# HUMAN PILOT DYNAMIC RESPONSE IN SINGLE-LOOP SYSTEMS WITH COMPENSATORY AND PURSUIT DISPLAYS

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# SYSTEMS TECHNOLOGY, INC.

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R. J. Wasicko D. T. McRuer R. E. Magdaleno

ERRATA

The data plots and controlled element designations on p. 29 should be interchanged with that on p. 52.

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#### FOREWORD

This report documents an analytical and experimental investigation of human pilot dynamics accomplished under Contract AF 33(657)-10835, BPS No. 5(6399-8219-62405364), sponsored by the Flight Control Division of the Air Force Flight Dynamics Laboratory. The research was performed by Systems Technology, Inc., at both its Hawthorne, California, and Princeton, New Jersey, offices, and, under subcontract, by The Franklin Institute Laboratories, Philadelphia, Pennsylvania. The project principal investigators were D. T. McRuer and D. Graham, of STI, and E. S. Krendel, of FIL. The Flight Control Division project engineers were Capt. J. E. Pruner and P. E. Pietrzak.

Many others besides the authors have contributed to the results reported here. All of the principal investigators participated in the detailed planning phases. An important contribution was made by William C. Reisener, Jr., of The Franklin Institute, who executed the experiments and data reduction phases. The authors would also like to thank their co-workers M. M. Solow, who was the subject, and Diane Fackenthal, who assisted in the data reduction phase; and A. V. Phatak, of STI, who assisted in pre-experiment calculations and predictions. Finally, the report has been significantly improved by the incorporation of many suggestions due to the careful review by R. O. Anderson and P. E. Pietrzak of FDCC. This manuscript was released by the authors in December 1966 for publication as an RTD Technical Report.

This technical report has been reviewed and is approved.

B. WESTBROOK

Chief, Control Criteria Branch Flight Control Division AF Flight Dynamics Laboratory

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#### ABSTRACT

The primary purpose of the experimental series reported here is to investigate, on a preliminary and exploratory basis, human operator performance differences between pursuit and compensatory displays. For each display type a wide range of forcing function bandwidths and controlled element dynamics was used. The effect of the additional information provided by separately displaying both forcing function and controlled element output (pursuit) rather than their difference (compensatory) was evaluated using the mean-squared error and a quantity called the "effective open-loop describing function" (Y<sub>B</sub>).

As a prelude to the new data, past pursuit/compensatory tracking results are reviewed, and then a tie-in is made between these and the current series.

iii

# TABLE OF CONTENTS

PAGE

I.	INTROD	UCTION.	• •	•	•	•••	•	•	•	•	•	٠	•	٠	٠	1
	Α.	Backgrou	nd ar	nd Mo	otiv	atio	n.	•	•	٠	•	•	٠	•	•	1
	В.	General I	Plan	•	•	•••	•	•	٠	•	•	•	•	•	•	4
	с.	Outline of	of th	e Re	epor	t.	•	•	٠	•	•	٠	٠	•	•	5
II.	SUMMAR	Y OF TRACI	CING	DAT	A, P	URSU	IT A	ND	COM	PEN	SAT	ORY				
	DT2LPU	I COMPARIS	SOINE	٠	•	• •	•	-	•	•	•	•	٠	•	٠	6
III.	BLOCK	DIAGRAM ST	RUCI	URE	, PUI	RSUI	r de	TEC	TIO	N T	ECH	NIQ	UES			
	AND EX	PERIMENTAI	CON	FIGU	JRAT:	ION.	٠	٠	0	•	٠	•	٠	٠	٠	11
	Α.	Block Dia	agrem	St	ructi	ure.	•	•	•	•	•	•	•	•	•	11
	Β.	Detection	n of	Pura	suit	Beha	avio	r.	0	•	•	•	٠	•	•	15
	С.	Experimen	ntal	Situ	atio	on .	•	•	•	•	•	•	•	•	•	17
IV.	EXPERI	MENTAL DAT	CA.	•	•	••	٠	•	•	•	•	•	•	٠	•	24
	Α.	Compensat	ory	Date	a Cor	npari	son	s.	•	٠	•	•	•	•	•	26
	в.	Compariso of Other	on of Expe	Per rime	form ents	ance	e Me °	asu	res	wit	th I	Resi	ilt:	5	•	35
	с.	Compariso	on of	Pur	suit	: and	l Co	mpe	nsat	tor	r Pe	erfo	orma	ance	e	
		Measures.	•	•	• •	• •	٠	•	•	•	•	٠	•	•	•	37
	D.	Compariso	on of	Pur	suit	; and	l Co	mpei	nsat	cory	r De	esci	ib	ing		1
		Functions	•	٠	• •	• •	٠	•	•	٠	•	•	•	•	٠	47
v.	DATA IN	TERPRETAT	ION	¢	• •	• •	•	•	•	•	•	•	•	•	•	57
VI.	GENERAI	CONCLUSI	ons	•	• •	•	•	٠	•	•	•	•	•	•	•	63
REFERE	NCES.		•	•	• •	• •	•	•	•	•	•	•	•	•	•	65
												-	-	-	-	-

The second second second second

# ILLUSTRATIONS

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ΞŽ

<u>تت</u>ر

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FIGURE		PAGE
1.	Compensatory and Pursuit Manual Control Systems	2
2.	Compensatory and Pursuit Displays	3
3.	General Block Diagram of the Human Operator in a Pursuit Display Tracking Task	12
4.	General Measurements and Task Variables	19
5.	Measured Forcing Function Power Spectra Magnitude	21
6.	Pursuit Display	22
7.	Stick Manipulator	23
8.	Comparison of Compensatory Display Data with Population of Pilots; $\Upsilon_c = K_c$ , $\omega_i = 2.5$ .	27
9.	Comparison of Compensatory Display Data with Population of Pilots; $Y_c = K_c/s$	28
10.	Comparison of Compensatory Display Data with Population of Pilots; $Y_c = K_c/s^2$	29
11.	Comparison of Compensatory Display Data with Population of Pilots; $Y_c = K_c/s(s - \lambda) \cdot $	30
12.	Long Term Variability of the Current Subject's Describing Function; $Y_c = K_c$ , R14 Input	31
13.	Long Term Variability of the Current Subject's Describing Function; $Y_c = K_c/s(s-1.5)$ , $\omega_1 = 1.5$ .	32
14.	Comparison of Current $Y_C = K_C$ Data with Elkind's Data .	34
15.	Performance Measure Comparisons with Elkind's Results, $Y_c = K_c \cdot \cdot$	36
16.	Relative Performance Measure Comparisons with kefs. 1, 2, and 9; $Y_c = K_c$ , $K_c/s$ , and $K_c/s^2$ .	38
17.	Performance and Control Deflection Data: $Y_c = K_c$	39
18.	Performance and Control Deflection Data; $Y_c = K_c/s^2$	40
19.	Performance and Control Deflection Data; $Y_c = K_c/s(s-\lambda)$	41

and the second second

FIGURE

;

١

÷

E

20.	Performance and Control Deflection Data; $Y_c = K_c/s$	÷	٠	42
21.	Performance and Control Deflection Data; $Y_c = K_c(s+0.25)/(s+5)^2$	•	•	43
22.	Comparison of Pursuit and Compensatory Display Describing Function; $Y_c = K_c \cdot \cdot$	•	•	48-49
23.	Comparison of Pursuit and Compensatory Display Describing Function; $Y_c = K_c/s^2 \dots \dots$	•	·	50-51
24.	Comparison of Pursuit and Compensatory Display Describing Function; $Y_c = K_c/s(s - \lambda) \cdot \cdot \cdot \cdot \cdot$	•	•	52
25.	Comparison of Pursuit and Compensatory Display Describing Function; $Y_c = K_c/s$ .	•	•	54 <b>-</b> 55
26.	Comparison of Pursuit and Compensatory Display Describing Function; $Y_c = K_c(s + 0.25)/(s + 5)^2$ .	٠	•	56
27.	Single-Axis Pursuit Tracking	•	•	57
58°	Implied Human Operator Feedforward Characteristics, $Y_c = K_c \cdot \cdot$	٠	•	60
29.	Implied Human Operator Feedforward Characteristics, $Y_c = K_c/s^2$ .	•	•	61
30.	Implied Human Operator Feedforward Characteristics, $Y_c = K_c/\epsilon(s - \lambda), \omega_i = 1.5$	•	ta ta	62

# TABLES

TABLE		PAGE
I.	Summary of Human Operator Tracking Deta: Pursuit and Compensatory Display Comparsons	'710
II.	Possible Human Operator Response Structures for Pursuit Displays	13
IUI.	Run Sequence (Three Replications per Configuration)	25
IV.	Summary of Performance Measure Differences	45

. . .

