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TECHNICAL REPORT NO. 65-127
DIGITAL EVALUATION OF A CALIBRATION TECHNIQUE
FOR MULTIPLE-ELEMENT ARRAY SYSTEMS

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TELEDYNE INDUSTRIES
GEOTECH DIVISION
GARLAND TEXAS

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TECHNICAL REPORT NO. 65-127

DIGITAL EVALUATION OF A CALIBRATION TECHNIQUE
FOR MULTIPLE-ELEMENT ARRAY SYSTEMS

by

Y. T. Huang

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8 December 1965

IDENTIFICATION

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ABSTRACT

A quantitative analysis is made using techniques of digital spectral analysis. Two different approaches are compared -- statistical and deterministic.

For the statistical approach, the computer program "BLACKY," which computes power density spectra, was adopted. A program which uses the Numerical Transform Theorem (Huang, 1965a) was written for the CDC 160A digital computer. This program, "NUTRAN," relies on a deterministic consideration of a time series.

The agreement between the field calibration and the estimation of the magnification curve is quite good up to 1.5 cps, considering the simplicity of the mathematical model used in our study. Above this frequency, the scatter is quite erratic, and further refinement of the technique is desirable. The reason for this disagreement comes essentially from a lack of the high signal-to-noise ratio which is the basis of our treatment. Our technical success seems to be limited by our ability to discriminate noise which contaminates the signal arrivals.

DIGITAL EVALUATION OF A CALIBRATION TECHNIQUE
FOR MULTIPLE-ELEMENT ARRAY SYSTEMS

1. INTRODUCTION

In the previous report (Huang, 1965b), we made a qualitative study of a calibration technique (Whalen, 1965) using the Geotech Analog Spectrum Analyzer. It was found that good agreement existed between the estimated and the field calibrated values of the relative magnification curve for frequencies below 1.5 cps. For the higher frequencies, the agreement was poor.

In order that a more quantitative evaluation of the technique could be made, we performed digital analyses. Two different approaches were taken; one was statistical, the other nonstatistical (or deterministic). Since we base our analyses on a high signal-to-noise ratio, a deterministic approach using numerical transforms seems more logical. On the other hand, since we are unable to determine what is really our signal, as it is invariably contaminated by random noise, a statistical method using power density spectra seems quite desirable. Both results are studied in this report.

It so happened that the events studied in the previous report did not fall on the same dates as the end of the month calibrations, which had more calibration information. However, from the day-to-day fluctuation of the motor constant (G), which was found to be within 5%, we construed that end of the month calibration results could be used reliably throughout the month. Since we had both controlled calibration data and a significant signal from an event on June 2, 1965, we concentrated our efforts on this set of data for more quantitative analyses.

The Jerome, Arizona, station (JR-AZ) is situated on the Verde formation, which is primarily a conglomerate of lake deposits consisting of white limestone, gravel, sand, clay, and evaporites. The Z_3 seismometer (figure 1) is on a thin section of a limestone cap which sits on the mesa of the lake deposits. The Z_1 , Z_2 , Z_6 , and Z_7 seismometers are on a fragmented limestone cap approximately 100 feet thick, and the Z_4 and Z_5 seismometers are on a relatively uniform river bed consisting of unconsolidated, incompetent sands, gravels, and clays.

