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Effect of CW-396A Radome on the Radiation Pattern of Rectangular Antennas

RONALD L FANTE PETER R. FRANCHI RICHARD L. TAYLOR

RADC-TR-76-397 IN-HOUSE REPORT DECEMBER 1976



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. 1

Effect of CW-396A Radome on the Radiation Pattern of Rectangular Antennas

1. ANALYSIS

In this report we will present some results showing the effect of a CW-396A radome on the radiation pattern of a rectangular aperture. It is well known that in the absence of the radome the radiated power can be written as

$$P_{v} = A (1 - \sin^{2} \theta \sin^{2} \phi) \left| \iint_{S} \frac{K_{s}}{-s} e^{ik(x \sin \theta \cos \phi + y \sin \theta \sin \phi)} dx dy \right|^{2}, \quad (1)$$

where A is a constant, \underline{K}_{s} is the electric or magnetic surface current on the radiator, S is the radiator surface area, and θ and ϕ are defined in Figure 1.

Now suppose this antenna is placed inside a CW-396A radome. It has been demonstrated elsewhere that for this type of radome only the radome ribs significantly influence the radiation pattern, and the effect of the radome panels and bolts can be neglected (at least, for frequencies up to and including S-band). For purposes of calculation we have found it convenient to assume that the near fields of the antenna are still collimated as they pass through the radome. Furthermore, we assume that over the intersection of this collimated beam and the radome, the

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Figure 1. Antenna Geometry

radome surface is nearly flat so that it can be approximated by a rib structure which is periodic in x and y, as shown in Figure 2.

In order to calculate the effect of this approximate radome structure on the radiation field we employ the equivalent current method. Using this method we calculate the field which is scattered by each radome rib and then approximate that rib by a current sheet which produces the same radiation field. This current sheet is then projected back onto the antenna so that the net antenna current is the original surface current K_g plus the surface currents K_R due to all the radome ribs that are projected back onto the antenna. Therefore, in place of Eq. (1) we get

$$P_{R} = A' (1 - \sin^{2} \theta \sin^{2} \phi) \bigg| \iint_{S} K_{s} e^{ik(x \sin \theta \cos \phi + y \sin \theta \sin \phi)} dx dy$$

$$+ \iint_{S} K_{R} e^{ik(x \sin \theta \cos \phi + y \sin \theta \sin \phi + h \cos \theta)} dx dy \bigg|,$$
(2)



Figure 2. Approximate Planar Radome Geometry: for CW-396A we have δ = 117.88" and $\theta_{\rm R}$ = 32°

where the equivalent current K_R is nonzero only over those portions of the surface S upon which a radome rib is projected and the factor exp (ikh cos θ) accounts for the phase shift between the actual location of the radome ribs and their projection onto the antenna. The quantity h is the distance between the antenna and the radome surface. Also, A' can be determined from the requirement that

$$\int_{0}^{2\pi} d\phi \int_{0}^{\pi} \sin \theta \, d\theta \, P_{v}(\theta, \phi) = \int_{0}^{2\pi} d\phi \int_{0}^{\pi} \sin \theta \, d\theta \, P_{R}(\theta, \phi) \ .$$

The equivalent currents K_R have been calculated previously for both L- and S-band signals in Eq. (18) and Tables 1 to 4 of report RADC-TR-76-379. For the horizontal struts

$$\frac{\underline{K}_{\mathbf{R}}}{|\underline{K}_{\mathbf{S}}|} = \hat{\mathbf{x}} |\mathbf{K}_{\parallel}|, \qquad (3)$$

whereas for the diagonal struts

$$\frac{\underline{K}_{R}}{|\underline{K}_{s}|} = \hat{x} [0.53 K_{\parallel} + 0.85 K_{\perp}] , \qquad (4)$$

where K_{\parallel} and K_{\perp} are given by Eqs. (2) and (22), respectively, in RADC-TR-76-379, and \hat{x} is a unit vector along the x-axis (note that a horizontally polarized E field gives a magnetic current $\hat{z} \times \underline{E}$ which is directed along the vertical axis).

We have not yet specified what we shall use for the antenna current distribution \underline{K}_s ; we choose

$$K_{s} = \hat{x} \cos\left(\frac{\pi x}{2x_{o}}\right) \cos\left(\frac{\pi y}{2y_{o}}\right) \exp\left\{-ik(x\sin\theta_{B}\cos\phi_{B} + y\sin\theta_{B}\sin\phi_{B})\right\}.$$
 (5)

The current has a cosine taper in both x and y planes and a phase tilt such as to give a beam maximum at $\theta = \theta_B$ and $\phi = \phi_B$.