# SYMBOLS

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B5	Forcing function designation defined on page 18					
c,c(t)	Human operator output					
ci	Portion of human operator output that is linearly correlated. with forcing function					
<sup>c</sup> n	Portion of human operator output that is not linearly correlated with forcing function					
C	Compensatory display					
C	Fourier transform of c(t)					
d(t)	Second input defined on page 16					
e(t)	Operator stimulus; error					
e <sub>i</sub>	Portion of error that is linearly correlated with forcing function					
en	Portion of error that is not linearly correlated with forcing function					
E	Fourier transform of e(t)					
Fs	Stick force gradient					
i(t)	System forcing function					
I	Fourier transform of i(t)					
ງ່ພ	Imaginary part of the complex variable, $s = \sigma \pm j\omega$					
К <sub>с</sub>	Controlled element gain					
К <sub>в</sub>	Control sensitivity — inches (display)/stick motion					
m(t)	System output					
М	Fourier transform of m(t)					
n	$T_R \omega_n / 2\pi$ defined on page 20					
n <sub>c</sub> (t)	Remnant injected at pilot's output					
N	Number of runs					

Nc	Fourier transform of n <sub>c</sub>
P	Pursuit display
R.24	Ref. 3 forcing function designation defined on page $41$
R.4	Ref. 3 forcing function designation defined on page $41$
<b>R.6</b> 4	Ref. 3 forcing function designation defined on page $41$ .
R14	Forcing function designation defined on page 18
ន	Laplace transform variable, $s = \sigma \pm j\omega$
τ̈́	Time
$\mathbf{T}_{\mathbf{R}}$	Run length
¥c	Controlled element
Y <sub>pc</sub>	Human operator describing function with compensatory display
Y <sub>Pe</sub>	Human operator describing function operating on $e(t)$ (Pursuit display)
Ypi	Human operator describing function operating on i(t) (Pursuit display)
r <sup>*</sup> pi	Implied $Y_{p_i}$ on assumption that $Y_{p_e} = Y_{p_c}$ and $Y_{p_m} = 0$
Ypm	Human operator describing function operating on m(t) (Pursuit display)
¥β	Effective open-loop describing function
λ	Controlled element parameter
σ <sub>e</sub>	rms value of the error
σi	rms value of the forcing function
$\phi_{i}$	Peak amplitude of an input sine wave
<b>\$</b> dc	Cross spectral density between d and c
$\Phi_{ ext{de}}$	Cross spectral density between d and e
¢ <sub>ee</sub>	Error spectral density
$\Phi_{\texttt{ic}}$	Cross power spectral density between i and c
₽ie	Cross power spectral density between i and e

₽ii	Forcing function power spectral density
Φ <sub>im</sub>	Cross spectral density letween i and m
Ф <sub>лат</sub>	System output power spectral density
$\Phi_{n_cc}$	Cross spectral density between $n_c$ and c
$\Phi_{n_ce}$	Cross spectral density between n <sub>c</sub> and e
<sup>∲</sup> n <sub>c</sub> m	Cross spectral density between n <sub>c</sub> and m
<sup>Φ</sup> nn	Closed-loop remnant spectral density, at pilot's output
•ncnc	Power spectral density of n <sub>c</sub>
ധ	Angular frequency, rad/sec
ω <sub>c</sub>	System crossover frequency, i.e., frequency at which $ Y_pY_c  = 1$
ω <sub>cβ</sub>	Crossover frequency of $Y_{\beta}$
പ്	Forcing function bandwidth
ω <sub>n</sub>	Frequency of the "nth" sinusoidal component of the forcing function
<b>:</b>	Approximately equal to
<b>ዻ</b>	Angle of
đb	Decibels; 10 log <sub>10</sub> if a power quantity, e.g., spectrum; 20 log <sub>10</sub> if an amplitude quantity, e.g., Y <sub>p</sub>
	Magnitude
lab	Magnitude in db
(¯)	Mean value ensemble average
$\mathcal{L}^{-1}$	Inverse Laplace transform

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#### CHAPPER I

#### INTRODUCTION

#### A. BACKGROUND AND MOTIVATION

The usefulness of a control engineering approach to the study and design of manual vehicular control systems has grown rapidly in the past decade and is now well established. This approach requires models of pilot dynamic characteristics which or we applied in conjunction with the formal methods of control engineering. The models are based on experimental measurements of dynamic ci acteristics exhibited by pilots in a variety of control situations. A great many past experiments in which human dynamic measurements were taken have considered so-called compensatory conditions, i.e., those in which the operator's actions are based solely on an error indicating the difference between system command and system output (Fig. 1a). In contrast, the more complex pursuit situation, i.e., that in which the system forcing function and output are displayed to the operator and he can utilize both, and/or their difference (system error) as his basis for control action, has received very little attention (Fig. 1b). In fact, the only experiments in which human operator dynamics have been obtained are those of Ref. 3. While this experimental series comprehensively covered effects of forcing function variations on operator dynamics, the controlled element dynamics were made the simplest possible (a pure gain,  $Y_c = 1$ ) and were held fixed. A variety of other experiments in which performance measures only we e taken (Refs. 1, 2, 9) has demonstrated the overwhelming importance of controlled element characteristics as task variables. However, the nature of the dynamic characteristics adopted by the operator which underlie the performance differences noted with different controlled elements has been unknown. In the absence of an experimental data base, the models of pilot dynamics in pursuit situations have been only conjectures. The experimental series described here was undertaken primerily to provide such data. Then, as these and other new data enhance our quantitative understanding of pursuit display systems, the next step is the evolution of a mathematical pilot model which is suitable for predictive purposes.



a) Functional Block Diagram for Compensatory Behavior



b) Functional Block Diagram for Pursuit Behavior

Figure 1. Compensatory and Pursuit Manual Control Systems

The compensatory and pursuit manual control systems indicated in Fig. 1 are intended to represent different behavioral situations. In compensatory operation only the system error acts as a stimulus for operator action, whereas in pursuit the system forcing function and output are separately observable. These two different situations are ordinarily defined in terms of pursuit and compensatory displays (Fig. 2) which actually present the appropriate system signals as visual stimuli. However, it is important to recognize that presentation of the signals does not necessarily imply pilot action thereon; for instance, in a pursuit display the operator may act only on the error, thereby performing in a compensatory fashion in spite of the presence of the forcing function and output. Conversely, under certain conditions with a compensatory display (e.g., a predictable forcing function) the operator



#### Compensatory Display

Pursuit Display

Figure 2. Compensatory and Pursuit Displays

can mentally separate input from output in the displayed error signal. Then by using a reasonable facsimile of the system forcing function and the error as information inputs the operator may function in a pursuit fashion. With random-appearing inputs and complex controlled element dynamics, the usual way to induce compensatory or pursuit behavior is by virtue of the display. This always works in the compensatory situation, although, as indicated above, the provision of a pursuit display does not guarantee pursuit operation by the operator.

There are many practical reasons for our interest in obtaining a quantitative understanding of the human operator response in pursuit situations. These include, but are not limited to, the following:

• In tasks where the external reference is present (e.g., VFR approach and landing), a pursuit model may be an appropriate representation for the human's operation.

- For director (as opposed to null steering) displays in IFR operations, the type of behavior desired and the appropriate pilot model correspond to those of the pursuit situation.
- As hypothesized by the Successive Organizations of Perception (SOP) theory (Ref. 6), a pursuit mode of response by the operator is an interim phase in the development of exceptional skill. In this hypothesis, pursuit behavior will occur both in the progression toward higher skills and in the regression to lower levels (compensatory) under stress.
- For many situations, tracking performance (rms error) with a pursuit display is superior to that for a compensatory display. Determination of a simple, usable model for operator response behavior with pursuit-like visual presentation of information (from the external field of view or displays) is required for system synthesis activities for which improved performance is desired.

#### B. GENERAL PLAN

The intent of the experiments reported here was to explore the nature of differences between pursuit and compensatory behavior for a wide cross section of forcing function characteristics and controlled element dynamics. To accomplish this the same forcing function and controlled element task variables were used for two different display configurations. The human pilot dynamics were characterized by describing function measurements, and the average system performance by appropriate performance measures. Differences in dynamic behavior between the two situations were detected by comparing (1) the mean-squared errors and (2) an effective open-loop describing function for the pursuit display with the actual open-loop describing function exhibited with the compensatory display. The latter provided a very sensitive measure of dynamic changes induced by the display differences. The remnant differences between the two display situations were found by examining the two components of the mean-squared error, 1.e., that due to pilot/system dynamics operating on the forcing function and that due to the effects of pilot remnant operating within the closed-loop system.

#### C. OUTLINE OF THE REPORT

The preceding introduction indicates that the purpose of this experimental series is to study the means used by the pilot to enhance performance in pursuit systems over that in compensatory systems. A quantitative understanding is desired, preferably to the extent that a mathematical model suitable for predictive purposes is obtained.

Chapter II reviews and summarizes past pursuit/compensatory studies for later comparison with the results of the present experiments. Consideration is restricted to those previous studies that utilized randomappearing inputs which allow direct comparison with our data.

Chapter III discusses the system relationships that result from the pilot's utilization of the additional information provided by the pursuit display. Of major importance is the section "Detection of Pursuit Behavior," where it is shown that performance measures alone are inadequate. Supplementing these with an effective open-loop describing function is shown to provide sufficient information to detect pursuit behavior. This chapter concludes with a description of the experimental configuration.

Chapters II and III lead directly to the desired experimental plan given in Chapter IV. This is followed by the performance measure and describing function data, with extensive comparisons of the present subject's compensatory display data with a large population of pilots to indicate that he is a representative sample. In addition, his data are tied in with those for yet another population for which pursuit data were also taken. Finally, the present pursuit and compensatory display data are compared using both performance measures and the effective open-loop describing function.

Chapter V is devoted to data interpretation. Implications are drawn as to a plausible description of the pilot's utilization of the pursuit display.

Finally, Chapter VI summarizes the general conclusions and findings of the study.

#### CHAPTER II

# SUMMARY OF TRACKING DATA, FURSUIT AND COMPENSATORY DISPLAY COMPARISONS

There have been many studies in which tracking performance for pursuit and compensatory systems has been compared. Initially, perhaps, these were motivated by a desire to demonstrate a clear-cut superiority of one display type over the other. Because the pursuit display provides more information, it was presumed that some advantage would thereby accrue. Indeed, average tracking performance with pursuit systems is often, perhaps even usually, better than that with a compensatory system; but this is by no means the rule. This is attested to by the summary cf past pursuit/compensatory comparisons presented in Table I. This is one of our two key starting points. The other, which is a notable omiscion from the table, is the work of Elkind which will be covered at much greater length as tie-in data for the present series in Chapter IV.

The experiments summarized in Table I can be divided into two categories, corresponding to simple and complex inputs. Only the latter, comprising inputs made up of a minimum of three sinusoids, are of further interest in connection with our data. Of these, the ones of primary interest are the two papers by Chernikoff, et al (Refs. 1, 2), and the comprehensive paper of Obermayer, et al (Ref. 9). Since these consider some of the same controlled elements as examined here, actual comparisons between performance measure data from these sources and those obtained in the present series are given later (Chapter IV).

