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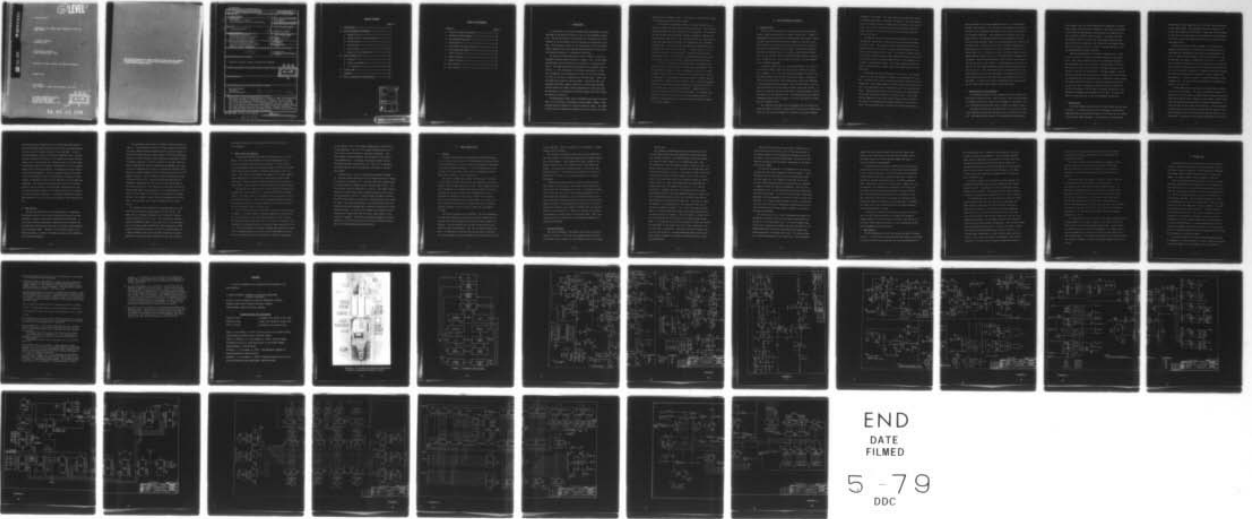
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ELECTRONICS FOR A ROCKET BORNE QUADRUPOLE CLUSTER ION
MASS FILTER

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TABLE OF CONTENTS

	Page No.
I. INTRODUCTION-----	1
II. MASS SPECTROMETER ELECTRONICS-----	3
A. Digital Control-----	3
B. Analog Control of the Mass Filter-----	5
C. Bias Circuits-----	6
D. RF Generator-----	8
E. Data Circuits-----	10
F. Power Control and Supplies-----	12
III. MANUAL CONTROL UNIT-----	14
A. Overview-----	14
B. Control and Display-----	15
C. Data Circuits-----	18
IV. BALLOON CIMS-----	21
V. PERSONNEL-----	26
VI. RELATED CONTRACTS AND PUBLICATIONS-----	26

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TABLE OF ILLUSTRATIONS

Figure No.	Page No.
1. Cluster Ion Mass Spectrometer-----	27
2. Electronics Block Diagram-----	28
3. Bias and Sweep Circuits-----	29
4. RF Oscillator-----	31
5. Electrometer and Other Circuits-----	33
6. Power Supply and Relays-----	34
7. Control Circuits -1-----	35
8. Control Circuits -2-----	36
9. Display Circuits-----	37
10. Data Circuits - 1-----	38
11. Data Circuits - 2-----	39

I. INTRODUCTION

A rocket-borne low altitude quadrupole mass spectrometer has been developed by the Aeronomy Division of the Air Force Geophysics Laboratory. The helium cryopump, the structures of the mass filter and the housing for the instruments have been designed and constructed by the AFGL. The electronic systems for the spectrometer and for the ground support equipment have been developed and constructed by the Electronics Research Laboratory of Northeastern University.

The Cluster Ion Mass Spectrometer shown in Figures 1 and 2 was designed to measure ambient positive or negative ions. In both modes, the instrument could be operated as a band pass or a high pass mass filter. The range covered by the quadrupole filter extended from 1 to 255 amu. To cover that range an RF generator operating at 2.3 MHz with a capability of peak amplitude in excess of 700 volts was used to excite the two sets of quadrupoles of the instrument. Twelve milliseconds were allowed for each amu domain. During that time, the quadrupole excitation signals completed a 16 level staircase sweep centered around the nominal voltages appropriate for that mass unit. The generation of the quadrupole excitation signals and bias voltages was digitally controlled.

The spectrometer flight control information was stored in an erasable and electrically programmable read-only-memory (EPROM). Eight programs could be sequentially selected during the flight. The number of times that each program was repeated could be set before the final

elevation of the sounding rocket. This required a simple wiring change of a connector provided for that purpose.

An electrometer logarithmic current to voltage converter was employed to process the data from the electron multiplier. These circuits were referenced to the anode potential of the multiplier. To translate the data to the vehicle potential, voltage to frequency converter and an optoisolator were used. At the vehicle potential the signal was converted into analog and PCM signals for transmission over a PCM/FM/FM telemetry link. The spectral data was sampled and converted into a 10-bit binary word once every millisecond. A complete frame of the PCM signal was devoted to each amu domain. Aside from the nine 10-bit spectrum data words, the frame also contained information on the mode of operation, the status of the bias signals and amu identification.

To facilitate laboratory experimentation and calibration procedures, a control unit to replace the flight EPRCM was designed. This unit enabled the operator to set via keyboard the limits for a sweep through a range of amu's, to control the ratio between the ac and the dc components of the quadrupole excitation signal, to choose a mode of operation and to adjust the bias voltages. Also, a simple PCM decommutator was included which allowed the operator to display the PCM data in an analog form. Selection and display of a single amu out of a sweep over several amu domains was also possible. Seven segment and single LED indicators were used to verify the keyboard entries, to display the selections and to confirm the mass spectrometer responses to the given commands.

II. MASS SPECTROMETER ELECTRONICS

A. Digital Control

Control over the operation of the Cluster-Ion Mass Spectrometer during a flight was exercised by a digital control unit. A number of preprogrammed digital signals were converted into the necessary analog voltages to provide bias and excitation signals for the mass filter. Also included in this general group were circuits for data conditioning and for the generation of a PCM signal. Detailed description of this unit has been published as a scientific report AFGL-TR-78-0106 listed in the "Related Contracts and Publications" section of this report. Therefore, only a brief summary of the unit will be presented in the following paragraphs.

The flight instructions for the mass spectrometer were stored in a one kilobyte EPROM. Eight bytes were used to define one complete set of commands. Two 12 bit instructions designated the limits of an amu domain through which the mass filter was to sweep. An eight bit instruction controlled the ratio between the dc and the RF components of the quadrupole excitation signal. Three bits were used to set the draw-in potential of the cone where the inlet orifice was located. The positive or negative ion and the high pass or band pass modes of operation were set by two additional bits. The remaining bits were used for housekeeping instructions, as spares or, for the convenience of design, were not utilized at all.

Past experience has shown that one program, which was usually repeated throughout the flight, required less than 16 sets of instructions. Therefore, this unit was designed to accommodate up to eight different

programs of that length. Only seven jumper wires between the pins of a connector were needed to determine the number of times each program was to be repeated during the flight before advancing to the new program. The connector could be wired and installed just before the final elevation of the sounding rocket, when the flight trajectory has been confirmed.

Although not directly associated with the control of the mass spectrometer, PCM and data conversion circuits were included within the control unit. The data from the mass filter section of the instrument arrived in the form of a pulse train whose repetition frequency varied between 20 and 70 kpps. A frequency-to-voltage converter produced an analog signal varying between zero and minus ten volts. Inverted and halved, the analog signal was sent to the subcarrier oscillator of the telemetry system. The original negative analog voltage was converted into a digital signal and transmitted over a PCM/FM link.

The PCM NRZ-L data was transmitted at 10 kHz rate. One complete frame of 120 bits was assigned to transmit data associated with each amu. It started with a 16-bit frame synchronization pattern which was followed by a 5-bit status monitor word and a 9-bit amu identification word. The remaining bits were divided into nine 10-bit spectrum data words. Complete synchronization was maintained at all times between the PCM and the control functions of the mass filter. The spectral data was sampled once every millisecond and 200 μ s before the MSB of the binary word representing that sample appeared in the PCM pulse train. One millisecond before the MSB of the frame synchronization

pattern appeared, a sync pulse signaled the start of a voltage sweep through a new amu domain. At this time, if any changes in the control functions of the instrument were required, the control unit provided the necessary information. Any transients in the analog control voltages caused by the updating process had 3.8 milliseconds to settle before the next meaningful sample of the analog signal was converted into the digital word. During this time the D/A converter generating the sweep signal in the analog section of the instrument continued to be updated every 750 μ s. Therefore, in these 3.8 ms, the sweep staircase signal advanced through five out of the sixteen levels assigned to one amu domain. Thus the nine data words in the PCM pulse train covered only the last 75% of that interval. By an appropriate offset bias the peaks of the mass spectrum were shifted into the intervals where the data was sampled. This offset bias could be introduced either in the sweep generator section of the instrument or by manipulating the four least significant bits of the control word specifying the start of an amu domain. In either case, the data transmitted as an analog signal was not effected and always covered the full amu domain.

B. Analog Control of the Quadrupole

The immediate control over the quadrupole excitation signals was exercised by analog circuits. These circuits in turn received commands in a digital form from a control unit programmed for the flight. The D to A interface and the analog control circuits are shown in Figure 3.

The digital control over the amplitude of the dc and the RF excitation signals was maintained through the two D/A converters (U1 and U2). The twelve bit D/A converter (U2) established the amplitude of

these signals, while the multiplying DAC (U1) maintained the desired ratio between the RF and the dc components. The required adjustments of the offset and the amplitude were provided by the amplifier circuit (A3) which followed the D/A converter. The same circuit had the provision (U3B) to increase the output signal during the high pass filter mode of operation. This allowed the same digital designation to be used for the cut-off mass in the high pass mode as in the more prevalent band pass operation.

Inverted and readjusted to establish a nominal RF to dc ratio by A4, the output of A3 became the control signal for the RF generator. Also, the A3 signal was used as the reference voltage for the multiplying ratio control DAC (U1). The two MSB's of this unit were held at +5 volts while the rest of the inputs were used to establish a fine control over the ratio. The output signal of the DAC was used to generate the +DC and the -DC voltages for the mass filter. The quadrupole bias (Q1) was injected as a common mode voltage through the output amplifiers A5 and A6. During the high pass filter mode of operation, the DC component was reduced to zero by a signal at the TOT ION terminal. The FET gate U3 increased the gain of A3 and reduced the voltage at the two MSB's of U1 to zero. Of course, the LSB's of U1 were also set to zero by the digital control unit.

C. Bias Circuits

Four bias voltages were provided to the main filter and associated structures. The conical front end of the instrument containing the sampling orifice was biased with respect to the vehicle by the "draw-in front end bias" (FEB) potential. For this purpose, sixteen preset

voltage levels in the range from +40 to -20 volts could be selected by the flight program. The remaining three bias voltages Q_1 , Q_2 and ACC were floated on this potential. These were the bias signals for the rods of the mass filter, for focusing rods located between the orifice and the main quadrupole, and for the accelerating electrode respectively.

The bias circuits are shown in Figure 3. The amplifier A1 generated the FEB signal. The polarity and the amplitude of that bias signal were determined by the status of the multiplexer circuit U1. When the control signal at D of U1 was high one of the R_p resistors (not shown) was connected to pin 5 of A1. A low signal at terminal D connected one of the R_n resistors to the amplifier. In the former case the output of A1 was a positive voltage while in the latter case either a positive or a negative output was obtained. When the selected resistor was returned to ground a positive FEB was generated. A resistor returned to a positive reference caused a negative FEB. The selection was accomplished by the appropriate signal at B_0 , B_1 and B_2 .

The FEB signal had to pass through two relays before reaching the cone of the instrument. The relays isolated and protected the amplifier from vehicle and/or instrument ground during the laboratory tests and during the initial phases of rocket flight. The latching relay (S_2) provided protection mainly during the testing phase. Before the flight it was latched into ON position. Then S_1 interrupted the signal flow to the cone, which at that time was shorted to the vehicle ground by the cap which sealed the entrance orifice of the mass filter. Only after the cap has been removed and the HV and RF

circuits have been activated, the relay allowed the FEB signal to reach the cone.

The same signal that controlled the selection of Rp/Rn resistors determined the polarity of the other three bias signals. Transistors Q₁ through Q₄ formed a switch which selected the polarity of the voltage supplied to the current source Q₇ and Q₈. The current source was used as a power saving measure. The current through the diodes CR7 through CR9 generated the ACC and Q1 bias signals.

During the testing and the calibrating process of the instruments, it was determined that in the negative ion mode the bias level to the focussing rods (Q2) had to be one half the level of Q1 bias. In the positive ion mode both the Q1 and the Q2 biases had approximately the same amplitude. To generate the Q2 bias the simple voltage divider circuit consisting of CR 22, 23 and R47, 48 was used. The three biases Q1, Q2 and ACC were referenced to the front end bias signal.

The monitor circuits consisting of amplifier A2 and the associated components converted the bipolar bias signals to the positive zero to 5 volt signals for transmission through telemetry.

D. RF Generator

A modified Hartley Oscillator was used to generate the RF component of the quadrupole excitation signal. The oscillator operated in the vicinity of 2.3 MHz when loaded with the 81 pF capacitance presented to the circuit by the two sets of rods of the prefocusing quadrupole and the mass filter. In order to cover the ion clusters up to 255 amu at that frequency, the circuit had the capability to drive

the two quadrupoles with nearly 700 volts of peak signal.

The oscillator and its associated control circuits are shown in Figure 4. A 40-volt battery was used to supply power to the oscillator. The supply voltage was pre-regulated to 37 volts by the series pass transistor TIP 142. This was done to avoid a possible secondary breakdown in the power transistors of the oscillator by an excessive voltage of a power supply. Also, for protection of the same transistors a current sensing shut-down circuit consisting of two-741 operational amplifiers was employed. Whenever the current exceeded a preset level of approximately one ampere the circuit interrupted the base drive to the oscillator transistors. The ensuing reduction in the collector current released the base drive, which in turn increased the collector current once more. The resulting oscillation continued until the base drive originating in the RF control circuit was reduced. This type of current limiting generated a transient which ensured a restart of the oscillator. The more conventional current limiting circuits which were tried had the tendency to stop the RF oscillator while still permitting considerable current to flow.

The control over the amplitude of the RF signal was exercised by a CA3140 op-amp providing the base drive current to the power transistors. This amplifier received signals from the RF control circuit of the sweep generator and from the RF feedback circuit employing another CA3140 operational amplifier. The RF feedback signal was derived from one of the windings of the output transformer. A capacitive divider attenuated the signal which in turn was processed by a dc restoration circuit followed by an integrator.