2. **RESULTS**

If we assume that the antenna lies at the center of the CW-396A radome it is appropriate to choose h = 324 in. Also, the separation $\delta/2$ between horizontal ribs is 58.94 in. and the angle $\theta_{\rm R}$ defined in Figure 2 is 32° . For the case when

there is no beam displacement (that is, $\theta_B = \phi_B = 0$) we have calculated the effect of the radome on the antenna pattern for the cases listed in Table 1.

Case	Frequency (GHz)	2x _o (Inches)	² y _o (Inches)	ϕ (Degrees)	See Figure
	1.35	360	180	0	3
	1.35	360	180	32	4
A	1.35	360	180	58	5
	1.35	360	180	90	6
	3. 15	240	144	0	7
-	3.15	240	144	32	8
в	3.15	240	144	58	9
	3.15	240	144	90	10
	3.30	312	168	0	11
С	3.30	312	168	58	12
	3.30	312	168	90	13

Table 1. Effect of Radome on Antenna Pattern

In plotting P_R in Figures 3 to 13 we have normalized P_R so that $P_R(\theta = 0) = P_V(\theta = 0)$. This of course ignores the power loss in the main beam caused by power being scattered by the radome into the sidelobes. By using the conservation equation

$$\int_{0}^{2\pi} d\phi \int_{0}^{2\pi} \sin \theta \, d\theta \, [P_{R}(\theta, \phi) - P_{v}(\theta, \phi)] = 0$$

And a state of the state of the

it is possible to estimate the loss in gain caused by the radome, as shown in Table 2.

Case	Approximate Gain Loss (dB)
А	0.1
в	1.0
С	1.3

Table 2. Estimated Loss in Gain

It is interesting to note from Figures 3, 7, and 11 that the grating lobes caused by the horizontal radome ribs are about 30 dB below the main beam level at L-band and about 23 to 24 dB below at S-band. The diagonal radome ribs produce grating lobes at $\phi = 58^{\circ}$, as is clear from Figures 5, 9, and 12. These are about 35 dB below the main beam level at L-band and about 26 to 30 dB below at S-band. There are no appreciable grating lobes for any other values of ϕ in the range $0 \le \phi \le 90^{\circ}$ (of course there are grating lobes in the $\phi = 122^{\circ}$, 180° , 238° , and 302° planes), as is clear from Figures 4, 6, 8, 10, and 13.

We have also studied the radiation pattern in the $\phi = 0^{\circ}$ plane for the case when there is a 5° main beam tilt in the vertical plane (that is, $\theta_{\rm B} = 5^{\circ}$, $\phi_{\rm B} = 0$). These results are shown in Figures 14 to 16.

3. CONCLUSIONS AND DISCUSSION

The CW-396A radome can produce grating lobes in the vertical plane which are about 30 dB below the main beam at L-band and about 23 to 24 dB below the main beam at S-band. The radome produces about a 0.1 dB loss in gain at L-band and about a 1 dB loss in gain at S-band.

Finally, we should note that all the panel sizes on the CW-396A radome are not exactly the same, as we have assumed in our approximate model. These slight size differences would tend to broaden and lower the grating lobes by a small amount (about a dB). We have also considered the effect of radome curvature. This effect is shown in Figures 17 to 19. Upon comparing Figures 17 to 19 with Figures 3, 7, and 11 we see that the effect of the radome curvature is small.



Figure 3a. Vacuum Radiation Pattern for Case A, $\phi = 0^{\circ}$, $\theta_{B} = \phi_{B} = 0^{\circ}$



Figure 3b. Radiation Pattern with Flat Radome for Case A, $\phi = 0^{\circ}$, $\theta_{\rm B} = \phi_{\rm B} = 0^{\circ}$









Figure 5a. Vacuum Radiation Pattern for Case A, $\phi = 58^{\circ}$, $\theta_{\rm B} = \phi_{\rm B} = 0^{\circ}$



Figure 5b. Radiation Pattern with Flat Radome for Case A, $\phi = 58^{\circ}$, $\theta_{\rm B} = \phi_{\rm B} = 0^{\circ}$



Figure 6a. Vacuum Radiation Pattern for Case A, $\phi = 90^{\circ}$, $\theta_{\rm B} = \phi_{\rm B} = 0^{\circ}$







Figure 7a. Vacuum Radiation Pattern for Case B, $\phi = 0^{\circ}$, $\theta_{\rm B} = \phi_{\rm B} = 0^{\circ}$



