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# OBJECTIVE MEASUREMENT OF VOICE CHANNEL INTELLIGIBILITY

K. J. Gamauf W. J. Hartman

U.S. Department of Commerce Office of Telecommunications Institute for Telecommunication Sciences Boulder, Colorado 80302



October 1977 Final Report



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AD NO.

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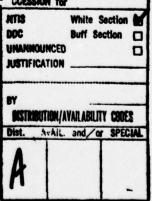
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#### OBJECTIVE MEASUREMENT OF VOICE CHANNEL INTELLIGIBILITY

K.J. Gamauf W.J. Hartman\*

Following the results of a feasibility study (Hartman and Boll, 1976) an objective intelligibility measure is developed using a large data base consisting of 8-50 word phonetically balanced word groups with twelve different kinds of distortion. Justification for the use of this particular measure is included, with mathematical derivations and physical interpretations.

A discussion of the feasibility of a hardware implementation of the software developed here is also included.

Key words: intelligibility measurements; linear predictive coding; voice systems.

#### 1. INTRODUCTION

There has long been a need for an inexpensive, reliable, and efficient method to evaluate the quality of speech sent over voice communication channels. Few voice communication systems today are judged by the quality and intelligibility of the speech received by the listener. Instead, system performance is generally specified by some engineering parameter, such as the signal-to-noise ratio of the receiver output.

The most common procedure for determining the intelligibility of a voice channel is a subjective method that involves trained speakers and listener panels that directly score the percentage of speech that is intelligible. These schemes have the desirable property that they produce repeatable results. Unfortunately, subjective scoring methods are expensive and time consuming and as a result, are not widely used. What is needed is an

\*The authors are with the Institute for Telecommunication Sciences, Office of Telecommunications, U.S. Department of Commerce, Boulder, CO 80302. inexpensive and efficiently applied objective evaluation of speech intelligibility that is comparable to subjective methods.

This paper develops a method of obtaining an objective intelligibility measure that gives good results for speech sent through both analog and digital, noise-corrupted communication channels. The distortion measure is obtained using Linear Predictive Coding (LPC), a mathematical technique widely known for its application to the analysis and synthesis of speech. The feasibility of using LPC to develop an objective intelligibility measure has been demonstrated by Hartman and Boll (1976).

#### A. The Articulation Index

A well known objective measure that is used for voice communication channel evaluation is the Articulation Index (AI) (ANSI, 1969; Kryter, 1962). The AI is a physical measurement that is highly correlated to speech intelligibility under certain conditions. The AI is obtained by evaluating the signal-to-noise power ratios in 20 specific frequency bands. These power ratios are then summed and normalized to give a score between zero and one. An AI equal to one signifies perfect intelligibility, while a value of zero represents a complete lack of intelligibility. The AI can be computed using a general purpose computer once the required spectral data are obtained from the speech.

An automated technique to obtain the AI is achieved through the Speech Communication Index Meter (SCIM) (Kryter and Ball, 1964), which uses 9 frequency bands, instead of 20, to obtain a modified AI. The SCIM system can be used to perform on-line measurements and has been found to be highly correlated with the standard 20 band AI. The SCIM system's AI can also be directly related to speech intelligibility as long as the noise present is generally additive white gaussion noise. Clipping of the speech by the voice channel or noise that is intermittent or colored distorts the AI, and a correction factor must be employed. Multiplicative noise requires a complete recalibration of the SCIM system. Therefore, the type of noise or distortion present

in the voice channel must be known in order to obtain accurate results. When digital voice systems were tested on the SCIM scheme, very poor estimates of speech intelligibility were obtained. This was true even when quantization noise was the only distortion present. Reliable correction factors for digital voice systems have as yet not been found to compensate for this poor performance. A more detailed description of the AI and the SCIM system of objective voice channel evaluation can be found in the work of Hubbard and Hartman (1974).

#### B. Chapter Summaries

Chapter 2 gives a brief discussion of linear prediction of the speech waveform. Chapter 3 describes the analog voice tapes used in this study. Chapter 4 describes the data processing of the analog voice tapes in order to obtain the LPC information from the speech. Chapter 5 discusses the distance or distortion measures that were used to predict objectively the intelligibility of noise corrupted speech. A comparison between objective distance measures and subjective intelligibility scores is given. Chapter 6 gives a block diagram for hardware implementation of the scoring techniques. Appendix A contains listings of the computer programs used to obtain the numerical results found in this report.

#### 2. LINEAR PREDICTIVE CODING

Linear predictive coding (LPC) has long been used in communication theory. More recently, it has found applications in speech analysis and synthesis, speaker identification, and word recognition, to name just a few new areas. In this study, LPC is used to develop an objective intelligibility measure of speech corrupted by noise.

#### A. Linear Prediction

LPC models the vocal tract as an all-pole digital filter and estimates the filter parameters (predictor coefficients) using

the time domain speech waveform itself, rather than the waveform's short-term frequency spectrum. This makes LPC a relatively efficient method for encoding speech compared to frequency domain techniques.

The vocal tract is assumed to be modeled as a discrete, time-varying filter with parameters changing slowly enough so that they can be considered fixed over a specified time interval. Hence, the vocal tract can be approximated by a series of stationary shapes. Atal and Hanauer (1971) have shown that this all pole model can account for the glottal volume flow and radiation of sound from the mouth in addition to vocal tract sounds.

The transfer function, H(z), used to describe the digital model over each analysis frame is given by

$$H(z) = \frac{1}{1 - \sum_{i=1}^{P} a_{i} z^{-i}}$$
 (2.1)

for a model with P poles.

The time sequence  $S_n$  corresponding to the output of the recursive filter can be written as

$$S_n = \sum_{i=1}^{P} a_i S_{n-i} + \delta_n \qquad n = 0, 1, ...$$
 (2.2)

where  $(a_i)$  are the predictor coefficients that completely describe the characteristics of the filter and  $(\delta_n)$  is the driving function or input to the filter.

While there have been several formulations for the estimation of the linear prediction coefficients, two least squares methods have become prominent, the autocorrelation method and the covariance method. The autocorrelation method, which will be justified in Chapter 4, was chosen for use in this study and further discussion will be restricted to that scheme. The autocorrelation method can be considered as estimating the filter coefficients by approximating the spectrum of the speech waveform by an all-pole model.

The portion of the signal to be analyzed is first multiplied by a finite window of length N, changing the signal to

$$s_n = \begin{cases} \text{Windowed speech samples, } o \le n \le N \\ o, n \le o \text{ and } n > N \end{cases}$$
 (2.3)

Using this windowed signal, the prediction error sequence is defined as

$$e_n = s_n - \sum_{i=1}^{p} a_i s_{n-i}$$
  $n = 0, 1, ...$  (2.4)

and the total squared error is then

$$E = \sum_{n} e_{n}^{2} = S_{0}^{2} + \sum_{n=1}^{N-1+P} (S_{n} - \sum_{i=1}^{P} a_{i} S_{n-i})^{2}.$$
 (2.5)

The predictor coefficients are selected so as to minimize the total squared error. This is accomplished by setting the partial derivative of the total squared error with respect to each predictor coefficient equal to zero. The system of equations that results is

$$\sum_{k=1}^{P} r_{|i-k|} a_{k} = r_{i} \qquad i = 1, 2, ..., P \qquad (2.6)$$

where

Ro

$$r_{i} = \frac{1}{R_{o}} \sum_{n=0}^{N-1-|i|} S_{n} S_{n+|i|}$$
 (= 0, 1, 2, . . . , P (2.7)

are the normalized short-term autocorrelation values of the speech signal and

$$= \sum_{n=0}^{N-1} s_n^2$$
(2.8)

is the normalization factor for these values. The normalized total squared error can be defined by making use of equations (2.5) and (2.6), yielding

$$E_{\min} = 1 - \sum_{i=1}^{P} a_i r_i$$
 (2.9)

The predictor coefficients are obtained by inverting a positive definite Toeplitz matrix

$$\begin{bmatrix} r \\ i-k \end{bmatrix}_{i, k = 1, 2, ..., P}$$
 (2.10)

This system of equations can be solved by using Levinson's recursion method, which will be expounded upon further in Chapter 4. The Toeplitz matrix is sometimes called the auto-correlation matrix, and the coefficients obtained from this linear system result in a recursive filter, H(z), which is guaranteed to be stable, (all of its poles lie inside the unit circle), as shown by Grenander and Szego (1958).

#### B. Spectral Approximation

Further insight can be gained by looking at the frequency domain approximation to the above system. Taking the z-transform of equation (2.4), one obtains

$$E(z) = S(z) [H(z)]^{-1}$$
 (2.11)

where H(z) is defined in equation (2.1) and E(z) and S(z) are the z-transforms of  $E_n$  and  $S_n$  respectively. Rearranging, equation (2.11) can be written as

$$S(z) = E(z) H(z).$$
 (2.12)

Minimizing the total squared prediction error is equivalent to approximating the error sequence,  $(e_n)$ , by

$$\hat{e}_n = \begin{cases} A, n = 0 \\ 0, n \neq 0 \end{cases}$$
(2.13)

in a least squares sense. This implies that E(z) is being approximated by the function A, a constant, and S(z) is being approximated by a spectrum corresponding to an all-pole transfer function, i.e.,

$$\hat{\mathbf{E}}(\mathbf{z}) = \mathbf{A} \tag{2.14}$$

$$\hat{S}(z) = \hat{E}(z) H(z) = \frac{A}{1 - \sum_{i=1}^{p} a_i z^{-i}}$$
 (2.15)

The value of A is determined by the application of energy conservation between  $\hat{e}_n$  and  $e_n$ . Using equations (2.9) and (2.13), one obtains

$$A^2 = E_{\min} = 1 - \sum_{i=1}^{P} a_i r_i$$
 (2.16)

thereby showing that  $A^2$  is equal to the minimum total squared error of the system.

This approach of estimating filter coefficients so as to minimize the energy of the output of the inverse of a system driver by its impulse response is sometimes called deconvolution or inverse filtering. Considerable work has been done in this area of linear prediction of speech in the past few years and many good references are available that give more detailed discussions of this subject. Some particularly good ones are Markel and Gray (1976 and 1973); Makhoul (1975 and 1973); Makhoul and Wolf (1972); and Boll (1973).

#### 3. THE VOICE TAPES

In order to develop an objective intelligibility measure for corrupted speech, a comparison must be performed between the distorted speech and the original noise free speech. A subjective intelligibility measure of the distorted speech must also be available in order to judge the quality of the objective measure being used. Both of these requirements are met by first making a noise free master tape of pre-selected speech and then sending it through voice communication channels to be tested and making a recording of the speech at the channel output. This recording can be subjectively scored for intelligibility and also compared to the original speech by some mathematical technique to obtain an objective measure.

#### A. Description of Voice Tapes

The pre-selected speech to be sent over a voice channel for intelligibility scoring are phonetically balanced (PB) groups of isolated words as opposed to complete sentences or nonsense syllables. These PB words were used because subjective scores have been shown to be repeatable, which is a necessary criterion for this study because the objective measure will be repeatable. Eight PB word groups, each containing fifty isolated words were selected as the test speech. A list of the fifty words in each word group is given in Tables 3-1 through 3-8, with their designated word group numbers.

An analog tape containing all eight word groups and using both male and female trained speakers was obtained from the Army Electronic Proving Ground Electromagnetic Environment Test Facility at Fort Huachuca, Arizona. From this tape, a master analog tape was made that would be sent over voice channels and later compared with the recorded output of the channel. In order to perform this comparison, the two tapes would have to be aligned, which meant synchronization information must be included on the master tape before being sent across the voice channel. Because

Table 3-1 PB Word Group 361

1.40

1.	STAB	26.	RUG
2.	TUCK	27.	CLIFF
3.	DRAPE	28.	LOUSE
4.	PITCH	29.	GAB
5.	INK	30.	RYE
6.	AID	31.	SANG
7.	KIND	32.	CLOSED
8.	STRESS	33.	THREE
9.	TURN	34.	MAP
10.	DROOP	35.	GAS
11.	PUMP	36.	SHEEP
12.	SUIT	37.	CREWS
13.	BARGE	38.	THRESH
14.	KNEE	39.	NAP
15.	DUB	40.	HAD
16.	WIELD	41.	SHEIK
17.	ROCK	42.	TIRE
18.	BOOK	43.	DAME
19.	THOU	44.	NEXT
20.	LAY	45.	HASH
21.	FIFTH	46.	SOAR
22.	ROGUE	47.	TON
23.	CHEESE	48.	DIN
24.	LEASH	49.	PART
25.	FRIGHT	50.	HOSE

# Table 3-2 PB Word Group 312

1.	JAB	26.	DIP
2.	ARC	27.	URGE
3.	JAUNT	28.	MOUTH
4.	ARM	29.	NET
5.	SHOP	30.	WAVE
6.	BEAM	31.	FINE
7.	KIT	32.	PURSE
8.	BLISS	33.	GOAT
9.	SPRIG	34.	HOG
10.	LAG	35.	RISK
11.	CHUNK	36.	DOUBT
12.	LATCH	37.	PUNK
13.	CODE	38.	DRAKE
14.	LOW	39.	WOOD
15.	TAB	40.	FEEL
16.	SHOT	41.	PROD
17.	SIGN	42.	FRISK
18.	CRUTCH	43.	DULL
19.	SAP	44.	MOST
20.	LOSS	45.	FUDGE
21.	CLASH	46.	POND
22.	SNOW	47.	HAVE
23.	CRY	48.	REEF
24.	SPY	49.	PROBE
25.	STIFF	50.	RICE

15

5.5

Table 3-3 PB Word Group 291

1.	ARCH		26.	NUTS
2.	BEEF		27.	WIPE
3.	KEY		28.	ODD
4.	SIP		29.	WITH
5.	BIRTH		30.	FLAG
6.	SMART		31.	NERVE
7.	SPUD		32.	FLUFF
8.	CLUB		33.	FOE
9.	TEN		34.	NOOSE
10.	CROWD		35.	FUME
11.	THAN	1 int	36.	WEAK
12.	BIT	1.172	37.	FUSE
13.	THANK	No.	38.	WILD
14.	CUD		39.	GIVE
15.	THRONE		40.	PHONE
16.	CARVE	in	41.	GATE
17.	TOAD		42.	HOOF
18.	LIT	de la	43.	YEAR
19.	CHESS		44.	ICE
20.	TROOP		45.	REED
21.	CHEST		46.	ITCH
22.	BOOST		47.	ROOT
23.	CLOWN		48.	GRACE
24.	DITCH	1.16	49.	PACT
25.	MASS		50.	RUDE

## Table 3-4 PB Word Group 265

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1.	AS	26.	CLOTH
2.	BEST	27.	GROPE
3.	EAT	28.	KEPT
4.		29.	RAY
5.	EYES	30.	FORGE
6.	SCAN	31.	CLOTHES
7.	COB	32.	ROOMS
8.	FALL	33.	LAG
9.	DAD	34.	THIGH
10.	ODE	35.	WAIT
11.	SHANK	36.	WIFE
12.	MASH	37.	JAG
13.	HITCH	38.	NIGH
14.	ROUGH	39.	CRIB
15.	FEE	40.	PRIG
16.	CHART	41.	FLOP
17.	WASP	42.	SUP
18.	HULL	43.	GAGE
19.	TONGUE	44.	WRIT
20.	PUN	45.	PRIME
21.	REAP	46.	FOWL
22.	PUS	47.	BOG
23.	BADGE	48.	GAP
24.	DEEP	49.	FLICK
25.	SLOUCH	50.	RAISE

Table 3-5 PB Word Group 275

1.	AM			26.	SLEDGE
2.	GRADE			27.	RANGE
3.	GASP			28.	WOO
4.	MOTE			29.	DOPE
5.	MUD			30.	FLING
6.	BY			31.	NINE
7.	PHASE			32.	SCOUT
8.	RASH			33.	OFF
9.	RICH			34.	PIG
10.	POUNCE			35.	FORT
11.	SHAFT			36.	WOE
12.	ROAR			37.	СНОР
13.	ACT			38.	PLOD
14.	AIM			39.	KNIT
15.	HIM			40.	WHIFF
16.	COAST	•		41.	PENT
17.	DOSE			42.	THOUGH
18.	BUT			43.	JUG
19.	SOUTH			44.	SNIFF
20.	SIEGE			45.	QUIZ
21.	DWARF			46.	GUN
22.	FAKE			47.	COOK
23.	CUT		1	48.	SAG
24.	COMES			49.	WIRE
25.	SIN			50.	RAID

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	Tab	Le 3-6	
PB	Word	Group	305

1.	STAFF		26.	TREE
2.	BASH		27.	GOOSE
3.	HAT		28.	PAGE
4.	WADE		29.	MAZE
5.	CHAMP		30.	FLIGHT
6.	ETCH		31.	PINK
7.	SLUG		32.	BUG
8.	CHANCE		33.	RAPE
9.	WAKE		34.	EARS
10.	VALVE		35.	SCRUB
11.	YOUTH		36.	COW
12.	FLAUNT		37.	TAG
13.	RUSH		38.	JAY
14.	GULL		39.	VOID
15.	DAUB		40.	EARTH
16.	REAL		41.	THOSE
17.	AIL		42.	LAP
18.	PUT		43.	SNIPE
19.	NUDGE		44.	FIR
20.	BACK		45.	CLOTHE
21.	PLUS ·		46.	MOPE
22.	BOB		47.	CORD
23.	THUG		48.	RIP
24.	CUE		49.	HURT
25.	LINE		50.	FORCE