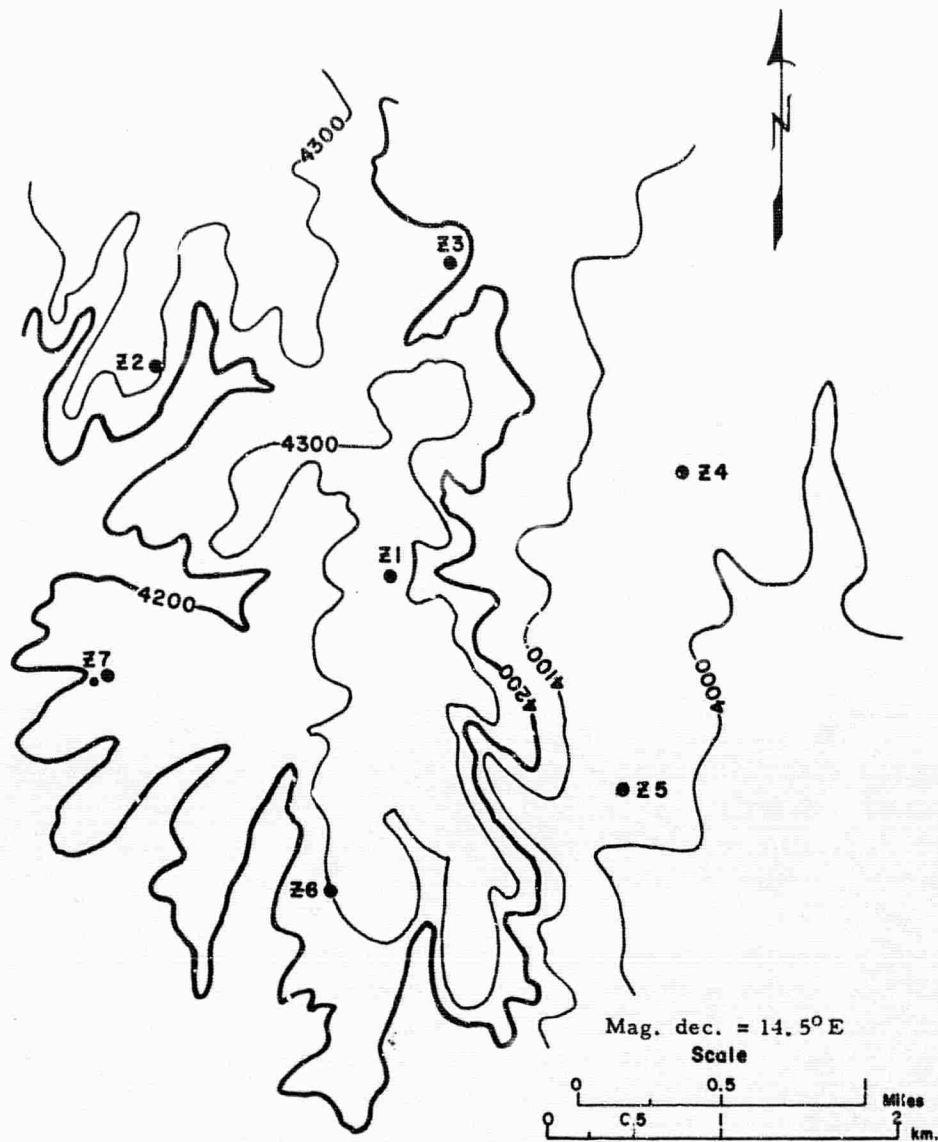


Figure 1. Jerome hexagonal subarray (JR-AZ)

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The Winslow, Arizona, station (WO-AZ) on the other hand, is on the Moenkopi formation which consists of nonmarine, locally gypsiferous, sandy, and silty redbeds interfingering with some calcareous strata. All seven seismometers are situated on this formation (figure 2). These two stations, therefore, made it possible to study the effects of both heterogeneity and homogeneity in local geological structures.

Since our qualitative study indicated the feasibility of taking short record lengths for signal analysis, the current study was confined to 2- and 3-second signal segments. The origin time of the June 2, 1965, Pacific event utilized was 13:57:51z. The epicenter was located at 4.6S, 105.6W (near Northern Easter I. Cordillera) and had a body-wave magnitude (m_b) of 4.8. The epicentral distance from these two stations was approximately 4500 km, which is within the teleseismic window (Carpenter, 1965) for signal transmission.

2. TECHNIQUES OF SPECTRAL ANALYSIS

The mathematical model used for our calibration study was explained in the theoretical considerations of the previous report (Huang, 1965b). The following assumptions were made with regard to our imperfect model:

- a. The data contain high signal-to-noise ratios to the extent that noise can be ignored.
- b. The effect of dispersion and scattering is very slight from seismograph to seismograph so that station correction for the attenuation is unnecessary.
- c. Local geology is isotropic and homogeneous.

The resulting calibration curves are given by

$$A(\omega) = A_c(\omega) \left| \frac{M(\omega)}{M_c(\omega)} \right|, \quad (1)$$