This approach had an advantage over the conventional peak detector where the feedback signal was derived either from a separate winding or from a tap in the output winding of the transformer. In both cases the harmonic content of the derived signal was high. The peak detector followed the distortion of the signal which varied with the output level. This resulted in a nonlinear control of the RF signal. The harmonic content of the signal derived directly from the resonant output circuit was greatly reduced. The integrator further removed any spurious disturbances. Since the average voltage of the signal in the dc restoration circuit is proportional to the peak of the RF voltage, an effective control over the amplitude of the RF signal was maintained. The nominal ratio between the RF and the dc was established by the two resistors (16.2K and 20K) connecting the RF control and the feedback signals to the summing junction of the oscillator driver operational amplifier. The feedback signal when isolated and attenuated by the -741 circuit also served as the RF monitor signal.

E. Data Circuits

The electron multiplier current was measured by a logarithmic electrometer similar to the circuits that have been used to process multiplier outputs in earlier mass spectrometers flown by AFGL. In this Cluster Ion Mass Spectrometer application the electrometer and its associated circuits were referenced to the high voltage potential of the multiplier anode. Therefore, to translate the data to the vehicle potential, a voltage to frequency converter and an optoisolator circuit was employed.

The electrometer and the data conversion circuits are shown in Figure 5. The logarithmic current to voltage converter was constructed using a MOSFET operational amplifier (A1) with bias currents at least two decades below the minimum current (10^{-12} A) to be measured. One half of a dual transistor (Q1A) produced the logarithmic current-to-voltage conversion. The other transistor of the pair (Q1B) was used to establish the customary temperature independent pivot point. To set this operating point current, an operational amplifier (A2B) produced a constant voltage (within the stability limits of the -15V supply) across R_2 . Thus a constant current was established in the collector circuit of Q1B. With this stable operating point set, temperature compensation at other current levels was provided by the silicon resistor R_5 . This resistor in conjunction with R_6 and R_4 also established the necessary gain in the log amplifier to provide a one volt change in the output for each decade of the input current. Six decades of current could be accommodated by this circuit. The lower limit of the input current was set by the Q2 circuit.

The negative output voltage of the log amplifier was translated into the positive voltage range by the A2A circuit. The lower limit of the signal presented to the voltage-to-frequency converter (U1) was set at +2 volts. The converter was adjusted to produce approximately 20kHz at this input level. Each additional volt of the input signal increased the output frequency by 10kHz. A CMOS inverter/buffer served as a driver for the optocoupler U3. A single transistor (Q3) circuit was used to amplify the output signal and to drive the cable connecting the mass spectrometer to

the signal processing circuits located in the control section of the instrument.

F. Power Control and Supplies

Four separate supplies were used to provide power to the instruments. Two of the units supplied the high voltages needed by the electron multiplier and the positive ion target. The low voltages for the quadrupole control and excitation signal circuits were obtained either directly from a battery or through a multi-output dc-dc converter. The needs of the floating electrometer circuits were supplied from a separate dc-dc converter with a high input-to-output breakdown rating. Magnetic latching DPDT relays were used to control the power transfer from external or internal sources to the power supplies.

The positive ion target and the electron multiplier voltages were obtained from Venus Scientific, Inc. supplies K-30Z and MG-12 respectively. The output levels of both supplies were controlled by the circuits shown in Figure 5. While the K-30Z required external control to set the output voltage, the remote capability of the MG-12 was utilized for accessibility and convenience only.

The low voltage needs were supplied by the dc-dc converters shown in Figure 6. To reduce voltage transients generated by the power supply, an oscillator and the pulse shaping circuits (U1 and U2) drove the low power inverter stage (Q1, Q2, T1) at 25kHz. The output of the transformer T1 provided the drive to the power output stage (Q4, Q5, T2, T3). To limit the number of secondary windings per transformer as well as for packaging purposes, two potcore output transformers were used. Since a pre-regulator (U3, Q3) was employed to supply the input power

to the inverter, most of the output voltages did not require additional regulation. Only the $\pm 15V$ outputs supplying the more critical circuits were provided with additional regulators. Also, where needed, on-board regulators were included in other parts of the instrumentation. For example, the digital control and data processing circuits contained their own $\pm 15V$ and $+ 5V$ regulators which were supplied from the unregulated $\pm 20V$ output of the main power supply.

External power to the dc-dc converters was supplied through S1. The high voltage supplies could be activated only when a voltage was present at the output of S2. The same relay (S2) provided means to switch between the external and the internal power sources. The RF oscillator power was controlled by S3 and S4. The former selected external or internal power source operation while the latter turned the RF excitation on and off. Provisions were included to activate the RF oscillator and HV circuits with a positive pulse at the "timed on" input. This pulse was generated by the vehicle timer after the rocket burn has been completed and a relatively vibration-free flight has been achieved. At the end of the flight the instrument could be turned off by another pulse from the vehicle timer applied to the "timed off" terminal. The S5 relay connected the negatively charged capacitor C2 to the control coils of the S2 and S3 relays to switch them to then nonexistent external power source.

III. MANUAL CONTROL UNIT

A. Overview

A digital instrument was constructed which allowed the operator to exercise manual control over the operation of the mass filter. Its main purpose was to provide flexibility in laboratory experimentation and calibration procedures. It also provided means to test the operation of the flight control unit, to decommutate the PCM signal and to convert it into an analog form.

The control commands were entered via a keyboard. The operator could select any amu region, within the capability of the instrument, which the mass filter had to scan. Ratio between the ac and the dc components of the quadrupole excitation signal, the modes of operation and the bias voltages could be manipulated at will. A provision was included to select and to convert into an analog form any one amu domain within the sweep range. The internal clock of the flight control unit could be substituted with an external one whose frequency could be varied in binary form from 80kHz to 1.25 kHz.

The selected parameters were displayed. The amu's designating the limits of a sweep, the ratio and the selected single amu to be extracted from the PCM signal were indicated by seven segment LED displays. The bias selection and the mode of operation were displayed by single LED indicators. Two sets of these indicators were used. One displayed the selected bias and the mode of operation; the other displayed the same information but the signals were derived

from the PCM data. Thus the response of the instrument to a given command could be verified.

The circuits of the manual control unit could be subdivided into three groups. The first group translated the keyboard entries into binary commands. These commands were either relayed to the flight control unit or to the other circuits of the manual controller. The second group consisted mainly of various LED indicators and drivers. The third group was dedicated to handle the PCM data and to convert it into an analog signal for recording and display.

The keyboard continued numerical entry keys ZERO through F and a number of function keys. Three digit decimal numbers were used to enter data designating amu or ratio. A hexadecimal number was used to enter the last four LSB's which represented the 16 levels within one amu domain. Also, a hexadecimal entry controlled the bias and the band pass/high pass status of the mass filter. Numbers 0 through 7 selected appropriate bias levels in the band pass mode of operation while numbers 8 through F set the same bias levels in the high pass mode. Positive or negative ion operation was selected by a separate key in the function group. Other function keys routed the entered data into appropriate registers for execution and display.

B. Control and Display

The circuit drawings of the manual control unit are shown in Figures 7 through 11. The first three drawings show the control, the display and the interface with the EPROM circuits in the flight

control unit.

The necessary housekeeping waveforms and the external clock for the flight control unit were generated by crystal oscillator and a counter (U28,U29). The numerical entry's from the keyboard (X1 through Y4) were incoded into the hexadecimal code. When the function key • (DOT) was depressed before a numerical entry, the encoded digit was stored in U12. Otherwise the encoded digits were stored in latches U13, 14 and 15. The outputs of these latches were converted into an eight bit binary number by U10. Also, the outputs of all latches were routed to the keyboard display for verification (Figure 9). Three seven segment and one hexadecimal LED displays were used to accomodate data from U13, 14, 15 and U12 respectively. Whenever data was to be entered into U12 by first depressing the DOT key, the seven segment displays were blanked out. Only BCD numbers were valid entries into the other three latches whose outputs were converted into binary code. Therefore, to alert the operator to an impending error, an invalid entry was also indicated by blanking the display. Once a valid data entry has been made, it could be transferred by depressing an appropriate key into the output latches U1 through U5. Thus, the STOP key loaded U1 and U2 with data which set the end point of the sweep through the amu range. If instead the ratio key were depressed, it would have loaded the eight MSB's of the entered data into U4 thus converting it into the ratio control information. The data stored into an output circuit was also latched as a decimal/hexadecimal number into an appropriate display (Figure 9) to indicate the operating status of the mass spectrometer.

Each time the flight control unit ended an execution of a given instruction set the data stored in the output latches of the manual control unit was sequentially transmitted to the output lines of the now replaced flight EPROM. The control of that sequence was derived from information originating in the address lines of that EPROM.

The circuits responsible for the generation of the various latching and other control signals are shown in Figure 8. Bounce eliminators (U23,30) were used on all the function keys except the clear signal key. The command to latch data into U12 and to blank the BCD keyboard display was generated by U21 connected to toggle everytime the DOT key was depressed. The latch signal appeared at pin 2 of U20 (2U20) when the encoder (12U17) generated data ready signal. At the same time the counter (U34) which generated the latching signals for the U13, 14 and 15 circuits was inhibited. The blanking command emerged at 15U32. The flip-flop (U21) was reset and thus the blanking was removed when any of the keys transferring data from the keyboard registers to the output latches were depressed.

The circuit group U7, 8, 9 and 35 generated the sequence (S1 through S6) in which the data from the output latches were made available to the flight control unit. The control signals (A0E, A1E and A2E) to generate the sequence were the three LSB's of the EPROM address which in turn were synchronized with appropriate latching signals in the flight control unit. The supply voltage for U7 also was obtained from the flight unit. This arrangement

together with the resistor network (U9) and the tristate capability of the output latches protected both the flight unit and the manual control unit from a possible damage when power to one of the units was disconnected.

The selection of an internal or external clock operation and the provision to stop the sweep on a selected amu were implemented by the U31 and U33 circuits. INT/EXT key toggled U31B producing a ONE or ZERO at 10U33 and forcing the flight unit to operate on external or internal clock respectively. The HOLD key toggled U13A. When pin 1 of 13A was in the high state, command to go into the external clock operation was transmitted to the flight control unit, while at the same time external clock was inhibited. This occurred only when A1E address line of the EPROM was high. This situation in turn existed only during the updating process when the amu sweep was returned to the starting point. Therefore, the flight control unit stopped within the amu domain displayed as the starting point of a sweep.

U32 served as a LED driver to monitor the status of the system. The LED's indicated whether the systems were running or stopped, whether it was on external or internal clock, whether the mass spectrometer power was turned on or off, and whether the PCM system decommutator was synchronized.

C. Data Circuits

The PCM decommutator and the data circuits are shown in Figures 10 and 11. The PCM train of pulses entered shift register (U3 through U7) which was driven by an inverted clock from the PCM generator of

the flight control unit. Comparator (U12 through U15) was used to detect the frame synchronization. Once the frame synchronization pulse was generated, data present in the first three shift registers were transferred into the latches (U8 through U11). The spectrum data were stored in U8 and updated every millisecond with the new information carried by the PCM signal. The output of this register was continuously converted into an analog signal by the AD7521 unit.

To select a single amu for display on a chart recorder, the information identifying that unit was transferred from keyboard latches into U22. The comparator U16 and 17 generated a pulse whenever a match occurred. The strobes and their timing required to control the latches were generated by the circuits of Figure 11.

The incoming PCM clock was converted into a series of narrow pulses centered within the clock period by the monostable U36. These pulses in conjunction with other waveforms were used to generate strobe signals. This allowed sufficient propagation delay time for the circuits driven directly by the clock waveform to settle before their data was transferred; thus any "race" problems were avoided. In coincidence with the frame synchronization pulse two strobe pulses were generated. One, the DATA strobe, latched the spectrum data into U8. The other, ID strobe, latched the status and the amu identification data into U10 and 11. It also generated, through 9U37, a pulse to dump a sample and hold circuit. This circuit (not shown in the diagrams) could be used, if desired, as a peak detector during the single amu selection

process. When in this single amu mode (2U27 low) the ID strobe was generated only during coincidence of the frame and the amu synchronization pulses.

The decade counters U31 and 32 monitored the number of spectrum data words to be latched into the U8 during a PCM frame. Reset was generated to clear the U8 register in the single amu mode after the nine words associated with that amu have been counted.

The circuit group U24 through U30 and D1, D2 and D3 displayed the selected amu when synchronization has been achieved. For that purpose a servo type binary to BCD converter was used. The latching amu identification bits A1 through A8 were compared with the binary count in U26. This counter as well as the BCD counter U28, 29 and 30 were reset by the ID strobe and then were driven from the same 80 KHz clock. When the binary count agreed with the amu identification word, the clock was inhibited. Both counters, the binary and the BCD, had the same count; thus the conversion was accomplished. The BCD number was displayed by the seven segment displays D1, 2 and 3.

Two additional counters, U33 and U38, were used to generate multiplexing signals for the various displays of the control unit and to detect loss of the PCM synchronization respectively. Both circuits are rather primitive and require no explanation. The manual control unit was packaged into a small metal suitcase and was powered from +28V external supply. The digital data latched into the storage registers was made available for further processing through interface connectors.

IV BALLOON CIMS

Preliminary work has been carried out on the definition and design of control electronics for a balloon borne mass filter. The filter, similar in principle to the rocket borne instrument, will measure ambient positive and negative ions in the stratosphere. Its range, however, will be extended to 1000 amu.

Some of the analog circuits developed to control the rocket borne mass spectrometer will be adapted for the new instrument. The circuits which may be used with only minor modifications include the various bias and monitor circuits and the sweep generator. The RF excitation generator may have to be modified for a two frequency operation. Some switching arrangement will have to be implemented to provide an RF signal above 3 MHz for a sweep through the lower amu range and to drop the frequency to one MHz for a sweep through the upper mass units. The basic design of the oscillator using the air core toroidal transformer and its housing will be maintained. Although the oscillator has been developed as a part of a masters thesis and was incorporated into the rocket borne instrument, a complete description of the unit could not be provided in time for this report. The thesis write-up will be completed during the follow-on contract and at that time will be issued as a scientific report.

The digital control and the data processing subsystems will be redesigned. The long duration of a balloon flight and the relatively slow changes in its environment offer an opportunity of

modifying the experiment based upon results currently being obtained. A microprocessor based control system provides the flexibility needed to control the instrument, to process the data and to furnish the facilities for an interaction with a ground based operator.

A microprocessor based control unit has been developed as part of an engineer's thesis. The complete thesis will be submitted in September, 1979, and subsequently will be issued as a scientific report during the follow-on contract. The μ P based unit has been limited to provide only the basic control functions for a mass filter. A set of instructions allow the user to assemble various programs from a library of stored repertoires of modes and use them in a pre-arranged sequence during the flight. These programs then exercise control over the various bias signals, establish the range of amu's over which the filter has to sweep and sets the maximum time interval during which data may be accumulated into a counter within each amu domain. The time interval during which the data was collected, the data and some other pertinent information such as the real time and the outputs of some environmental monitors are then stored into a temporary memory. No provisions have been included for an interaction with a telemetry system or a bulk data storage system.