# Table 3-7 PB Word Group 214

1.	TOE		26.	HID
2.	ARE		27.	SUCH
3.	RUB		28.	CRASH
4.	GROVE	1	29.	BOX
5.	PANTS		30.	THERE
6.	DEATH		31.	END
7.	BAD		32.	MANGE
8.	PAN		33.	PLUSH
9.	USE		34.	IS
10.	SLIP		35.	FORD
11.	BASK		36.	HUNT
12.	FRAUD		37.	RAG
13.	NOT		38.	FEAST
14.	DEED		39.	NO
15.	SMILE		40.	CLOVE
16.	DISH		41.	FERN
17.	RISE		42.	PILE
18.	FUSS		43.	STRIFE
19.	WHEAT		44.	CANE
20.	DIKE		45.	FOLK
21.	PEST		46.	RAT
22.	CREED		47.	CLEANSE
23.	HEAP		48.	THEN
24.	BAR		49.	RIDE
25.	NOOK		50.	HIVE

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And the second second

## Table 3-8 PB Word Group 283

1.	US	26.	BIND
2.	SHACK	27.	CHEW
3.	CRACK	28.	WHEEZE
4.	CHANT	29.	FREAK
5.	YEAST	30.	PINT
6.	ASK	31.	GUESS
7.	EASE	32.	QUEEN
8.	REST	33.	CLOD
9.	JELL	34.	LOOK
10.	BOLT	35.	FRONT
11.	KILL	36.	NIGHT
12.	LICK	37.	WIG
13.	CALF	38.	ROPE
14.	САТСН	39.	DAY
15.	TILL	40.	RHYME
16.	EACH	41.	SLIDE
17.	ROT	42.	FROCK
18.	ROLL	43.	LEFT
19.	BID	44.	FOOD
20.	COD	45.	SPICE
21.	DEUCE	46.	BORED
22.	DUMB	47.	THIS
23.	FAD	48.	THREAD
24.	HUM	49.	FORTH
25.	ROD	50.	FLIP

the tapes would be processed in a digital state, the alignment procedure would also have to work in a digital format. It was found that a shift of plus or minus 10 samples of a 256 sample analysis window caused the predictor coefficients to vary less than 0.1% in all cases. Therefore, the synchronization procedure to be used was required to align two segments of digitized speech to within 10 samples.

It should be noted that the bound on the variation of the predictor coefficients cannot be translated into a bound on the distortion measures described in Chapter 5. However, the alignment method described in the next section was tested extensively, and never produced a variation in the distortion measures larger than that produced by the normal round off error.

#### B. Voice Tape Alignment

A synchronization procedure that was found to meet the required 10 sample variation specification, made use of a binary pseudo noise (PN) sequence. The binary PN sequence was sent through a phase-continuous frequency shift keying modem using the two frequencies 1.2 kHz and 2.2 kHz. Several different length binary PN signals and modem bit rates were tested to determine the best combination for alignment capability. The test consisted of cross-correlating the PN sequences under different noise and distortion conditions and looking for an impulse like correlation function. A further requirement was to have the PN sequence as short as possible. It was found that a length 127 binary PN sequence sent through the FSK modem operating at a 635 Hz bit rate followed by a low pass filter with a cutoff of 2.5 kHz met all the requirements necessary to insure the alignment of two PN sequences distorted by noise. The low pass filter was used to make certain that the frequency spectrum of the PN signal was in the range required for input to most voice systems. A PN signal was then placed before each word and after the last word of all eight word groups thereby creating the master analog tape with alignment capabilities.

In order to align a distorted tape with the master tape, the location of all the PN sequences and words on the digitized master tape had to be known. This was done by blocking the quantized samples into 125 sample records and computing the mean and standard deviation (SD) for each record. The SD was used as an energy criterion to determine the midpoints of the PN sequences and words and the length of each word. The distances between the midpoints of each PN sequence and the word following it were then determined. The corresponding midpoints of the words from a distorted tape are now all that remains to be found.

Each of the eight word groups on a digitized tape made up one file and corresponding files between the master and distorted tape were aligned independently from the other seven sets of files. Using the SD energy criterion, the midpoints of the first and last PN sequences of the distorted word groups are estimated. The cross-correlation between these PN sequences and the corresponding ones from the master tape are then computed, thereby obtaining the midpoints of the two distorted PN sequences with respect to the master word group's PN sequences. From this computation, the slight drift between the samples of the two tapes can be calculated. Using this drift and the PN sequence . midpoints of the master tape, an estimate of the midpoints of the PN sequences of the distorted tape can be made taking into account the shift between the two tapes. The true midpoints of the PN sequences of the distorted tape with respect to the master tape can now be found by again computing the cross-correlation between each pair of PN sequences. Using the distances between the midpoints of the PN sequences and the words following them of the master word group, and the drift between the two tapes, the midpoints of the words of the distorted word group with respect to the master can be obtained using the midpoints of the PN sequences of the distorted word group. This procedure is repeated for all eight word groups of each distorted tape.

The alignment of two words from two different tapes to within 10 samples was the goal of the synchronization procedure.

This can safely be assumed using the above alignment scheme. The PN sequences on either side of each word are lined up to within one sample in all cases. The drift between two consecutive PN sequences was never more than 15 samples, usually quite a bit less. Since the word to be aligned is roughly midway between the two PN sequences and the drift is taken into account, the 10 sample synchronization specification is always met, generally to within a sample or two. The drift between two tapes was verified to be linear, with only small (+1 sample) fluctuations.

Three computer programs were used in the synchronization procedure discussed above. "Words" was the program that was used to find the locations of the PN sequences and words through the SD energy criterion. FFTCOR4 computed the cross-correlation between the PN sequences of the two tapes. Finally, WRDMIDP calculated the midpoints of the words of the distorted tape with respect to the master tape. A listing of all three programs can be found in Appendix A.

#### 4. DATA PROCESSING OF THE VOICE TAPES

Once the master analog tape of eight 50-word groups was made, it could then be sent over various voice communication channels to obtain distorted tapes. Copies of the distorted tapes were sent to Fort Huachuca to be scored for intelligibility. The intelligibility score for a single word was the percentage of the listener panel that correctly identified it, and the intelligibility score for the entire word group was the average score of all 50 words in the word group. The subjective intelligibility scores were used later for comparison with the objective intelligibility measure. The distorted tapes and the master tape were then processed to obtain the LPC information necessary to develop the objective distance measure. No filtering was used on the tapes used for subjective scoring.

#### A. Digitization of the Voice Tapes

Before the master or distorted tapes were digitized, they were first sent through a pre-emphasis filter, and then lowpass filtered to 3.2 kHz. Pre-emphasis was used because it enhances the high-frequency formants of the speech which is important for speech comparison. Pre-emphasis also limits the effects of the glottal waveform and lip radiation and therefore enhances the spectral properties of the speech due to the vocal tract.

Based upon an average vocal tract length of 17 cm, the first three formant frequences will be found in the frequency range of about 250 - 2800 Hz. Shorter vocal tracts will shift this range up slightly. Low-pass filtering at 3.2 kHz would therefore pass the first three formant frequencies. This (3.2 kHz) is generally also the high frequency cut-off for most voice communication channels. Any noise above 3.2 kHz picked up by the distorted tapes will also be filtered out which will help the accuracy of the objective distance measure.

The analog tapes were sampled at 10 kHz and then quantized to 12 bits. The sampled signal was then stored on digital magnetic tape for future processing.

#### B. Analysis Conditions

Once the tapes are digitized and the distorted tapes are aligned to the master tape, LPC processing of each word can be done. First, however, the decision must be made regarding which least squares method to be used to obtain the predictor coefficients. For this study, the autocorrelation method of linear prediction was chosen over the covariance method because it requires fewer calculations, it is assured of producing a stable filter, (i.e. all the poles are within the unit circle), and it allows a meaningful spectral matching term to be computed. Also, as mentioned before, the autocorrelation method can be considered as estimating the predictor coefficients by approximating the spectrum of the speech waveform by an all-pole filter or model. A detailed comparision between the two methods is given in Makhoul and Wolf (1972).

The analysis interval in which the speech waveform's spectrum is estimated should be short enough so that vocal tract movement is negligible, but long enough to insure stable spectral estimates. The vocal tract can, in general, be assumed to be stationary on the order of 15 to 20 ms. Since in the autocorrelation method of linear prediction the approximation is to model a short-term signal spectrum, it is necessary to window in order to guarantee spectrally accurate results. By using a nonrectangular window, a larger analysis interval can be used without sacrificing spectral accuracy. Therefore, the length of the analysis frame was chosen to contain 256 samples, which means an analysis interval of 25.6 ms because of the 10 kHz sampling The non-rectangular window used on each analysis frame was rate. a Hamming window of the form

$$W_{\rm H}(129-n) = W_{\rm H}(128+n) = \frac{1}{1.08 \cdot N} \cdot (5.04+0.46 \cos \frac{2\pi n}{N}) n = 1, 2, ..., 128$$
(4.1)

where N=256, the analysis frame length. The Hamming window was chosen because of its desirable spectral properties and its widespread use in linear prediction literature.

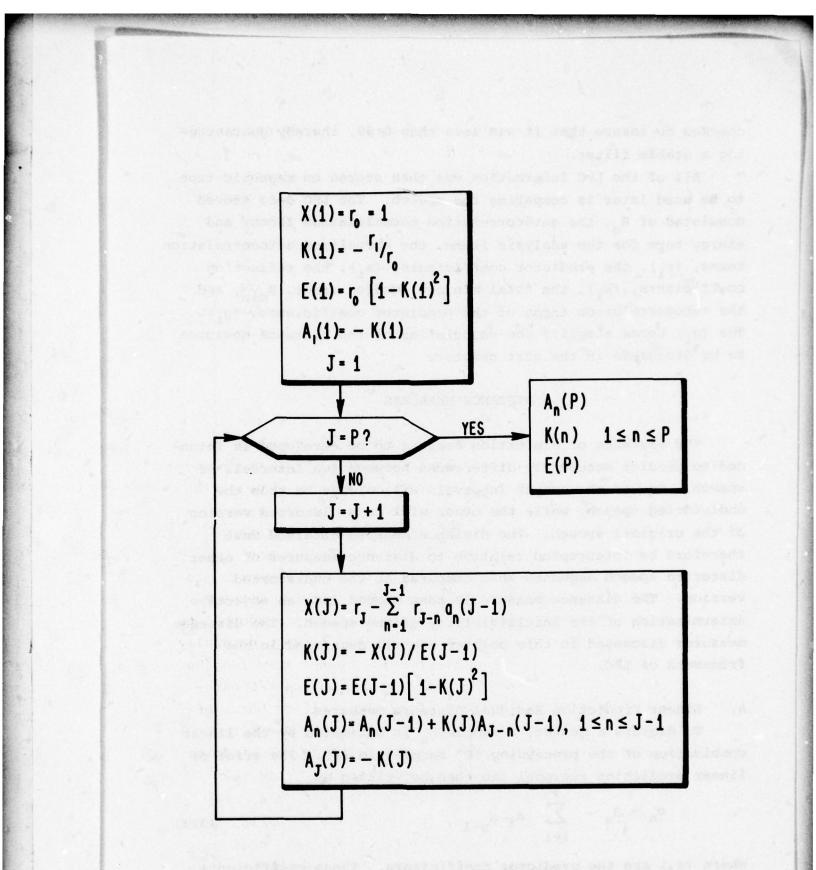
Another important consideration was the number of predictor coefficients to be used. As a practical matter, it is best to choose the number of coefficients as small as possible because it saves computation time and there is less chance of filter instability due to finite arithmetic effects. It has been shown that to represent adequately the vocal tract under ideal circumstances, the memory of the model or filter must be equal to twice the time required for sound waves to travel from the glottis to the lips, i.e., M = 2L/C, where L is the length of the vocal tract and C is the speed of sound. Using the average vocal tract length of L = 17 cm and the speed of sound, C = 34 cm/ms, the memory required is 1 ms with a 10 kHz sampling rate, the number of predictor coefficients needed is equal to the sampling rate times the memory required, or 10 coefficients. In order to take into account the influences of the glottal waveform flow and lip radiation characteristics an additional two coefficients are necessary. Hence, twelve predictor coefficients were computed for each analysis frame thereby producing a twelve pole filter that accurately models the spectral properties of each speech interval.

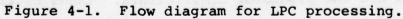
The information necessary to choose the various parameters and methods for LPC processing were obtained from experimental results and several references, which include Markel and Gray, (1973) and Makhoul (1973), and Boll (1973).

#### C. LPC Data

With all the necessary parameters and methods chosen to analyze the speech, LPC processing of each word could then be undertaken. Computer program LPC was used to perform the speech analysis. A listing of the program can be found in Appendix A. The 256 point window was moved along each word at 256 sample shifts. Originally, the window was shifted by 128 points to create overlapping analysis frames. However, due to the averaging, significant differences in the distance measures, to be described in Chapter 5, were not found when the two methods were compared. Consequently, the 256 point shift was adopted to save processing time. The distorted word lengths were naturally defined to be the same as the corresponding master word lengths.

Processing of the windowed signal to obtain the LPC parameters was done using the Levinson Algorithm. A flow chart of the algorithm is shown in Figure 4-1. In order to start the algorithm, the normalized autocorrelation terms of the windowed signal had to be obtained. This was accomplished using the direct method of equation (2.7) as opposed to using the Fast Fourier Transform (FFT) because of the small number of lag terms needed (thirteen terms). The Levinson Algorithm was used to obtain the twelve predictor coefficients,  $(a_i)$ , and the minimum total squared error,  $E_{min}$ . Each reflection coefficient  $(k_i)$  was





checked to insure that it was less than 0.99, thereby guaranteeing a stable filter.

All of the LPC information was then stored on magnetic tape to be used later in comparing the speech. The LPC data stored consisted of  $R_0$ , the autocorrelation normalization factor and energy term for the analysis frame, the normalized autocorrelation terms,  $(r_i)$ , the predictor coefficients,  $(a_i)$ , the reflection coefficients,  $(k_i)$ , the total minimum squared error,  $E_{min}$ , and the autocorrelation terms of the predictor coefficients,  $(g_i)$ . The  $(g_i)$  terms simplify the calculation of the distance measures to be discussed in the next chapter.

#### 5. DISTANCE MEASURES

The distance or distortion measure to be developed is intended to predict accurately differences between two intervals of speech. One of the speech intervals will always be from the undistorted speech, while the other will be a distorted version of the original speech. The distance measure obtained must therefore be interrupted relative to distance measures of other distorted speech segments when compared to the undistorted version. The distance measure is then mapped into an objective determination of the intelligibility of the speech. The distance measures discussed in this chapter are all developed in the framework of LPC.

#### A. Linear Prediction Residual Distance Measures

To digress a moment, a sample  $S_n$  is estimated by the linear combination of the preceeding "P" samples in LPC. The error or linear prediction residual can then be written as

$$e_n = s_n - \sum_{i=1}^{P} a_i s_{n-1}$$
, (5.1)

where (a<sub>i</sub>) are the predictor coefficients. These coefficients are obtained by choosing them so as to minimize the total squared error. The total squared error or linear prediction residual energy can be considered to be the output of an inverse filter  $H(z)^{1}$ , where

$$[H(z)]^{-1} = 1 + \sum_{i=1}^{P} a_{i} z^{-1} .$$
 (5.2)

 $[H(z)]^{1}$  is the filter that minimizes the residual energy and H(z) corresponds to a smoothed spectral estimate of the data sequence  $(S_{n})$  up to a scale factor representing the gain.

If  $(S_n)$  is passed through a different inverse filter,  $[H'(z)]^{-1}$ , of the form

$$[H'(z)]^{-1} = 1 + \sum_{i=1}^{P} a'_{i} z^{-1}$$
 (5.3)

which minimizes the residual energy for some other data sequence  $(S_n')$ , the residual energy D, must be greater than or equal to the minimum residual energy E, i.e.,  $D \ge E$ , with the equality holding if and only if H(z) = H'(z). Assuming the data sequence  $(S_n)$  is obtained from an analysis frame of speech from the undistorted tape and  $(S_n')$  is the corresponding analysis frame from a distorted tape the difference between D and E is a measure of the distance between the two speech segments. (Unless otherwise identified, unprimed variables represent data from the master or undistorted tape.)

The dual of the above situation is also true. If  $(S_n)$  is sent through  $[H(z)]^1$ , the output will be D', while E' is the output of  $[H'(z)]^1$  when  $(S_n')$  is sent through it. Again,  $D \ge E'$ with equality if and only if H'(z) = H(z). As before, the difference between D' and E' can be considered to be a distance measure between the two speech segments.