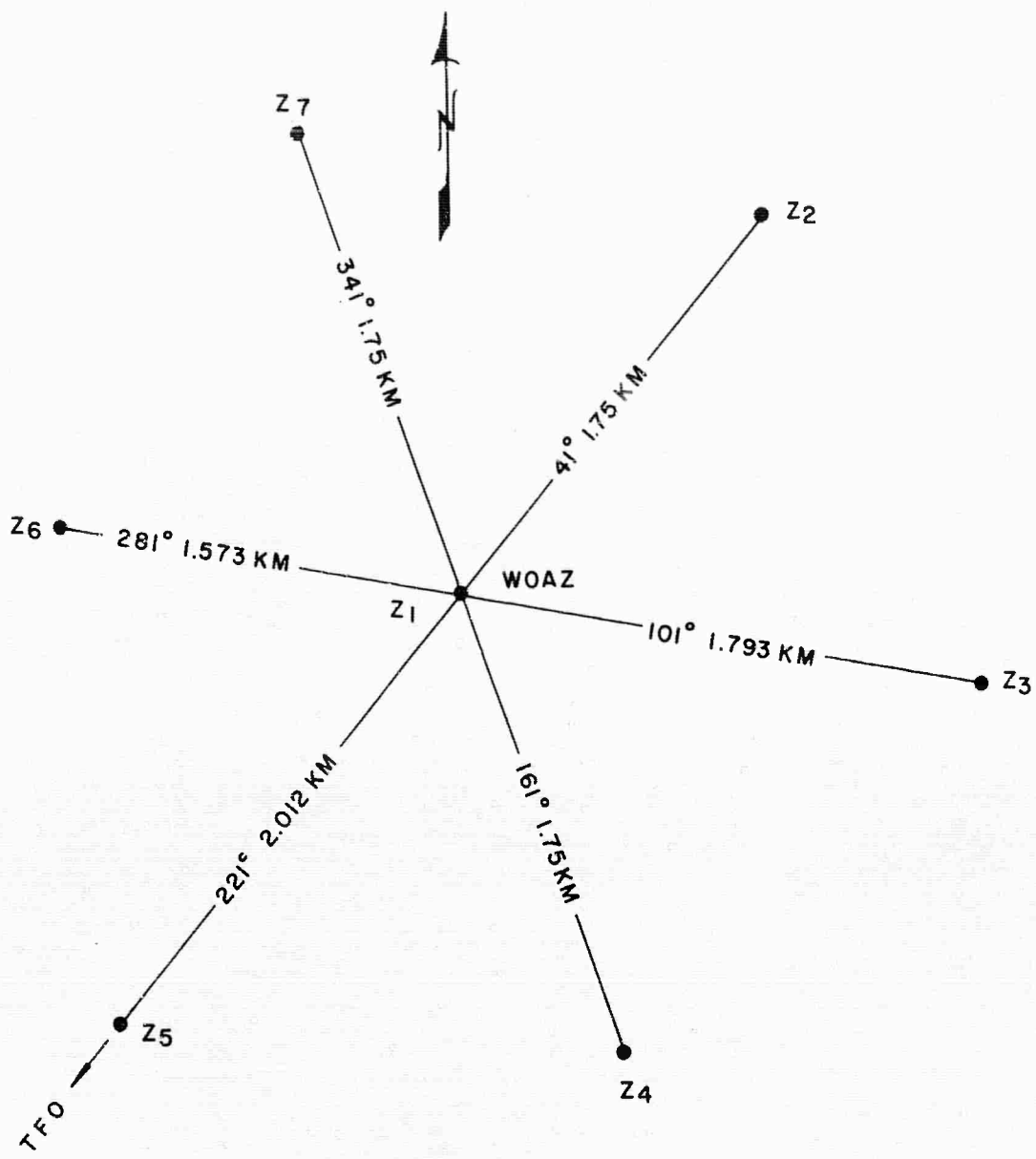


Figure 2. Winslow hexagonal subarray (WO-AZ)

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and

$$B(\omega) = B_c(\omega) + \frac{M(\omega)}{M_c(\omega)}, \quad (2)$$

where $A(\omega)$ and $B(\omega)$ are the amplitude and phase response curves of the estimated calibration results. $\left| \frac{M(\omega)}{M_c(\omega)} \right|$ and $\frac{M(\omega)}{M_c(\omega)}$ are the amplitude and phase curves of the spectral ratios of the received signals, and $A_c(\omega)$ and $B_c(\omega)$ are the magnification and phase-response curves of the center instrument, which serves as a standard. ω is circular frequency.

From equations (1) and (2), it is obvious that spectral analysis constitutes the core of our evaluation scheme. Digital spectral analysis is used in this report. The magnetic tapes from JR-AZ and WO-AZ, which recorded the Pacific event of June 2, 1965, were digitized using the Geotech digitizing system.

Two computer programs were used; the program "NUTRAN," which performs numerical transforms was written by Paul Kozsuch of Geotech. The program "BLACKY," which was written by personnel of the Seismic Data Laboratory (SDL) of Alexandria, Virginia, performs power density spectral analysis (Blackman and Tukey, 1958). A brief description of these two programs is given below.

2.1 PROGRAM NUTRAN

For a digitized trace of finite length, the numerical transform theorem (Huang, 1965a) can be conveniently applied. This relationship is given by

$$f_i = \frac{1}{m} \sum_{n=0}^{m-1} \sum_{k=0}^{m-1} f_k e^{-j \frac{2\pi n}{m} (k-i)}, \quad i = 0, 1, 2, \dots, m-1, \quad (3)$$

where j is an imaginary unit.

The relationship is similar to the well-known Fourier Transform Theorem for a continuous function of an infinite length, that is,

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} d\omega \int_{-\infty}^{\infty} f(\xi) e^{-j\omega(\xi-t)} d\xi. \quad (4)$$

The complex expression for Fourier analysis is given by

$$f(t) = \sum_{n=-\infty}^{\infty} C'_n e^{-j \frac{\pi n t}{T}}, \quad (5)$$

where the coefficients C'_n are determined by

$$C'_n = \frac{1}{2T} \int_{-T}^T f(t) e^{-j \frac{\pi n t}{T}} dt. \quad (6)$$

$2T$ in the expression (6) represents the period.