Figure 7b. Radiation Pattern with Flat Radome for Case B, $\phi = 0^{\circ}$, $\theta_{\rm B} = \phi_{\rm B} = 0^{\circ}$



Figure 8a. Vacuum Radiation Pattern for Case B, $\phi = 32^{\circ}$, $\theta_{\rm B} = \phi_{\rm B} = 0^{\circ}$





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Figure 9a. Vacuum Radiation Pattern for Case B, $\phi = 58^{\circ}$, $\theta_{\rm B} = \phi_{\rm B} = 0^{\circ}$







Figure 10a. Vacuum Radiation Pattern for Case B, $\phi = 90^{\circ}$, $\theta_{\rm B} = \phi_{\rm B} = 0^{\circ}$







Figure 11a. Vacuum Radiation Pattern for Case C, $\phi = 0^{\circ}$, $\theta_{\rm B} = \phi_{\rm B} = 0^{\circ}$







Figure 12a. Vacuum Radiation Pattern for Case C, $\phi = 58^{\circ}$, $\theta_{\rm B} = \phi_{\rm B} = 0^{\circ}$















Figure 14b. Radiation Pattern with Flat Radome for Case A, $\theta_{\rm B} = 5^{\circ}$, $\phi = 0^{\circ}$, $\phi_{\rm B} = 0^{\circ}$











Figure 16a. Vacuum Radiation Pattern for Case C, $\theta_{\rm B} = 5^{\rm o}$, $\phi = 0^{\rm o}$, $\phi_{\rm B} = 0^{\rm o}$









Figure 17. Radiation Pattern with Spherical Radome for Case A, $\phi = \theta_B = \phi_B = 0^0$



Figure 18. Radiation Pattern with Spherical Radome for Case B, $\phi = \theta_{B} = \phi_{B} = 0^{\circ}$



Figure 19. Radiation Pattern with Spherical Radome for Case C, $\phi = \theta_B = \phi_B = 0^0$

Appendix A

Fortran Listing of the Computer Program Used for Calculating Radome Effects

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C	PF1	TWICE THE SEPARATION BETWEEN HORIZONTAL ETDS (THOUSE
	×n	1/2 THE AFERTIRE VERTICAL DIMENSION (THOMES)
C	YC	1/2 THE DEFETLEE FORTZONTAL STMENSTON (TACHES)
		WAVENUMEEE (RECTOGORAL TARWES)
c	FTHETA	1/2 THE ANGLE PETKEEN DIACONAL ETES AT A HEDTTCHTAL
		THE THE ANGLE THETH STREOMPL FILS PT P HERIZENTEL
c	TF9	FEFRIENCY BANK - 1 FOR L-FAND 2 FOR S- FAND
	TTYPE	- + FOR DI ANAR. 2 SCP SPUESTCAL
c	н	DISTANCE FOR ANTENNA TO FATCHE (INCHES)
	THETAL	PEGTNNING VALUE OF THETA IDECESS
c	THETAL	ENDING VALUE OF THETA (DEGREES) 1
Ċ	PH1	ATTMUTHAL ANGLE (CECREES)
-0-		PEAM
С	THETAE	TILT (DEGREES)
C		
-0		
	NAREAY(3)	= NARRAY(2) = 1
	CALL SYST	EMC (115. NARPAY)
	DEC2FAT =	PT / 180.0
	- EPS = 4.1	
	REAT XINF	ידשי
	CALL SSHT	CH(3, MSW)
	IF (NSW .F	C. 2) PRINT XINFUT
	AFCTIN =	X9 / H
	IFITYPE	.EA. 2) ARCSIN = ASIN(ARCSIN)
	NN = INTE	APRSINTH/REL)
	ENN = FLO	AT (NN)
	SINPHI -	STY (PHI*REG2RAR)
	COSENT -	COS(PHI®DEG2RAD)
	0000111 -	CONTRACTOR OF CONTRACT

PRCCRAM MATH (INFUT, OUTPUT)
REAL K, LAMBDA1, LAMBDA2, LAMBDA3, LAMBDA4
 CIMENSION MARRAY(6), AMPL(14,2), ANGLE(14,2), OP(7)
CIMENSICH WW(3), TT(2), LU(3), SS(3)
DATA (AMFL = 0.9199, 0.9578, 1.0398, 1.1651, 1.2929, 1.3873.
 1 1.3049, 1.3334, 0.2782, 0.2755, 0.2765, 0.2805, 1.2867, 0.3452
2 1.0364, 9.6293, 1.6219, 1.0056, 1.2736, 1.5761, 2.1450, 0.3625.
 3 0.3302, 9.3468, 0.3400, 0.3229, 0.3336, 0.4166)
PATA (ANGLE = -95.0, -102.8, -110.1, -114.1, -114.7, -113.3.
 1 -110.29 -04.59 -09.29 -09.49 -92.99 -94.28 -95.38 -99.08
2 -83.1, -132.5, -146.7, -136.3, -145.5, -180.4, -176.5, -84.1.
 3 -95.19 -96.99 -97.59 -101.29 -106.39 -114.69
NAMFLIST/YINFUT/D, CEL, XJ, YJ, K, RTHETA, IFE, ITYPE, H, THETAL,
 1 THETAU, DELTH, PHI, PHT8, THETAB