E, E', D, and D' can all be written as a combination of the autocorrelation terms of  $(S_n)$  and  $(S_n')$  and the corresponding linear prediction coefficients  $(a_i)$  and  $(a'_i)$ . Let

$$\underline{\mathbf{A}}^{\mathrm{T}} = (1, -\mathbf{a}_{1}, -\mathbf{a}_{2}, \ldots -\mathbf{a}_{n})$$

(5.4)

be the transpose of the linear prediction coefficient vector  $\underline{A}$ , and

$$\frac{R}{2} = r_{[i-k]} \qquad i, k = 0, 1, 2, ..., P \qquad (5.5)$$

the normalized autocorrelation matrix. The four error terms can then be written as

 $\mathbf{E} = \underline{\mathbf{A}}^{\mathrm{T}} \ \underline{\mathbf{R}} \ \underline{\mathbf{A}} \tag{5.6}$ 

 $\mathbf{E}' = \underline{\mathbf{A}}'^{\mathrm{T}} \underline{\mathbf{R}}' \underline{\mathbf{A}}' \tag{5.7}$ 

$$D = \underline{A}^{T} \underline{R} \underline{A}^{T}$$
(5.8)

$$\mathbf{D}' = \underline{\mathbf{A}}^{\mathrm{T}} \underline{\mathbf{R}}' \underline{\mathbf{A}}$$
(5.9)

where the primes signify variables from a distorted tape. The derivation of this can be found in Market and Gray (1973) and Boll (1974).

E and E are calculated for each analysis frame through the Levinson Algorithm, but D and D are calculated when a distorted tape is compared to the master tape. These calculations can be simplified because of the structure of <u>R</u> and <u>R</u>, the autocorrelation matrices, by calculating the autocorrelation terms of <u>A</u> and <u>A</u>, the linear prediction coefficient vectors. Using the symmetry of the autocorrelation terms of <u>A</u>, and <u>A</u>, D and D', can be written as

$$D = \sum_{i=0}^{P} g_{i} r_{i}$$
(5.10)

$$D' = \sum_{i=0}^{n} g_i \dot{r}'_i$$
 (5.11)

where

$$g_{i} = 2 \cdot \sum_{k=0}^{P-|i|} \alpha_{k} \alpha_{k+i}$$
 (i = 1, 2, ..., P (5.12)

$$g_{0} = \sum_{k=0}^{P} \alpha_{k}^{2}$$
 (5.13)

 $g_{i} = 2 \cdot \sum_{k=0}^{P-|i|} \alpha_{k} \alpha_{k+1}$  (i = 1, 2, ..., (5.14)

$$g_{0}' = \sum_{k=0}^{n} \alpha_{k}'^{2}$$
 (5.15)

 $(\alpha_i)$  and  $(\alpha_1)$  are the P+1 terms in the vectors <u>A</u> and <u>A</u> respectively.  $(g_i)$  and  $(g'_i)$  can be calculated for each analysis frame right after the predictor coefficients are obtained and stored on the LPC data tape.

Each of the four error terms can be interpreted in the frequency domain. Using Parseval's Theorem, the total squared error, E, can be written as

$$E = \sum_{n} e_{n}^{2} = \frac{T}{2\pi} \int_{-\pi/T}^{\pi/T} |E(\omega)|^{2} d\omega , \qquad (5.16)$$

where  $E(\omega)$  is obtained by substituting  $z = e^{i\omega t}$  into E(z). From Chapter 2, the minimum linear prediction error was found to be

$$E(z) = S(z) [H(z)]^{-1}$$
 (5.17)

while the least squares estimate can be written as

$$\hat{E}(z) = \hat{S}(z) [H(z)]^{-1}$$
 (5.18)

substituting  $z = e^{i\omega t}$  into (5.17) and (5.18), one obtains

$$E(\omega) = S(\omega) [H(\omega)]^{-1}$$
 (5.19)

$$\widehat{\mathbf{E}}(\omega) = \widehat{\mathbf{S}}(\omega) \left[\mathbf{H}(\omega)\right]^{-1} \qquad (5.20)$$

Rearranging (5.20) and substituting in E(z) = A,

$$[H(\omega)]^{-1} = \frac{A}{\hat{S}(\omega)} \qquad (5.21)$$

Inserting (5.21) into (5.19),

$$E(\omega) = A \frac{S(\omega)}{S(\omega)} .$$
 (5.22)

substituting (5.22) into (5.16)

$$E = \frac{T A^2}{2\pi} \int_{-\pi/T}^{\pi/T} \frac{|S(\omega)|^2}{|S(\omega)|^2} d\omega$$
(5.23)

But  $|S(\omega)|^2$  and  $|\hat{S}(\omega)|^2$  are just the corresponding power spectra,  $P(\omega)$  and  $\hat{P}(\omega)$ , of the speech signal and its least squares linear prediction estimate. Therefore,

$$E = \frac{TA^2}{2\pi} \int_{-\pi/T}^{\pi/T} \frac{P(\omega)}{\hat{P}(\omega)} d\omega \qquad (5.24)$$

Similarly E can be shown to be

$$E' = \frac{TA'^{2}}{2\pi} \int_{-\pi/T}^{\pi/T} \frac{P'(\omega)}{\hat{P}'(\omega)} d\omega .$$
 (5.25)

D and D can also be obtained by the same method and written as

$$D = \frac{TA'^2}{2\pi} \int_{-\pi/T}^{\pi/T} \frac{P(\omega)}{\hat{P}'(\omega)} d\omega \qquad (5.26)$$

$$D' = \frac{TA^2}{2\pi} \int_{-\pi/T}^{\pi/T} \frac{P'(\omega)}{\hat{P}(\omega)} d\omega . \qquad (5.27)$$

The distance measures D and D are not all that pleasing when defined in the frequency domain. They compare the ratio differences between a true speech power spectrum and estimated power spectrum. A much more desirable measure would compare the ratio differences between the estimated power spectra of the undistorted speech and the distorted version of it. This can be done by taking the ratio of D to E and D to E. The ratio of each of these pairs of residual errors, D/E and D'/E', then defines two new distance measures which are much more appropriate. In both cases, the ratios are greater than or equal to one, with equality if and only if H(z) = H'(z).

The ratios D/E and D /E are sometimes called likelihood ratios because under certain circumstances, they have been shown to be true likelihood ratios by Itakura (1975). As mentioned before, the frequency domain interpretation of the likelihood ratios gives a good justification for using them as distance measures. In the time domain

$$\frac{D}{E} = \frac{\sum_{n}^{n} D_{n}^{2}}{\sum_{n}^{n} e_{n}^{2}}, \qquad (5.28)$$

where

$$\sum_{n} D_{n}^{2} = \sum_{n} \left\{ s_{n} - \sum_{i=1}^{p} a_{i}^{*} s_{n-i} \right\}^{2}$$
(5.29)

(5.30)

$$\sum_{n} e_{n}^{2} = \left\{ s_{n} - \sum_{i=1}^{P} a_{i} s_{n-i} \right\}^{2}$$

Gray and Markel (1976) have shown that D/E can be written in the frequency domain as

$$\frac{\mathbf{D}}{\mathbf{E}} = \frac{\mathbf{T}}{2\pi} \int_{-\pi/\mathbf{T}}^{\pi/\mathbf{T}} \frac{\left|\mathbf{H}(\omega)\right|^2}{\left|\mathbf{H}'(\omega)\right|^2} d\omega , \qquad (5.31)$$

where the substitution  $z = e^{j\omega T}$  is made in the filters H(z) and H (z). Inverting (5.21) and its dual, one obtains

$$H(\omega) = \frac{\hat{S}(\omega)}{A}$$
(5.32)

$$H'(\omega) = \frac{\hat{S}'(\omega)}{A}$$
(5.33)

Substituting (5.32) and (5.33) into (5.31)

$$\frac{D}{E} = \frac{T}{2\pi} \int_{-\pi/T}^{\pi/T} \frac{|\hat{S}(\omega)|^2}{A^2} \cdot \frac{A'^2}{|\hat{S}'(\omega)|^2} d\omega.$$
(5.34)

$$\frac{D}{E} = \frac{TA'^2}{2\pi A^2} \int_{-\pi/T}^{\pi/T} \frac{\left|\hat{s}(\omega)\right|^2}{\left|\hat{s}'(\omega)\right|^2} d\omega$$
(5.35)

Once again, the magnitude squared of the signal's spectrum is just its power spectrum, therefore

$$\frac{D}{E} = \frac{TA'^2}{2\pi A^2} \int_{-\pi/T}^{\pi/T} \frac{\hat{P}(\omega)}{\hat{P}'(\omega)} d\omega$$
(5.36)

Similarly

$$\frac{\mathbf{D}}{\mathbf{E}} = \frac{\mathbf{T}\mathbf{A}^{\prime 2}}{2\pi\mathbf{A}^{\prime 2}} \int_{-\pi/\mathbf{T}}^{\pi/\mathbf{T}} \frac{\hat{\mathbf{P}}^{\prime}(\omega)}{\hat{\mathbf{P}}(\omega)} d\omega \qquad (5.37)$$

As can be seen from (5.36) and (5.37), D/E and D/E compute the differences between the estimate power spectra of the undistorted and distorted speech, while D and D compared the true power spectra of the speech to estimates of the power spectra.

D/E and D'/E' will be used as the basic components in the distance measures discussed in B, below. Several methods are used to normalize D/E and D'/E' so that two distorted tapes can be compared relative to each other by comparing them only to the master or undistorted tape. For more information concerning the likelihood ratios above, see Gray and Markel (1976).

One additional measure that is considered here is derived as follows. Let  $S_n + N_n = S_N'$ . Then,

$$\sum_{n} \left[ s_{n} - \sum a_{i} s_{n-i} + s_{n} - \sum a_{i} s_{n-i} \right] = \left[ s_{N}' - \sum a_{i} s_{n-i} \right] \quad (5.38)$$

This can be rewritten as

$$E + D_N + [cross products] = D$$
,

where  $D_N$  is given by  $\underline{A}^T \underline{R}_N \underline{A}$  with the  $\underline{R}_N$  the autocorrelation matrix of the noise. Assuming the signal  $S_N$  and the noise  $N_n$  are uncorrelated, this simplifies to

$$E + D_N = D$$
. (5.39)

Assuming the noise for any frame  $(N_n)$  is the same (statistically) except for a constant factor as the noise for a period during which voice is not present  $(\tilde{N}_n)$ , we have

$$D_{N} = k D'_{\widetilde{N}}$$
(5.40)

where k is a different constant for each frame. Since k must be positive, we have

$$k = \frac{\left| D - E \right|}{\left| D \right|_{N}}$$
(5.41)

A large number of calculations showed that  $E/D_N$  was very small compared to  $D'/D_N$ , and consequently, k was calculated using

$$k = D / D \tilde{N}$$
(5.42)

k was then averaged over all frames of all words to obtain  $\overline{k}$ .

A signal to noise ratio was determined using the peak signals during a PN sequence and the noise N from a quiet period between PN sequences, for each word group. The quantity

$$SNR = 10 \log_{10} S/N - 10 \log_{10} \overline{k}$$
 (5.43)

(5.44)

was then used to calculate an AI score,

$$AI = \begin{cases} 0 & SNR \le 0\\ \frac{SNR}{30} & \text{if } 0 < SNR < 30\\ 1 & SNR > 30 \end{cases}$$

B. An Objective Intelligibility Measure

The quantities D / E and D / E were computed on a frame-byframe basis using the computer program DISTMEA and stored along with other LPC data for use in developing an objective intelligibility measure. (See Appendix A.) The natural logarithms of D / E and D / E are respectively labeled El and E2 and relate directly to a decibel (dB) scale.

Under the assumption that the errors  $e_n$  are independent Gaussian variables, Itakura derives the result that  $N_{eff}$  El is a chi-squared variable with P(=12) degrees of freedom. Here, because of the windowing,  $N_{eff} \simeq 101$ . However, since in general the  $e_n$  are correlated, it is assumed that the actual  $N_{eff}$  is smaller than this (101).

Instead of modifying N<sub>eff</sub>, the procedure given in the next paragraphs was used to modify two thresholds. First, a lower threshold for El was taken as 0.82, based on an average of approximately three times the "barely perceptible" difference of Flanagan (1972) and three times the "barely perceptible" threshold used by Sambur and Jayant (1976). Values of El below 0.82 mean the frame is understood. An upper threshold of 2.46 (3x0.82) was used to decide that the frame was completely misunderstood. A linear relationship was used between 0.82 and 2.46. A sample of noise was taken from the distorted tape being analyzed (actually the same samples used in the previous section to derive k). From this sample several frames were analyzed from which two frames were selected using the criteria of the largest and smallest values for the sum of the squares of the predictor coefficients, these two frames were used with the master tape to calculate (frame by frame) values of  $ElN_1$  and  $ElN_2$ , where N signifies noise. These two values (for each frame) were then averaged to obtain ElN. If ElN <2.46 the thresholds were not changed. If ElN >2.46 the thresholds were changed to

0.82 + 0.82 (ElN - 2.46) and 2.46 + 0.82 (ElN - 2.46).

To summarize, two thresholds Tl and T2 are defined. Using these, a linear measure is defined for each frame as

(5.45)

LM1 =	1	if El < Tl	
	0	if El > T2	a vaco, ciderable non h
	m2 m1		
lade Te	$\frac{T2-E1}{T2-T1}$	otherwise.	

The linear measure LM1 of the above method was calculated for all frames of each word. Further, for each word an average of this measure was calculated for those frames for which  $R_{O} \ge \overline{R}_{O}/2$ , where  $\overline{R}_{O}$  was the average of  $R_{O}$  for the word. This was designated LM1H. Similarly, an average for frames for which  $R_{O} < \overline{R}_{O}/2$  was calculated and called LM1L. This divides the measure into two groups, one for frames with higher power and one for frames with lower power compared to the average power in the word. Frequently, although not always, the low power frames correspond to the unvoiced speech and the high power to the voiced speech. Finally, LM1H and LM1L were averaged over fifty words to form LM1H and LM1L.

In a similar fashion, high and low values  $\overline{ElH}$  and  $\overline{ElL}$ , and  $\overline{ElNH}$  and  $\overline{ElNL}$  were calculated in order to modify the average of the linear measures in the following way. If  $\overline{\text{ElH}}$  (resp  $\overline{\text{ElL}}$ ) is less than .82 no modification is made. If  $\overline{\text{ElNH}}$  (resp  $\overline{\text{ElNL}}$ ) is greater than 2.46 no modification is made. Otherwise  $\overline{\text{LMH}}$  (resp  $\overline{\text{LML}}$ ) is multiplied by

#### $\frac{\overline{\text{E1NH}} - 0.82}{2.46 - 0.82} \quad (\text{resp} \ \frac{\overline{\text{E1NL}} - 0.82}{2.46 - 0.82} ) .$

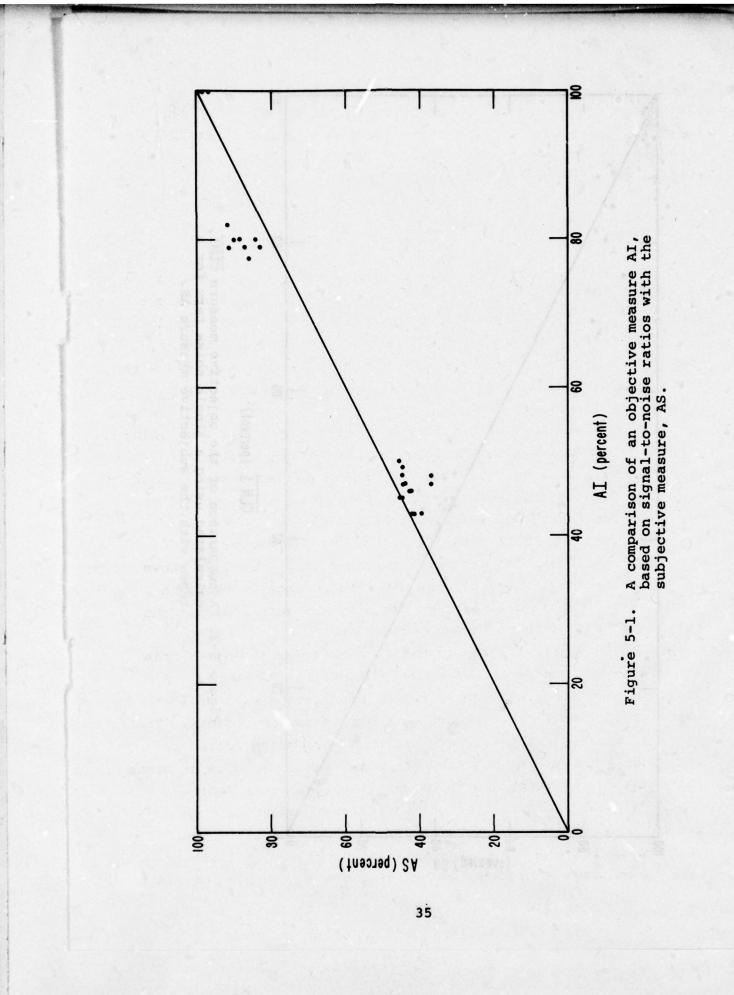
The modified high and low measures were then averaged to form  $\overline{LM1}$ . This has the effect of weighting the low values more than the high values since only about ]/3 of the frames are low.