Independently of this control unit development, a study has been conducted by Dr. Stuart of Northeastern University to establish guidelines for an overall control system for the balloon borne experiment. The study has been presented as an informal eight part

report to the contract monitor. Therefore, only the recommendations and future work suggestions presented in the summary of the report are included below.

RECOMMENDATIONS

1. A standard flight-proven complete system with the addition of the microprocessor system should be flown so that the existing hard-wired controller can act as a back-up to the microprocessor.
2. One signal on the tone command link should be allocated to change-over from microprocessor to hardwired control.
3. The overall software system should be based on a simple run-to-completion executive.
4. All software tasks including the executive should be table driven and consist of a program in ROM and a data base in RAM, the data base being modifiable by operator intervention.
5. One signal of the tone command link should be allocated to RESET to return the microprocessor to the beginning of the executive program.
6. A standard teletype should be used as a ground terminal (this provides only start and stop bits in addition to standard ASCII).
7. For error control on the teletype link reliance should be placed principally on echo checking.
8. Operator interaction should be restricted to those modifications which can be implemented by changing a small number of bytes in RAM.
9. Recognition of an operator request should normally be by polling but in addition some form of recognition via interrupt or reset is desirable.

FUTURE WORK

The immediate task is to firm up the specification of the first flight unit. Some of the areas requiring attention are listed below:

1. An overall plan for address space use should be drawn up. Some addresses must be reserved e.g., those used for reset and interrupt purposes. There may be more than one person involved in coding and they will be concerned with address allocation. This obviously needs coordinating.

2. For similar reasons a plan for the use of interrupts is also needed. Correct priorities need establishing.
3. A decision should be made whether to include the noise protection scheme involving the inhibiting of reading during periods of low signal strength. It will complicate the interrupt scheme and could probably be left as a future refinement. The decision will depend upon the expected performance of the radio link.
4. Precise requirements for the input to telemetry need defining. These should probably match as closely as possible those currently provided by the present hardwired system. There may however be complications. For example, if the TM requires continuous input for sync purposes the microprocessor system bus cannot provide it as it must be time shared with other tasks.

Quite a number of questions arise in connection with the recording of data. Consideration needs to be given throughout to maximizing the efficiency of memory use since as already noted this is a scarce resource.

5. The acquisition of the data from the mass spectrometer is the subject of a separate study as noted earlier. The range precision and format of what is to be recorded needs to be specified.
6. Four parameters are to be recorded along with the data - altitude, ambient temperature, vehicle/space potential probe and real-time clock. All need specifications for range precision and frequency of recording.
 - The formatting and correlation with MS data needs defining.
 - Interfacing should be examined, e.g., A/D converters may be needed in some instances which may or may not be provided by existing equipment.
 - Initialization should also be examined, e.g., for the real-time clock.
7. Five parameters are to be monitored - pump pressure/ion pump, cryogen level, cryogen vapor flow, package temperature and package pressure. The form of monitoring needs defining. Typically each could be tested to check that it is within a prescribed range, in which case the ranges need defining.
 - Action to be taken if an out of range value is encountered needs defining. Possibly this could be recorded and sent to the terminal by the equipment checkout routine. Again range precision and format of each parameter needs defining. It would also need identifying, e.g., by assigning a 2 letter identifier to each

parameter. The identifier could be stored in the program and transmitted to the terminal with the value of the parameter concerned. Much depends here upon what facilities are available to correct such faults.

8. The mass spectrometer can be operated in various modes such as positive or negative ion, total ion etc. These may be selected from the ground. It is also possible to control to some extent other parts of the equipment e.g., by switching motors or heaters etc. Three methods are available - tone command "ASCII push button" and software. All such operations need listing and specifying. If it merely entails assigning one of the tone commands, the system currently under discussion is not affected. If ASCII push button is used, code words should be assigned and I/O ports allocated. Any interfacing hardware should be defined. If software is involved, the routines should be specified and developed.
9. The parameters to be checked by the initialization routines need defining. This routine should be concerned principally with hardware operations. Software initialization should be confined to that needed by the executive and anything needed by the overall system (e.g., the real-time clock). The individual tasks should take care of their own initialization as far as possible.

PERSONNEL

A list of the engineers and technicians who contributed to the work reported:

J. Spencer Rochefort, Professor of Electrical Engineering,
Principal Investigator.

Robert D. Stuart, Professor of Electrical Engineering, Engineer.

Raimundas Sukys, Senior Research Associate, Engineer.

Thomas Palasek, Research Assistant, Engineer.

RELATED CONTRACTS AND PUBLICATIONS

F19628-74-C-0042	1 September 1973 through 31 July 1976.
F19628-76-C-0256	1 August 1976 through 31 October 1978.
F19628-78-C-0218	15 September 1978 through present.

Sukys, R. and Goldberg, S. (1974), Control Circuits for a Rocket Payload Neutralization and Other Topics", AFCRL-TR-74-0580.

Sukys, R., Rochefort, J.S. and Goldberg, S. (1975), "Bias and Signal Processing Circuits for a Mass Spectrometer in the Project EXCEDE: SWIR Experiment," AFGL-TR-76-200.

Rochefort, J. S. and Sukys, R. (1976), "Instrumentation Systems for Mass Spectrometers", AFGL-TR-76-200.

Rochefort, J. S. and Sukys, R. (1978), "A Digital Control Unit for a Rocket Borne Quadrupole Mass Spectrometer, AFGL-TR-78-0106.

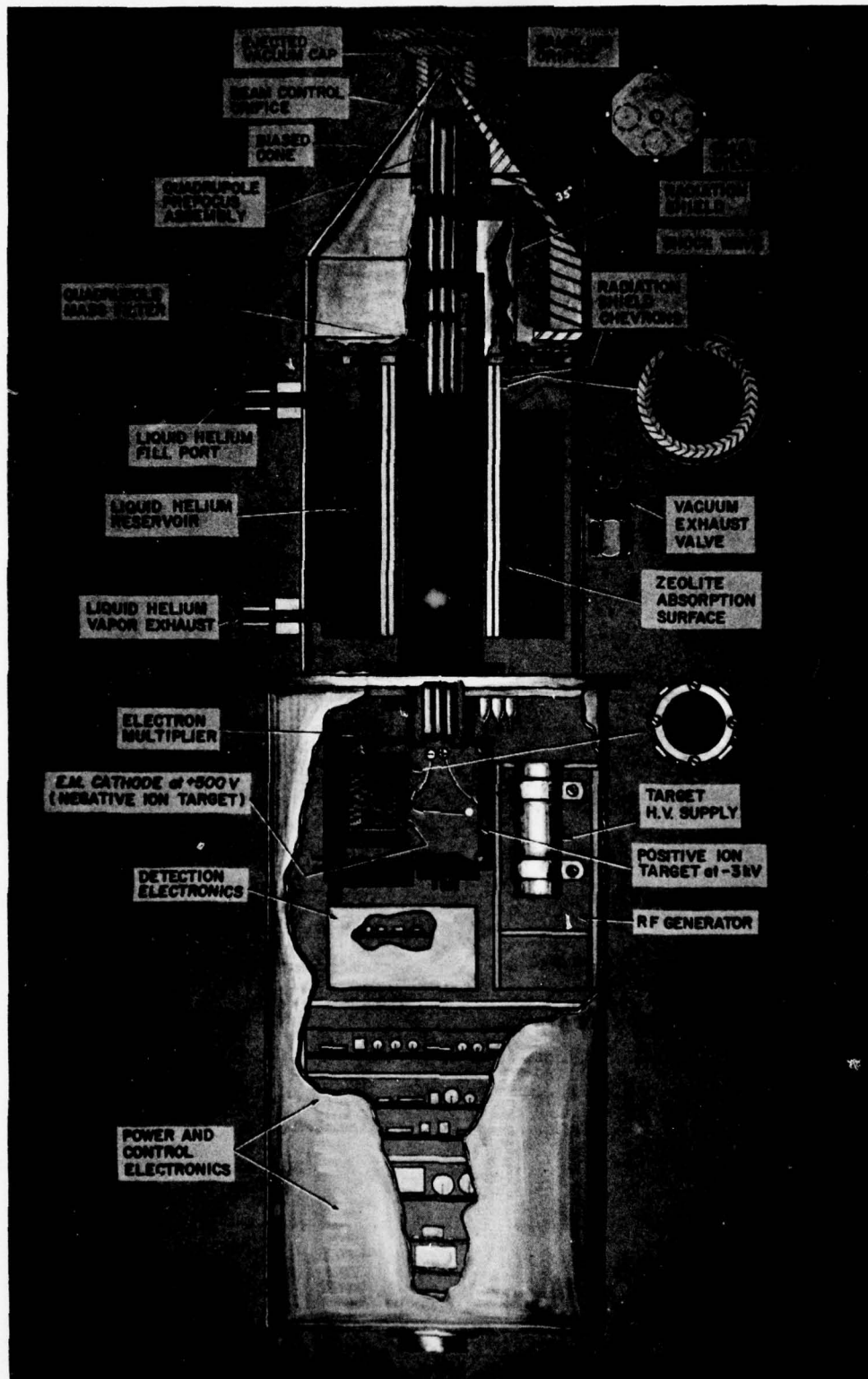


FIGURE 1. CLUSTER ION MASS SPECTROMETER
 (Courtesy of Aeronomy Division of AFGL)

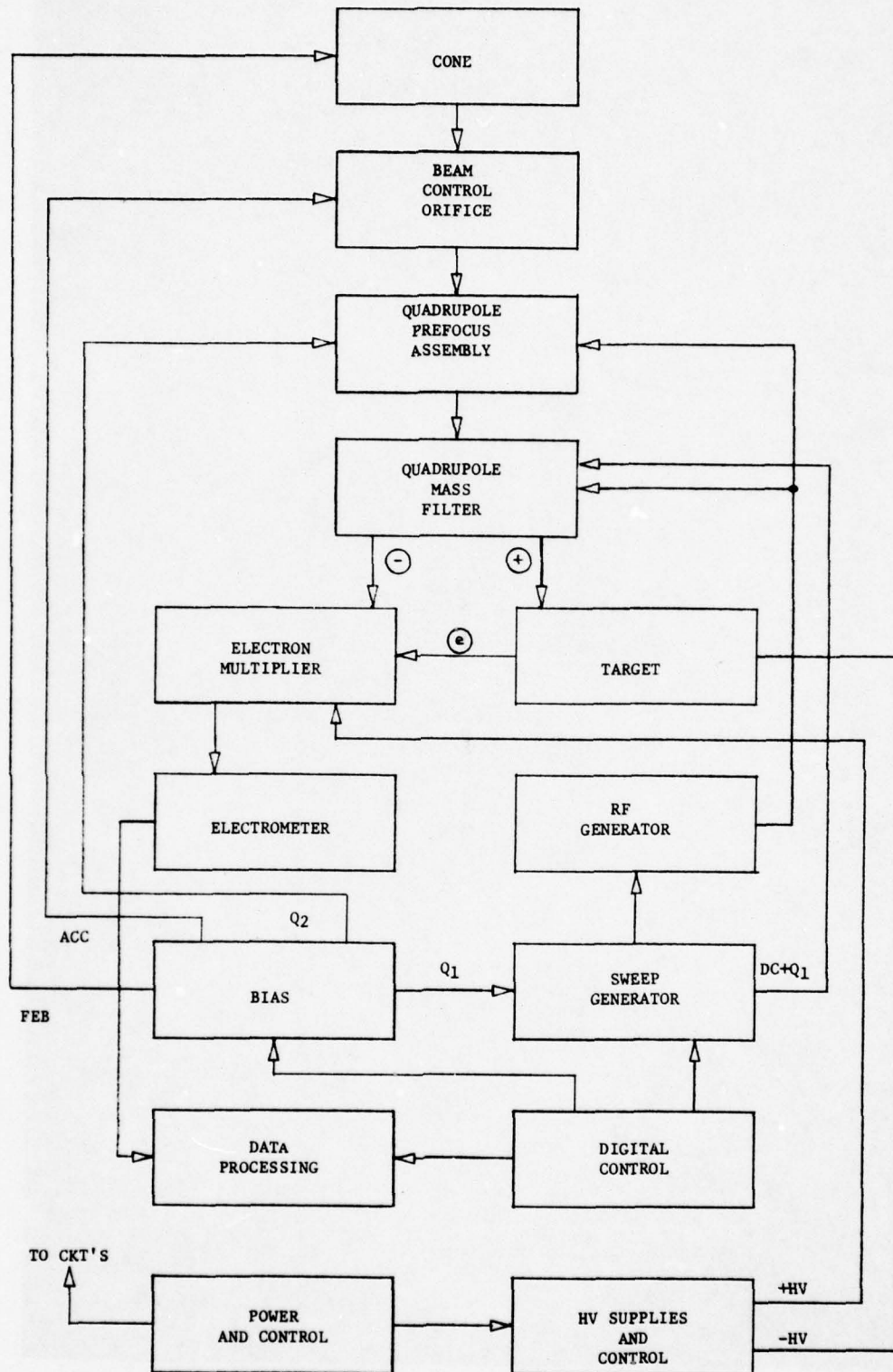
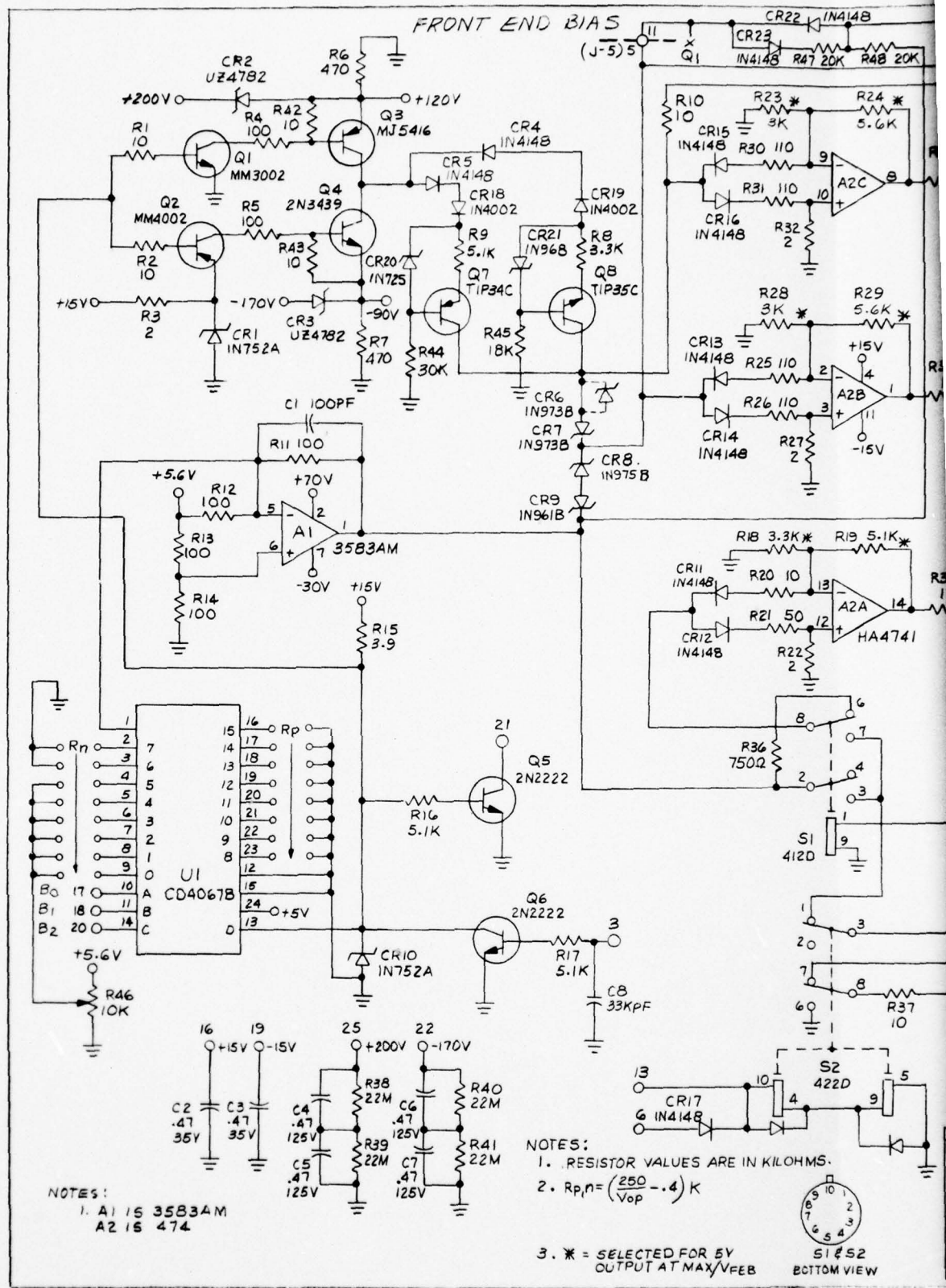


