANT VELOCITYAZIMUTH LINE THROUGH F-K SPACE VIELDED AN APPARENT I COULTY DF 4:592 ANYS AT 335 DEGRESS WITH AN RPS EREGR TO F 10 018 CYCLES/KM. COULTY DF 4:592 ANYS AT 335 DEGRESS WITH AN RPS EREGR TO F 10 018 CYCLES/KM. THE FOLLENTYS LIST REPEATS TY AZIMUTHL DATE ALOVE VECTORS AT THE RIGHT MANING F-STATISTICS GREATER THAN A RACH VECTOR THE LAST CHUCH DISEAST THE DISTANCE FROM THE PRECEDING VECTORS.				ŀ						
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