The correlation  $C_{R,R}$ , of  $R_{O}$  and  $R_{O}$  was then calculated, and multiplied by  $\overline{LM1}$  to form  $\overline{CLM1}$ . This value was averaged with the AI measure of the previous paragraphs to form ASQ, the objective articulation score.

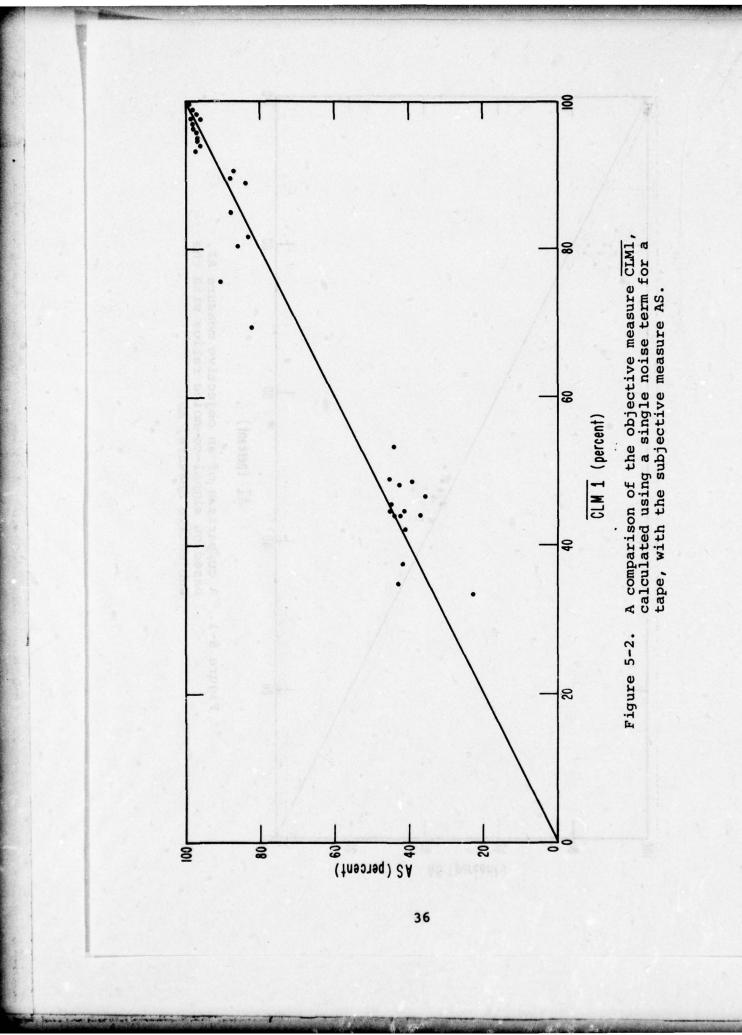
The correlation correction was applied to account for fading signals. Several other methods were used which compared the signal levels on a word-by-word and frame-by-frame basis. These required considerably more computing time and gave essentially the same results as using the correlation factor.

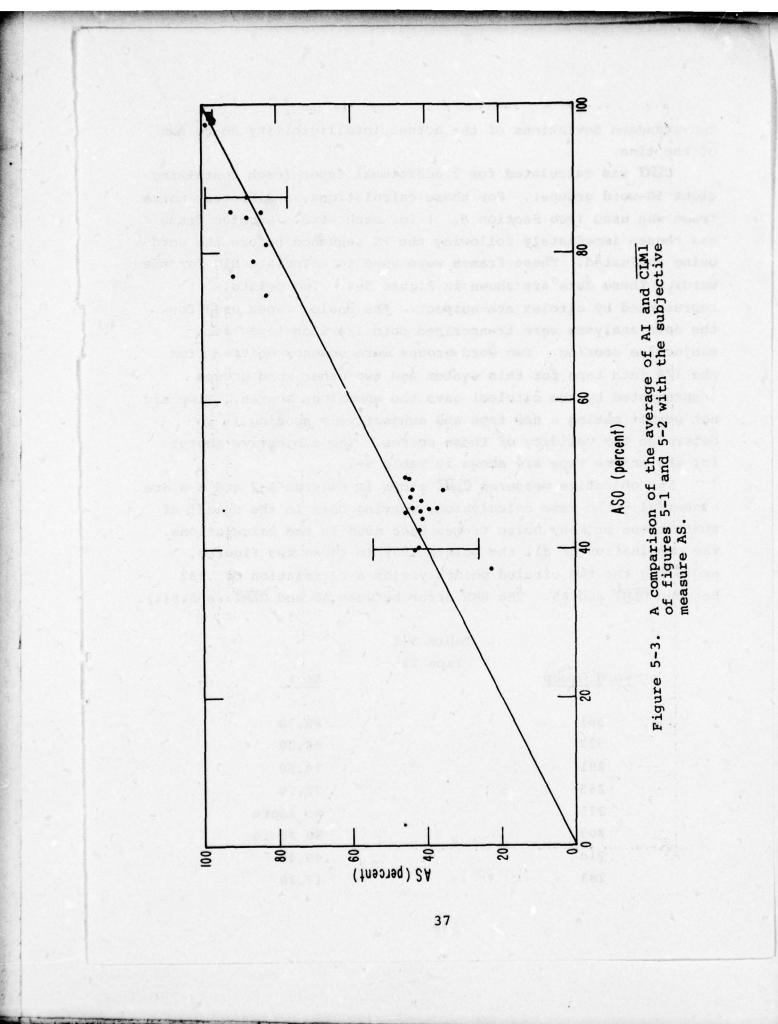
The quantity AI is shown in Figure 5-1 plotted vs the subjective articulation scores (AS), the quantity CLM1 is shown in Figure 5-2 plotted vs AS, and ASO, the average is shown in Figure 5-3 plotted vs AS. In Figure 5-3, the bars indicate the confidence limits about the subjective score.

While subjective intelligibility scores for isolated words are repeatable, there are some fluctuations in the scores. A listener panel is trained on a set of word groups with well established intelligibility scores and standard deviations. The average intelligibility of each training word group scored by individuals that make up a listener panel is always plus or minus one standard deviation of its actual intelligibility score. Also, the standard deviation of the listener panel for each training word group is approximately the same as its actual standard deviation. Using the training procedure, a listener panel will produce an intelligibility score within plus or minus



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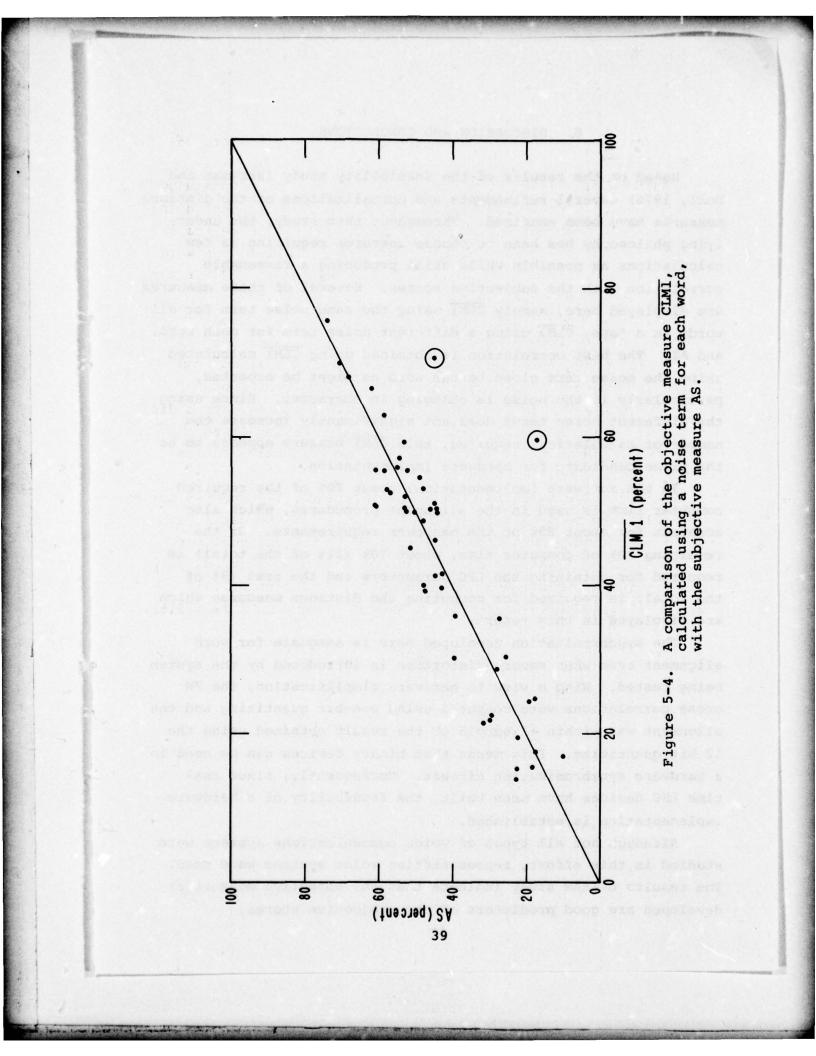


two standard deviations of the actual intelligibility score 95% of the time.

**CLMT** was calculated for 7 additional tapes (each containing eight 50-word groups). For these calculations, a different noise frame was used (see Section 5A ) for each word. A noise frame was chosen immediately following the PN sequence before the word being evaluated. These frames were used to calculate ElN for the words. These data are shown in Figure 5-4. Two points, represented by circles are suspect. The analog tapes used for the data analyses were transcribed onto 1/4 inch tapes for subjective scoring. Two word groups were somehow omitted from the 1/4 inch tape for this system and two other word groups (represented by the circles) gave the anomalous scores. Time did not permit making a new tape and subjectively scoring it to determine the validity of these scores. The subjective scores for the entire tape are shown in Table 5-1.

The objective measures  $\overline{\text{CLMI}}$  shown in Figures 5-2 and 5-4 are essentially the same calculation, varying only in the detail of whether one or many noise frames were used in the calculations. The combination of all the points (90) in these two figures, excluding the two circled points yields a correlation of .982 between  $\overline{\text{CLMI}}$  and AS. The RMS error between AS and  $\overline{\text{CLMI}}$  is 5.5(%).

	Tabl	le 5-1		
	Тар	ре ТЗ		
Word Group			AS	<u>s</u>
261			<i>co</i> .	
361			60.7	/0
312			68.0	)0
291			70.5	50
265			72.7	70
275			No s	score
305			No s	score
214			45.0	00
283			17.3	20



#### 6. DISCUSSION AND CONCLUSIONS

Based on the results of the feasibility study (Hartman and Boll, 1976) several refinements and normalizations of the distance measures have been examined. Throughout this study, the underlying philosophy has been to choose measures requiring as few calculations as possible while still producing a reasonable correlation with the subjective scores. Several of these measures are displayed here, namely CLMI using the same noise term for all words on a tape, CLMI using a different noise term for each word, and AI. The best correlation is obtained using CLMI calculated using the noise term close to the word as might be expected, particularly if the noise is changing in character. Since using the different noise terms does not significantly increase the number of calculations required, this CLMI measure appears to be the prime candidate for hardware implementation.

In the software implementation, about 70% of the required computer time is used in the alignment procedures, which also accounts for about 85% of the manpower requirements. Of the remaining 30% of computer time, about 70% (21% of the total) is required for obtaining the LPC parameters and the rest (9% of the total) is required for computing the distance measures which are displayed in this report.

The synchronization developed here is adequate for word alignment even when severe distortion is introduced by the system being tested. With a view to hardware simplification, the PN cross correlations were computed using one-bit quantizing and the alignment was within <u>+</u>1 sample of the result obtained using the 12 bit quantizing. This means that binary devices can be used in a hardware synchronization circuit. Consequently, since real time LPC devices have been built, the feasibility of a hardware implementation is established.

Although not all types of voice communications systems were studied in this effort, representative voice systems were used. The results of the study indicate that the objective measure(s) developed are good predictors of the subjective scores.

In order to enlarge the data base further, or to investigate other uses and modifications of these methods it appears that the most economical procedure is to develop a flexible hardware system.

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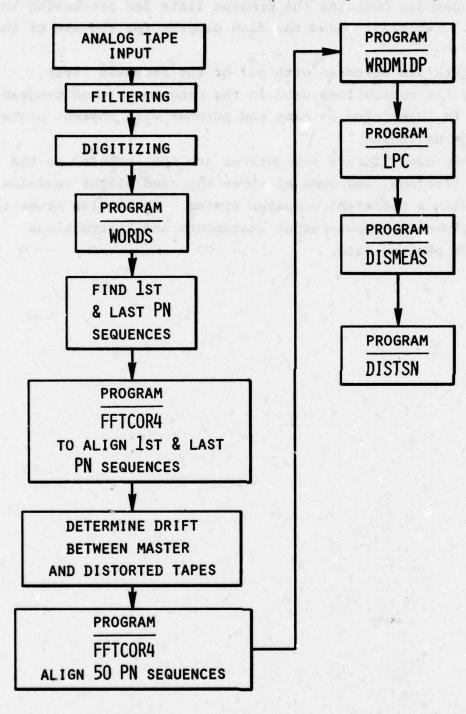
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#### APPENDIX

This appendix contains the program lists for processing the voice data. Table A-1 shows the flow diagram for the use of the programs.

The lists are arranged with all of the programs first, followed by the subroutines used in the programs. Each program or subroutine is identified by name and purpose with comment statements at the beginning.

Commonly used library subroutines are not included in the subroutine listings, and some of these may need slight revision when used with a different computer system. Particular attention should be given the input-output statements and instructions dealing with packing data.



#### Table A-1. Flow Chart for Processing Voice Tapes

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the strategies and

1		PROGRAM WORDS(INPUT,OUTPUT,TAPE1)
•	c	PROGRAM WORDS FINDS THE LOCATION OF THE PN SEQUENCES AND JORDS OF
	c	A PARTICULAR WORD GROUP BY FINDING THE MEAN AND STANDARD DEVIATION
	C	OF CONSECUTIVE 125 SAMPLE BLOCKS. THE STANDARD DEVIATION IS USED
5	c	AS AN ENERGY CRITERION IN ORDER TO LOCATE THE PV SEQUENCES AND WORDS.
	c	DIMENSION IN(90), IRECORD(400), ID(125), ITEMP(2300) INITIALIZE VARIABLES. +THOLD+ IS A THRESHOLD USED TO DETERMINE IF
	č	PART OF A PN SEQUENCE OR WORD IS PRESENT IN THE DATA. +ISKIP+ AND
	c	*IEND* DETERMINE THE STARTING AND ENDING PDINTS OF THE RECORDS TO
10	c	BE PROCESSED. +KOUNT+, +ICYCLE+, AN" +IPARTS+ ARE COUNTING VARIABLES.
	c	«ILENGTH» IS THE NUMBER OF SAMPLES TO BE PROCESSED AT A TIME AND «Iwords» is the number of samples in a tape record»
	•	
		ISKIP=0
15		IEN0=3451
		KOUNT=1 ICYCLE=0
		IPARTS=16
		TI ENGTH-12E
20		Idons = 4.0
		141C=1SKIP
	c	THE TAPE IS POSITIONED TO THE STARTING POINT FOR PROCESSING.
		IF (ISKIP .E3. 0) 60 TO 5 DO 1: K19=1.1SKIP
25		PJFFER IN (1+1) (IN(1)+IN(40))
		IF (UNIT(1)) 10-20-10
		CONTINUE
	55	PRINT 53.ISKIP Format (1H1.IS.+ RECORDS SKIPPED +)
36		IRECHICAL
	C	FIVE CONSECUTIVE BO WORD RECORDS ARE BUFFERED IN FOR PROCESSING.
	c	EACH 60 BIT WORD IS UNPACKED INTO FIVE 12 BIT WORDS USING SUBROUTINE
	c	JUPACK. THIS YELLDS A TOTAL OF 2000 SAMPLES TO BE PROCESSED AT A TIME.
35		BJFFER IN (1+1) (IN(1)+IN(B+)) IF (UNIT(1)) 15+20+25
	25	PRINT 12C+IREC
	120	FORMAT (+ RECORD ++15++ HAS A PARITY ERROR +)
	•	CO TO 3
41	15	LL=LENGTH(1) CALL UNPACK (IRECORD+IN+LL)
		KINDEX=ICYCLE+IWORDS
		30 12- K12=1+IWORDS
	12	ITEMP(K12+KINDEK)=IRECORD(K12)
45		ICYCLE=ICYCLE+1 IF (ICYCLE +LT+ 5) 30 TO 30
•.	с	THE 2006 SAMPLES ARE BROKEN UP INTO SIVTEEN 125 SAMPLE BLOCKS. FOR
	C	EACH BLOCK. THE MEAN AND STANDARD DEVIATION ARE COMPUTED. IF THE
	c	STANDARD DEVIATION IS GREATER THAN OR EQUAL TO THE PRESET THRESHOLD.
	C C	THOLD, THE FIRST TEN SAMPLES, THE MEAN, THE STANDARD DEVIATION, AND THE PLOCK NUMBER FOR THE SAMPLE PLOCK ARE PRINTED. OTHERWISE THE
50	č	PROGRAM PROCEEDS TO THE NEXT SAMPLE BLOCK.
		00 35 K35=1+1PARTS
		VINOFX=(K35-1)+JLENGTH
		DO 4C K4C=1+ILENGTH ID(K4C)=ITEMP(K4C+NINDEX)
55		SUM1=SUM2=0.0
		DD 45 K45=1+ILENGTH
		SUM1=SUM1+ID(K45)
	45	SUM2 = SUM2 + ID(K + 5) + ID(K + 5) A ME AN = SUM1 / ILENG TH
60		STODEV=SQRT(ABS((1LENGTH+SUM2-SUM1+SUM1)/ILENSTH++2))
		IF (STDDEV .LT. THOLD) GO TO 50
		PRINT 136+KOUNT+(ID(I)+I=1+14)+AMEAN+STDDEV+KOUNT
		FORMAT (+ +,16+1019+2F12+4+19)
65		KOUNT=KOUNT+1 Continue
	c	THE NUMBER OF THE NEXT RECORD TO BE PROCESSED IS COMPARED TO THE
	c	NJMBER OF THE LAST RECORD TO BE PROCESSED, IEVD. IF IT IS LESS THAN
70	c	IEND. THE PROGRAM CONTINUES. OTHERWISE THE PROGRAM TERMINATES. IF (IREC .LT. ISND) 50 TO 30
	20	PRINT 100
		FORMAT (1H1)
_		PRINT 115.IREC
75	115	FORMATCIS) DECT AVAILADIE CONV
	110	FORMATCISS PAINT 110 FORMAT C. NORMAL TERMINATION -> BEST AVAILABLE COPY
		EV)