29

-

STATHE = STATHETAR+CEG2RAD)
C1 = K*SINTHB*COSCHIB
C2 = K*SINTHR*SINPHIP
FR = F. 4286
$CF = n_{*}F*FR*Kt$
F102X0 = F1 / (2.0*X0)
PIO2VC = PT / (2.6*Y0)
HC = C / 2.0
CALL SSNTCH(4, MSN)
IF(MSN .ER. 2) PPINT 607, PHI
00 = (0.0, 0.0)
AFGE = AMFL(N+7, IFR)*COS(#NGLE(N+7, IFP)*DEG2RAC)
APGI = AMPL(N+7, TFRI*SIN(ANGL=(N+7, IFB)*0-G2RAU)
AC = CA + CAPLX (ARGR, ARGI)
AFEF = BWFL(N, IFS) GUS (ANGLE(N, IFS) DEGZKAD)
ARGI = AMFLICK, IFM) *SINCANGER (N, JFM) *LEGZMACJ
$-10 \text{ Gr(p)} = (\text{GPE} \times 10^{\circ} \text{ gr}) + 10^{\circ} \text{ gr}$
C_{0} C_{0
FET TA = AMAY14 (XY) - YFCOSETH/SINRTH) = 0.0)
13 = INT(OFLTA/OFL)
L1 = -L4
Wk(2) = -Y0 / SIARTH + Wk(2) = -Y0 / STARTH
TT(2) = YO / SINRTH $TT(3) = YO / SINRTH$
$\frac{UU(2) - 40}{SINRTH} SUU(3) = 40 / SINRTH$
SS(1) = YO / SINPTH $SS(2) = YO / SINFTH$
C REGIN THETA LCOP
C(S) = C(S) (T + T A)
$O^{KH} = (O_* O_* O_* O_*)$
30 OKH = OF(N) +CFXP (CMPLX(0.C, -CF+FLOAT(2+N-1)+ (1.0-COSTH))) + QKH
0KH = (EPS - 1.0) * CF * 9-1
0K = 0.53+0KH + 0.85+0KV
ALPHA1 = K*SINTH*COSPPI + FIC2X0 = C1
ALFFA2 = K#SINTH#COSFFI - FTC2X0 - C1
PETA1 = K*SINTH*SINPFI + \$79240 = C2
BETA2 = K*SINTH*SINFHI - TTO2YD - C2
Nh2 = 2 ⁷ NN
NZ = -NNZ
SI = (CSTENNALPHAITTEL) - GUST TENNTUSJIALPHAITUEL) //(1.0 - C
A CONCINCTAL THAP LELY - COST LENNING FULLAR LELY FILLO -
1 10510, 5.4EPHAC.VEL/ /