In a sense the data in this chapter is of limited usefulness, since only performance measures can be compared and as shown in Chapter III this is an insufficient indication of the effects of pursuit versus compensatory displays. In addition, Refs. 1, 2, and 9 used sine waves with harmonic relationships. It is possible that this produced a recognizable pattern which the pilot could utilize to reduce tracking errors below that expected for random appearing inputs.

TABLE I

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SUMMARY OF HUMAN OPERATOR TRACKING DATA: PURSUIT AND COMPENSATORY DISPLAY COMPARISONS

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	Results	Statistical analysis of data for last four ses- sions of the 10 daily session experiment (6 sub- jects, 6 trials per condition per day) showed: 1. For the unaided Y <sub>C</sub> , pursuit tracking perform- ance was superior to compensatory performance 2. For the aided Y <sub>C</sub> , pursuit and compensatory tracking performance vere not significantly different 3. For compensatory display, tracking perform- moded Y <sub>C</sub> was superior to that with unaided Y <sub>C</sub> was superior to that with unaided Y <sub>C</sub> was vorse than that' with waided Y <sub>C</sub> 6. The unsided Y <sub>C</sub> was vorse than that' with worlded Y <sub>C</sub> 7. The unsided Y <sub>C</sub> was vorse than that' with worlded Y <sub>C</sub> 7. The unsided Y <sub>C</sub> was vorse than that' with worlded Y <sub>C</sub>	Amalysis of data for last four sessious of the 18 daily seation experiment (18 subjects, 6 for each forcing function, 3 trials per condition per day) showed: For the medium and fast forcing functions, pursuit tracking performance vas superior to compensatory performance for all three $Y_c$ 's 2. For the slow forcing function, pursuit track- ing performance, compared to comperisory ing performance, was superior with the position $Y_c$ but was worse with the rate and aided rate $Y_c$ 's 3. The best performance was obtained with a compensatory display, aided rate $Y_c$ , and slow 4. The worst performance was obtained with a forcing function
	Performance Measure	$\int_{15}^{60}  e  \ \text{at for a} \\ 1 \ \text{min. run}$	J <sup>60</sup> [e] dt for a 15 1 min run
EXPERIMENTAL CONDITIONS	Controlled Element	Unaided: $Y_c = K_c$ Aided: $Y_c = K_c \left[ 1 + \frac{2(0.576)}{0.576} = + \left( \frac{8}{0.576} \right)^2 \right]$ $K_c = \frac{0.01}{deg} \frac{110.}{(manipulator)}$	Position: $Y_c \sim K_c$ $K_c = \frac{2 \text{ in. (contpulator)}}{\text{in. (contpulator)}}$ Rate: $Y_c = K_c/s$ $K_c = \frac{4 \text{ in. per sec (display)}}{\text{in. (manipulator)}}$ Added rate: $Y_c = K_c(1 + 0.5s)/s$ $K_c = \frac{4 \text{ in. per sec (display)}}{\text{in. (manipulator)}}$
	Manipulator	6 in. dis. 41sk hand- wheel, 3/4 in. dis. knob at rin, located 10 in. to right of CRT centerline at subject's elbow height	24 in. long flexible steel bar joystick, located 10 in. to right of CRT centerline and 2 in. above subject's arm rest, deflected laternlly, spring gradient of 0.5 lb/in.
	Dîsplay	5 in. GRU, 12 in. from subject, 1/16 in. dia. dot for input or fixed reference, 1/4 in. vertical line for out- put or error, horizon- tal displacements	5 in. CRU, 12 in. from subject, 1/16 in. dia. dot for input or fixed reference, 1/4 in. vertical line for out, put or error, horizon- tal displacements
	Forcing Function	Cum of three sinusoids, $\alpha = 0.628, 1.05,$ 1.57 md/sec Relative sumplitudes, $2^{n:10:1}$ Maximum deflection, 4 in.	The three-sinusoid frequency content of the three forcing functions was: a. (mad/sec) Slow 0.28. 0.467, 0.70 Med. 0.70, 1.17, 1.75 Reat 1.12, 1.86, 2.79 Relative amplitudes, 2.5:1.5:1 MatTmin deflection, 4 in.
	AUTHORS	Chernikoff, Birningham, and Taylor (Ref. 1)	Chernikoff and Thylor (Ref. 2)

TABLE I

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				EXPERIMENTAL CONDITIONS		
AUTHORS	Forcing Function	D1 splay	Manipulator	Controlled Element	Performance Measure	Results
Senderr and Cruren (Ref. 11)	Sum of two sinusoids, frequencies not speci- fied	CRT, 14 in. from sub- Ject's eyes, with a 10-to-the-inch grid, a dot for input or fixed reference, circle for output or error, circle size corro- sponded to scoring (TOT) area, horizontal distlacement	J-1/14 in. fluted kmob, rotating in a plane parallel to the display surface, located below and to the right (left for the one left-banded aubject) of the CKT center	Position: Y <sub>c</sub> = K <sub>c</sub> K <sub>c</sub> = <u>0.0171 in. (display)</u> K <sub>c</sub> = deg (manipulator)	Time-on-target (TOT) for a 1 min. run	Amelysis of data (5 subjects, 5 trials per condi- tion per day for 5 days) for compensatory, com- bined (25, 50, and 75 percent pursuit) and pure pursuit displays showed that performance laproves significantly when the amount of the pursuit component is increased from zero (companent or display) to 75 percent, but does not continue to change as the amount of the pursuit component is increased from 75 to 100 percent (pure pursuit).
Pulton (Ref. 10)	<pre>Simple harmonic input: Single ainusoids of 1.77, 3.14, 4.71, and 6.28 hd/acc complex harmonic input: Sum of constitutions Jower frequency com- powent same as above, higher frequency com- powent vith one-half the and 1.57 times the frequency</pre>	Exposed portions of a mechanical pulley/cable system. For pursuit, input and output vere triangule " pointers; for compensatory, rhombus pointer moving over a horizontal fixed line. Vertical dis- placements for both displays.	Wheel with attached horizontal handle, upward movements rais- ing corresponding display pointer	Position: Y <sub>C</sub> = K <sub>2</sub>	$\int_0^{20}$  e  dt for 0.5 zin. runs	Relative integrated absolute error data for Z trialo (latin square experimental data with 16 subjects) showed: 1. For all inputs, the pursuit display perform- ance was superior to that for the compensa- tory display 2. With a pursuit display, the performance was better for the simple input compared to the complex harmonic input complex harmonic input o significant difference in the relative per- formance for both types of inputs at all speeds
Walaton and Warren (Ref. 12)	A stryle simusoid of 3.14 rad/sec and a ran- den junction with Gaumitan arglitude distribution, spectrul density pack at 3.14 rad/sec and one- half pover points at 1.57 and 4.71 rad/sec harinum arglitudes, 1/8, 1/6, 1/2, 1, and 2 in.	GRT, 5/8 1n. long and 1/32 1n. wide vertical line for input or fixed reference, vertical line of same size for output or error, output line below input with slight overlap, hori- zontal displacements	17 in. lever, low inertim and friction, motion in a horizontal plane with operator resting his elbow on the pivot end of the lever	<pre>Position, Ye = Ke Five Ke's for eaci input ampli- tude. For imax = 2 in., Ke = 0.025, 0.05, 0.1, 0.2, and 0.4 in. (display) per deg (manipulator) Gains decreased proportional to decrease in imax</pre>	$\frac{1}{60}\int_{15}^{75} e^2 dt \text{ for a}$ Average error, time-on- target, and target hits for several target to bands measured but not used	For individual subjects, mean square error per- formance with the single-sinusoid input is con- sistently better with the pursuit display than with the compensatory display for all combina- tions of input amplitude and controlled element gain. This is not true for the randow input, although avenaged data for four subjects show the pursuit display performance consistently better than that with the compensatory display.

rabre r (Continuea)

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	Results	<ol> <li>Average data for six subjects, 10 trials per condition showed:</li> <li>There is no statistically significant dif- ference bettern purnult and compensatory display for three cases: the si ' input vith mite and acceleration control</li> <li>The pursuit display performance var signifi- cantly vorse than inst with the compensatory display for three cases: the slow input with rate and acceleration control and the fast input with acceleration control and the fast input vith acceleration control and the fast input vith position and rate con- trol; the vorst is the sense, and the input, pursuit display and rate con- trol; the vorst performance with the fast input, pursuit display and rate con- trol; the vorst performance with the fast input, pursuit display and controlled element control</li> <li>For all display and controlled element control</li> <li>For all forcing function speed</li> <li>For all forcing function display condime- tiony display, performance deterion to display to control</li> </ol>
	Performance Measure	$\frac{1}{T} \int_0^T e^{dt}, \frac{1}{T} \int_0^T  e  dt,$ $\frac{1}{T} \int_0^T e^2 dt,  time-on-target (TOT) with the tolercnee band of it/10 in., and hite (number of times  e  <1/10 in.),$ T = (0.4) 3  sec
EQTERIMENTAL CONDITIONS	Controlled Element	Position: $Y_c = K_c$ $K_c = \frac{0.5 \text{ in. (display)}}{1.6 \text{ (manipulator)}}$ Bate: $Y_c = K_c/a$ $K_c = \frac{1 \text{ in. (manipulator)}}{1.6 \text{ (manipulator)}}$ $K_c = \frac{4 \text{ in. per sec}^2 (\text{display})}{1.6 \text{ (manipulator)}}$
	Manipulator	22 in. long center- positioned joystick, deflected fore and aft, maximum deflection6 i6 in. at top, breakout forre of 0.5 lb, spring gradient of 1 lb/in., i1/8 in. deadband in circuitry circuitry
	Display	5 in. CRT, 28 in. from subject, 2 in. jong horizontal line for input or fixed refer- ence, inverted "g" vith 1 in. long horizontal line and 0.5 in. vertical line for out- put or error, vertical displacements displacements
	Fowling Punction	The three-strugold frequency content of the cline forcing fructions was: $\frac{\nu (ind frequency 0.14, 0.25, 0.55)}{Med. 0.28, 0.47, 0.70}$ Slov 0.14, 0.23, 0.25 Med. 0.28, 0.47, 0.70 Past 0.70, 1.16, 1.75 Relative amplitudes, 2.5 1.5:1 Average absolute and root mean square medit- tude: $\frac{n}{1}$ (in.) [1] (in.) Slov 0.94 0.81 Med. 0.92 0.77 Past 0.88 0.74
	AUTRORS	Oberranyer, Swartz, and Muckler (Ref. 9)