PROGRAM FFTCOR4 (INPUT.OUTPUT.PUNCH.TAPE1.TAPE2) 1 PROGRAM FFTCORA COMPUTES THE CROSS-CORRELATION BETWEEN TWO SEQUENCES OF DATA SAMPLES. WHICH IN THIS CASE ARE THE PV SEQUENCES. C COMMON /FFT1 /A (4200), 5 (4200), C (4200), D (4200), D1 (5000) Common /FFT2 /ICOR (2100) 5 DIMENSION ID(503) . ITEMP(100) . IN(100) INITIALIZE VARIABLES. • MPRINT. IS A PRINTING CONTROL VARIABLE AND C \*NUMREC\* IS THE NUMBER OF SAMPLES IN A TAPE RECORD. \*IASIZE1\* AND \*IASIZE2\* ARE COUNTING VARIABLES. \*NRECRED\* AND \*NRECIAA\* KEEP C C TRACK OF THE POSITION OF THE UNDISTORTED AND DISTORTED TAPES. 10 MPRINT=0 NUMREC=400 IASIZE1 = A200IASIZE2 = 5000PRINT 1504 Format (141) NRECRED = NREC1AA = 0 15 1504 100 DO 110 K10 = 1. IASIZE1 135 A (K10) = B (K10) = C (K10) = D (K10) = C. 110 CONTINUE 20 00 115 K20 = 1. IASIZE? 01 (K20) = 0 CONTINUE 115 THE CONTROL CARD IS READ THAT GIVES THE MIDPOINTS OF THE TWO SEQUENCES С TO BE PROCESSED AND THEIR LENGTH. IF VARIABLE MIDPNT1 IS EQUAL TO ZERO, THE PROGRAM TERMINATES. INDEXES ARE THEN COMPUTED THAT POSITION THE TWO TAPES IN ORDER TO GET THE DESIRED DATA. 25 C C C READ 1502. MIDPNT1. MIDPNT2. NUMEXP 1502 FORMAT(314) 30 IF (MIDPNT1 .EQ. 0) GO TO 495 NJMBER = 2 . . NUMERP INDEX1 = MIDPNT1 - NUMBER / 2 INDEX2 = MIDPNT2 - NUMBER / 2 INDER2 = HIDPATA - NUMBEC - NRECRED ISTRTR1 = INDEX1 / NUMBEC - NRECRED ISTRTR2 = INDEX2 / NUMBEC - NRECLAA IORIGIN = INDEX1 - NUMBEC + (INDEX1 / NUMBEC) IORIJAA = INDEX2 - NUMBEC + (INDEX2 / NUMBEC) 35 THE THE TAPES ARE POSITIONED TO THE START OF THE SEQUENCES TO BE C C PROCESSED. 40 IF(ISTRTR1 .LT. 1) GO TO 500 DO 140 K40 = 1. ISTRTR: 3JFFER IN (1.1) (ITEMP(1).ITFMP(80)) IF (UNITE 1))141. 435. 135 PRINT 1508+ K40 Format (\*RFAD Error Tapf 1 ++ I10) Convinue 135 45 1508 140 500 CONTINUE NRECRED = NRECRED + ISTRTR1 IF(ISTRTR2 .LT. 1) GO TO 505 DO 155 K60 = 1, ISTRTR2 SUFFER IN (2.1) (ITEMP(1).ITEMP(P))) 50 IF (UNITE 2))155. 445. 150 PRINT 1510. K65 150 1510 FORMAT (+READ ERROR TAPE 2 ++ IIC) 155 CONTINUE 55 505 CONTINUE NRECIAL = NRECIAA + ISTRTR2 THE TWO DATA SEQUENCES ARE SUFFERED INTO THE PROGRAM. UNPACKED BY SUBROUTINE UNPACK, AND STORED IN INDIVIDUAL ARRAYS. NRECIN = 2 + NUMBER / NUMREC IOFFSET = 0 C C 60 DO 180 K100 = 1, NRECIN SUFFER IN (1.1) (IN(1),IN(8C)) IF (UNIT( 1))170, 485, 165 KPARITY = K40 + K100 - 1 PRINT 1516,KPARITY 155 65 FORMAT (+PARITY ERROR IN DATA RECORD. RUN ABORTED +. 110) 1516 50 TO 475 CONTINUE 170 LL=LENGTH(1) 70 CALL UNPACK(ID.IN.NUMREC.LL) 00 175 K95 = 1. VUMREC 01 (K95 + IOFFSET) = ID (K95) BEST AVAILABLE COPY 175 CONTINUE IDFFSET = IOFFSET + NUMREC 75

190 CONTINUE WRECKED = NRECRED + NRECIN 00 185 K102 = 1. NUMPER A (K102) = D1 (IORIGIN + K102) 80 195 CONTINUE 00 190 4103 = 1. IASIZE? D1 (K103) = 7 190 CONTINUE IOFFSET = 0 100031 = 0 0035 K130 = 10 NRECIN 3JFFER IN (201) (IN(1)0IN(PO)) IF (JNIT(2))215, 485, 205 205 KPARITY = K60 + K100 - 1 PRINT 1516.KPARITY GO TO 475 85 91 215 CONTINUE LL=LiNGTH(2) CALL UNPACK (ID.IN.NUMREC.LL) 00 225 K125 = 1. NUMSEC D1 (K125 + 10FFSET) = 10 (K125) 95 225 CONTINUS IDFFSET = IDFFSET + NUMREC CONTINUE 235 VRECIAA = NRECIAA + NRECIN 00 255 K150 = 1. NUMBER 160 P (K150) = 01 (IORI144 + K157) CONTINUE 255 THE MEANS AND THE STANDARD DEVIATIONS OF THE THO DATA ARRAYS ARE С COMPUTED. C SUMAIA = SUMAIR = SUMAZA = SUMAZE = / FNUMBER = NUMBER 165 DENOM = FNUMBER + (FNJMBER - 1.) DO 405 K152 = 1. NUMBER SUMX1A = SUMX1A + A (K152) SUMX18 = SUMX18 + B (K152) SUMX2A = SUMX2A + A (K152) + A (K152) SUMX28 = SUMX28 + B (K152) + B (K152) 110 405 CONTINUE AMEAN = SUMX1A / FNUMBER BMEAN = SUMX1B / FNUMBER 115 STDA = (FNUMBER + SUMX2A - SUMX1A + SUMX1A) / DENOM STDA = SORT (ABS (STDA)) STOB = (FNUMBER + SUMX2B - SUMX1B + SUMX13) / DENOM STDB = SORT (A3S (STD3)) THE FORWARD FFT FOR BOTH DATA ARRAYS ARE COMPUTED USING SUBROUTINES 120 С C REVBIN. CFFTRC. AND RTRANZT. NUMERP = NUMERP + 1 NUMBER = 2 + + NUMERP FNJMBER = NUMBER M = NUMEXP MM = NUMEXP - 1 125 INV = 1 NOTE = - 1 SC = .5 CC = +5 CALL REVBIN (A, B, M) CALL CFFTRC (A, B, M, SC, NDIR) CALL RTRANZT (A, B, M, INV) THE TRANSPOSE OF THE PRODUCT OF THE FFTS OF THE DATA ARRAYS IS 130 C С CALCULATED. 135 IUPLIN = 1 + NUMBER / 2 00 415 K160 = 1. IUPLIM L160 = 4160 + IUPLIM REAL = A (K160) + A (L160) + 9 (K160) + B (L150) COMP = 3 (K160) + A (L160) - A (K160) + P (L150) A (K160) = REAL B (K160) = COMP 140 CONTINUE 415 THE INVERSE FFT IS TAKEN OF THE FFT PRODUCT USING SUBROUTINES REALTRA. CFFTS. AND REORDER. C C 145 IVV = - 1 BEST AVAILABLE COPY NDIR = 1 127 3 SC = 1. / FNUMBER CALL REALTRA (A. B. MM. NOIR. INV) CALL CFFTS (4. 3. MM. SC. NDIR) 150 CALL REORDER (A, S, MM)

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155	C C C	THE LOCATION AND VALUE OF THE MAXIMUM CROSS-CORRELATION TERM IS FOUND, ALONG WITH THE MIDPOINT OF THE DISTORTED DATA SEQUENCE THAT LINES IT UP WITH THE UNDISTORTED DATA SEQUENCE. NUMHALF = NUMBER / 2 NJMOTR = NUMBER / 4 FNUMHALF = NUMBER / 4 FNUMHALF = NUMBER / 4 FNUMHALF = S. FCOR = 0.
160		DO 435 K170 = 1+ NUMOTR FK170 = K170 BLOWER = B (K170 + NUMOTR) / (FNUMOTR + FK170 - 1+) D1 (K170 + NUMOTR) = (BLOWER - AMEAN + BMEAN) / STDA / STDB RATID1 = (FNUMOTR + FK170 - 1+) / FNUMMAL
165		$ \begin{array}{l} \textbf{Q} \textbf{UAN1} = \textbf{RATIO1} + \textbf{DI} (\textbf{K170} + \textbf{NUMQR}) \\ \textbf{IF} (\textbf{Q} \textbf{UAN1} \bullet \textbf{LE} \bullet \textbf{FMAXCOR}) \textbf{GO TO 425} \\ \textbf{FMAXCOR} = \textbf{Q} \textbf{UAN1} \\ \textbf{FCDR} = \textbf{D1} (\textbf{K170} + \textbf{NUMQTR}) \\ \textbf{LOCATIO} = \textbf{K170} + \textbf{NUMQTR} - \textbf{NUMBER} / 2 - 1 \\ \end{array} $
170	425	$ \begin{array}{llllllllllllllllllllllllllllllllllll$
175	435 C	FMAXCOR = QUAN2 FCOR = D1 (K170 + NUMHALF) LOCATIO = K170 + NUMHALF - NUMBER / 2 - 1 CONTINUE THE LOCATION AND VALUE OF THE MAXIMUM CROSS-CORRELATION TERM AND
180	с с с	THE NEW DISTORTED TAPE SEQUENCE MIDPOINT ARE PRINTED ALONG JITH THE INPUT VARIABLES. THE NEW DISTORTED TAPE SEQUENCE MIDPOINT IS ALSO PUNCHED ON A HOLLERITH CARD FOR FUTURE USE. PRINT 1502,MIDPNT1,MIDPNT2,NUMEYP IF(MPRINT .NE. 03 CALL PRINT(3,NUMBER)
185	1522	PRINT 1522. FMAXCOR, FCOR, LOCATIO Format (2F12.3, 10x, 15) MIDPN=MIDPNT2-LOCATIO PRINT 2010.MIDPN Format (110)
190		PUNCH 2010.MIDPN PRINT 2000 FORMAT (3%,//) The Program IS Sent Back to read the Next Control Card. 50 to 125
195	485 1512 C 495 1514	FRINT 1512 FORMAT (+FND OF FILE READJOB TERMINATED +) PROGRAM TERMINATION SEQUENCE + PRINT 1514 FORMAT (JH1+ + NORMAL TERMINATION +)
200		CONTINUE SND

1			PROGRAM WRDHIDP (INPUT-OUTPUT-PUNCH)
	C		PROGRAM WRDMIDP CALCULATES THE MIDPOINTS OF THE WORDS OF THE
	č		DISTORTED TAPE WITH RESPECT TO THE UNDISTORTED TAPE.
			DIMENSION TPN(50)
5	C		THE DRIFT PER SAMPLE OF THE DISTORTED TAPE IS CALCULATED FROM +DIFF+.
	č		THE DIFFERENCE IN THE NUMBER OF SAMPLES BETWEEN THE UNDISTORTED AND
	c		DISTORTED TAPE, AND +TOTAL+, THE TOTAL NUMBER OF SAMPLES IN THE
	č		DISTORTED WORD SROUP TAPE.
			READ 50.DIFF.TOTAL
10		50	FORMAT(2F10-1)
			DRIFT=DIFF/TOTAL
			PRINT 35
		35	FORMAT (1H1)
	C		THE 50 UNDISTORTED PN SEQUENCE MIDPOINTS ARE READ IN FOR THE
15	č		PARTICULAR WORD GROUP.
			20 10 J=1.50
			READ 15-MIDPN
		15	FORMAT (110)
		10	IPN(I)=HIDPN
20	C		USING THE UNDISTORTED PN SEQUENCE MIDPOINTS. THE DISTANCE BETWEEN
	C		THE PN SEQUENCE AND THE WORD OF THE UNDISTORTED TAPE. AND THE DRIFT
	c		PRESENT. THE WORD MIDPOINTS OF THE DISTORTED TAPE ARE COMPUTED.
			00 20 J=1+50
			READ 25. NWORDPN. N256
25		25	FORMAT (110+15)
			IDRIFT=DRIFT+NWDRDPN
			MIDJORD=IPN(J)+NWORDPN+IPRIFT
	C		THE RESULTS ARE PRINTED AND PUNCHED ON HOLLERITH CARDS FOR FUTURE USE.
			PRINT 30.MIDWORD.V256
30		33	FORMAT (110+15)
		1.17	
		20	CONTINUE
			END

11

V400 110

PROGRAM LPC (INPUT.OUTPUT.TAPE1.TAPE2=/680) 1 PROGRAM LPC DOES THE FRAME BY FRAME LPC PROCESSING OF EACH JORD OF A PARTICULAR WORD GROUP AND STORES THE INFORMATION ON MAGVETIC TAPE. The stored data consists of the energy term, the normalized Autocorrelation terms, the predictor coefficient autocorrelation C CCC 5 TERMS. THE PREDICTOR COEFFICIENTS. THE REFLECTION COEFFICIENTS. AND C THE MINIMUM SQUARED ERROR. COMMON A(256) . ACOR (13) . PCOEF (12.12) . GAM(13) . ERROR DIMENSION INCADD. IDCADD. ITEMP(8400). #(256) INITIALIZE VARIABLES. +NUMBER IS EQUAL TO THE AVALYSIS FRAME SIZE. +NPC+ IS EQUAL TO THE NUMBER OF PREDICTOR CDEFFICIENTS DESIRED. +IWORDS+ IS THE NUMBER OF SAMPLES IN A TAPE RECORD. AND +PI+ IS THE UNIVERSAL CONSTANT. 10 C CCC NUMBER=256 15 NUMHALF=NUMBER/2 NPC=12 IWORDS=400 THENUMBER PI=3.141592653589 THE HAMMING WINDOW TO BE USED ON THE SPEECH SAMPLES IS GENERATED. 20 С DO 5 I=1.NUMHALF WH=(1.0/(1.08+TW))+(0.54+0.46+COS(2.0+PI+I/TW)) 5 W(128+I)=W(129-I)=WH C MAGNETIC TAPE IDENTIFICATION VARIABLES ARE DEFINED. READ 3.NFILE.NTAPE.NTYPE 25 3 FORMAT (3110) 50 WRECORD=0 NHORD=1 PRINT 6 5 FORMAT (1H1) 30 THE MIDPOINT AND LENGTH. IN TERMS OF ANALYSIS FRAMES, OF THE WORD TO BE PROCESSED IS READ INTO THE PROGRAM. IF THE WORD MIDPDINT IS EQUAL TO ZERO. THE PROGRAM TERMINATES. INDEXES ARE COMPUTED THAT POSITION THE TAPE IN ORDER TO GET THE DATA SAMPLES FOR THE JORD. C C C 35 100 READ 105.MIDPNT. N255 105 FORMAT (110.15) IF (MIDPNT .EQ. 0) GO TO 20 MJLT=N256 4LENGTH=NJMBER+N256 PRINT 7.MIDPNT.MLENGTH.N256 40 7 FORMAT (2110.15./) INDEX="IDPNT-"LENGTH/2 ISKIP=INDEX/INORDS-NRECORD IDRIG=INDEX-IWORDS+(INDEX/IWORDS) ALPHA IS DEFINED, WHICH IS USED AS AN IDENTIFIER ON THE MAGNETIC TAPE. 45 С ENCODE (10+H0+ALPHA) NTYPE 30 FORMAT (17+3X) THE TAPE IS POSITIONED TO THE START OF THE 40RD TO BE PROCESSED. IF (ISKIP .LT. 1) 50 TO 12 C 50 33 15 K15=1.ISKIP BJFFER IN (1.1) (IN(1),IN(4")) IF (UNIT(1)) 15.20.25 NPARITY=K15-NRECORD 25 PRINT 30. NPARITY 30 FORMAT (+ PARITY ERROR IN RECORD ++15) 55 15 CONTINUE NRECORD=NRECORD+ISKIP THE SAMPLES OF THE WORD ARE BUFFERED INTO THE PROGRAM, UNPACKED BY SUBROUTINE UNPACK, AND STORED IN AN ARRAY. 60 10 NREC=2+MLENGTH/IWORDS IOFFSET=0 DO 35 K35=1.NREC BUFFER IN (1.1) (IN(1).IN(80)) IF (UNIT(1)) 33.20.40 40 KPARITY=K35+NRECORD 65 PRINT 45+KPARITY 45 FORMAT(+ PARITY ERROR IN RECORD ++15++ RUN ABORTED +) GO TO 75 33 LL=LENGTH(1) CALL UN-ACK (ID+IN+LL) D0 55 K55=1+IW07DS 55 ITEMP (IOFFSET+K55)=ID(K55) 70 35 IOFFSET=IOFFSET+IWORDS BEST AVAILABLE COPY NRECORD=NRECORD+NREC