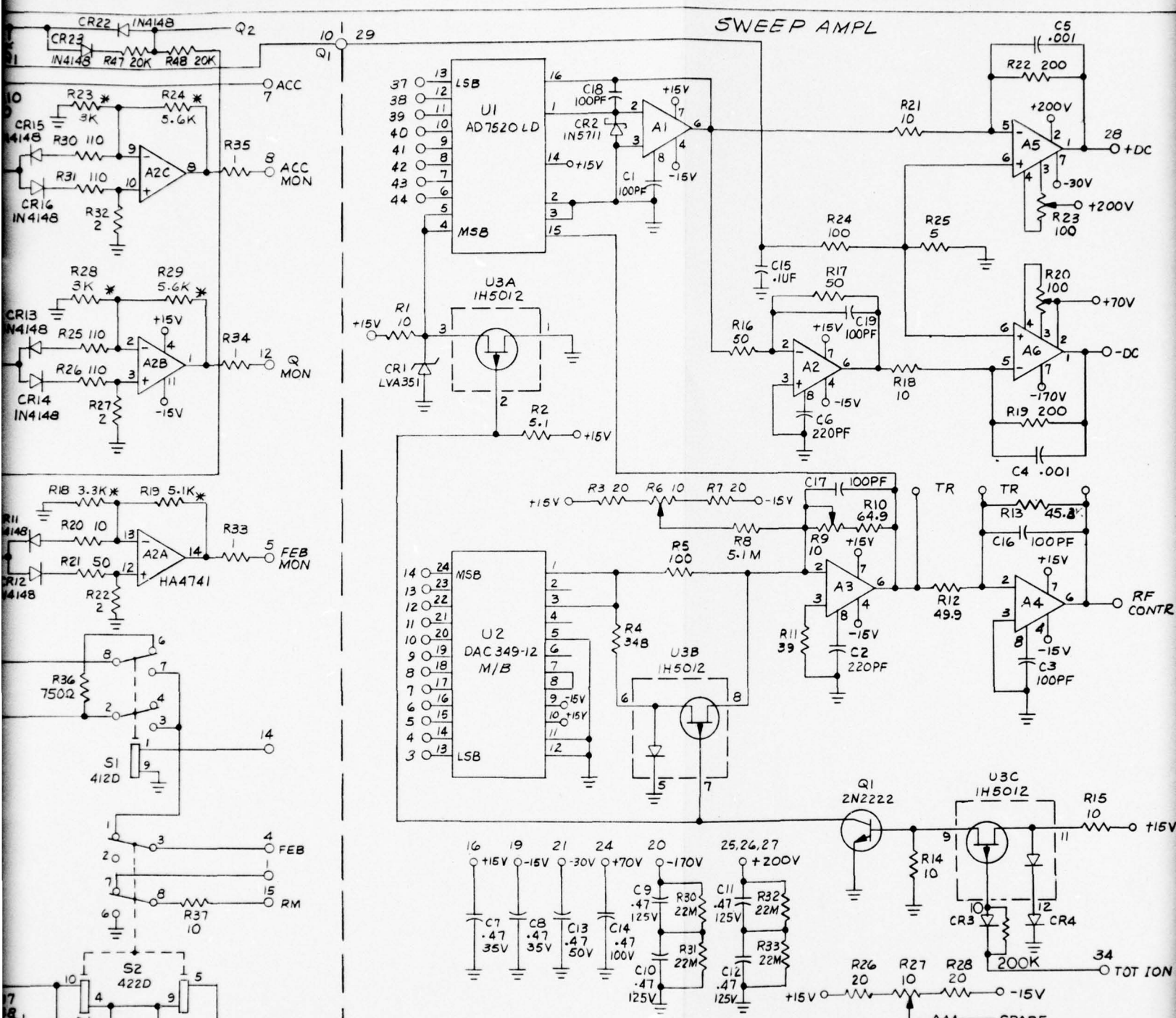
FIGURE 2. ELECTRONICS BLOCK DIAGRAM



NOTES:
 1. A1 IS 35B3AM
 A2 IS 474

NOTES:
 1. RESISTOR VALUES ARE IN KILOHMS.
 2. $R_{p,n} = \left(\frac{250}{V_{op}} - .4\right) K$
 3. * = SELECTED FOR 5V OUTPUT AT MAX/V_{EEB}





RESISTORS ARE IN KILOHMS.

S1 S2
 BOTTOM VIEW

HIGHEST REF DES

A2
 CB
 CR23
 QB
 R48
 S2
 U1

NOTES:

1. A1, A2, A3 AND
 A4 IS 208A
 A5 & A6 IS
 3583AM

HIGHEST REF DES

A6
 C19
 CR4
 Q1
 R33
 U3

FIGURE 3

TOLERANCE UNLESS OTHERWISE NOTED:	DRAWN	DATE	CONTRACT NUMBER
DECIMAL: 01	R.S. EVELYN	4-24-78	NORTHEASTERN UNIVERSITY
FRACTIONAL: 3/16	SCALE		ELECTRONIC RESEARCH LAB
ANGULAR: 3/16	EXTERNAL		COLLEGE OF ENGINEERING
SURFACE FINISH: 150			BOSTON, MASS. 02115
DRILL ALL SHARP EDGES AND GROUNDS			
FINISH			
APPLICATION			

CLUSTER ION MS BIAS & SWEEP CIRCUITS

FIGURE 3.

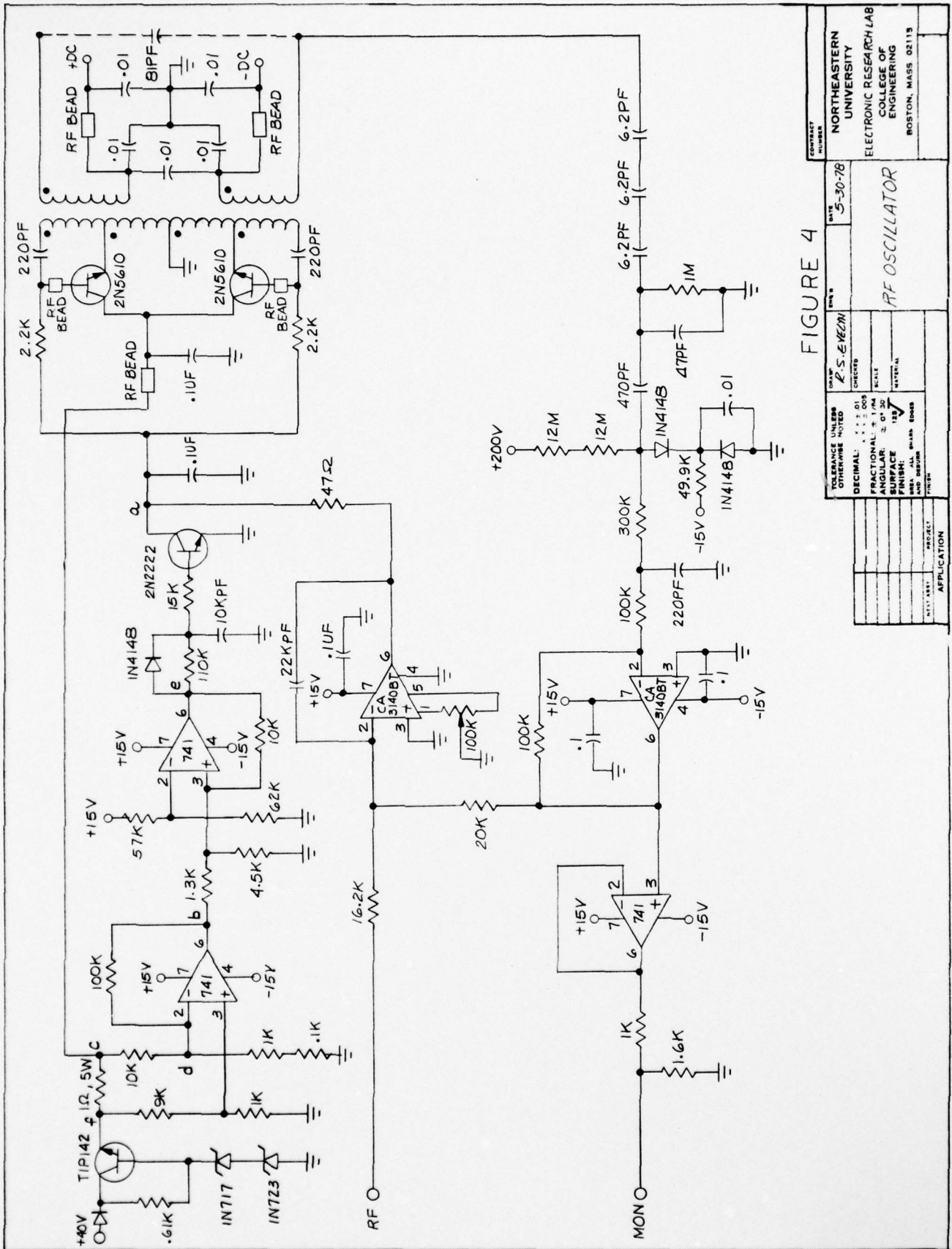
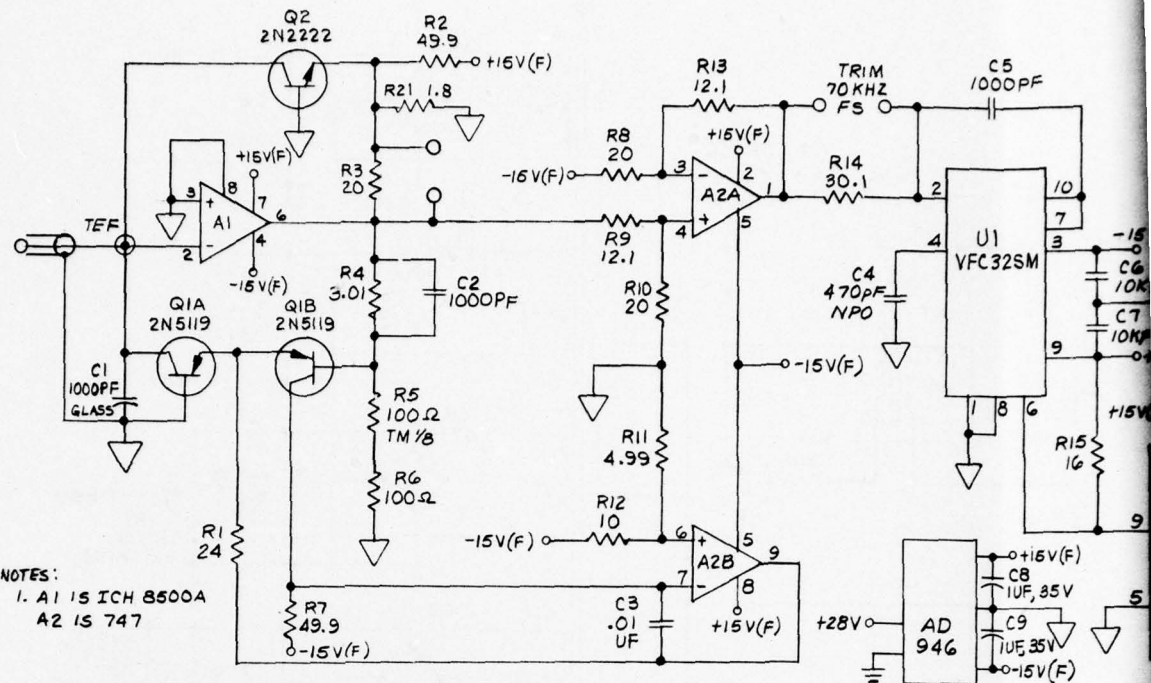


FIGURE 4

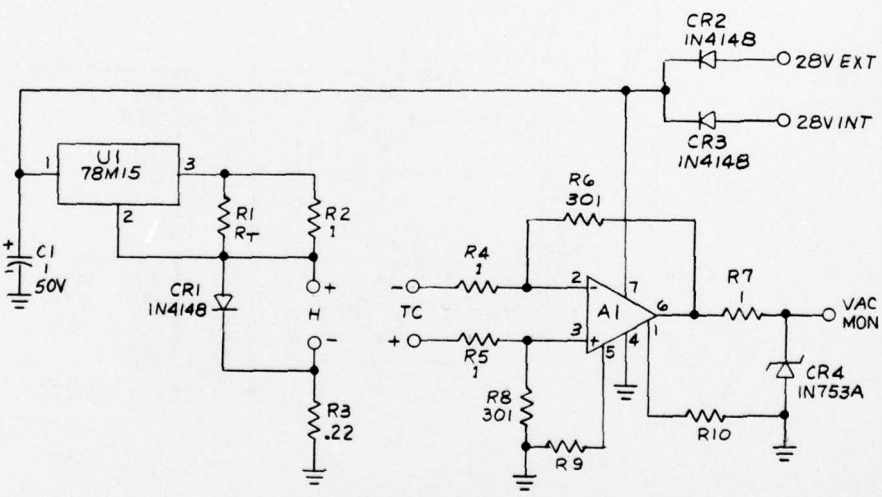
CONTRACT NUMBER	NORTHEASTERN UNIVERSITY
DATE	5-30-78
DRYER	2-S-EYE/IN
DRAWN	
CHECKED	
SCALE	
INSTRUMENT	
PROJECT	RF OSCILLATOR
DESIGNED BY	
DRAWN BY	
CHECKED BY	
DATE	
PROJECT	
APPLICATION	