With a properly chosen digitization rate and filter setup, the digital spectrum can be given by

$$F_n = \frac{1}{m} \sum_{k=0}^{m-1} f_k e^{-j \frac{2\pi n k}{m}}. \quad (7)$$

The frequency spectrum is related to (6) and (7) by

$$F\left(\frac{\omega}{2\pi}\right) = F\left(\frac{n}{2T}\right) = F_n = C'_n . \quad (8)$$

Detailed descriptions of the above relationships are given in a separate report (Huang, 1965a). In order to perform a Fourier transform, expression (7) is replaced by

$$F_n = \frac{1}{m} \sum_{i=0}^{m-1} f_k e^{-j\frac{2\pi nk}{m'}} , \quad (9)$$

where m' is larger than m .

To perform a spectral stacking, the m' in equation (9) should be taken from a subsection of $2T$. The record length $2T$ and m are related by

$$2T = m\Delta t , \quad (10)$$

where Δt is a digitization increment.

The NUTRAN program listing is contained in Appendix 2. Since a signal is invariably contaminated by random noise, spectral smoothing is performed just before estimation of the calibration results. In order for these results to be comparable to BLACKY outputs, smoothing of the Hanning type (Blackman and Tukey, 1958) is incorporated.

2.2 PROGRAM BLACKY

This program computes power density spectra through Fourier transform of the autocorrelation function.

For a digitized time series X_i , the bias-removed autocorrelation function is given by

$$Y(k) = \frac{1}{n-1} \sum_{i=k+1}^n X_{i-k} X_i - \frac{1}{(n-1)^2} \sum_{i=k+1}^n X_{i-k} \sum_{i=k+1}^n X_i. \quad (11)$$

The power density spectrum is given by

$$F(\omega) = \frac{\delta\omega}{m} \left[\sum_{k=1}^{m-1} 2W(k)Y(k) \cos \frac{\omega k\pi}{m} + Y(0) \right], \quad (12)$$

where

$$\delta\omega = \begin{cases} \frac{1}{2} & \text{for } \omega = 0 \text{ or } m \\ 1 & \text{otherwise} \end{cases}$$

and

$$2W(k) = 1 + \cos \frac{k\pi}{m}.$$

Expressions (11) and (12) are approximations of the results given in "The Measurement of Power Spectra" (Blackman and Tukey, 1958). $W(k)$ is the lag window.

3. RESULTS OF SPECTRAL ANALYSIS

In figures 3 and 4, playbacks of the Northern Easter Island Cordillera event at JR-AZ and WO-AZ are shown. These signals, together with the noise preceeding them, were digitized at a rate of 25 samples per second with a low-pass filter set at 8 cps and 18 dB per octave attenuation. The folding frequency falls at 12.5 cps.