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while the part that a manual part in the

75	c	ONE BY ONE THE ANALYSIS FRAMES OF THE WORD ARE WINDOWED BY THE HAMMING
17	C	JINDOW AND PROCESSED BY SUBROUTINES AUTOCOR, PRECOEF, AND GAMMA.
	THE ALL BRIDE	D0 60 K60=1.4ULT
		VINDEX=IORIG+(K50-1)+NUMBER
		D0 65 K65=1.NUM3E3
80		65 A(K65)=ITEMP(K65+NINDEX)+W(K65)
		CALL AUTOCOR (NJABES.NPC)
		CALL PRECOEF (NPC)
		CALL GAYMA (NPC)
	c	THE LPC INFORMATION FOR EACH FRAME IS STORED ON MAGNETIC TAPE. THE
85	c	FIRST SIX VARIABLES ARE FOR IDENTIFICATION PURPOSES. ACOR(1) IS THE
	c	ENERGY TERM. ACOR(J). J=2.13 ARE THE 12 NORMALIZED AUTOCORRELATION
	c	TERMS. GAM(J). J=1.13 ARE THE PREDICTOR COEFFICIENT AUTOCORRELATION
	Ċ	TERMS. PCDEF(J.12). J=1.12 ARE THE PPEDICTOR COEFFICIENTS, PCDEF(J.J.).
	C	J=1.12 ARE THE REFLECTION COEFFICIENTS. AND ERROR IS THE MINIMUM
90	C	SQUARED ERROR.
		WRITE (2.200) ALPHA.NFILE.NTAPE.NWDRD.MULT.K6D.(ACDA (J).J=1.13).(6
		144(J) .J=1.13) . (°COEF(J.12) .J=1.12) . (°COEF(J.J) .J=1.12) .FR303
	2	00 FORMAT (A10.515.616.13.12F12.10.13E14.8.12E14.8.12F10.8.E12.6)
		SO CONTINUE
95		END FILE 2
		BACKSPACE 2
		NUDRD=NUORD+1
		NTYPE=NTYPE+1
	C	THE PROGRAM IS SENT BACK TO READ THE NEXT WORD CONTROL CARD.
100		30 TO 1°0
	C	PROGRAM TERMINATION SEQUENCE.
		20 PRINT 73
		73 FORMAT (1H1++ NORMAL TERMINATION +) .
		75 CONTINUE
165		END FILE 2
		END

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PROGRAM DISMEAS(INPUT,OUTPUT,TAPE1=/690,TAPE2=/680,TAPE3=/175) PROGRAM DISMEAS COMPUTES THE RESIDUAL DISTANCE MEASURES D. D. D./E.. AND D/E AND STORES THEM ON MAGNETIC TAPE ALONG WITH THE SUM OF THE SQUARES OF THE PREDICTOR COEFFICIENTS, THE ENERGY TERMS, AND THE MINIMUM SQUARE ERROR TERMS OF THE UNDISTORTED AND DISTORTED TAPES ON A FRAME BY FRAME BASIS. 1 C C С 5 cc ON A FRAME BY FRAME BASIS. DIMENSION ACOR(13),GAM(13),A(12),RC(12) DIMENSION DACOR(13),DGAM(13),DA(12),DRC(12) DIMENSION E(4),ASUM(B),B(51,R) INITIALIZE VARIABLES. +MWORD+ IS A COUNTING VARIABLE. 10 C MWORD=0 PRINT 42 42 FORMAT(1H1) THE LENGTH, IN ANALYSIS FRAMES, OF THE WORD TO BE PROCESSED IS READ INTO THE PROGRAM. IF THE WORD LENGTH IS EQUAL TO ZERO, THE PROGRAM 15 JUMPS OUT OF THE PROCESSING LOOP. READ 10.N256 30 10 FORMAT(15) IF (N256 .EQ. 0) 60 TO 20 20 HWORD=HWORD+1 NHULT=N255 DO 38 I=1.8 38 ASUM(1)=0.0 FRAME BY FRAME, THE UNDISTORTED AND DISTORTED SPEECH LPC DATA IS READ INTO THE PROGRAM. C C 25 00 15 K15=1.NMULT READ (1+200) ALPHA,NFILE, VTAPE, VWORD, MULT, K50, (ACOR(J), J=1+13), (GA 14(J)+J=1+13)+(A(J)+J=1+12)+(RC(J)+J=1+12)+ERROR 200 FORMAT (A10+515+E16+10+12F12+10+13E14+8+12E14+8+12F10+8+E12+5) READ (2.200) ALPHA, NFILE, NTAPE, NWORD, MULT, KSG, (DACOR(J), J=1.13), (D 30 IGAM(J)=J=1+13)+(DA(J)+J=1+12)+(DR(J)+J=1+12)+DERROR THE RESIDUAL DISTANCE MEASURES D++ D+ D+/E++ AND J/E ARE CALCULATED AND STORED ON MAGNETIC TAPE ALONG WITH SIX IDENTIFICATION VARIABLES+ THE SUM OF THE SQUARES OF THE PREDICTOR CDEFFICIENTS+ THE ENERGY TERMS+ AND THE MINIMUM SQUARE ERROR TERMS OF THE UNDISTORTED AND C C С 35 C DISTORTED TAPES FOR EACH ANALYSIS FRAME. SUM1=GA4(1) SUM2=DGAM(1) DO 30 1=2.13 SUM1=SUM1+GAM(I) +DACOR(I) 40 30 SJM2=SUM2+DGAM(I)+ACOR(I) E(1)=SUM1 E(2)=SU42 E(3)=E(1)/DERROR E(4)=E(2)/ERROR 45 DO 32 I=1.4 32 ASJM(I)=ASUM(I)+E(I) ASUM(5)=ASUM(5)+ACOR(1) ASUM(6)=ASUM(6)+DACOR(1) ASU"(7)=ASU"(7)+ERROR 50 ASUM(3)=ASUM(8)+DERYOR WRITE (3.310) ALPHA.NFILE.NTAPE.NWORD.MULT.(60.E(1).E(2).E(3).E(4) 1.SAM(1).DGAM(1).ACOR(1).DACOR(1).ERROR.DERROR 300 FORMAT(A10,515,10E14.8) 55 15 CONTINUE END FILE 3 BACKSPACE 3 С THE WORD AVERAGE OF THE RESIDUAL DISTANCE MEASURES, THE ENERGY C TERMS. AND THE ERROR TERMS ARE COMPUTED. 60 DO 78 I=1.8 78 B(MWORD, I) = ASUM(I) / NMULT IF (EOF(1)) 20+25 25 IF (EOF(2)) 20+5 C THE PROGRAM IS SENT BACK TO READ THE VEXT JORD CONTROL CARD. 45 GO TO 50 65 20 CONTINUE IF (MWORD .EQ. D) MWORD=1 WH=HWORD+1 THE WORD GROUP AVERAGE OF THE RESIDUAL DISTANCE MEASURES. THE ENERGY C

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10	-	TERMS. AND THE ERROR TERMS ARE CALCULATED. D0 90 I=1.8
		SU#=0.0
		DO 91 J=1.MWORD
		91 5J4=SUM+5(J+1)
75		90 B(4W.1)=SJM/4W0RD
	ç	THE JOPD AVERAGE AND THE WORD GROUP AVERAGE OF THE RESIDUAL DISTANCE MEASURES. THE EVERGY TERMS. AND THE ERROR TERMS ARE PRINTED.
		PRINT 75
		75 FORMATCOUM COUNT ARAA AARAA E1/E. E2/E
80		1 RS RS+ EMAG EMAG+)
		00 54 J=1+MW0R0
		64 PRINT 65+J+(R(J+I)+I=1+)
		65 FORMAT(110.4510.4)
		PRINT 53
85		53 FORMAT(/)
		PRINT 65
		65 FORMAT(+ WORD SROUP AVERAGE +)
		PRINT 67.(8(4W.T).I=1.5)
		67 FORMAT(10X+8-10-4)
9,		PROGRAM TEPMINATION SEQUENCE.
		PRINT 7
		TO FORMAT (141++ NORMAL TERMINATION +)
		END FILE 3

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1 PROGRAM DISTKNCINPUT, OUTPUT, TAPE2=/135, TAPE3=/175) PROGRAM DISTRICT COMPUTES THE \*K\* FACTOR USED IN THE \*AI\* DISTANCE MEASURE AND PRIVIS IT FOR EACH WORD AND THE WO'D GROUP AVERAGE FOR EACH PAIR OF NOISE AND SIGNAL FILES READ IN. DIMENSION E(2).NW(51).FNAME(5) C C C 5 DIMENSION FDAME(5).D(2).S(30) DIMENSION A(30). R(30). DM(2). NCT(2) DIMENSION G1(51,5),42(51,5),43(51,5) DIMENSION NG1(51,5),NG2(51,5),NG3(51,5) INITALIZE VARIABLES. . MUORD. IS A COUNTING VARIABLE. С 1 . MUORD=0 THE LENGTH. IN ANALYSIS FRAMES. OF THE 5C WORDS IN THE WORD GROUP BEING PROCESSED ARE READ INTO THE PROGRAM. C 51 READ 10.N256 15 10 FORMAT(15) IF (N256 .EQ. 0) 60 TO 2. MWORD=MWORD+1 NMULT=N256 NH(MNORD)=NMULT GO TO 5: 20 21 CONTINUE Mu=MUORD+1 A TOTAL OF .IFILE. NOISE AND SIGNAL GATA FILE! APE READ INTO THE PROGRAM. С IFILE=5 READ 5+(FNAME(I)+I=1+5) READ 5+(FDAME(I)+I=1+5) 5 FORMAT(5A7) 25 C EACH OF THE PAIRS OF NOISE AND SIGNAL DATA FILES ARE PROCESSED. 00 15 K25=1.IFILE LFN=LFORM(FNAME(K25)) 30 CALL PEMATCH(SLTAPE3.LEN. (.0.0.0.1) LEN=LFORM(FDAME(K25)) CALL PEMATCH(SLTAPE2+LEN+0+0+0+C+1) EACH OF THE WORDS IN THE WORD GROUP FILES ARE PROCESSED. C 35 DO 3 KBG=1.MWORD NMULT=NJ(K3.) SUM1=SUM2=0.0 00 15 K15=1.NMULT FRAME BY FRAME, THE NOISE AND SIGNAL LPC DATA IS READ INTO THE PROGRAM. READ (2+200) ALPHANNFILE.NTAPE.NWORD.MULT.K6(.71.22.D(1).D(2).Y1.Y C 4 1 12.73.74.75.76 200 FORMAT(A10+515+10E10+4) READ (3+300) ALPHA+NFTLE+NTAPE+NWORD+MULT+K6(+Y1+X2+F(1)+E(2)+A2+D 1A2+H .. DA :. ERROR. DERFOR 500 FORMAT(A10.515.1.214.4) 45 C THE +K+ FACTOR IS COMPUTED FOR EACH FRAME. A(K15)=x1 5(4:5)=7: P(K15)=8 SUM1=SUM1+P. 5 SU% = SU\*2++1/21 15 CONTINUE HALF THE AVERAGE ENERGY AND THE AVERAGE .K. CTOR OF THE WORD BEING PROCESSED ARE CALCULATED. C 55 RAVG=SUM1/NMULT KAVG2=RAVG/2.0 31(K30,K25)=SUM2/NMULT NQ1 (K30. K25) =NMULT SUBROUTINE HILDW IS CALLED TO COMPUTE THE A' AGE \*K\* FACTOR FOR THE HIGH AND LOW ENERGY FRAMES OF THE WORD BEING ROCESSED. C 60 CALL HILOW(NMULT + RAVG2 + A + R + S + D" + NCT) 92(x30+x25) = DM(1) Q3(K30.K25)=DM(2) NQ2 (K30, K25) =NCT (1) NQ3 (K 30. K25) = NCT (2) 65 30 CONTINUE THE AVERAGE \*K\* FACTOR OF THE WORD GROUP FILE FOR THE HIGH, LOW, AND TOTAL AVALYSIS FRAMES ARE FOUND. C C SUM1=SUM2=SUM3=0.C 70 NC1=NC2=NC3=0 DO 23 J=1.MWORD SUM1=SUM1+Q1(J+K25) SUM2=SUM2+02(J.K25) SUM3=SU43+Q3(J+K25)

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75		NC1=NC1+NQ1(J+K25)
		NC2=NC2+NG2(J+K25)
		NC3=NC3+N43(J+K25)
	,	3 CONTINUE
	•	Q1 ( MW + K2 5) = SUM1/MWORD
80		92 ( WW + K2 5) = SUM2/MWORD
0.1		33 (MW, K25) = SUM3/MWORD
		NG1 (Md+K25) =NC1
		NQ2(MW,K25)='IC?
		N33(MW+K25)=11C3
85		5 CONTINUE
	c	THE 3 AVERAGE .K. FACTORS FOR EACH WORD OF EACH WORD GROUP FILE
	c	ARE PRINTED.
		PRINT 4
	•	G FORMAT(141)
9		PRINT BUONFILE
	4	■ FO4MAT(+ JORD SROUP NO. ++I <sup>e</sup> )
		PRINT 7°
	7	5 FORMATCHIN VARIALLE 3844 19.2 9.6 FM
		1 44)
95		DO 64 J=1+MJORD
		PRIVIT 69.J
	6	3 FORMATCH WORD NUMBER ++151
		PRINT 61+(G1(J+I)+I=1+F)
		PPINT 62+(02(J+I)+I=1+")
1		PEINT 63.(23(J.)).I=1.5)
		+6147 86. (4.1 (J.1).1=1.")
		PRINT OT CHG2CLATIALEIS
		PRINT FRONT SCUDIDOIE1053
		PR117 35
115		5 FORMAT(/)
		4 CONTINUE
		1 FORMAT(1CH K .5E10.4)
		2 FORMAT(10H KH +5E10+4)
110		6 FORMAT(10H NUMBER +5110)
		7 FORMAT(10H NUM3ERH +5110)
	8	8 FORMATCION NUMBERL +51103
		PRINT 35
		PRINT 35
115	C	THE 3 AVERAGE .K. FACTORS FOR EACH WORD GROUP FILE ARE PRINTED.
		PRINT 45
		5 FORMAT(* WORD GROUP AVERAGES *)
		PRINT 610(91(MW0I)0I=105)
		PRINT 62+(Q2(MW+I)+I=1+5)
120		PRINT $63 \circ (Q3(MW \circ I) \circ I = 1 \circ 5)$
		PRINT 86+(N91(MW+I)+I=1+5)
		PRINT 87+(NQ2(MJ+I)+I=1+5)
		PRINT 88+(NQ3(MW+1)+1=1+5)
	c	PROGRAM TERMINATION SEQUENCE.
125		PRINT 7
	S. A. S. S. S. S. S.	C FORMAT (1H1++ NORMAL TERMINATION +)
		END