FIGURE 4



NOTES:
 1. A1 IS ICH 8500A
 A2 IS 747

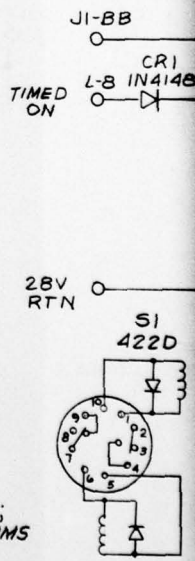
LOG AMPL



NOTES:
 1. A1 IS CA 3040
 2. R9 + R10 ≈ 100K
 SELECT TO NULL OFFSET

VAC MONITOR

NOTE:
 UNLESS OTHERWISE SPECIFIED;
 RESISTOR VALUES ARE IN KILOHMS



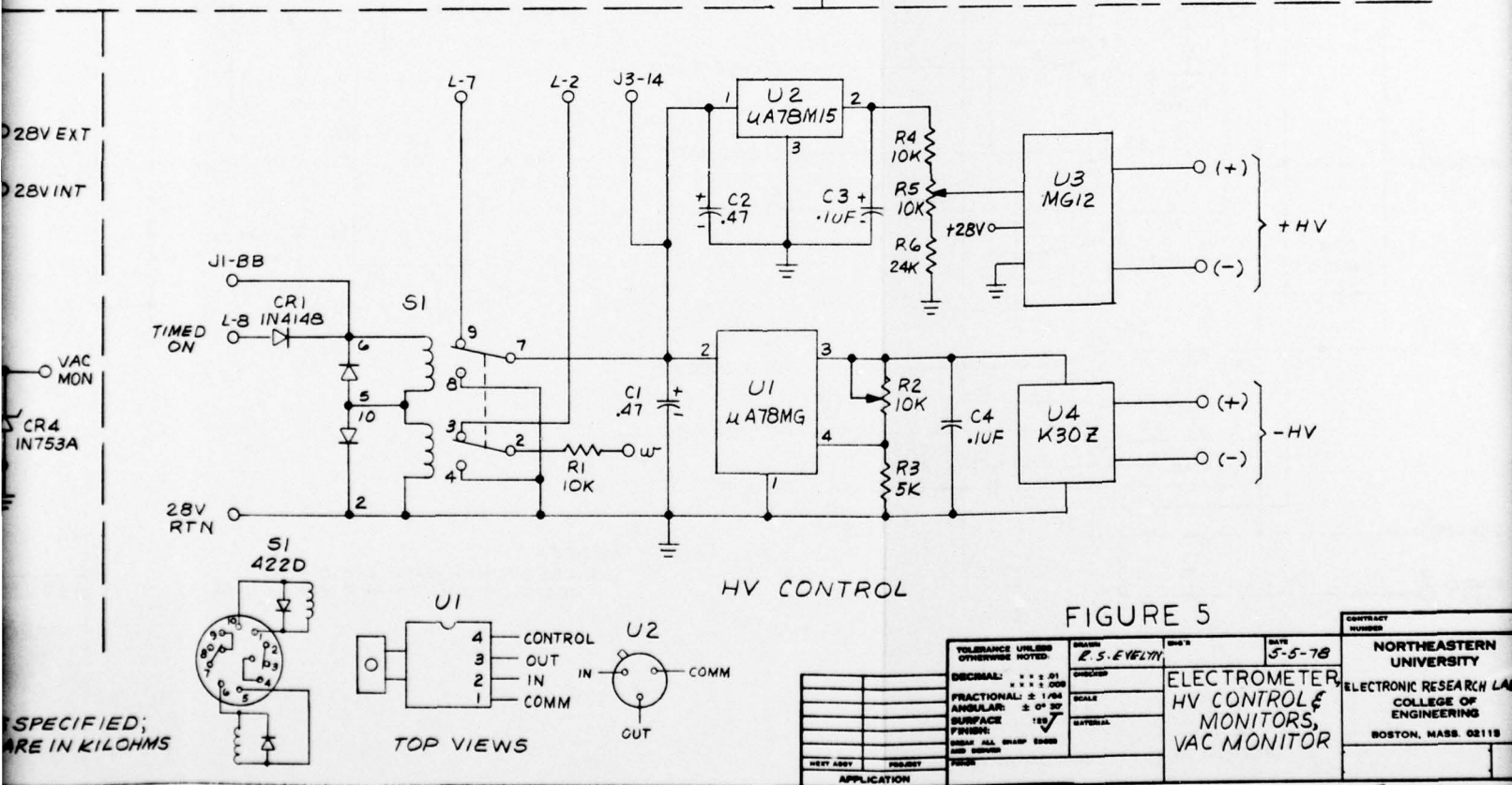
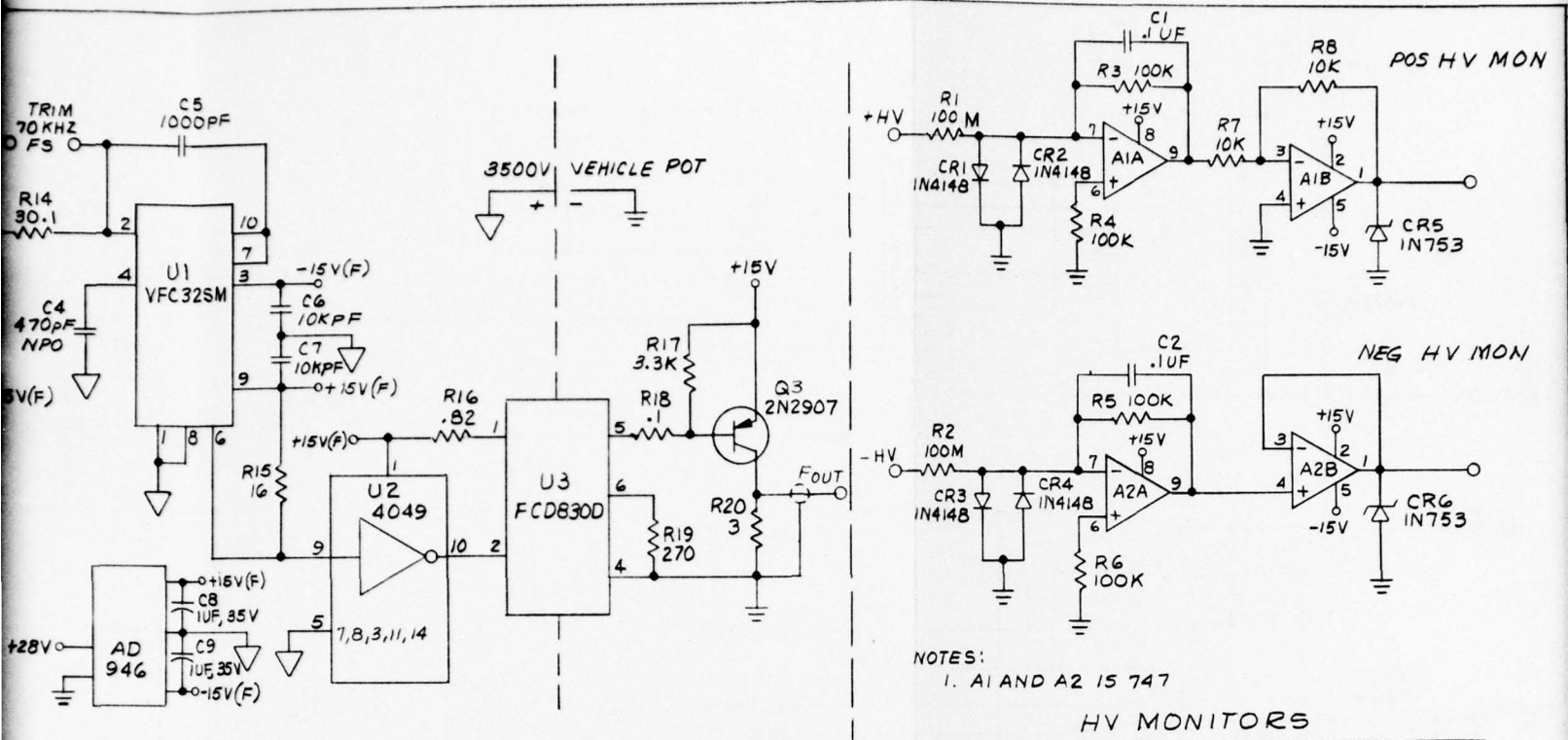


FIGURE 5.

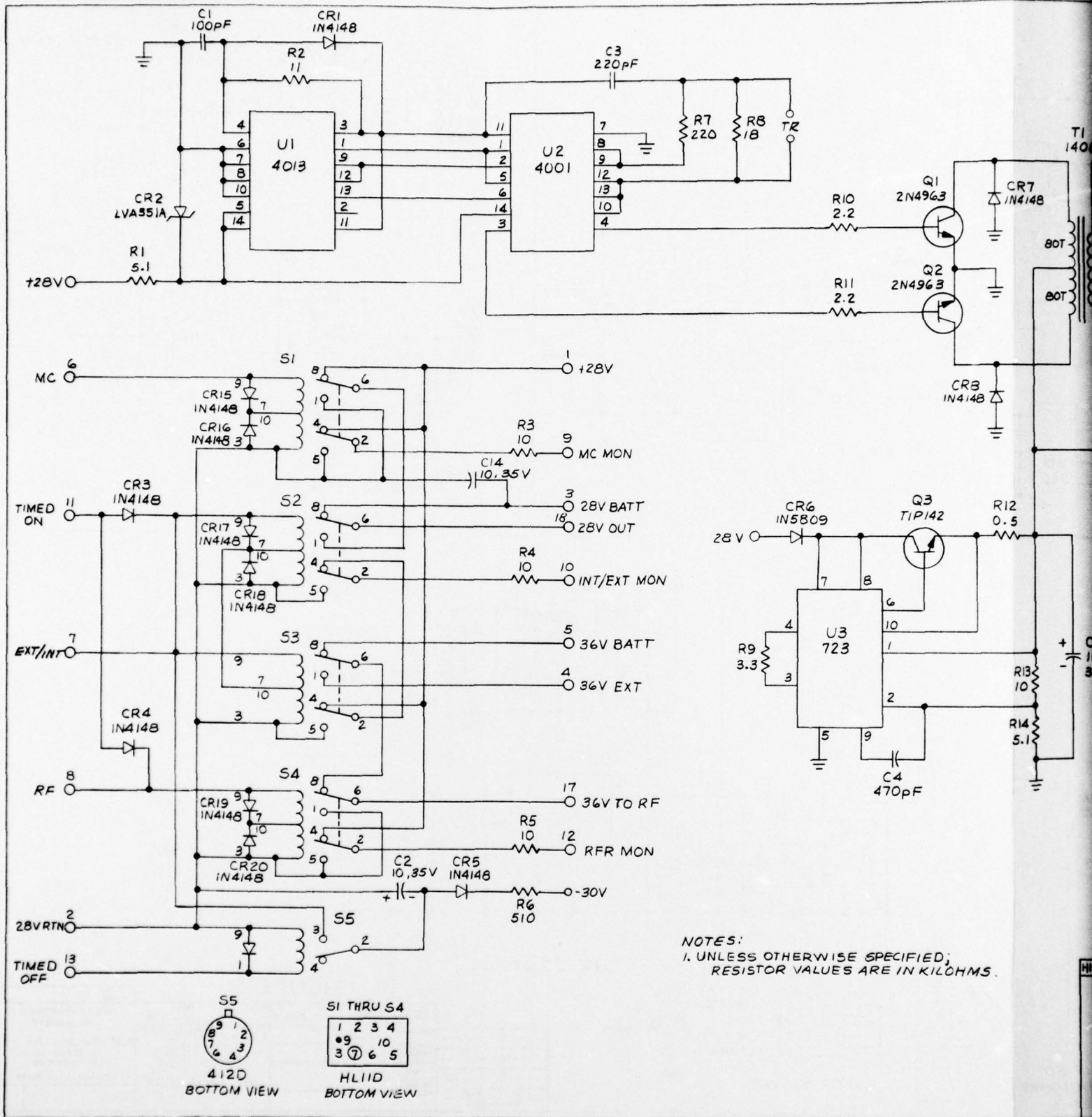
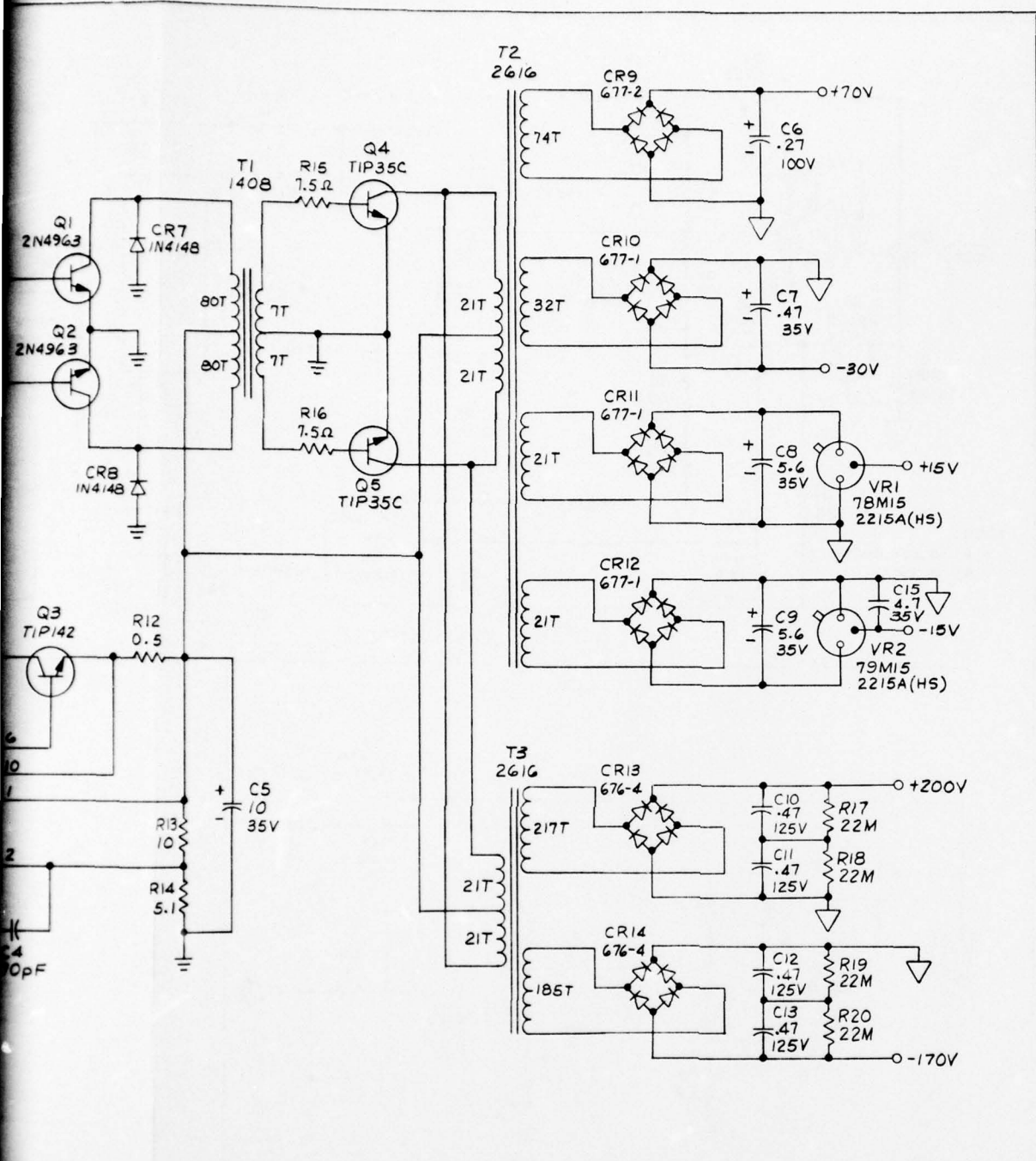


FIGURE 6.