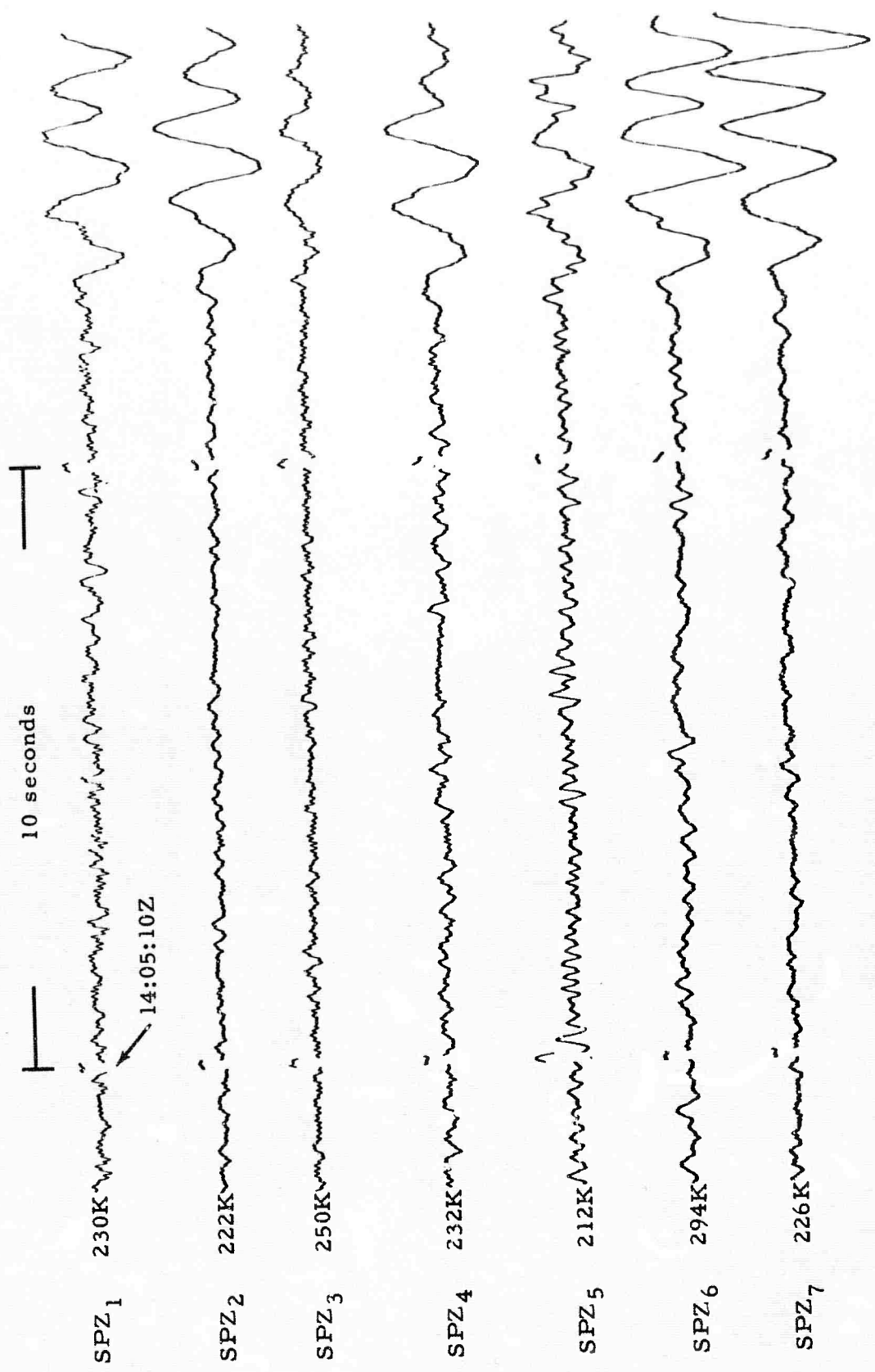


Figure 3. JR-AZ, 2 June 1965. Event from the Northern Easter Island Cordillera. Epicentral data: $\Delta \approx 39.7^\circ$, $h \approx 33\text{km}$, azimuth $\approx 170^\circ$, magnitude ≈ 4.8