		$01 = (MPLX(S1, 0.0) \\ SC2 = CMPLX(S2, 0.0) \\ C2 = CMPLX(S2, 0.0) \\ C2 = CMPLX(S2, 0.0) \\ C3 = CMPLX(S1, 0.0) \\ C4 = CMPLX(S1, 0.0)$
-		SEGIN N2 LOOPI
-	-35	ARC = # * SIN(8,5*FLOAT(N2)*CFL/H)
		Q1 = CEXP(CMPLX(0.0, ALPHA1*ARG)) + 01
-		02 = CEXP(CMPLX(0.0, ALPHA (* ARG)) + C2
		N2 = N2 + 1
		IF (N2 .LE. NN?) GO TC 35
_	- 78-	TEEMI = STN(RETAINYA) /RETA/ + STN(RETAPHVA) /RETAP
	50	QEH = (C1 + SIN(HD+ALEHA1) / A) EHA1 + 02 + SIN(HD+ALEHA2) / ALEHA2) +
_		TERMI + OKH
		PU = (SIN(ALPHA1*XC)/ALPHA1 + SIN(ALPHA2*XC)/ALPHA2) * TERM1
-		GATMA1 - ALPHA1*SINRTH - PETA1*00SPTH
		PHC1 = ALFHA1*SINFTH + EETA1*CCSRTH
		CAMMA2 = ALPHA1*SINGTH - CETA2*COSETH
		PHC2 = ALPHA1*SINFTH + 'ETA2*COSRTH
-		GAMMA3 = ALPHA2*SINRTH - PETA1*COSRTH
		PHC3 = ALFHA2*SINRTH + ETA1*COSRTH
		RFU4 = ALPHA2*SINFIF + ETA2*COSRIF
		AMARAZ = ALCHALICCORTH + CETASINEIN
		XT2 = -ALPHA1+CCSETH + FTA2+SINETH
_		
		XI3 = -ALPHA2*COSFTH + "FTA1*SINFTH
-		LAMPDAL = ALPHA2*COSPTH + "ETA2*SINRTH
		XI4 = -ALFHA2*CCSRTH + "ETA2*SINRTH
-		N = -kN
		0FM = (C, C, 0, 0)
-		OFF = (0.0, 0.0)
		REGIN N LOOO
	40	
		$\frac{1}{1} + \frac{1}{1} + \frac{1}$
		TE (N .GE. 12 . AND. N .IT. 13) GO TO 50
_	_	TFIN -6F- 11 -AND- N -17- 121 - 60 TP 20
	50	
		TT(1) = (X0 - XN) / COSPTH
		UU(1) = -TT(1)
-		66 TC F0
	60	J = 2
-		6C 10 Pt
	70	J = 7
-		Wh (3) = - (XA + XN) / COSQTH
		SS(3) = -WW(3)
	90	DE1 - CEXP(TPP1X(0.0, ALPH) TXN) }
		$ne2 = CFXP(CMPLX(0.C, ALPH^*XN))$
-		s = sstu

	U = UU(J)
	QPS = (CMPLX(0.0,(SIN(HD*GAMMA1)/(GAMMA1*LAMBDA1)))*GF(LAMBDA1,T,W)
	1+0#FLX(0.0,(SIN(HD*GAPPA2)/(GAMPA2*LAPBCA2)))*OF(LAPECA2,T,W))*QE1
:	2+(CMFLX(0.0,(SIN(HD*GAMMA3)/(GAMMA3*LAMBCA3)))*QF(LAMBCA3,T,W)
	\$+C*FLX(%-9,(SIN(40*GAMMA4)/(GAMMA4*LAMPDA4)))*QF(LAMEDA4,1,W))*QE2 OFF = OFP - OFS
	QPM = -GF1*(CMPLX(0.0,(SIN(H0*RH01)/(RHC1*XI1)))*OF(XI1,S,U) +
1.1.1.1	CMPLX (0, 0, (SIN (H0* RH02) / (FHC2* XI2)) +0 F(XI2, S, U))
	-GE2*(CMPLX(0,0,(SIN(HC*RH03)/(RHC3*XI3)))*GF(XI3.S.U) +
	CMPLX(0,0,(SIN(HC*RH04)/(RH(4*XI4)))*0F(XI4.S.U))+0FM
	N = N + 1
	IF (N .LE. NI) GO TO 40:
c	
C	ENF CE N LOOPI
	OFF = CMPLX(0.5, 0.0) + CK + OPP
	GFW = CMPLX(0.5, 0.0) * GK * GPM
	TEFM2 = 1.0 - (SINTH*SINPHI)**2
	IF(TERM2 .LT. 1.0E-20) TERM2 = 1.0E-20
	PV = TFRM2 * P0**2
	$PV = 1^{\circ} \cdot C + ALOG10(FV)$
	PF = TERM2*CABS(CMPLX(P0, 0.0) + CEXP(CMPLX(0.0, K*H*COSTH))*
	(0PH + 0PH + 0PP)) **2
	FR = 10.0 + ALOG10 (FP)
	PRINT 610. THETAD. PV. PR
	THETAD = THETAD + DELTH:
	IF (THETAD .LE. THETAU) GO TO 20
C	
	END OF THETA LOOP
600	FCFMAT(1H1.T15.6HPHT = .F6.2/1H0.T5.5HTHFTA.124.2HPV.1F4.2HPV.
610	FORMAT (FIG. 3. 192518. 7)
010	

MISSION of Rome Air Develor (C^3) activities, and in the C^3 areas of information sciences and intelligence. The principal technical mission areas are communications, electromagnetic guidance and control, surveillance of ground and aerospace objects, intelligence data collection and handling, information system technology, ionospheric propagation, solid state sciences, microwave physics and electronic reliability, maintainability and competibility.



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