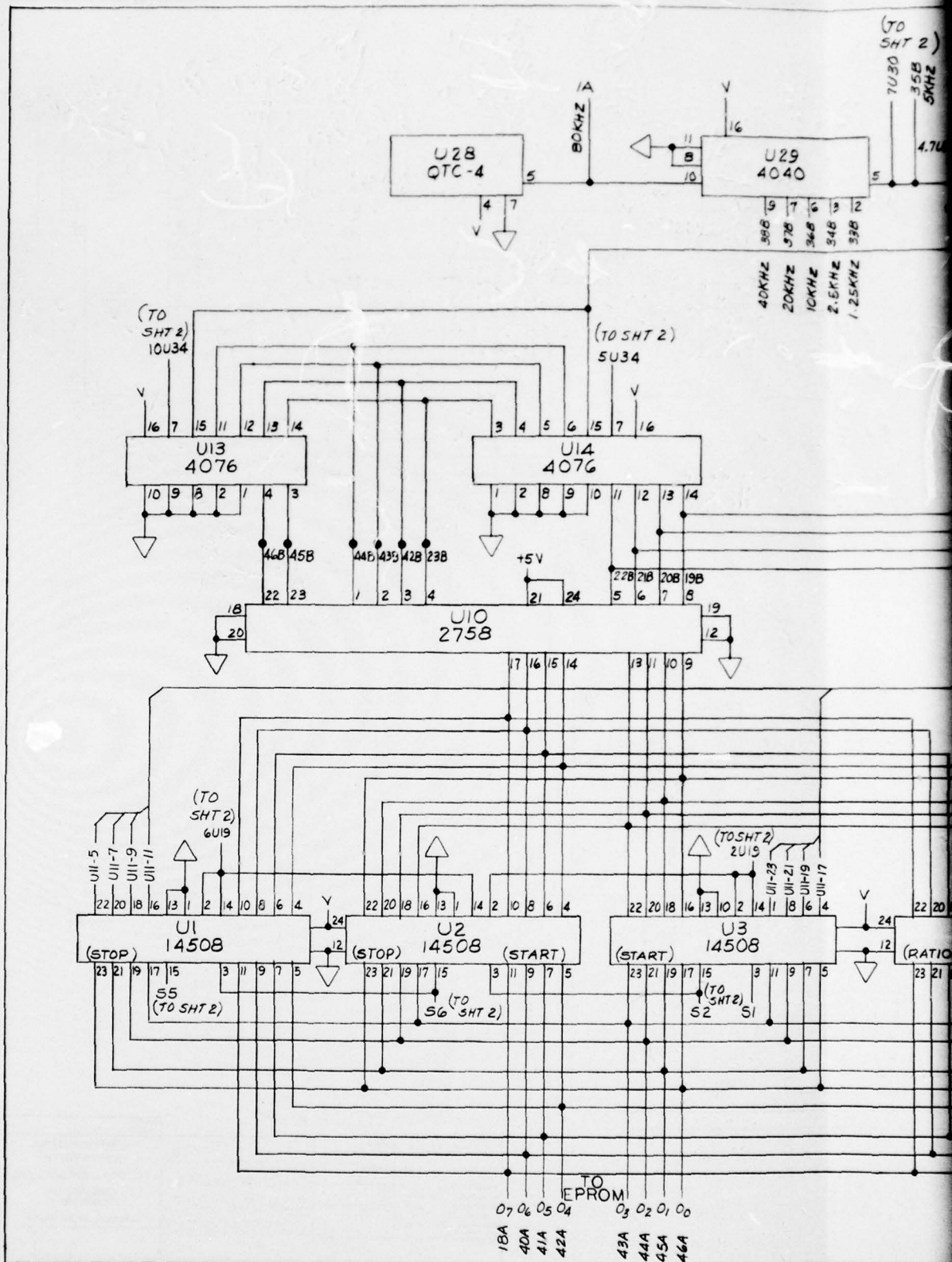


RESISTORS SPECIFIED IN KILOHMS.

HIGHEST REF DES.	
C15	
CR20	
Q5	
R20	
S5	
T3	
U3	
VR2	

FIGURE 6

TOLERANCE UNLESS OTHERWISE NOTED DECIMAL: .1% ± .01 .1% ± .005 FRACTIONAL: ± 1/100 ANGULAR: ± 0° 30' SURFACE FINISH: 150 V HONE ALL SHARP EDGES USE DIMENSIONS	DESIGNER R. S. EVELYN	DATE 4-27-78	CONTRACT NUMBER NORTHEASTERN UNIVERSITY ELECTRONICS RESEARCH LAB COLLEGE OF ENGINEERING BOSTON, MASS. 02115
	CHECKED SCALE MATERIAL	PROJECT CLUSTER ION MS POWER SUPPLY & RELAYS	
NEXT ASSY PROJECT APPLICATION	FIGURE		



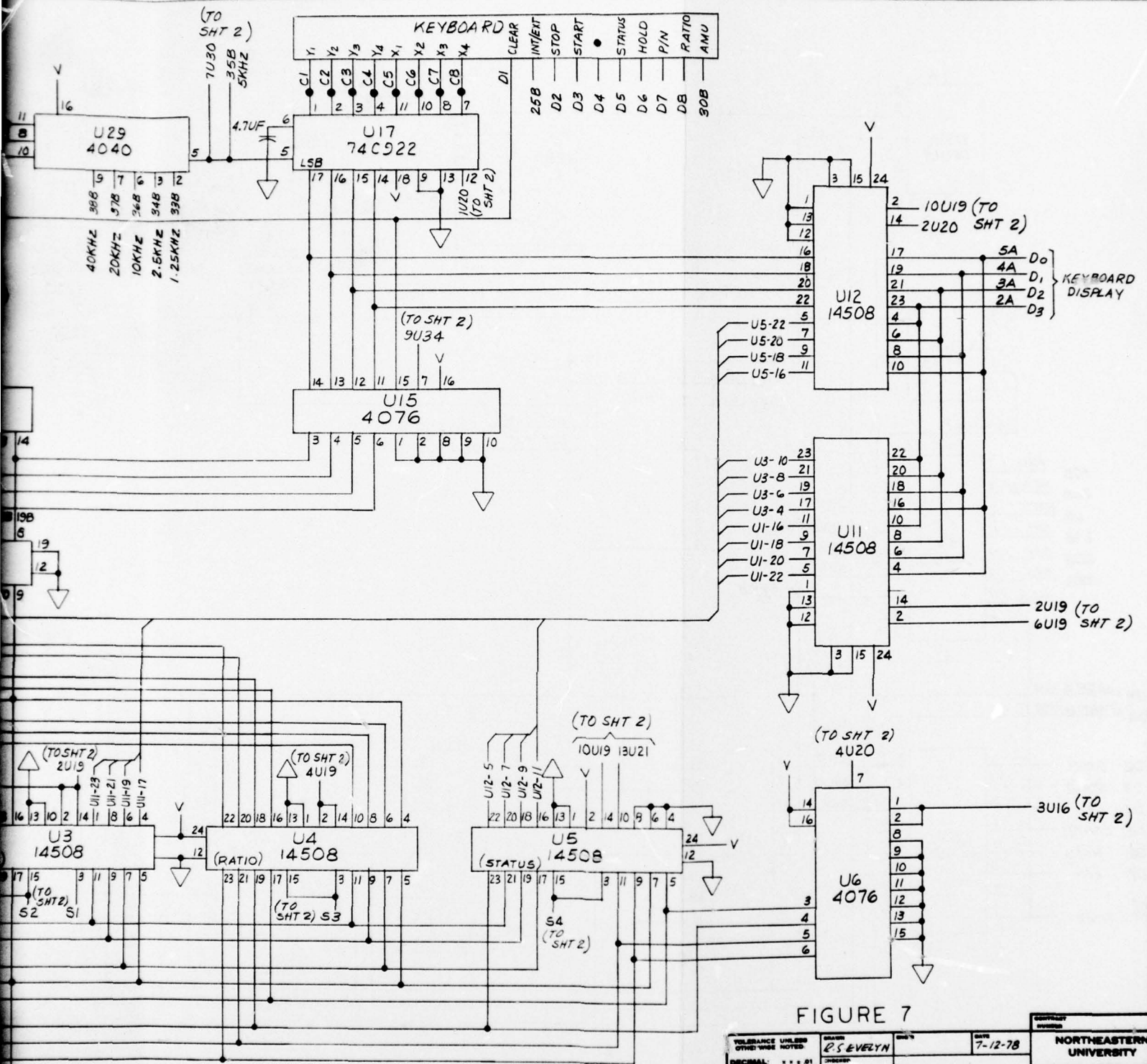


FIGURE 7

TOLERANCE UNLESS NOTED DECIMAL: .0005 FRACTIONAL: 1/64 ANGULAR: ± 9° 30' SURFACE FINISH: 125 BREAK ALL SHARP EDGES AND RADIUS TYPED	DRAWN: P. S. EVELYN CHECKED: [] DATE: 7-12-78	CONTRACT NUMBER: NORTHEASTERN UNIVERSITY ELECTRONIC RESEARCH LAB COLLEGE OF ENGINEERING BOSTON, MASS. 02115
---	--	---

FIGURE 7.

2

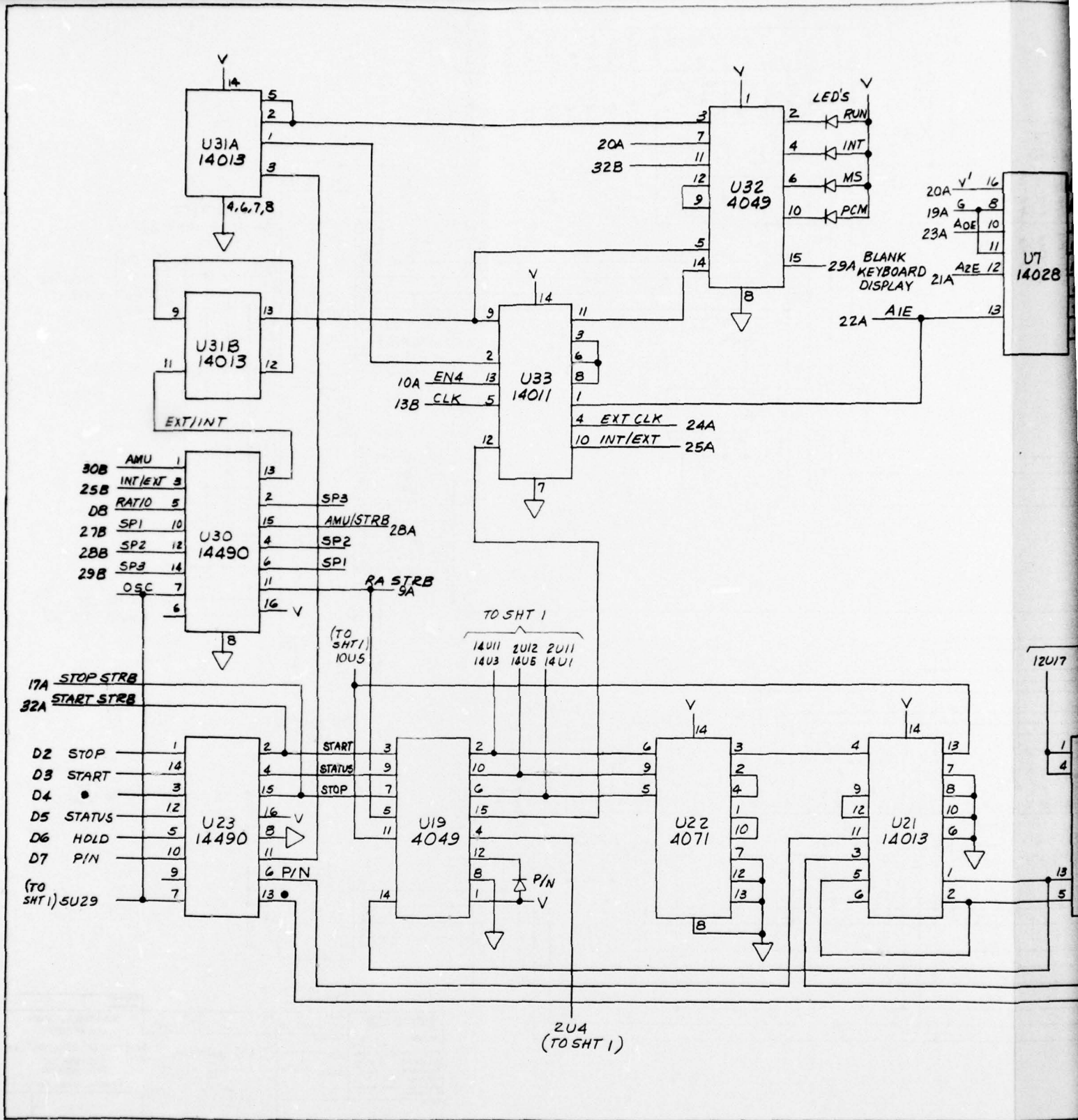


FIGURE 8.