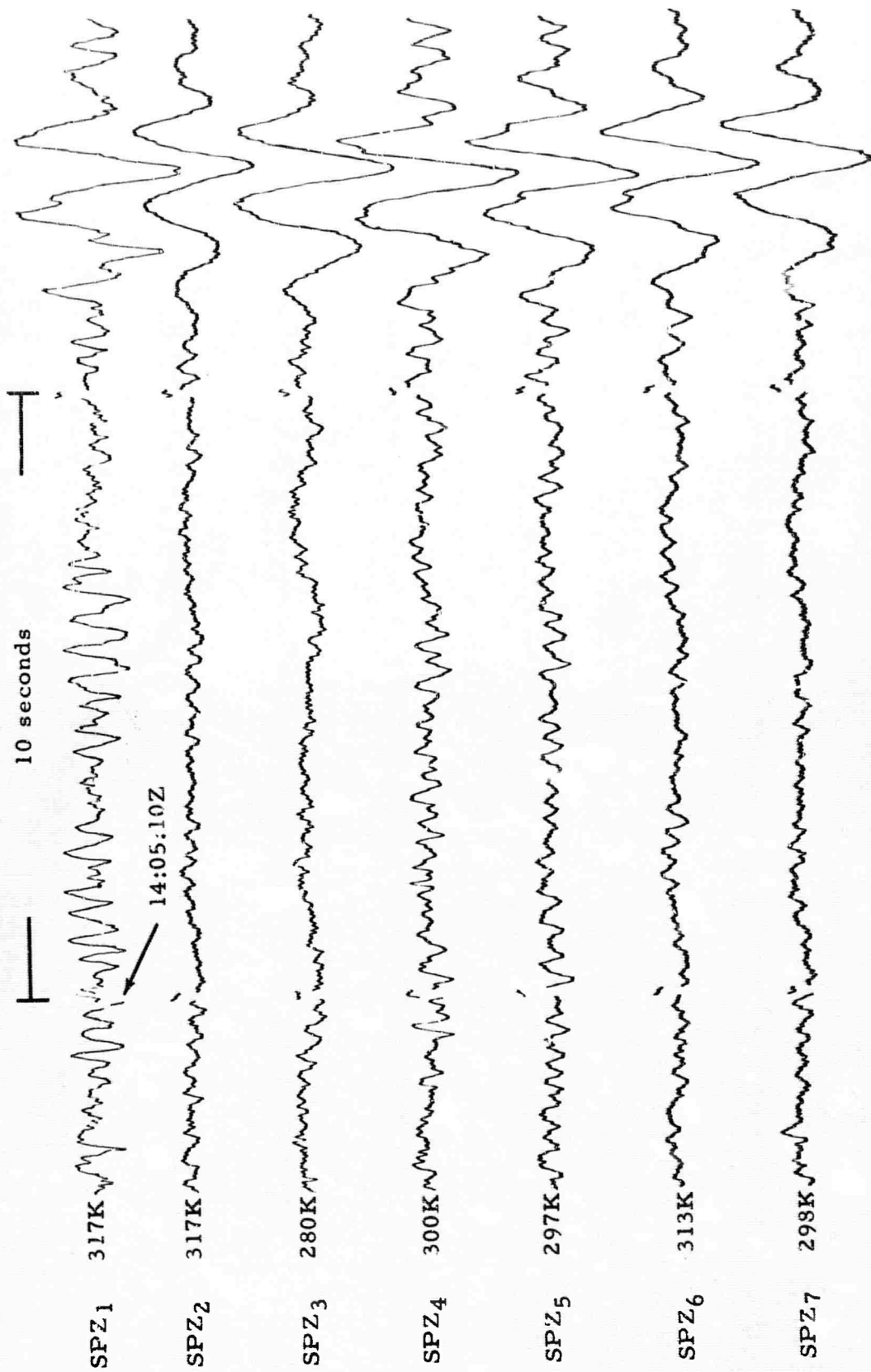


Figure 4. WO-AZ, 2 June 1965. Event from the Northern Easter Island Cordillera.
 Epicentral data: $\Delta \approx 39.6^\circ$, $h \approx 33\text{km}$, azimuth $\approx 172^\circ$, magnitude ≈ 4.8

For the 2-second (50 samples) cases, a maximum lag of 25 samples was used in computing the power density spectra. We assumed that the received signal had a low noise content, and so tried to improve our resolution at the expense of the confidence level of the power spectrum (Blackman and Tukey, 1958). The maximum lag was changed to 10 samples for the 3-second (75 samples) cases, and this difference is reflected in the output of the BLACKY program.

Figures 5 through 18 are the plotted results of the NUTRAN computations. The corresponding BLACKY results are shown in figures 19 through 32.

4. CALIBRATION RESULTS

Application of formulas (1) and (2), together with the reference calibration curves at the hexagonal subarray centers (Appendix 1), yielded the results shown in figures 33 through 80.

4.1 AMPLITUDE RESPONSE

Both smoothed and unsmoothed cases were studied. As already mentioned in section 2 of this report, Hanning smoothing was applied in the frequency domain for the NUTRAN output. Figures 33 through 44 combine the results of the 2- and 3-second unsmoothed cases and figures 45 through 68 show the 2- and 3-second smoothed cases. In general, the 2-second cases agree better with the field calibrations.

A comparison of smoothed and unsmoothed cases indicates that the smoothed spectra give better results. This is probably due to the random nature of the noise contamination. The BLACKY estimates for the calibration were inferior to the NUTRAN estimates. A possible interpretation of this result is that the deterministic approach is more suited for treating signals.

As previously concluded (Huang, 1965b), the agreement between the estimated and the calibrated results is quite good for frequencies below 1.5 cps; the technique requires refinement beyond 1.5 cps.

It is also of interest to note that the geologically homogeneous station, WO-AZ, shows a better agreement than the heterogeneous station, JR-AZ.