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PROGRAM DISTSN(INPUT.OUTPUT.TAPE2=/135.TAPE3=/175) PROGRAM DISTSN COMPUTES THE LINEAR DISTANCE MEASURES LM1H. LM1L. LM2H. AND LM2L FOR EACH WORD AND THE WORD GROUP AVERAGE FOR EACH PAIR OF NOISE AND SIGNAL FILES READ IN. C C C DIMENSION E(2) . NW(51) . FNAME(5) DIMENSION FDAME(5),D(2),B(30,2) DIMENSION A(30,2).R(30).DM(4).NCT(2) DIMENSION G1(51,5),42(51,5),43(51,5),64(51,5) DIMENSION NQ1(51,5),NQ2(51,5) INITALIZE VARIABLES. +MWORD+ IS A COUNTING VARIABLE. C MUORD=0 THE LENGTH. IN ANALYSIS FRAMES. OF THE 50 WORDS IN THE WORD GROUP Being processed are read into the program. C 50 READ 10.N256 10 FORMAT(15) IF(N256 .EQ. 0) GO TO 25 WORD=MWORD+1 NMULT=N256 NW(MWORD)=NMULT GO TO 5. 21 CONTINUE MW=MWORD+1 C IFILE=5 READ 5. (FNAME(I).I=1.5) KEAD 5.(FDAME(I).1=1.5)
5 FORMAT(5A7) C DO :5 K25=1.5 LFN=LFORM(FNAME(K25)) LFN=LFORM(FDAME(#25)) C

RAVG2=(SUM/NMULT)/2.0

Q1(K30+K25)=DM(1) Q2(K30+K25)=DM(2) Q3(K30+K25)=DM(3)

Q4(K30,K25)=DM(4) NQ1(K30,K25)=NCT(1) NQ2(K30,K25)=NCT(2)

SUM1=SUM2=SUM3=SUM4=0.0

SUM4=SUM4+Q4(J+K25) NC1=NC1+NQ1(J+K25) NC2=NC2+NQ2(J+K25)

#### A TOTAL OF +IFILE+ NOISE AND SIGNAL DATA FILES ARE READ INTO THE PROGRAM. EACH OF THE PAIRS OF NOISE AND SIGNAL DATA FILES ARE PROCESSED. CALL PEMATCHISLTAPE3.LEN.0.0.0.0.1) CALL PFMATCH(SLTAPE2+LFN+0+0+0+0+1) LACH OF THE WORDS IN THE WORD GROUP FILES ARE PROCESSED. DO 3' K3C=1.Ma020 WHULT=Na (K30) SUM=C.O 00 15 K15=1.NMULT FRAME BY FRAME, THE NOISE AND SIGNAL LPC DATA IS READ INTO THE PROGRAM. READ(2.2.0) ALPHA.NFILE.NTAPE.NWORD.MULT.K60.21.22.0(1).D(2).Y1.Y2 PEAD (3.366) ALPHA.NFILE.WTAPE.NWORD.MULT.K66.X1.X2.E(1).E(2).A.F.O. 1A2.H0.DRG.ENROR.DERROR 30C FURMAT(A10. -15.1 -14.P) A(#15,1)=ALOG(E(1)) A(K15,2)=ALOG(E(2)) - (K15.1) = ALOG(D()) +(K15+2)=AL05(7(2)) R(#15)=5: SU"=SUM+F. 15 CONTINUE

HALF THE AVERAGE ENERGY OF THE WORD BEING PROCESSED IS COMPUTED.

SUBROUTINE ENERGY IS CALLED TO COMPUTE THE DISTANCE MEASURES LM1H, LM1L, LM2H, AND LM2L FOR THE WORD BEING PROCESSED. CALL ENERGY(NMULT;RAVG2;A,B;R,DM;NCT)

THE 4 AVERAGE DISTANCE MEASURES FOR THE WORD GROUP FILE ARE FOUND.

C

C

C

c

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CONTINUE

NC1=NC2=0 D0 24 J=1 +MWORD SUM1=SUM1+Q1(J+K25) SUM2=SUM2+Q2(J+K25) SUM3=SUM3+Q3(J+K25)

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75		24	CONTINUE	
	Star Lag		01 (MW+K25) = SUM1/MWORD	
			Q2(MW+K25)=SUM2/MWORD	
			Q3(MW+ K25)=SUM3/MWORD	
			Q4(MW,K25)=SUM4/MWORD	
80				
			NQ2 (MW+K 25) =NC2	
		25	NQ1 (MW+K25) =NC1	
	c		THE AVERAGE DISTANCE MEASURES FOR EACH WORD OF EACH WORD GROUP F	TIF
	c			1.1
85			ARE PRINTED. PRINT 4C	
		40	FORMAT(1H1)	
			PRINT BC.NFILE	
		80	PRINT BC.NFILE Format(+ word group NO. +.15)	
			PRINT 75	
90		75	FORMATIGOH VARIABLE 38.4 19.2 9.6 FM	
			AMI	
			DO 64 J=1.0MWORD PRINT 69.J	
		69	FORMAT(+ WORD NUMPER +.15)	
95			PRINT 61 (01(Jal) I=1.5)	
			PRINT 62+(G2(J+I)+I=1+5)	
			PRINT 63.(33(J.1).1=1.5)	
			PRINT 66+(G4(J+1)+1=1+5)	
			PHINT 86+(NG1(J+1)+1=1+5)	
100			PRINT 87.(N92(J.I),I=1.5)	
			PRINT 35	
			FORMAT(/)	
		64	CONTINUE	
			Frank Frank	
105		62	FORMATCI M LM14 5521.48 FORMATCI M LM24 5521.64) FORMATCI M LM14 552.5.49	
		63	FOR 44T(124 LM1L +FF.C.4)	
		66	FORMAT(10H L42L +5E10+4)	
		86	FORMAT(10H NUMBERH +5110)	
		87	FORMAT(10H NUMBERL +5110)	
110			PRINT 35	
			PRINT 35	
	c		THE AVERAGE DISTANCE MEASURES FOR EACH WORD GROUP FILE ARE PRINT	IED.
			PRINT 45	
		45	FORMAT(+ WORD GROUP AVERAGES +>	
115			PRINT 61.(Q1(MV.I).1=1.5)	
			PRINT 62. (Q2(MV.I).I=1.5)	
			PRINT 63. (Q3(MU, I).I=1.5)	
			PRINT 66+(Q4(MV+I)+1=1+5)	
			PRINT 86.(NQ1(MW.I).I=1.5)	
120			PRINT 87.(NQ2(MW.I).I=1.5)	
	c		PROGRAM TERMINATION SEQUENCE.	
			PRINT 70	
		70	FORMAT (1H1++ NORMAL TERMINATION +)	
			END	

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1 c	SUBROUTINE PRINTL (N. NUMBER) Subroutine printl prints all of the cross-correlation terms	
č	CALCULATED BY FITCORA FOR A PARTICULAR SET OF HIDPOINTS.	
	COMMON /FFT1 /A (+200) - 8 (+200) - 2041 (+2(3) - 2042 (+2)0) - C (500	
	10) Common /FFT2 /ICOR (2100) Nummalf = Number / 2 Nummatr = Number / 4	
	COMMON /FFT2 /ICOR (2100)	
	NUMALF = NUMAFA / 2 NUMATA = NUMAFA / 4	
	KUPLIM = 1 + NJMOTR / 16	
	KUPLIN = 1 + NJNGTR / 16 LINIT = NUMGTR - 1 C (LINIT + 1) = 0.	
	C (LIMIT + 1) = 0. Do 115 K25 = 1. LIMIT	
	DO 115 K25 = 1, LIMIT C (K25) = 0	
	C (K25 + NUMAL F + NUMATR) = 0	
	15 CONTINUE	
	DO 120 K30 = 1. KUPLIM	
	LCNTR = - 10 + (KUPLIM - K30) - 1 MCNTR = LCNTR - 9	
	MCNTR = LCNTR - 9 LL = NUMMALF + 1 + LCNTR	
	PRINT 1500+ LCNTR+ C (LL)+ C (LL - 1)+ C (LL - 2)+ C (LL - 3)+ C (	
	1LL - 4), C (LL - 5), C (LL - 6), C (LL - 7), C (LL - 8), C (LL - 9	
	2), "CNTR 20 CONTINUE	
	PRINT 1502+ C (NUMHALF + 1)	
	00 125 K4C = 1, KUPLIM	
	LCNTR = 16 + (K+3 - 1) + 1	
	MCNTR = LCNTR + 9	
	LL = NUMHALF + 1 + LCNTR LUP = LL + 9	
	PRINT 1500, LONTR, (C (1), I = LL, LUP), MONTR	
	25 CONTINUE	
	RETURN	
	500 FORMAT (15, 2%, 10F12.3, 2%, 15) 502 Format (* 0 *, 12%, F12.3)	
	END	
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	Called A Submitted Articles	
	and the Association of the Assoc	
	The second s	
	SUBROUTINE UNPACK (ID,IN,LL) Subroutine unpack takes a 60 bit word and seperates it into five 12 bit samples. This is done for the each 30 word record.	
	DIMENSION 10(400)+IN(80) LS=-5	
	00 60 I=1+LL	
	LT=IN(I)	
	DO 50 L=1+5 LTI=LT +AND+ 37778	
	LTJ=LT .AND. 40005	
	IFELTJ .NE. 0) LTI= .NOT. LTI	
	ID(LS)=LTI	
157. 19	•SHIFT(LT,+12)• SHIFTS THE BITS OF VARIABLE *LT+ TO THE RIGHT 12 PLACES. THIS IS PERFORMED IN ORDER TO OBTAIN THE FIVE 12 BIT	
	SAMPLES FROM EACH 6" BIT WORD.	
Sec. and	LTISHIFT(LT,-12)	
	LS=LS-1	
	50 CONTINUE	
	SO CONTINUE	
	RETURN	
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5		HODIFIED FROM.	DAVID LEWIS AND MARIE JEST, E Or inspired by the algol pro	CEDURE
	c	REVERSEBINARY.	BY R. C. SINGLE	TON. SRI.
10		DIMENSION A (16384), B	16384)	
	•	COMMON /FFTCC /M. JC (15	34 ST (15)	
		CALL EFTE	Was at the periods	
		TE CH		
15		N = 10 (N + 1)	Contain Barbara and States	
		NP = N + 1	MALE OF A CAREFUL A REAL PROPERTY	
		4 = 44 CALL FFTC IF (M .LE. 1) RETURN N = JC (4 + 1) NP = N + 1 K = 1		
		1 = 2		
		J = N - 1		
20	100	LC = M		
	105	K = K + JC (LC) JC (LC) = - JC (LC)		
		JC (LC) = - JC (LC)	and the state of a glat sure	
		IF (JC (LC) .LT. 0) 60 1	10 110	
		IF (LC +E9 +2) RETURN LC = LC - 1 GO TO 105		
25			the fly willing and willing the	
	110	GO TO 185 IF (K .LE. I .OR. J .LT.	KA 60 TO 115	
	110			
		A(I) = A(K)		
50		A (K) = T		
			aburan to 123 million	
		B (I) = R (K) B (K) = T	1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	
		IF (J .EQ. K) 60 TO 115	aparte a sparte canita	
55		KK = NP - K		
		T = A (KK)		
		A CKK) = A CJ)	CONTRACTOR CONTRACTOR	
		A (J) = T		
1		T = 3 ((K))		
0		3 (KK) = 8 (J)		
		B(J) = T		
	115	I = I + 1 J = J - 1		
		J = J - 1 60 TO 100		
5		END		

1		SUBROUTINE CFFTRC (A. B. 44, SCALE, N		
	c	DISCRETE COMPLEX FAST FOURIE Call C=FTRC(A+B+M+SC+NX)		
5	c	INPUT A(J) + I+B(J) IN REVER Output A(K) + I+B(K) IN NORM		
	C C	SEQUENCE LENGTH IS H = 2**4 SC IS REAL SCALING MULTIPLIE	R.	
	c	NX IS THE SIGN OF THE EXPONE Inner Loop Sines and Cosines		EFINITION.
10	c	RECURSIVELY BY SINGLETONS 2N INITIALIZED FROM A 24TA TABL	D DIFFERENCE ALGORITH	H•
	Ċ	WRITTEN BY L. DAVID LEWIS AN Modified from, or inspired b	D MARIE JEST, ESSA.	
15	č	REVERSEFOURIERC. DIMENSION A (16384), B (16384)	BY R. C. SINGLETON. S	RI.
13		COMMON /FFTCC /M. JD (15). 5 (15)		
		4 = MM Call FFTC		
20		N = JD (M + 1) K = N / 4	X 2 23 512	
		NQ = K VN = N - 1 ISBAN - 1		
		JSPAN = 1 SC = SCALE		
25		IF (ABS (SC - 1.) .LT. 1.E-17) GO TO DO 100 JC = 1. N		
		DO 100 JC = 1+ N A (JC) = SC + A (JC) B (JC) = SC + B (JC)		
30	160	CONTINUE IF (M .EQ. S) RETURN		
		DO 110 KK = 1. N. 2 KS = KK + 1		
		RF = A (KK) - A (KS)	Carefornia de	
35		A (KS) = RE FIM = B (KK) - 9 (KS) 3 (KK) = B (KK) + B (KS)		
	110	9 (KS) = FIM Continue		
40		IF (M .EQ. 1) RETURN EXPS = 1SIGN (1, NEXP)		
		VP = 1 D0 125 JB = 2, 4		
45		$\begin{array}{l} \text{D0 } 125 \ \text{JB} = 2 \cdot \text{M} \\ \text{SD} = - \text{S} \ \text{(JB} - 1) \\ \text{CD} = 2 \cdot \text{S} \ \text{(JB}) + \text{S} \ \text{(JB}) \end{array}$		
		R = - 2. * CD CN = 1.		
		C4 = 0. SN = 0.		
50		JJ = C KK = 1		
		SM = + EXPS JSPANH = JSPAN		
55	115	JS <sup>d</sup> an = Jspan + Jspān KS = KK + Jspan		
		RE = CN + A (KS) - SN + B (KS) FI4 = SN + A (KS) + CN + B (KS)		
		A (KS) = A (KK) - RE A (KK) = A (KK) + RE		
60		B (KS) = B (KK) - FIM 3 (KK) = B (KK) + FIM		
		KK = KK + JSPAN4 KS = KS + JSPAN4		
65		FIN = SM + A (KS) + CM + B (KS) RE = CM + A (KS) - SM + B (KS)		-N
Van	71	(ARR) I A (RK) JRE		MABLE COPY
ITUJ	11	B (KS) = B (KK) - FIM B (KK) = B (KK) + FIM	•	. of E Co.
70		KK = KS + JSPAN4 IF (KK .LT. N) 30 TO 115		1 ADLL
		KK = KK - NN JJ = JJ + K	-NIA	In.
		1F (JJ .SE. NQ) GO TO 120	ord Mar.	
			RE2.	
		60	-	

75	CD = R + CN + CD
	CN = CD + CN
	SM = CN + EXPS
	SD = R + CM + SD
	C4 = SD + C4
80	SN = - CM + EXPS
	GO TO 115
120	0 K = K / 2
125	5 CONTINUE
	RETURN
85	END

....

......