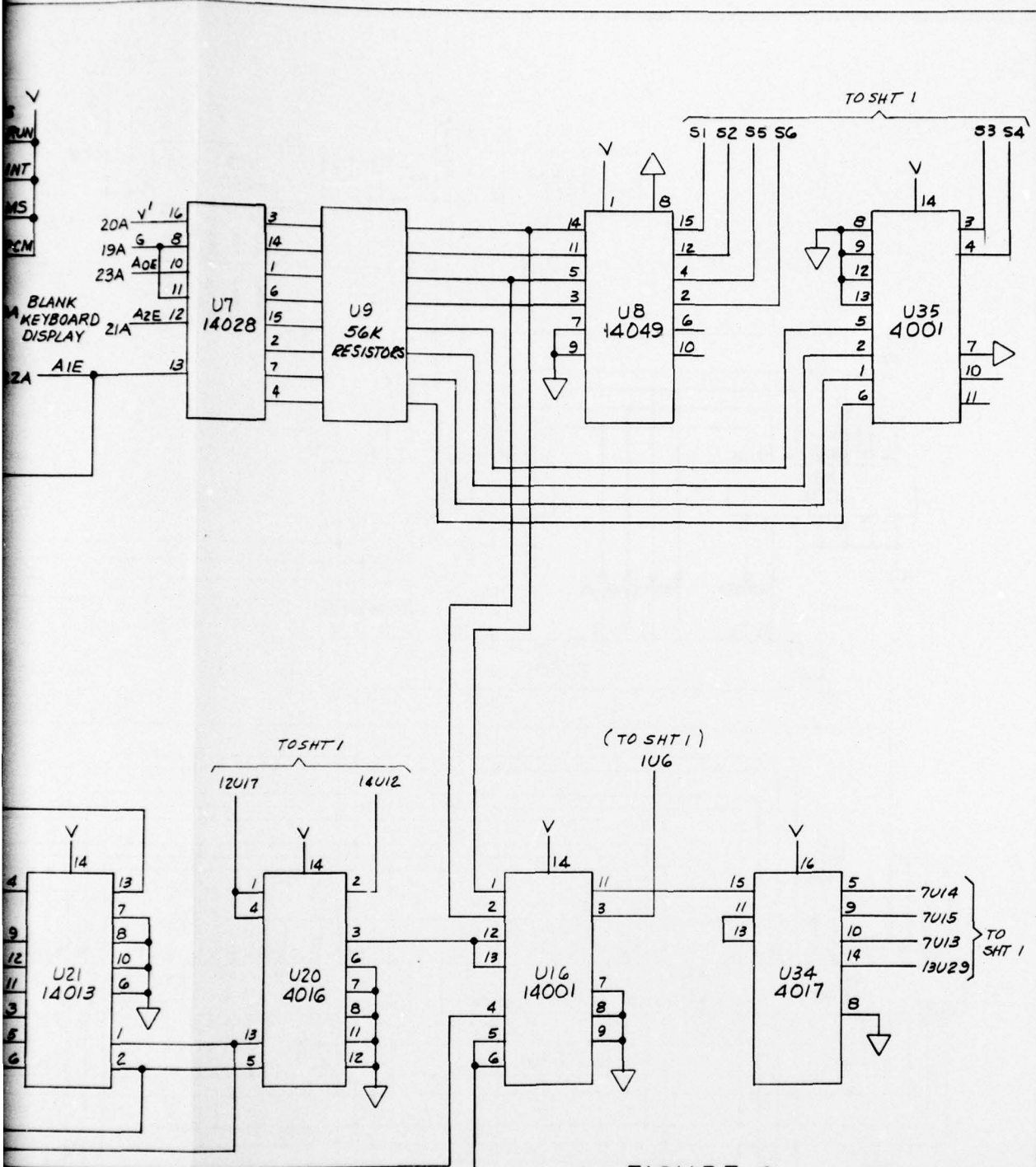
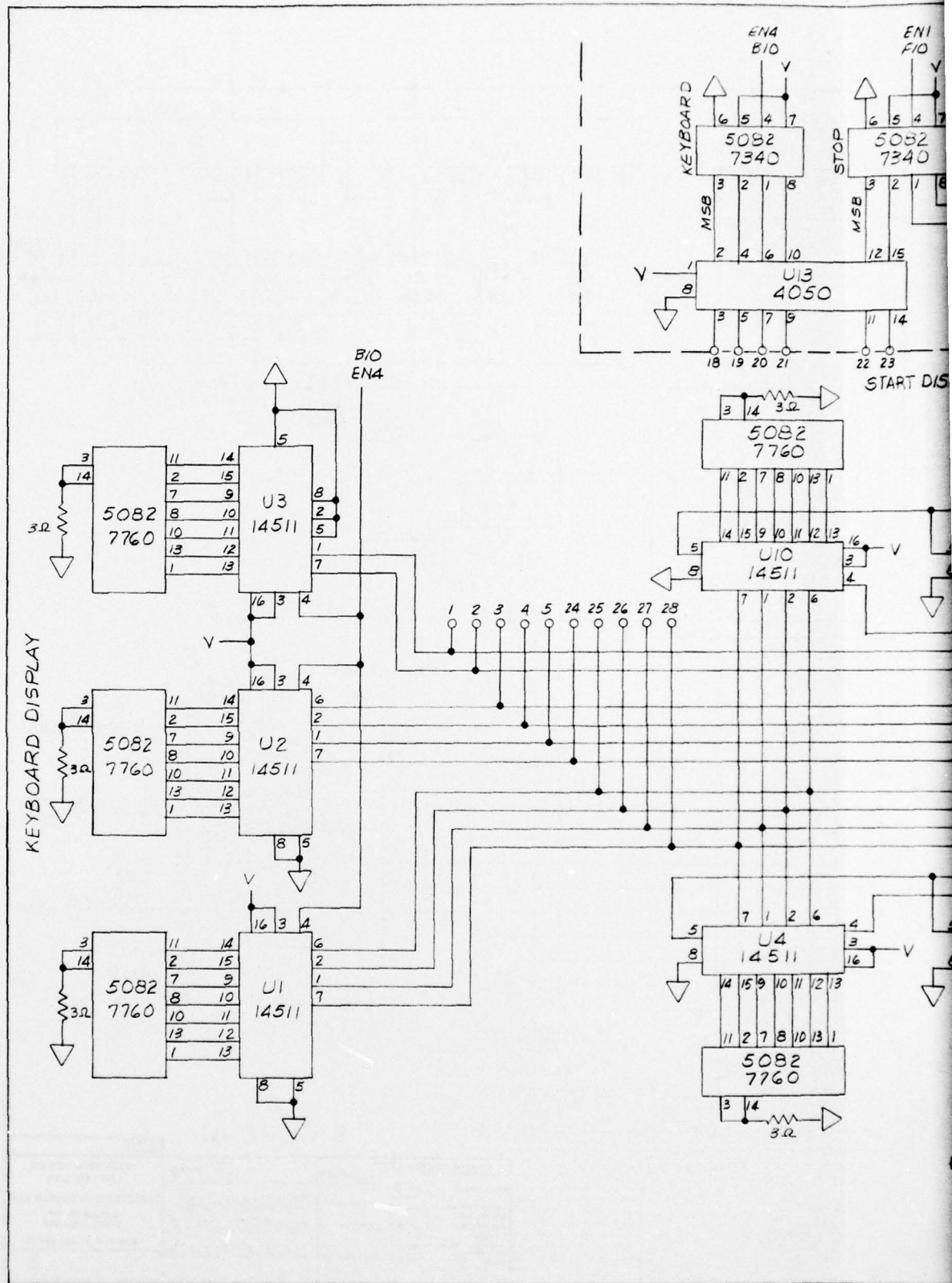


FIGURE 8

TOLERANCE UNLESS OTHERWISE NOTED: DECIMAL: ± 0.1 FRACTIONAL: ± 1/100 ANGULAR: ± 0° 30' SURFACE FINISH: 125 DIMENSIONS UNLESS OTHERWISE NOTED	DESIGNER: R.S. EVELYN	DATE: 7-12-78	NORTHWESTERN UNIVERSITY ELECTRONIC RESEARCH LAB COLLEGE OF ENGINEERING BOSTON, MASS. 02118
CIMS MANUAL CONTROL UNIT CONTROL CIRCUITS - 2			



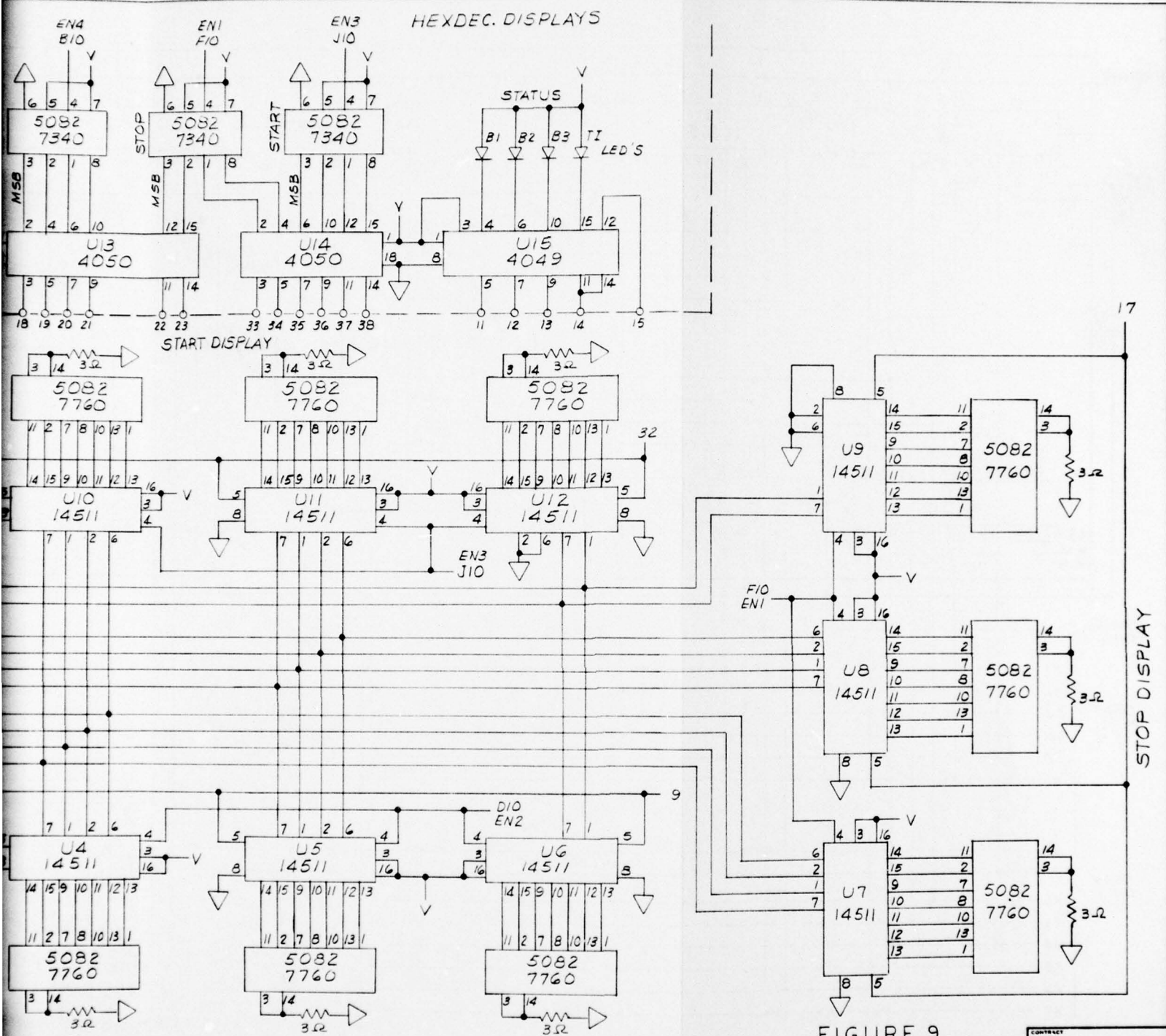


FIGURE 9

TOLERANCE UNLESS OTHERWISE NOTED		DESIGNED BY	DATE	CONTRACT NUMBER
DECIMAL: .01		R.S. EVELYN	7-26-78	NORTHEASTERN UNIVERSITY
FRACTIONAL: 1/64		CHECKED		ELECTRONIC RESEARCH LAB
ANGULAR: 0° 30'		SCALE		COLLEGE OF ENGINEERING
SURFACE: 125		MATERIAL		BOSTON, MASS. 02115
FINISH: V				
BREAK ALL SHARP EDGES				
AND FILL				
PLATE				
APPLICATION				

CIMS MANUAL
CONTROL
DISPLAY
CIRCUITS

FIGURE 9.

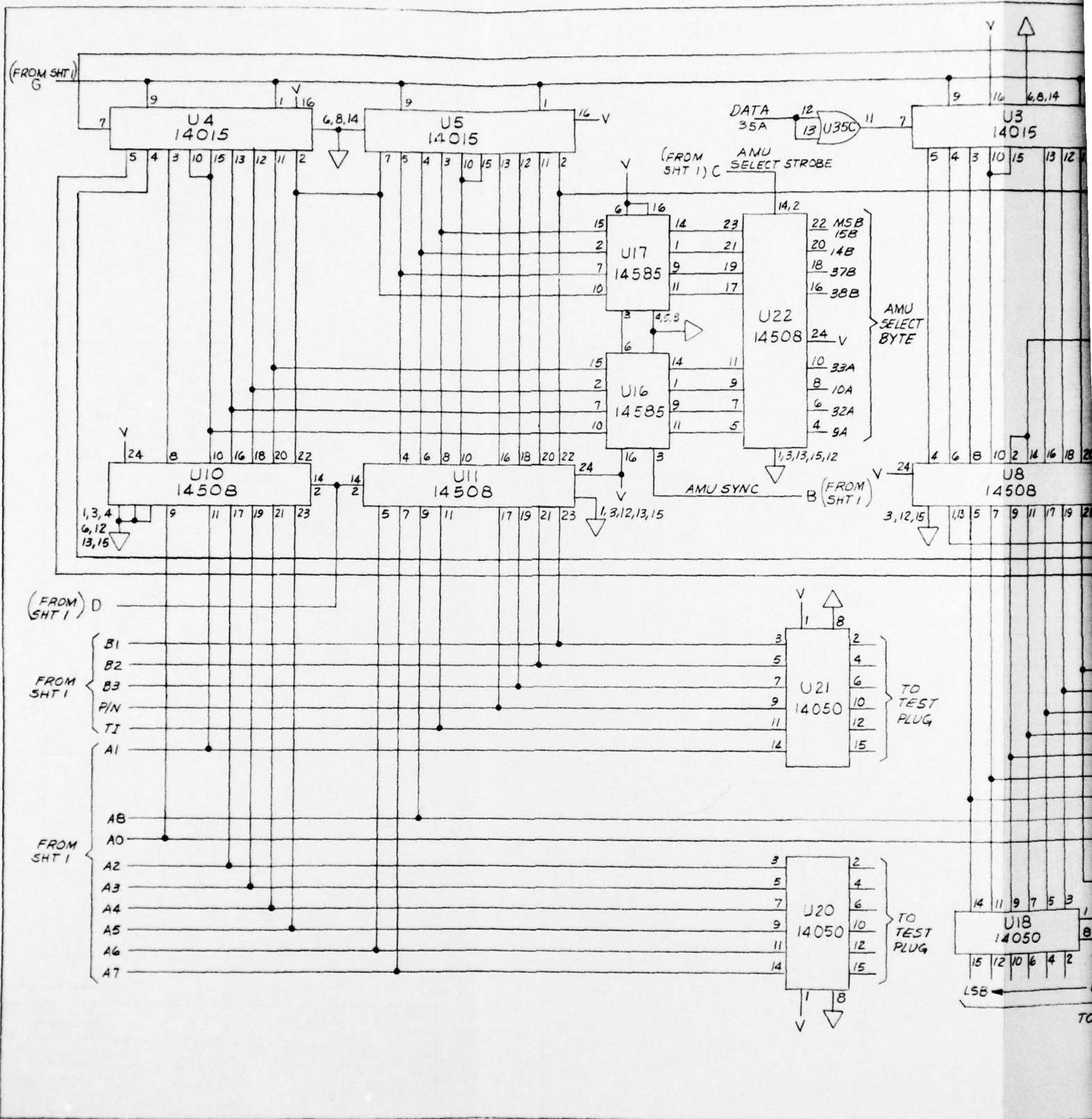


FIGURE 10.