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TABLE I (Concluded)

	80.1% 	<ul> <li>A. M. M. M. M. Bht subject', three russ per way, you is you with the russ per way.</li> <li>F. J. M. M.</li></ul>	
	Performance Massure	Time-on-target (KM) and number of hits for last 75 sec of 105 sec tracking run, scoring and +0.7 in.	
EXPERIMENTAL CONDITIONS	Controlled Element	Fosition: Y <sub>c</sub> = K <sub>c</sub> Z <u>2 in. (display)</u> K <sub>c</sub> = <u>in. (muipulator)</u>	
	Mulpulator	12-1/2 in. springless Joystick free to move in two dimensions, side to-side mwements for control, maximum deflection ±3-3/4 in.	
	Dirplay	21 in. CRT, 101 in. from subject, 5/8 in. long and 1/16 in. wide vertical line for input or fixed refer- ence, vertical line of same size for output or error, input line just abowe horizontal mid- line of scope, output mounted below, hori- zontal displacements	
	Forcing Turstion	Single sinusuida, fre- quencies 1.0, 2.09, 3.14, 4.19, 7.23, a-1 6.25 rud/sec Maximum deflection, 7 in.	
	SAURIUM	Hartaan (Refs. 4, >)	

#### CHAPTER III

# BLOCK DIAGRAM STRUCTURE, PURSUIT DETECTION TECHNIQUES AND EXPERIMENTAL CONFIGURATION

The first purpose of this chapter is to discuss the system relationships that can result from the pilot's utilization of the information in the pursuit display. Possible block diagram organizations are presented and their implications for the experimental detection of pursuit organization, as well as its measurement, are examined. Awkward measurement problems arise because pursuit behavior implies two or more pilot describing functions operating with a single forcing function. The explicit determination of the pilot's describing functions requires two or more independent inputs — unfortunately, the addition of a second input increases the number of task variables and therefore modifies the control task. Finally, the chapter concludes with a discussion of the experimental configuration.

#### A. BLOCK DIAGRAM STRUCTURE

With a pursuit display the operator sees both the input, i, and output, m, of the system, and has as his task the minimization of their difference, the system error, e. The possible block diagram structure representing the operator's response in this situation is shown in Fig. 3, where  $Y_{p_i}$ ,  $Y_{p_e}$ , and  $Y_{p_m}$  are describing functions indicating the operations on i(t),  $\epsilon(t)$ , and m(t), respectively. An integral part of the quasi-linear system description is the remnant,  $n_c(t)$ , injected at the pilot's output to account for the portion of the response that is not linearly correlated with the system input, i(t).

The presence of a pursuit display does not guarantee that the pilot will utilize all the information presented. Table II summarizes the six possibilities.

Using Fourier transforms, the equations of motion are

$$C = IY_{p_1} + EY_{p_e} + MY_{p_m} + N_c$$
(1)

$$E = I - M$$
 (2)

$$M = CY_{c}$$
(3)





TITLE	SYSTEM HARAMETERS USED*
Compensatory	e
Compensatory and feedforward	e,i
Compensatory and inner loop	e, m
Compensatory, inner loop, and feedforward.	e,m,i
Inner loop and feedforward	m,i
Pure feedforward	i
 *For a pu display only two parameter since ج m.	rs are independent

POSSIBLE HUMAN OPERATOR RESPONSE STRUCTURES FOR PURSUIT DISPLAIS

TABLE II

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Solving for the pilot's output and system error and system output yields

$$C = \left[ \underbrace{\frac{Y_{p_{i}} + Y_{p_{e}}}{1 + Y_{c}(Y_{p_{e}} - Y_{p_{m}})}}_{\frac{\Phi_{ic}}{\Phi_{ii}}} \right] I + \left[ \underbrace{\frac{1}{1 + Y_{c}(Y_{p_{e}} - Y_{p_{m}})}}_{\frac{\Phi_{n_{c}c}}{\Phi_{n_{c}n_{c}}}} \right] N_{c}$$
(4)

$$E = \left[ \underbrace{\frac{1 - Y_{c} (Y_{p_{m}} + Y_{p_{1}})}{1 + Y_{c} (Y_{p_{e}} - Y_{p_{m}})}}_{\frac{\Phi_{ie}}{\Phi_{ii}}} \right] I - \left[ \underbrace{\frac{Y_{c}}{1 + Y_{c} (Y_{p_{e}} - Y_{p_{m}})}}_{\frac{\Phi_{n_{c}e}}{\Phi_{n_{c}n_{c}}}} \right] N_{c}$$
(5)

$$M = \left[ \underbrace{\frac{Y_{c}(Y_{p_{1}} + Y_{p_{e}})}{1 + Y_{c}(Y_{p_{e}} - Y_{p_{m}})}}_{\frac{\Phi_{im}}{\Phi_{ii}}} \right] I + \left[ \underbrace{\frac{Y_{c}}{1 + Y_{c}(Y_{p_{e}} - Y_{p_{m}})}}_{\frac{\Phi_{n_{c}m}}{\Phi_{n_{c}n_{c}}}} \right] N_{c}$$
(6)

In Eqs. 4-6 the parts of E, C, and M which are correlated with the input are noted by the ratios of the cross-spectra with respect to I.

Since i and  $n_c$  are uncorrelated, the error spectral density is given by

$$\Phi_{ee} = \left| \frac{1 - Y_{c} (Y_{p_{m}} + Y_{p_{i}})}{1 + Y_{c} (Y_{p_{e}} - Y_{p_{m}})} \right|^{2} \Phi_{ii} + \left| \frac{Y_{c}}{1 + Y_{c} (Y_{p_{e}} - Y_{p_{m}})} \right|^{2} \Phi_{n_{c}n_{c}}$$
(7)

and the mean-squared error by

$$\overline{e^2} = \frac{1}{2\pi} \int_0^\infty \Phi_{ee}(\omega) \, d\omega = \overline{e_1^2} + \overline{e_n^2}$$
(8)

where

$$\overline{e_1^2} = \frac{1}{2\pi} \int_0^\infty \left| \frac{1 - Y_c \left( Y_{p_m} + Y_{p_i} \right)}{1 + Y_c \left( Y_{p_e} - Y_{p_m} \right)} \right|^2 \Phi_{ii} dx$$
(9)

and

$$\overline{e_n^2} = \frac{1}{2\pi} \int_0^\infty \left| \frac{Y_c}{1 + Y_c (Y_{p_e} - Y_{p_m})} \right|^2 \Phi_{n_c n_c} d\omega$$
(10)

The ratio

$$\mathbf{Y}_{\beta} = \frac{\Phi_{im}}{\Phi_{ie}} = \frac{\mathbf{Y}_{c} \left( \mathbf{Y}_{p_{i}} + \mathbf{Y}_{p_{e}} \right)}{1 - \mathbf{Y}_{c} \left( \mathbf{Y}_{p_{m}} + \mathbf{Y}_{p_{i}} \right)}$$
(11)

nas the property that the portions of M and E linearly correlated with the input are described by

$$\frac{M}{I} = \frac{Y_{\beta}}{1 + Y_{\beta}}$$
(12)

$$\frac{E}{I} = \frac{1}{1 + Y_{\beta}}$$
(13)

independent of the type of display or pilot utilization. Thus  $Y_{\beta}$  is the effective open-loop describing function, and has essentially the same interpretation in the pursuit as in the compensatory situation. That is, a single loop closure about  $Y_{\beta}$  results in the closed-loop characteristics.

To reduce tracking errors at low frequency the pilot should adjust his describing function boxes such that  $Y_{\beta} \gg 1$  at  $\omega < \omega_c$ . For the compensatory display this has to be accomplished by a single describing function  $Y_{p_c}$  operating on the error, whereas  $Y_{p_m}$ ,  $Y_{p_i}$ , and  $Y_{p_e}$  are theoretically available in pursuit.

In terms of  $Y_{\mbox{\scriptsize B}}$  the characteristic equation is given by

$$1 + Y_{\beta} = 0$$
 (14)

or

$$\frac{1 + Y_{c}(Y_{p_{e}} - Y_{p_{m}})}{1 - Y_{c}(Y_{p_{m}} + Y_{p_{i}})} = 0$$
(15)

Setting the numerator of Eq. 15 to zero gives the same characteristic equation as implied by the denominators in Eqs. 4-6.

Using Eq. 13, the mean-squared error correlated with the forcing function,  $\overline{e_1^2}$ , is related to  $Y_\beta$  as

$$\overline{e_{i}^{2}} = \frac{1}{2\pi} \int_{0}^{\infty} \left| \frac{1}{1 + Y_{\beta}} \right|^{2} \Phi_{ii} d\omega \qquad (16)$$

An example of a simple change in  $Y_{\beta}$  is an increase in the effective crossover frequency, and if  $Y_{\beta} \doteq \omega_{c\beta} e^{-j\omega\tau}/j\omega$ , then in a fashion analogous to the compensatory situation (Ref. 7),

$$\frac{\overline{e_{1}^{2}}}{\sigma_{1}^{2}} \doteq \frac{1}{3} \left( \frac{\omega_{1}^{2}}{\omega_{c_{\beta}}^{2}} \right) \quad \text{if} \quad \omega_{1} \ll \omega_{c_{\beta}}$$
(17)

where the effective crossover frequency is defined by  $|Y_{\beta}(j\omega_{c_{\beta}})| = 1$ .