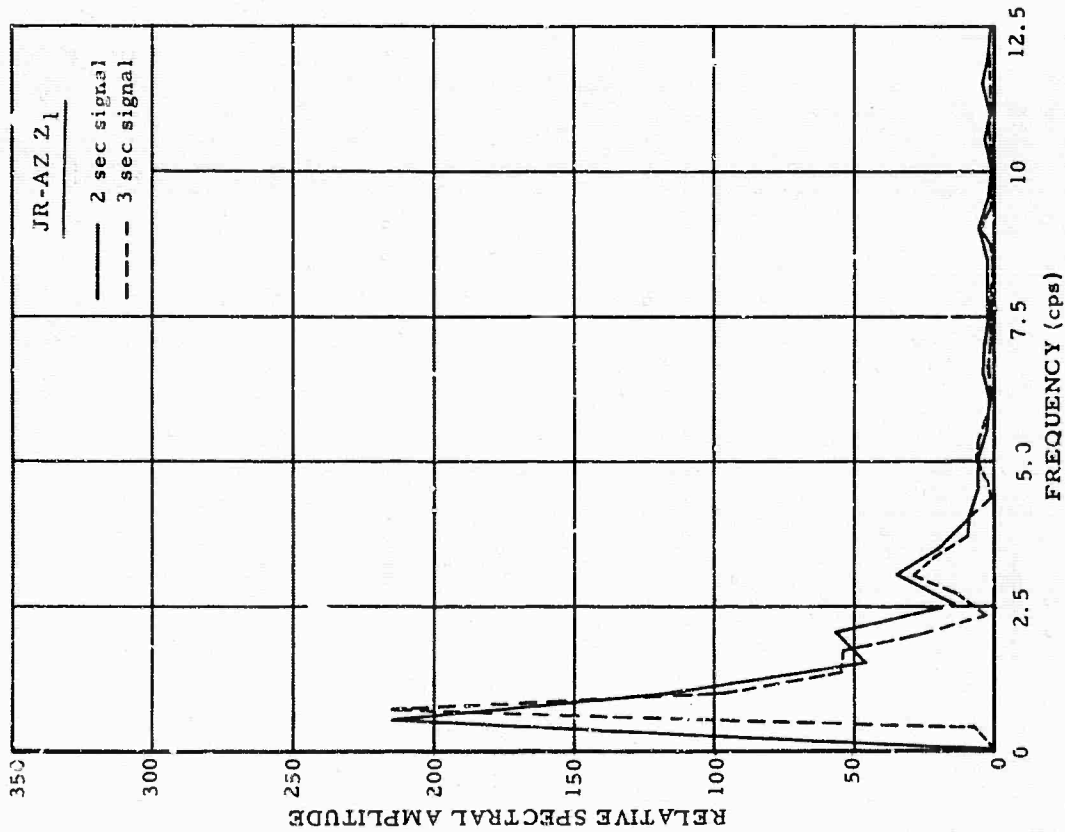


Figure 5. JR-AZ Z1, amplitude spectrum of
2 June 1965 event

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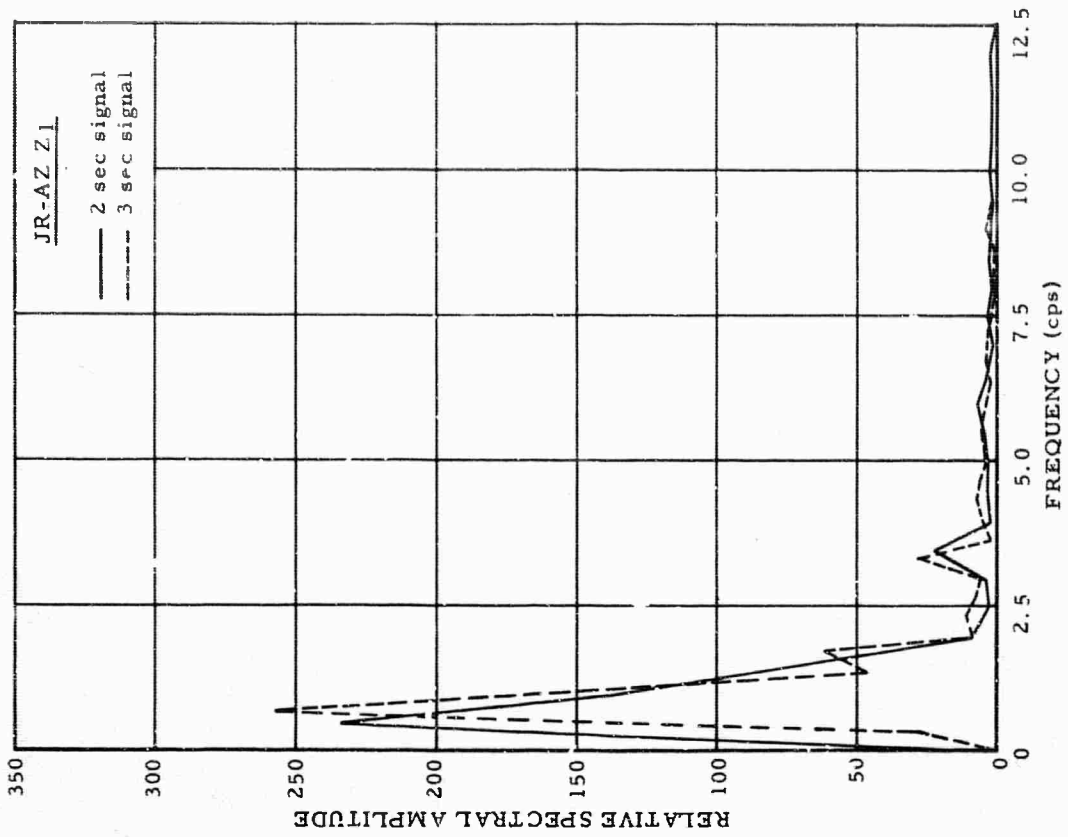