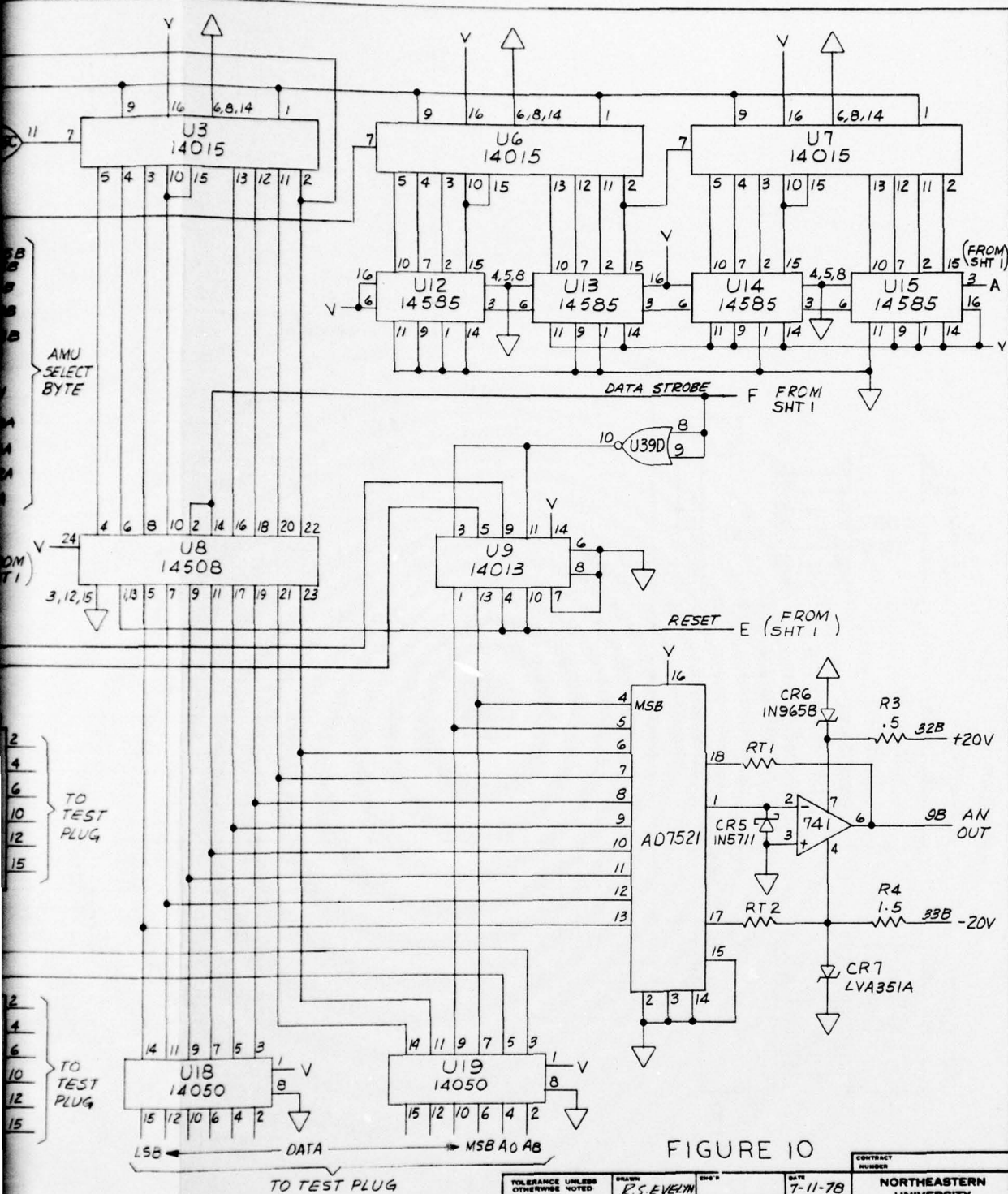
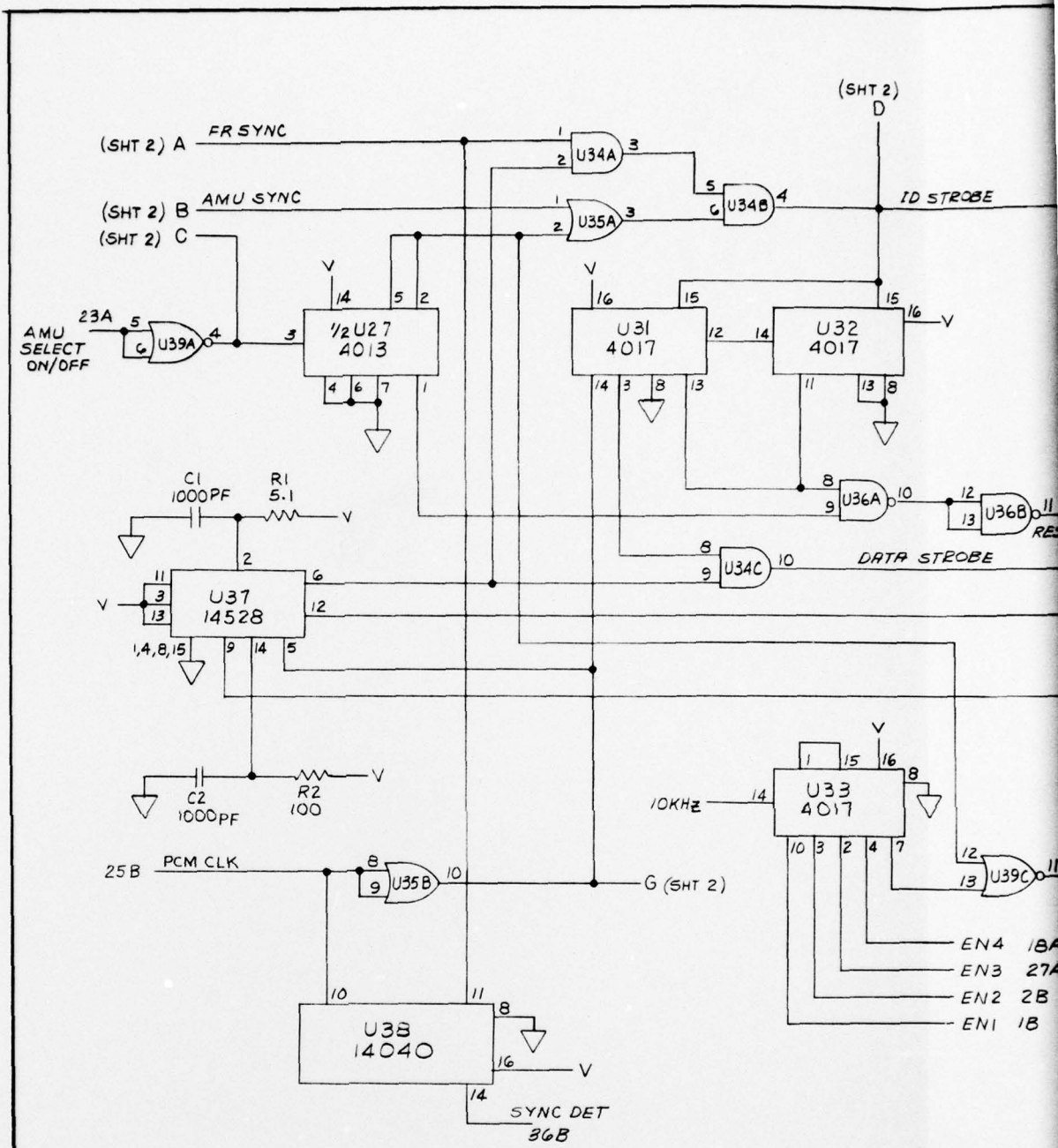


FIGURE 10

TOLERANCE UNLESS OTHERWISE NOTED		DRAWN	DATE	CONTRACT NUMBER
DECIMAL: 01		<i>P.S. EVERLIN</i>	7-11-78	NORTHEASTERN UNIVERSITY
FRACTIONAL: 2/160	CHECKED			ELECTRONIC RESEARCH LAB
ANGULAR: 2 OF 30	SCALE			COLLEGE OF ENGINEERING
SURFACE FINISH: 125	MATERIAL			BOSTON, MASS. 02115
HEAT: ALL SHARP EDGES AND SQUARE				
KEY: 1-15	PROJECT			
APPLICATION				

CIMS MANUAL
CONTROL UNIT
DATA CIRCUITS
-1



NOTES:

- U34 IS MC14091
- U35 IS MC14071
- U36 IS MC14011
- U39 IS MC14001

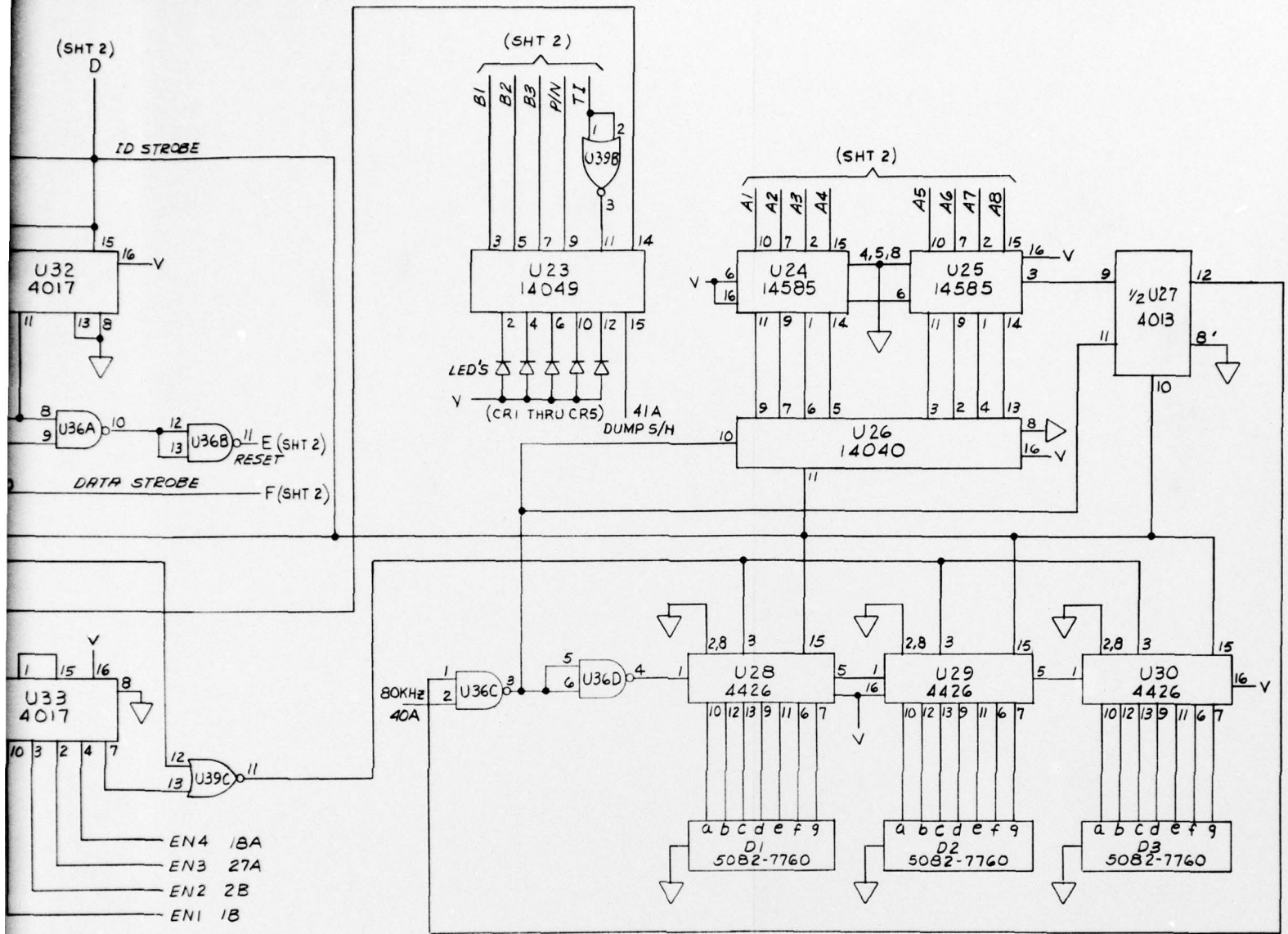


FIGURE 11

TOLERANCE UNLESS OTHERWISE NOTED		DRAWN	DATE	CONTRACT NUMBER
DECIMAL: .01		<i>S. EVELYN</i>	7-11-78	NORTHEASTERN UNIVERSITY ELECTRONIC RESEARCH LAB COLLEGE OF ENGINEERING BOSTON, MASS. 02115
FRACTIONAL: 1/64		CHECKED		
ANGULAR: ± 0° 30'		SCALE		
SURFACE FINISH: 125		MATERIAL		
BRIEF ALL SHARP EDGES AND CORNERS				
PROJECT	DATE	CIMS MANUAL CONTROL UNIT DATA CIRCUITS		
APPLICATION		-2		

FIGURE 11.

2