#### B. DEFINITION OF FURSUIT BEHAVIOR

For the same input and controlled element, the <u>detection</u> of differences between compensatory and pursuit behavior has, in the past, relied primarily on performance measures. A key feature of the experimental program reported here is that comparisons were also made in terms of the effective open-loop describing function,  $Y_{\beta}$ , which reveals the dynamic effects of the pilot's system organization. The inadequacy of  $e^2$  alone is revealed by Eq. 8, i.e., a change in  $e^2$  could be due to a change in either  $e_1^2$  or  $e_n^2$ . It is also possible that the change from a compensatory to a pursuit display would produce no change in  $e^2$  but equal and opposite changes in  $e_1^2$  and  $e_n^2$ . From these factors it is seen that the detection of pursuit/ compensatory differences requires the comparison of two quantities, such as  $e^2$  and  $Y_{\beta}$ , or both  $e^2$  and  $e_1^2$  (where the latter reflects  $Y_{\beta}$ ).

As opposed to <u>detection</u> there are two major unknowns which contribute to the problem of <u>directly measuring</u> human operator describing function characteristics for pursuit display tracking These are:

 The actual block diagram structure adopted by the operator, i.e., the system parameters used to generate his output, is not known, nor is there any knowledge that the same block diagram structure exists for all inputs and controlled elements. Note that of the three unknown boxes in Fig. 3 only two are independent.

2. A second input that is statistically independent of the primary input (forcing function) is required to obtain data for computing the describing functions of the two elements. It is desired that the control situation be characterized only by the forcing function i(t) and the controlled element. Therefore the second input must be such that it does not influence the operator's normal pursuit response characteristics to the forcing function, and at the same time must be of sufficient amplitude to permit accurate cross-spectral or equivalent measurements.

The requirement that the second input be uncorrelated with the forcing function further compounds the problem of directly computing both operator describing function elements. Without loss of generality, we can assume that  $Y_{P_m} = 0$ . If the second input, d(t), is injected downstream of the pilot's output, c(t), the cross-spectra ratios are

$$\frac{\Phi_{ic}}{\Phi_{ie}} = \frac{Y_{p_i} + Y_{p_e}}{1 - Y_c Y_{p_i}} \qquad (1$$

$$\frac{\mathbf{v}_{dc}}{\mathbf{\Phi}_{de}} = \mathbf{Y}_{p_e} \tag{19}$$

Thus  $Y_{p_e}$  is measured directly by the second input and  $Y_{p_i}$  can be calculated from Eq. 18. Note that for inputs which are sums of sine waves the crossspectra ratios exist only at the sine wave frequencies of each input. Thus for i(t) and d(t) to be independent there can be no common frequencies; therefore Eqs. 18 and 19 cannot te solved directly. Two procedures could be used in an attempt to resolve this difficulty.

- 1. Interpolate the  $\Phi_{ic}/\Phi_{ie}$  and  $\Phi_{dc}/\Phi_{de}$  data to the d and i frequencies, respectively.
- 2. Interchange some of the forcing function and second input frequencies, thus obtaining  $\Phi_{ic}/\Phi_{ie}$  and  $\Phi_{dc}/\Phi_{de}$  data at the same frequencies but from separate runs.

The problems discussed above apply to the direct measurement of the operator's describing function characteristics for pursuit tracking, and are applicable to the general multiloop measurement task. An alternate procedure for obtaining additional insight into the human operator's pursuit describing functions is to perform the experiments without the second input and apply the analysis technique discussed in Chapter V.

#### C. EXPERIMENTAL SITUATION

The experimental arrangement and measurement techniques are nearly identical to those in Ref. 7. Thus, the following sections will briefly discuss the situation for this experimental series only as it differs from that of Ref. 7.

#### 1. Physical Layout and Equipment

The experiments were performed in a laboratory area consisting of two connected rooms. The larger of the two rooms contained all of the electronic equipment for performing and analyzing the experiments. The smaller room contained the manipulator and display. In this way the operator is isolated from the measuring equipment and other disturbances. The describing function data were obtained using the watthour-meter analyzer described in Ref. 7. This machine evaluates the real and imaginary parts of the Fourier coefficient of the e, c, and m signals using each input frequency as a reference. In addition, each run was recorded on magnetic tare.

The general measurements and task variables involved in the experiments are shown in Fig. 4. The task variables used in this experimental series are shown in dashed boxes, i.e., the forcing function and the controlled element. The  $Y_c$ 's used were

$$K_{c}$$

$$K_{c}/s$$

$$K_{c}/s^{2}$$

$$\frac{K_{c}}{s(s-\lambda)} \quad \lambda = 0.5, \ 1.0, \ 1.5$$

$$\frac{K_{c}(s+0.25)}{(s+5)^{2}}$$

These controlled elements were selected to provide a very broad coverage of pilot equalization to close the loop, ranging over very low frequency lags to essentially pure gain to low frequency leads (Ref. 7).

#### 2. Forcing Function

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The forcing functions, i(t), used in the pursuit experiments were of the augmented rectangular input spectrum form but with two different frequency spacings. One set was identical to that in Ref. 7 and spectra using this frequency spacing are designated by  $\omega_i$ ,  $\sigma_i$ , where  $\omega_i$  is the cutoff frequency in radians/second and  $\sigma_i$  is the rms amplitude in inches, i.e., as seen on the display. The other set is designated by either E5,  $\sigma_i$ , or R14,  $\sigma_i$ , a notation similar to that in Ref. 3. The following paragraph describes the frequency content and bandwidth of the various inputs. The rms values used are discussed in Chapter IV and indicated in Table III.

The frequency setting,  $\omega_n$ , and the number of periods,  $n_r$ , for each component in the fixed 240-sec run length,  $T_R$ , are given on p. 20. For the  $\omega_i$ ,  $\sigma_i$  spacings three approximate  $\omega_i$  values were used -1.5, 2.5 and 4.0 rad/sec. To define these three inputs, the amplitudes at the lowest





six, seven, or eight frequencies were set equal, for cutoff frequencies of 1.49, 2.54, or 4.03 rad/sec, respectively. The amplitudes of the remaining frequencies were set to one-tenth of the low frequency amplitudes (20 db attenuation). The logarithmic spacing (which facilitates fitting ratios of rational polynomials to the measured values) was selected to insure the effective independence of the sine wave components over the run length of interest.

	ω <sub>i</sub> , σ <sub>i</sub>	B5,	Β5, σ <sub>i</sub> ; R14, σ <sub>i</sub>			
ω <sub>n</sub>	$n_r = T_{R}\omega_n/2\pi$	ω <sub>n</sub>	$n_r = T_{R} \omega_n / \epsilon$	\π		
0.157	6	0.31	4 12			
0.262	10	0.73	52 28			
0.393	15	1.15	51 44			
0.602	23	1.67	75 64			
0.969	37	1.98	39 76			
1.49	57	2.40	0 <sup>r</sup> i 92			
2.54	97	4.29	9 164			
4.03	154	5.17	7 236			
7.57	289	10.14	⊦ <u>3</u> 88			
13.8	527	14.03	536			

For the other forcing functions all the amplitudes were equal for the R14,  $\sigma_i$  input, while the four highest frequencies were attenuated 10 Åb for the B5,  $\sigma_i$  input. Figure 5a shows measured input power spectral magnitudes for the three  $\omega_i$  forcing functions, while Fig. 5b provides a similar picture for the B5 and the R14 spectra.

The forcing functions selected serve several purposes. The range of forcing functions is covered by the augmented  $\omega_1$  spectra; these are also appropriate for direct tie-in and extension of the Ref. 7 results. These would be sufficient except that, in the major past study (Ref. 3), two interesting pursuit/compensatory differences were found for high bandwidth forcing function spectrs. These were:

• Above a particular forcing function bandwidth, compensatory rms error becomes less than the pursuit rms error.



Figure 5. Measured Forcing Function Power Spectra Magnitude

• At a still higher forcing function bandwidth, a pursuit system is still operable, but compensatory cannot be controlled by the operator.

The high frequency bandwidth(s) serve to further explore these conditions.

#### 3. Display and Manipulator Sensitivities

The pursuit display is shown in Fig. 6, where the dot for the input and the vertical line for the system output move laterally. The pilot's output, c(t), consisted of lateral motions of a spring-restrained side stick (negligible inertia and damping). For a pure gain controlled element of unity, the display and stick are related by

 $K_s = 1$  in. (display)/6<sup>o</sup> (stick)  $F_s = 2.21$  oz/deg (stick) (applied at top of 4-in. stick)

With the 4-in. moment arm, lateral motion of the operator's hand amounts to about 0.07 in. (stick) per degree of stick rotation. Accordingly, the sensitivity



Figure 6. Pursuit Display

can be expressed in terms of the linear motion of the operator's hand by dividing the angular sensitivity (in inches per radian) by the moment

arm, i.e.,

$$K_s = 1 \text{ in.}/6^\circ \times 57.3^\circ/\text{rad} \times 0.25 \text{ in.}^{-1}$$
  
= 2.38 in. (display)/in. (stick)

For controlled elements other than unity, all of the above sensitivities are multiplied by  $Y_{c}(j\omega)$  and they become dynamic quantities.

The pilot/manipulator/display configuration is depicted in Fig. 7. Only the roll axis was used for this experimental series. Movements of the stick in one direction produced system output movements in the same direction.



#### OHAPTER IV

#### HXPERIMENTAL DATA

This chapter presents the experimental describing function and performance measure data. For this sort of exploratory study a design which includes a large number of forcing functions and controlled elements is essential. This would be excessively expensive if several subjects were used. Consequently, in the interests of economy and maximum coverage a single well-trained subject was used throughout with several special restrictions to increase the likely generality of the results. These were:

- Tie-in of this set of experiments with those involving a large population of pilots to indicate that the subject used was a representative sample (Ref. 7).
- Tie-in of this set with the only previous comprehensive describing function measurements (Ref. 3), thereby corroborating those findings and indicating a reasonable tie-in to yet another population.