1		SUBROUTINE RTRAN2T (A. B. MM. INV)
		CALL RTRANZT (A,B,M,INV)
	2	IF(INV.GT.O) UNSCRAMBLE THE TRANSFORMS OF TWO REAL SEQUENCES. IF(INV.LT.O) SCRAMBLE THE TRANSFORMS OF TWO REAL SEQUENCES.
5	-	INPUT AND OUTPUT ARE IN NORMAL SEQJENCE.
•	č	SEE WRITEUP FOR DETAILS.
	č	SEQUENCE LENGTH IS N = 2++M
		JRITTEN BY L. DAVID LEWIS AND MARIE JEST. ESSA.
	ĉ	NODIFIED FROM. OR INSPIRED BY THE ALGOL PROCEDURE
10	c	REALTRAN, BY R. C. SINGLETON, SRI.
-	Carrier Carlos Maria	DIMENSION & (16385), 3 (15386)
		COMMON /FFTCC /4, JC (15), ST (15)
		4 = 4M
		CALL FFTC
15		N = JC (4 + 1)
		NH = N / 2
	•	IF (INV .LT. 0) GO TO 120
		IF (M .LT. 2) GO TO 105
		K = N
20		DO 1:0 J = 2, N4
		A (K + 1) = B (J) + B (K)
		3 (K + 1) = A (K) - A (J)
		A(J) = A(J) + A(K) B(J) = B(J) - B(K)
25	100	3(J) = 9(J) - B(K)
23	105	A (1) = 2 + A (1)
	105	A(YH + 2) = 2 + + B(1)
		B (1) = B (NH + 2) = 0.
		IF (4 .= 2. 0) RETURN
30		A (NH + 1) = 2 + A (NH + 1)
		A(N+2) = 2 + 9(NH+1)
		B(NH + 1) = B(N + 2) = 0.
		IF (M .EQ. 2) RETURN
		K = N + 1
35		5 + HV = L
	110	NS = NH / 2 - 1
		D0 115 L = 1+ NS
		T = A (J)
		A(J) = A(K)
40		A(K) = Y
		T = B (J) B (J) = B (K)
		BEST AVAILABLE COPY
45	115	DLJ AVAILADLL COLI
		35TURN
	120	3 (1) = A (N4 + 2)
		IF (4 .12. 0) RETURN
		3 (VH + 1) = A (N + 2)
50		IF (* .ED. 1) RETJRN
		K = NH + 2
		30 125 J = 2+ NH
		A(K) = A(J) + P(K + 1)

	8	(K)	=		(K	+ 1	>	-	8	())
		(J)	=	4	(J)	-	R	**	+	1)
	3	(1)	=	B	(J)	•	A	**		1)
125	K :	= K	+	1						
	IF		•E	0.	2)	RE	TL	JRN		
		= N								
	1	- N+		2						
	GO	TO	11	0						
	EN	D	-							

Y900 102412-14

	SUBROUTINE FFTC
C	COMMON SUBROUTINE FOR FFT SUBROUTINES.
c	JC IS POWERS OF THO ARRAY. JC(M)=2++(M-1)
ç	ST IS SINE ARRAY ST(M)=SIN(PI/(24))
c	M IS TESTED FOR PROPER INPUT RANGE, J.LE.M.LE.14.
	COMMON /FFTCC /4, JC (15), ST (15)
	DATA (JC = 1, 2, 4, 3, 16, 32, 64, 128, 256, 512, 1024, 2043, 4096
	1. 8192. 16384)
	DATA (ST = 1.0000000000CE+000, 7.07106781187E-001, 3.82683432365E-
	1001. 1.95090322016E-001, 9.80171403295E-002. 4.90675743274E-002. 2
	2.45412285229E-002. 1.22715382857E-002. 6.13588464915E-003. 3.06795
	3676297E-003+ 1+33398018629E-003+ 7+66991318743E-004+ 3+83495187571
	42-004+ 1-917475973112-004+ 9-587379909602-005)
	IF (4 .LT. 0 .OR. M .GT. 14) PRINT 1530, 7
	RETURN
1500	FDRMAT (+CILLESAL VALUE FOR M+ M =++ I4) END

1		SUEROUTINE REALTRA (A. B. MM. NE. INV)	
	C	CALL REALTRAV(A+B+M+NE+INV)	
	Ċ	IF(INV.GT.3) UNSCRAMBLE THE TRANSFORM OF A REAL SEQUEN	CF.
	c	IF(INV.LT.C) SCRAMBLE THE TRANSFORM OF A PEAL SEQUENCE	
5	č	INPUT AND OUTPUT ARE IN NORMAL SEQUENCE.	•
	č	SEE WRITEUP FOR DETAILS.	
	C	SEQUENCE LENGTH IS N = 2++M	
	c	NE MUST AGREE WITH SIGN OF EXPONENT IN TRANSFORM DEFIN	ITION.
	c	INNER LOOP SINES AND COSINES COMPUTED	
10	C	RECURSIVELY BY SINGLETONS 2ND DIFFERENCE ALGORITHM.	
	c	INITIALIZED FROM A DATA TABLE.	
	c	ARITTEN BY L. DAVID LEWIS AND MARIE JEST, ESSA.	
	C	MODIFIED FROM, OR INSPIRED BY THE ALGOL PROCEDURE	
	c	REALTRAN, BY R. C. SINGLETON, SRI.	
15		DIMENSION & (16385) . B (16385)	
		COMMON /FFTCC /4, JC (15), ST (15)	
		4 = 44	
		CALL FFTC	
		N = K = JC (M + 1)	
20		VH = N/2	
		NK = NH + 1	
		CN = ISIGN (1, INV)	
		SN = ISIGN (1. VE)	
		A(NK) = 2 + A(NK)	
25		$B(NK) = 2 \cdot CN \cdot SN \cdot B(NK)$	
		IF (CN .GE. 0.) GO TO 10	
		FIM = A (1) - A (N + 1)	
		A(1) = A(1) + A(1) + 1	
		B (1) = FIM	
30		60 TO 1^5	
	100	$A(N + 1) = 2 \cdot (A(1) - P(1))$	
		A(1) = 2 + (A(1) + B(1))	
		B(1) = B(N + 1) = 0.	
	105	IF (M .Eg. D) RETURN	
35		SD = CN + SN + ST (M)	
		R = 2. * ST (M + 1)	
		R = - R + R	
		CD =5 + CN + 3	
		SN = 1.	
40		00 110 J = 2. NH	
		CD = R + CN + CD	
		CN = CD + CN	
		SD = R + SN + SD	
		SV = SD + SN	
45		$\Delta A = A (J) + A (K)$	
40			
		BA = B(J) + B(K)	
		33 = 8 (J) - 8 (K)	
		RE = CN + BA + SN + AB	
50		FIM = SN + BA - CN + AR	
		3 (K) = F1M - 33	
		3 (J) = FI4 + B3	
		A(K) = AA - PE	
		A (J) = AA + RE	
55	110	K = K - 1	
		RETURN	
		END	

1			SUBROUTINE CFFTS (A. B. MM. SCALE. NEXP)
	C C		DISCRETE COMPLEX FAST FOURIER TRANSFORM. Call CFFTS(A,B,M,SC,NX)
	č		INPUT A(J) + I+B(J) IN NORMAL SEQUENCE.
5	C		OUTPUT A(K) + I+B(K) IN REVERSE BINARY SEQUENCE.
	c		SEQUENCE LENGTH IS N = 2 ···M
	č		SC IS REAL SCALING MULTIPLIER. NX IS THE SIGN OF THE EXPONENT IN THE TRANSFORM DEFINITION.
	ć		INNER LOOP SINES AND COSINES COMPUTED
10	c		RECURSIVELY BY SINGLETONS 2ND DIFFERENCE ALGORITHM.
	c		INITIALIZED FROM A DATA TABLE. URITTEN BY L. DAVID LEWIS AND MARIE JEST, ESSA.
	c		MODIFIED FROM, OR INSPIRED BY THE ALGOL PROCEDURE
	C		FASTFOURIERS, BY R. C. SINGLETON. SRI.
15			DIMENSION & (16394), B (16384) Common /FFTCC /4, JC (15), SNT (15)
			MA = M = NM
			CALL FFTC
26			N = JC (M + 1) NH = N / 2
			N3 = NH / 2
			SC = SCALE
			IF (M •£3• 0) 30 TO 125 IF (M •£3• 1) 50 TO 115
25			EXPS = ISIGN (1, NEXP)
			NN = N - 1
			K = 1 D0 110 JA = 2, 4
			CE = SNT (MA)
30			AA = AA - 1
			CD = 2. + CE + CE SD = - SNT (MA)
			R = -2. * CD
			CN = 1.
35			CM = 0. SV = 7.
			JJ = 1
			KK = 1
4.9			SM = + EXPS JSPAN = N4
			NH = JSPAN / 2
	1	סיו	
			RE = A (KK) - A (KS) A (KK) = A (KK) + A (KS)
45			FIM = B (KK) - 3 (KS)
			B(KK) = B(KK) + B(KS)
			A (KS) = CN + RE - SN + FIM B (KS) = SN + RE + CN + FIM
			KK = KK + NH
50			KS = KS + NH
			RE = A (KK) - A (KS) $A (KK) = A (KK) + A (KS)$
			FIM = 8 (KK) - 3 (KS)
55			B (KK) = B (KK) + B (KS)
			A (KS) = CM + RE - SM + FIM B (KS) = SM + RE + CM + FIM
			KK = KS + NH
			IF (KK .LT. N) 30 TO 100 KK = KK - NN
60			
			1F (JJ .GE. NO) GO TO 115
			CD = R + CN + CD $CN = CD + CN$
			SD = R + CH + SD
65		1	
		1	SN = - CN + EXPS
		.1 -	60 TO 100
		105	K = K · K
70		10	DO 120 KK = 1, N, 2
			RE = A (KK) - A (KS)
			A (KK) = (A (KK) + A (KS))

SUBROUTINE PRECOEF (NPC) Subroutine precoef computes the 12 predictor coefficients, the 12 Reflection coefficients, and the minimum source error using the 1 C C LEVINSON ALGORITHM AND THE AUTOCORRELATION TERMS FROM SUBROUTINE AUTOCOR. SUBROUTINE STABLE IS USED TO INSURE A STABLE FILTER IS GENERATED. THE PREDICTOR COEFFICIENTS ARE STORED IN PCOEF(J+12), C 5 C c J=1+12+ THE REFLECTION COEFFICIENTS ARE STORED IN PCOEFF(J+J)+ J=1+12+ AND THE MINIMUM SQUARED ERROR IS THE VARIABLE D. С COMMON A(256).ACOR(13).PCOEF(12,12).GAM(13).D 10 C=-ACOR(2) D=1.0-C+C PCOEF(1,1)=-C DO 10 J=2.NPC J1=J-1 15 SUM=1.0 D0 15 I=1+J1 15 SUM=SUM+ACOR(J+1-I)+PCOEF(I+J1) G=ACOR(J+1)-SUM C=-6/3 20 IF (ABS(C) .GT. 0.99) CALL STABLE (J.C) D=D+(1.0-C+C) 00 20 1=1.J1 20 PCDEF(I,J)=PCOEF(I,J1)+C+PCOEF(J-I,J1) 10 PCDEF(J,J)=-C . 25 RETURN

FIM = B (KK) - 3 (KS) 3 (KK) = (B (KK) + 8 (KS)) B (KS) = FIM

120 CONTINUE 125 IF (ABS (SC - 1.) .LT. :.E-10) GO TO 135 DO 130 JB = 1. V. A (JB) = SC + A (JB) B (JB) = SC + B (JB)

75

80

85

A (KS) = RE

CONTINUE

CONTINUE

RETURN

END

END

120

130

135

1	C C C	SUBROUTINE REORDER (A, B, NM) Call Reorder(A,B,M) Reversible Permutation of Real Seduence From First-Last-Normal Sequence	
5	C C C C	TO ODD-EVEN-REVERSE BINARY SEQUENCE. SEQUENCE LENGTH IS N = 200M. WRITTEN BY L. DAVID LEWIS AND MARIE JEST. ESSA. Reorder. Dy R. C. Singleton. Sri. Dimension A (16384). 5 (15384). LST (15)	
10		COMMON /FFTCC /4, LC (15), ST (15) M = MM	
		CALL FFTC 1F (M .EQ. 0) RETURN JA = M + 1	
15		JB = 4, - 1 1 = KB = 0 KU = LC (JA) - 1 30 100 KA = 1, KU, 2 T = A (KA + 1)	
20	100	A (KA + 1) = B (KA) B (KA) = T CONTINUE	
		IF (M .EQ. 1) RETURN LIM = M / 2 + 1	
25	105	KS = LC (JB + 1) + KB KU = KS JJ = LC (JA - J3) KK = KB + JJ K = KK + JJ	
	1010 101000.21	KK = KB + JJ	
	110	$\kappa = \kappa \kappa + JJ$	
30		Y = A (KK + 1) A (KK & 1) = A (KS + 1)	
		T = R (KK + 1)	
		B (KK + 1) = B (KS + 1)	
35		9 (KS + 1) = T KK = KK + 1	
		15 (КК «LT» К) 30 ТО 115 КК = КК + ЈЈ	
40		KS = KS + JJ IF (KK .LT. KU) GO TO 110 IF (J9 .LE. LIM) GO TO 120	
		12 - 19 - 1	
45	120	50 TO 105 If (I +LE+ U) RETURN	
		$J_{3} = LST(1)$	
50		I = I - 1 KB = KS SO TO 105 END	

SMEAN=SUM/NUMBER DO 15 I=1.NUMBER 15 A(I)=A(I)-SMEAN SUM=0.0 DO 20 I=1.NUMBER 20 SUM=SUM+A(I)+A(I) ACOR(1)=SUM D0 25 J=2.N SUM=0.0 VUM=NUMBER-J+1 DO 30 I=1+NUM 30 SUM=SUM+A(I)+A(I+J-1) 25 ACOR(J)=SUM/ACOR(1) RETURN END SUBROUTINE STABLE (J.C) Subroutine stable forces the absolute value of each reflection cc COEFFICIENT TO BE LESS THAN OR EQUAL TO 0.99. THEREBY INSURING A C STABLE FILTER. CC=-C PRINT 10.J.CC 10 FORMAT (\* REFLECTION COEFFICIENT \*+12+\* IS UNSTABLE, = ++F10-5) IF (C +6T+ 0+99) C=0+99 IF (C +LT+ -0+99) C=-0+99 RETURN END SUBROUTINE GAMMA (NPC) Subroutine Gamma computes the predictor coefficient autocorrelation terms from the predictor coefficients. CC COMMON A(256), ACOR(13), PCOEF(12,12), GAM(13), ERROR DIMENSION PC(13) N=NPC+1 PC(1)=1.0 00 10 I=1.NPC 10 PC(I+1)=-PC0EF(I.12) SUM=0.0 DO 15 I=1+N 15 SUM=SUM+PC(I)+PC(I) GAM(1)=SUM DO 20 1=1.NPC SUM=0.0 BEST AVAILABLE COPY NN=N-I 00 25 J=1.NN 25 SUM=SUM+PC(J)+PC(J+I) SAM(I+1)=2.0+SU4 20 RETURN END

SUBROUTINE AUTOCOR (NUMBER, NPC) SUBROUTINE AUTOCOR COMPUTES THE ENERGY TERM AND THE 12 NORMALIZED AUTOCORRELATION TERMS FOR EACH 256 SAMPLE WINDOW. THESE VALUES ARE STORED IN ACOR(J), J=1,13 RESPECTFULLY. COMMON A(256), ACOR(13), PCOEF(12,12), GAM(13), ERROR

C C C

> N=NPC+1 SU#=0.0

DO 10 I=1.NUMBER 13 SUM=SUM+A(I)

15

1

5

19

20

1

5

10

1

5

10

15

20

SUBROUTINE HILOW(NMULTORAYG20AOROSODMONCT) SUBROUTINE HILOW COMPUTES THE AVERAGE \*K\* FACTOR FOR THE HIGH AND LOW ENERGY FRAMES OF THE WORD BEING FROCESSED. DIMENSION A(30)or(30)os(30)odm(2)onct(2) SUMI=SUM2=000 1 C. 5 NC1=NC2=0 DO 11 K10=1.WMULT IF (R(K10) .LT. RAV62) 60 TO 15 SUM1=SUM1+A(K10)/S(K10) 10 NC1=NC1+1 GO TO 2: 15 CONTINUE SUM2=SUM2+A(K10)/S(K10) NC2=NC2+1 20 CONTINUE 10 CONTINUE NCT(1)=NC1 NCT(2)=NC2 15 DM(1)=SUM1/NC1 DM(2)=SUM2/NC2 RETURN 20 END

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YGOD ATRADAVA TZAS

SUBROUTINE ENERGY(NMULT+RAV62+A+B+R+DM+NCT) SUBROUTINE ENERGY COMPUTES THE AVERAGE OF THE DISTANCE MEASURES LM1H+ LM1L+ LM2H+ AND LM2L FOR THE WORD BEING PROCESSED. DIMENSION A(30+2)+B(30+2)+R(30)+DM(4)+NCT(2) C C RLIM1=0.82 RLIM2=2.46 SUM1=SUM2=SUM3=SUM4=0.0 NC1=NC2=0 DO 1C K10=1.NMULT C THE RESIDUAL DISTANCE MEASURES ARE LINEARIZED FOR EACH FRAME. 00 5 K5=1+2 IF (B(K10.K5) .LE. RLIM2) 60 TO 25 RM2=B(K10+K5) RM1=(RM2-RLIM2+1.0)+RLIM1 GO TO 30 25 RM1=RLIM1 RM2=RLIM2 30 AA=1.0/(RM1-RM2) BB=-AA+RM2 IF(A(K10,K5) .LT. AM1) GO TO 35 IF(A(K10,K5) .GT. RM2) GO TO 40 A(K10,K5)=AA+A(K10,K5)+BB 60 TO 45 35 A(K10.K5)=1.0 GO TO 45 40 A(K10+K5)=C.0 45 CONTINUE 5 CONTINUE THE HIGH AND LOW ENERGY FRAMES ARE FOUND AND THE LINEAR DISTANCE MEASURES LMIH, LMIL, LM2H, AND LM2L ARE COMPUTED. IF(R(K1G) .LT. RAVG2) 60 TO 15 SUM1=SUM1+A(K1G,1) C C SUM2=SUM2+A(K10+2) NC1=NC1+1 GO TO 21 15 CONTINUE SUM3=SUM3+A (K1C+1) SUM4=SUM4+A(K16+2) NC2=NC2+1 20 CONTINUE 1-CONTINUE THE 4 AVERAGE DISTANCE MEASURES FOR THE WORD BEING PROCESSED ARE COMPUTED. С NCT(1)=NC1 NCT(2)=NC2 IF(NC1 .E3. 0) NC1=1 IF(NC2 .EG. )) NC2=1 DM(1)=SUM1/NC1 OM(2)=SUM2/NC1 DM(3)=SUM3/NC2 DM(4)=SUM4/NC2 PETURN END

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