The particular controlled elements, forcing functions, and display configurations tested are shown in Table III. Exceptions to the rms input values are indicated by an asterisk. The numbers in each cell indicate the run sequence. Each configuration ...s repeated three times using a single highly motivated subject who was a light-airplane-qualified civilian pilot with extensive tracking experience. Because of his experience the subject was able to rapidly approach asymptotic values of performance on any given configuration, thereby permitting a large number of configurations to be examined at minimal cost. While more subjects would have been desirable, the exploratory and limited-effort nature of the experimental series made this unrealistic. To make the results as representative as possible, considerable effort was made to tie them in to other data.

In what follows Section A presents the subject's compensatory display data to tie in with that in Ref. 7 for a population of pilots and that in Ref. 3 (where pursuit data were also taken). Also included is a comparison

TABLE III RUN SEQUENCE (THREE REPLICATIONS PER CONFIGURATION)

	σ <sub>1</sub> (in.)		1.0	1/2	1/2 *σ <sub>i</sub> = 1/4	1 /4	1 /4	1 /4	1.0
9		R1 4	4	ω	12*				
	Pursuit Display	Β5	N	6	10				
		4,0	18	2t	8				115
N		2.5	16	52	28				40
FUNCTIO		1.5	th 1	20	26	R	ネ	36	38
ORCING		R1 l4	5	7	11*				
Fi Fi	isplay	Β5	-	Ś	6				
	tory Di	4°0	71	23	29				41
	Jompense	2.5	15	21	27				39
	0	1.5	13	19	25	31	33	35	37
C:ONTROLLED	L L L L L L L L L L L L L L L L L L L		Kc	N N N N	<sup>R</sup> C <sup>8</sup>	Kc s(s - 0.5)	K <sub>c</sub> s(s - 1.0)	Kc <u>s(s - 1.5)</u>	$\frac{K_{c}(s + 0.25)}{(s + 5.0)^{2}}$

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of long term effects for the present subject to demonstrate the typical asymptotic performance levels achieved by human operators. These tie-ins and long term effects increase the generality of the pursuit/compensatory comparisons in later sections. Section B compares the performance measure results with those of other experiments. Section C presents a detailed comparison of Pursuit and Compensatory performance measures obtained in this study. Finally, Section D compares the describing functions for compensatory and pursuit display.

# A. COMPENSATORY DATA COMPARISONS

1. Population of Pilots (Ref. 7)

In general the describing function data, Figs. 8-11, indicate that the present subject is typical of a pilot population. Note that the data from Ref. 7 (circles) are plotted at the correct input frequencies, whereas the current data have been shifted slightly to the right to avoid overlapping. In each case the data points indicate mean values; the vertical lines indicate  $\pm 1\sigma$  spread. For easy controlled elements (K<sub>c</sub> and K<sub>c</sub>/s) and for low bandwidth inputs, there is no difference between the present subject and the other pilots (Ref. 7). For the harder controlled elements and higher bandwidth inputs, the present subject usually has a higher crossover frequency and low frequency gain, plus less effective time delay near crossover and above. These differences are probably due to the extensive tracking experience of the present subject (close to 500 data runs plus numerous practice runs) compared to the eight other pilots used in the Ref. 7 series.

To illustrate the long term effects of practice, consider Figs. 12 and 13 for  $Y_c = K_c$  and  $K_c/s(s-1.5)$ , respectively. Figure 12, for the relatively easy but extremely well-practiced case of  $Y_c = K_c$ , shows amazingly close agreement for test periods over one year apart. Even the subtler data trends are closely duplicated in this comparison, and we conclude that the subject's describing function had stabilized. (The slight improvement in tracking error is probably due to reduced remnant.) Figure 13 shows a similar comparison for the second-order unstable element in which  $\lambda = 1.5$  (near the uncontrollable limit). Even it this much more difficult case the agreement is quite good, except that there is less tracking error and high frequency phase lag, indicating a refined tracking technique over the intervening year.



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Comparison of Compensatory Display Data with Population of Pilots Figure 9.

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Comparison of Compensatory Display Data with Population of Pilots Figure 10.

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Figure 12. Long Term Variability of the Current Subject's Describing Function  $Y_c = K_c$ , R14 Input



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Figure 13. Long Term Variability of the Current Subject's Describing Function  $Y_c = K_c/s(s-1.5), \omega_1 = 1.5$ 

While skill level differences may explain some of the differences between the present subject and the population of pilots, it's quite likely that our subject's extensive tracking experience is the largest factor. Thus, our single subject can be considered typical of a pilot population in that they will reach an asymptotic level of performance in their well-trained flying tasks in analogous fashion to the present subject in tracking tasks.

# 2. Population of Students (Ref. 3)

In order to tie in with the body of data generated from  $Y_c = K_c = 1$ in Ref. 3, a tracking situation as similar as possible to Elkind's had to be considered in both Ref. 7 and the present experiments. Ideally, the tie-in experiments should be conducted with forcing functions and manipulators similar to those to be used in our other experiments, yet also similar enough to Elkind's to effect a reasonable connection. Fortunately, Elkind's B6 forcing function amounts, in our notation, to  $\omega_i = 3.0, 1$  in., so the  $\omega_i = 2.5, 1$  in., forcing function was thought to provide reasonably close approximation. The lightly restrained stick manipulator used in this series and Ref. 7 differs substantially in form from Elkind's freely moving pencil-like pip tracker, although the movements in both cases were generally lateral (with more rotation involved in Ref. 7 and the present series). Yet, in our past work we were able to show reasonable connections with Elkind's data even using an aircraft center stick (Ref. 8), so any differences due to the manipulators were expected to be slight.

In Ref. 7 three highly trained pilots tracked two runs each for  $Y_c = K_c = 1, 2$ , and 5, respectively. The differences between  $Y_p$  measurements for successive runs for each pilot were very slight; the two runs were averaged and are shown in  $Y_pY_c$  form in Fig. 14 along with the corresponding compensatory display data from the current experiments. Elkind's comparable data for  $\omega_1 = 3.0, 1$  in., are also shown in Fig. 14. These data are averages of four four-minute runs, two from one subject and one run from each of two other subjects. It is clear from Fig. 14 that the results are remarkably compatible with the Elkind data, with the current subject giving a slightly better match for low frequency



amplitudes. In fact, this extremely close correspondence between data taken years apart, by different experimenters at different locations, with different subjects, different analysis apparatus, and slightly different forcing functions and manipulators, etc., is very satisfying especially considering the subject selection procedures: ļ

- Reference 7 used three naval test pilots, each trained to a stable performance level (as measured by  $e^2/\sigma_1^2$ ) for each configuration.
- The current series used a single subject with extensive tracking experience.
- Reference 3, utilized highly trained students.

Thus, all indications are that the current single subject has reached a stable performance level against which to compare his pursuit display data in Subsections B, C, and D below, increasing the generality of the conclusions drawn there.

# B. COMPARISON OF PERFORMANCE MEASURES WITH RESULTS OF OTHER EXPLRIMENTERS

While our describing function data can only be compared with Elkind's, the performance measure results can be corpared with those of Refs. 1, 2, 3, 9. Figure 15 shows the normalized mean-squared error for pursuit and compensatory obtained in our series with those found by Elkind. Our augmented rectangular forcing function spectra are similar to his rectangular spectra, differing primarily by the addition of a few low amplitude waves to form a high frequency shelf. Thus, the Elkind R.24, R.4, and R.64 forcing function spectra (where the 94, .4, and .64 are in cycles/second) correspond roughly to our  $\omega_i = 1.5$ , 2.5, and 4.0 rad/sec, respectively. Also, our R14 (highest frequency, 14 rad/sec) spectrum is akin to his R1.6 or R2.4. As indicated in Fig. 15, the trends between compensatory and pursuit exhibited by the two sets of data are generally similar; both sets of data indicate lower mean-squared error for the pursuit display except at the higher forcing function bandwidths (B5, R14, and R2.4). Although Elkina had a forcing function similar to B5, the relative mean-squared errors for this input are not available,



Figure 15. Performance Measure Comparisons with ELmind's Results,  $Y_c = K_c$ 

Another comparison for  $Y_c = K_c$  is made on the basis of the data from Refs. 1 and 9. In both these experimental series only three sinusoids were used to make up the forcing function. These input spectra do not have a meaningful "bandwidth" as such. Consequently, to show a rough comparison between these and our results they are classified in terms of the maximum forcing function frequency (neglecting the high frequency shelf in our data). The result for  $Y_c = K_c$  is shown in Fig. 16. The indications are that all the data are reasonably comparable, and that ursuit performance measures are smaller than compensatory for all by the very lowest frequency input used in Ref. 9.

Similar comparisons for  $Y_c = K_c/s$  and  $K_c/s^2$  are also shown in Fig.16. These comparisons are both ambiguous and confusing. For  $Y_c = K_c/s^2$ , for example, the current data and those of Ref. 2 indicate that pursuit is superior, whereas the data of Ref. 9 show precisely the opposite. For  $Y_c = K_c/s$  both Ref. 2 and Ref. 9 are in opposition to our results for higher frequency forcing functions, although they both also indicate compensatory is better than pursuit at lower frequencies, a trend that would coincide with our results. Finally, the lowest frequency forcing function for Ref. 2 is in direct opposition for the medium frequency input of Ref. 9. The causes of such differences are subtle indeed, for critical examination of the conflicting data shows no defect in method or procedures. About the only conclusion that can be drawn from the  $Y_c = K_c/s$  data is that performance with either form of display can be superior to the other under highly restricted circumstances.

#### C. COMPARISON OF PURSUIT AND COMPENSATORY PERFORMANCE MEASURES

Figures 17-21 present the performance measures obtained with pursuit data plotted on the ordinates and compensatory data on the abscissas. Both the total relative mean-squared error,  $\overline{e^2}/\sigma_1^2$ , and its components  $(\overline{e_1^2}/\sigma_1^2, \overline{e_n^2}/\sigma_1^2)$  are presented. These figures also contain plots of  $\overline{c^2}$ ,  $\overline{c_1^2}$ , and  $\overline{c_n^2}$  (not normalized). The average performance measures are given at the intersections of the lines which have short hatch marks on their ends to indicate the range of the measurements.















= K<sub>c</sub>/s(s -Performance and Control Deflection Data;  $\mathbf{Y}_{\mathbf{C}}$ Figure 19.

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Figure 21. Performance and Control Deflection Data;  $Y_c = K_c(s + 0.25)/(s + 5)^2$ 

A summary of the performance measures  $(\overline{e^2}, \overline{e_1^2}, \text{ and } \overline{e_n^2})$  is given in Table IV, where the letters "P" (pursuit) or "C" (compensatory) indicate the configuration having the smaller measure. In Figs. 17-21 when the equivalent (i.e., P = C) line falls within the range of measures a "PC" or "CP" is used, with the first symbol indicating the smaller mean. The numbers in parentheses are the relative rms performance measures. The results for the five configurations have been grouped into those showing clear superiority of the pursuit display  $[Y_c = K_c, K_c/s^2, \text{ and} K_c/s(s-\lambda)]$ , e.g., the relative rms performance measure is about 0.8, and those showing little or no consistent difference  $[Y_c = K_c/s \text{ and} K_c(s+0.25)/(s+5)^2]$ . More detailed conclusions are for:

$$\mathbf{Y}_{\mathbf{C}} = \mathbf{X}_{\mathbf{C}} \text{ (Fig. 17)}$$

Pursuit is better than compensatory for both  $e^2$  and  $e_1^2$  for low bandwidth inputs (augmented  $\omega_1$  spectra), becoming essentially the same for high bandwidth inputs (B5 and R14) (This was discussed in Subsection B above.) Note that while  $e_n^2$  is erratic it is quite small, so that it has little effect on  $e^2$ .

 $Y_{c} = K_{c}/s^{2}$  (Fig. 18)

Generally pursuit is increasingly better than compensatory as  $\omega_1$  increases for both  $e^2$  and  $e_1^2$ . Results for  $e_n^2$  are scattered and generally small encugh not to affect  $e_1^2$  or  $e^2$ . An exception is the  $\omega_1 = 2.5$  compensatory results, where a small value of  $e_n^2$ makes the  $e^2$  pursuit nearly equal to  $e^2$  compensatory.

$$Y_c = K_c/s(s-\lambda)$$
 (Fig. 19)

Pursuit is increasingly better than compensatory as  $\lambda$  increases. This holds for both components of  $\overline{e_1^2}$ . Here  $\overline{e_n^2}$  is on the order of  $\overline{e_1^2}$ .

$$Y_c = K_c/s$$
 (Fig. 20)

Compensatory is slightly better at low and high bandwidth inputs for  $\overline{e^2}$  with essentially no difference for moderate bandwidth inputs.

Чc		FORCING FUNCTIONS				
		Augmented wi Spectra			Extended Band Spectra	
		1.5	2.5	4.0	B5	R14
к <sub>с</sub>	e <sup>2</sup> /o <sup>2</sup>	P (0.76)	P (0.81)	P (0.80)	C (1.1)	Same
	$\overline{e_1^2}/\sigma_1^2$	P (0.79)	P (0.81)	P (0.76)	Same	Same
	$\overline{e_n^2}/\sigma_1^2$	Р	PC	PC	С	CP
K <sub>c</sub> s <sup>2</sup>	$\overline{e^2}/\sigma_1^2$	P (0.78)	P (0.95)	P (0.78)	P (0.81)	P (0.71)
	$\overline{e_1^2}/c_1^2$	PC (0.93)	P (0.89)	P (0.82)	P (0.72)	P (0.77)
	$\overline{e_n^2}/\sigma_i^2$	Р	CP	P	Same	PC
		λ				
Υ <sub>c</sub>		0	0.5	1.0	1.5	
$\frac{K_{c}}{s(s-\lambda)}$	$\overline{e^2}/\sigma_1^2$	P (0.78)	P (0.71)	P (0.66)	P (0.59)	
	$\overline{e_1^2}/\sigma_1^2$	PC (0.93)	P (0.81)	P (0.61)	P (0. <sup>1</sup> :7)	
	$\overline{e_n^2}/\sigma_1^2$	P (0.65)	P (0.60)	P (0.72)	P (0.71)	
Ϋ́c		Augment d dy Spectra Extended Band Sp			and Spectre	
		1.5	2.5	4.0	B5	R14
K <sub>c</sub> s	e²/of	CP (1.1)	C (1.27)	Same	Same	CP (1.07)
	$\frac{1}{e_{1}^{2}/\sigma_{1}^{2}}$	¢ (1.2)	C (1.2)	CP	Same	P (0.87)
	$\frac{1}{e_n^2}/\sigma_1^2$	P	С	Same	Same	С
$\frac{K_{\rm c}(s + 0.25)}{(s + 5)^2}$	$\overline{e^2}/\sigma_i^2$	CP (1.05)	P (0.92)	Same		
	$\overline{e_i^2}/\sigma_i^2$	Same	P (0.91)	P (0.86)		
	$\overline{e_n^2}/\sigma_1^2$	CP	PC	С		

TABLE IV SUMMARY OF PERFORMANCE MEASURE DIFFERENCES\*

\*Letters indicate configuration with lower measure, P (pursuit) or C (compensatory) Numbers indicate relative rms performance measure  $X_c = K_c(s + 0.25)/(s + 5)^2$  (Fig. 21)

For the bandwidths tested there is no significant difference between displays.

In is, in general the performance improvement with pursuit display is largest for the more difficult controlled elements and inputs.

The mean-squared pilot output,  $\overline{c^2}$ , and its components,  $\overline{c_1^2}$  and  $\overline{c_n^2}$ , are also shown in Figs. 17 - 21. The  $\overline{c^2}$  data show quite consistent trends, as does  $\overline{c_1}$ :

 $Y_c = K_{-}$  (Fig. 17)

 $c^2$  and  $c_x^2$  are arger for pursuit display than for compensatory display. This increased pilot activity has resulted in less mean-squared error.  $c^2$  and  $c_1^2$  decrease as  $\omega_1$  increases.  $c_n^2$  tends to increase as  $\omega_1$  increases, but  $c_n^2$  is only in the order of 10 percent or icos of  $c^2$  for the low frequency inputs, so the run-to-run variability of  $c_n^2$  display than much.

 $Y_C = K_C/s^2$  and  $K_C/s(s-\lambda)$  (Figs. 18 and 19)

 $\overline{c^2}$  and  $\overline{c_1^2}$  are lower for pursuit than for compensatory. Thus the improved performance ( $\overline{e^2}$  and  $\overline{e_1^2}$ ) is accomplished with less pilot activity. This contrasts with the case for  $Y_c = K_c$  above.  $\overline{c_n^2}$  is somewhat lower for pursuit for the augmented  $\omega_i$  inputs, perhaps higher for B5 and R14.

$$Y_c = K_c/s$$
 (Fig. 20)

 $\overline{c_1^2}$  is lower for pursuit, whereas results for  $\overline{c^2}$  are mixed. It appears that slightly better performance with the compensatory display is obtained with slightly higher effort.

 $Y_c = K_c(s + 0.25)/(s + 5)^2$  (Fig. 21)

Results appear consistent.  $c^2$ ,  $c_1^2$ , and  $c_n^2$  decrease (both for pursuit and compensatory) as  $\omega_1$  increases. For  $c^2$  and  $c_1^2$ the trend appears to be that the operator works less with pursuit for  $\omega_1 = 1.5$ , about the same for  $\omega_1 = 2.5$ , and more for  $\omega_1 = 4.0$ . These trends roughly parallel those for  $e^2$  and  $e_1^2$ .

The mean-squared error and mean-squared pilot output results indicate that the pursuit display is desirable for the harder controlled elements,  $K_c/s^2$  and  $K_c/s(s-\lambda)$ , in that the performance improves and less pilot activity is required. For  $Y_c = K_c$  the pilot can also take advantage of the additional information presented, but he must use more activity to do so.

#### D. COMPARISON OF FURSUIT AND COMPENSATORY DESCRIBING FUNCTIONS

The describing function differences between the pursuit and compensatory displays are given in Figs. 22-26 in terms of the effective open-loop,  $Y_{\beta}$  (Eq. 11). Average data are shown with the range of the measurements indicated by the hatch marks. The pursuit data are plotted at the correct frequencies, while the compensatory data have been shifted slightly to the right. Detailed comparisons are:

 $Y_c = K_c$  (Fig. 22)

For the augmented  $\omega_i$  spectra the P has a slightly higher  $\omega_c$ and low frequency gain which leads to less  $e_i^2$  (Fig. 17). For the B5 and R1<sup>4</sup> inputs, P has a slightly smaller  $\omega_c$  but larger low frequency gain, such that  $e_i^2$  is about the same as C (Fig. 17). In general, P has more phase lag  $(10^\circ - 30^\circ)$  than C at mid-band

$$Y_{c} = K_{c}/s^{2}$$
 (Fig. 23)

The most significant difference is the much smaller low frequency phase lag for P for all inputs. In addition, the P amplitude, which is nearly the same at low  $\omega_i$ , becomes much smaller at mid-frequencies than the C as  $\omega_i$  increases. Ordinarily this would lead to greatly increased  $\overline{e_i^2}$  according to the one-third law approximation (Eq. 17). For this controlled element, phase lag has a large effect on  $\overline{e_i^2}$  because the phase margin is small for C. Thus the increased phase margin for P has a larger effect on  $\overline{e_i^2}$  than the reduced amplitude ratio.

 $Y_c = K_c/s(s-\lambda)$  (Fig. 24)

The effect of the increasingly unstable controlled element (at  $\omega_1 = 1.5 \text{ rad/sec}$ ) is a slightly smaller low frequency phase lag for P. For  $\lambda = 1.5$  the P low frequency amplitude becomes





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Figure 22. (Concluded)

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Figure 23. Comparison of Pursuit and Compensatory Display Describing Functions



Figure 23. (Concluded)

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slightly larger and the  $\omega_c$  becomes slightly smaller than the C. Here, as for  $Y_c = K_c/s^2$  above, the performance improvement is due mostly to the increased phase margin for the P.

# $Y_{c} = K_{c}/s$ (Fig. 25)

For the lower  $\omega_i$ 's, P has slightly less amplitude at low frequency and at crossover leading to a larger  $e_i^2$  (Fig. 20). The same amplitude trends hold for the B5 input but now the P has slightly less phase .ag, such that  $e_i^2$  shows no difference. For the R14 input the P amplitude ratio is much smaller and there is less phase lag, leading to an improvement in  $e_i^2$  (Fig. 20).

$$Y_c = K_c(s + 0.25)/(s + 5)^2$$
 (Fig. 26)

Very little difference between P and C.

Thus, the describing function and performance measure comparisons indicate that the pursuit display is superior to the compensatory display for  $Y_c = K_c$ ,  $K_c/s^2$ , and  $K_c/s(s-\lambda)$ . For  $Y_c = K_c$  the subject improves performance by increasing the bandwidth of the effective open-loop,  $Y_\beta$ , but must increase his activity slightly. For  $Y_c = K_c/s^2$  and  $K_c/s(s-\lambda)$ performance improvements are obtained by drastic changes in the phase of  $Y_\beta$  but with less pilot activity.



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Figure 26. Comparison of Pursuit and Compensatory Display Describing Functions

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## CHAFTER V

## DATA INTERPRETATION

The data in the previous chapter indicate the differences between pursuit and compensatory display situations. The purpose of this chapter is to interpret the data as to the possible nature of the pilot's organization. As indicated in Chapter III, the explicit determination of the pilot's describing functions is not possible with a single forcing function. Nevertheless, using reasonable assumptions it is possible to imply certain characteristics about the pilot's operation in system stabilization as well as his operation on the input.

The discussion in Chapter III indicates that only two of the pilot's describing functions  $(Y_{p_i}, Y_{p_e} \text{ and } Y_{p_m})$  are independent, i.e., any box can be zero and the other two can describe the data. Of the three possibilities, we will assume that  $Y_{p_m} = 0.*$  The pursuit situation then becomes as in Fig. 27, where the system error and output spectra are, from Eqs. 6 and 7 with  $Y_{p_m} = 0$ 

$$\Phi_{\rm mm} = \left| \frac{\left( Y_{\rm p_{1}} + Y_{\rm p_{e}} \right) Y_{\rm c}}{1 + Y_{\rm p_{e}} Y_{\rm c}} \right|^{2} \Phi_{\rm ii} + \left| \frac{Y_{\rm c}}{1 + Y_{\rm p_{e}} Y_{\rm c}} \right|^{2} \Phi_{\rm n_{c} n_{c}}$$
(20)

$$\Phi_{ee} = \left| \frac{1 - Y_c Y_{p_i}}{1 + Y_{p_e} Y_c} \right|^2 \Phi_{ii} + \left| \frac{Y_c}{1 + Y_{p_e} Y_c} \right|^2 \Phi_{n_c n_c}$$
(21)



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Figure 27. Single-Axis Pursuit Tracking

<sup>\*</sup>The data were also examined using the assumption that  $Y_{p_i} = 0$  thereby leaving  $Y_{p_e}$  and  $Y_{p_m}$  to describe the pilot's behavior. However, this approach did not result in any useful interpretation of the data.

Since system stability and low frequency error reduction are determined by  $Y_{p_e}$ , an appropriate model for the pilot to adopt is that which results for the same controlled element in the compensatory situation, i.e.,

$$Y_{p_e} = Y_{p_c}$$
(22)

If  $Y_{p_i}$  is selected so that

$$Y_{p_i}Y_c \doteq 1$$
 (23)

then the system output (neglecting the remnant) approximately equals the input resulting in smaller error. Thus, the pilot can take advantage of the additional information in the pursuit display by adopting equalization that is the inverse of the controlled element. The system then becomes nearly open-loop through the feedforward, with ' .e feedback acting as a vernier control and as a means of stabilizing the controlled element.

To test this hypothesis an implied feedforward,  $Y_{p_i}^*$ , can be calculated from the assumption that  $Y_{p_e}$  for the pursuit display is the same as the measured  $Y_{p_c}$  for the corresponding compensatory display. Thus,  $E_1$ , 11, repeated below, can be solved for  $Y_{p_i}^*$  on the assumption that  $Y_{p_e} = Y_{p_c}$ and that  $Y_{p_m} = 0$ .

$$X_{\beta} = \frac{\Phi_{im}}{\Phi_{ie}} = \frac{(Y_{p_{i}}^{*} + Y_{p_{c}})Y_{c}}{1 - Y_{p_{i}}^{*}Y_{c}}$$
 (24)

Solving for  $Y_{p_i}^* Y_c$  yields

$$Y_{p_{i}}^{*}Y_{c} = \frac{Y_{\beta} - Y_{c}Y_{p_{c}}}{1 + Y_{\beta}}$$
 (25)

The implied feedforward was calculated using averaged data for  $Y_{\beta}$ and  $Y_{P_{c}}$ . The results are given in Figs. 28-30 for the cases that showed large performance differences, i.e.,  $Y_{c} = K_{c}$ ,  $K_{c}/s^{2}$ ,  $K_{c}/s(s - \lambda)$ , respectively. The results are given in  $Y_{P_{i}}^{*}Y_{c}$  form to illustrate the extent that the ideal adjustment given by Eq. 23, in the light of the assumption given by Eq. 22, is approached. Detailed comments arc:  $Y_c = K_c$  (Fig. 28)

 $Y_{P_i}^* Y_c$  is less than one and approximately constant, but with lagging phase at high frequency similar to that of a time delay.

 $Y_c = K_c / s^2$  (Fig. 29)

 $Y_{P_i}^*Y_c$  is greater than one at low frequency, then becomes much less than one at high frequencies. The phase is positive  $(20^\circ - 60^\circ)$  at low band and mid-band frequencies becoming negative at high frequency.

 $Y_c = K_c / s(s - \lambda)$  (Fig. 30)

Generally  $Y_{p_i}^* Y_c$  stays closer to unity at all frequencies than for  $Y_c = K_c/s^2$ , although the general character is the same. As  $\lambda$  increases,  $Y_{p_i}^* Y_c$  becomes closer to unity, reflecting the relative reduction in performance measures (Fig. 19).

For the second-order controlled elements additional dynamics are present. These are partially obscured by data scatter at low frequencies probably due to taking differences between large numbers (numerator cf Eq. 25). Nevertheless, the pilot appears to be able to invert the controlled element at low frequencies up to just below crossover.

For the single-forcing-function case, the advantages of the above assumptions and procedures are:

- The division of the pilot's actions cannot be unique, so any combination of Yp<sub>j</sub>, Yp<sub>2</sub>, Yp<sub>m</sub> compatible with the data is equally appropriate on theoretical grounds.
- The pilot adjustment rules for the compensatory situation (Ref. 7) can be utilized for  $Y_{p_c}$ , and a first-cut estimate for  $Y_{p_i}$  can be made on the basis of  $Y_c$ .



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Figure 29. Implied Human Operator Feedforward Characteristics,  $Y_c = K_c/s^2$


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### CHAPTER VI

# GENERAL CONCLUSIONS

The data and analyses presented in this report indicate many differences between pursuit and compensatory systems. Unfortunately, many of these distinctions are subtle and the ones that are clear-cut are not general. Most of these features have been described in the local discussions of the data and need not be repeated here. The broader and more general conclusions are, however, summarized below:

- 1. The describing function data and performance measure trends for the  $Y_c = K_c$  series correspond very well with those of a previous investigation by Elkind.
- 2. The compensatory describing function data for the series reported here tie in fairly well with an earlier series in which the present subject was but one member of a population.
- 3. A sufficient indication of pursuit behavior is that  $Y_\beta$  pursuit differs from  $Y_\beta$  -compensatory.
- 4. The operator dynamics as measured by the effective open-loop describing function,  $Y_{\beta}$ , are different for all but one of the controlled-element/forcing-function combinations tested. Thus, in all but the exceptional case the presence of the pursuit display is sufficient to induce pursuit behavior.
- 5. The provision of a pursuit display loes not necessarily induce pursuit behavior. This is shown conclusively for  $Y_c = K_c(s+0.25)/(s+5)^2$ , for which  $Y_\beta$  is the same in the pursuit and compensatory conditions.
- 6. For those systems where pursuit and compensatory dynamic differences were present, in three cases  $[K_c, K_c/s^2, \text{ and } K_c/s(s-\lambda)]$  pursuit was superior and in one case  $(K_c, s)$  compensatory was better.
- 7. For the augmented  $\omega_1$  spectra the normalized mean-squared errors are smaller for pursuit display in all cases where  $Y_{\beta}$ -pursuit is superior to  $Y_{\beta}$ -compensatory.
- 8. The relative superiority of pursuit over compensatory, when present, increases as the controlled element becomes more difficult to control, i.e., higher bendwidths or larger instabilities.
- 9. The differences between pursuit and compensatory behavior are often subtle and confusing. A plausible description

of pursuit operations is:

The compensatory loop portion of the pursuit system has dynamics similar to those of the corresponding compensatory system. Thus the analytical/verbal model for compensatory operation can be applied to pursuit.

When pursuit behavior is actually present the pilot operates directly on the forcing function, thereby adding an additional describing function block,  $Y_{\text{pi}}$ , to the system structure. To a first approximation the feedforward,  $Y_{\text{pi}}$ , is adjusted such that  $|Y_{\text{pi}}Y_{\text{c}}| \doteq 1$ . Such adjustment has the net effect of making  $M/T \doteq 1$  and  $E/I \doteq 0$  over a wide frequency band.

This model for pursuit action is not a unique description of pursuit operation, although it becomes so when the constraint is applied that the compensatory portion of the pursuit system will be the same as that measured for the compensatory display.

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