Figure 6. JR-AZ Z2, amplitude spectrum of
2 June 1965 event

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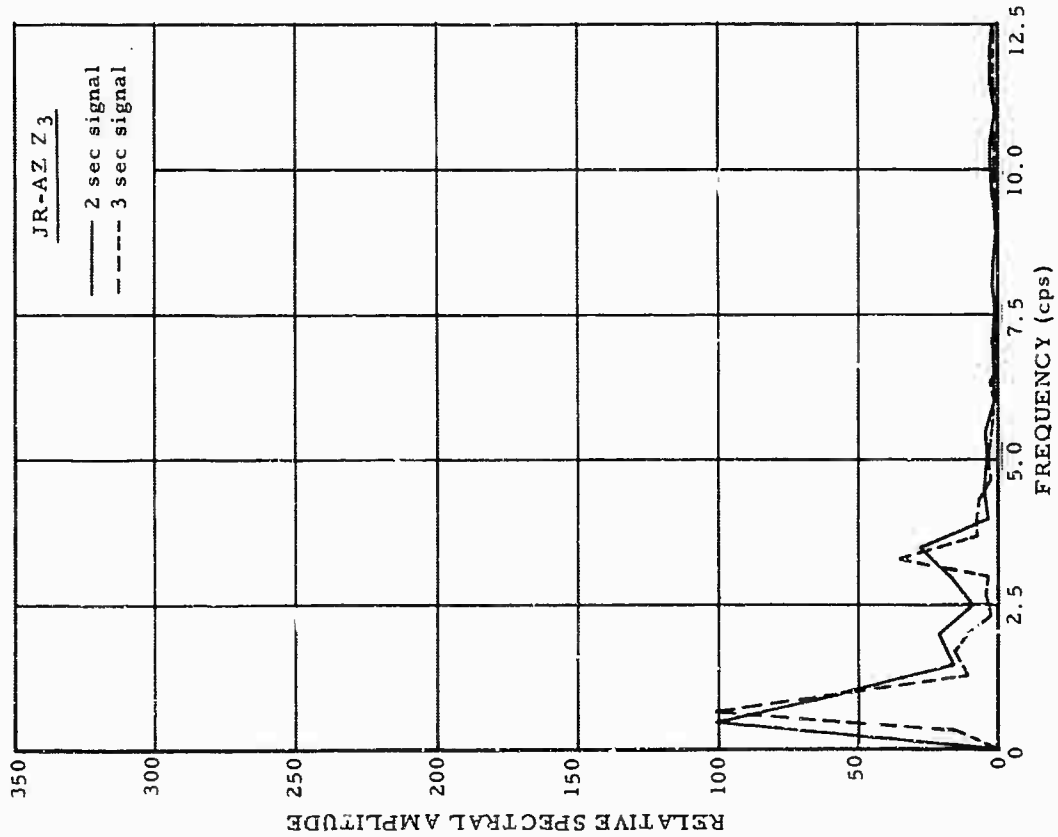


Figure 7. JR-AZ Z3, amplitude spectrum of
2 June 1965 event

G 346

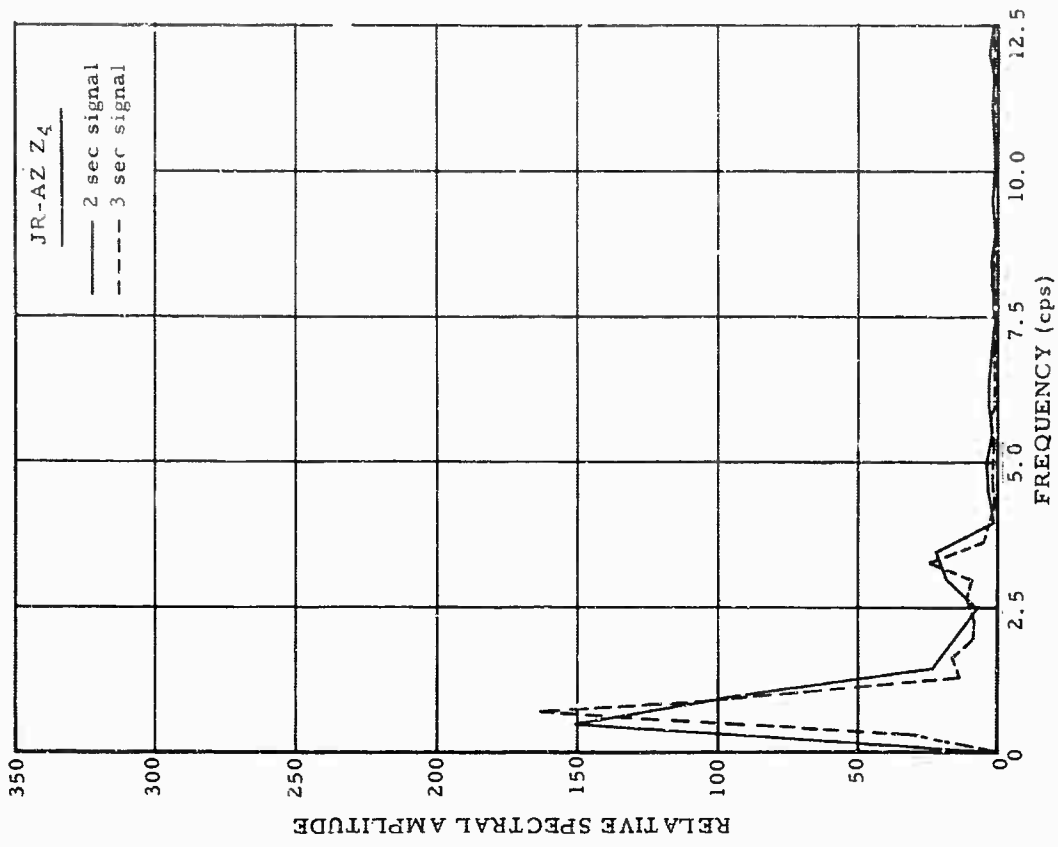


Figure 8. JR-AZ Z4, amplitude spectrum of
2 June 1965 event

G 347

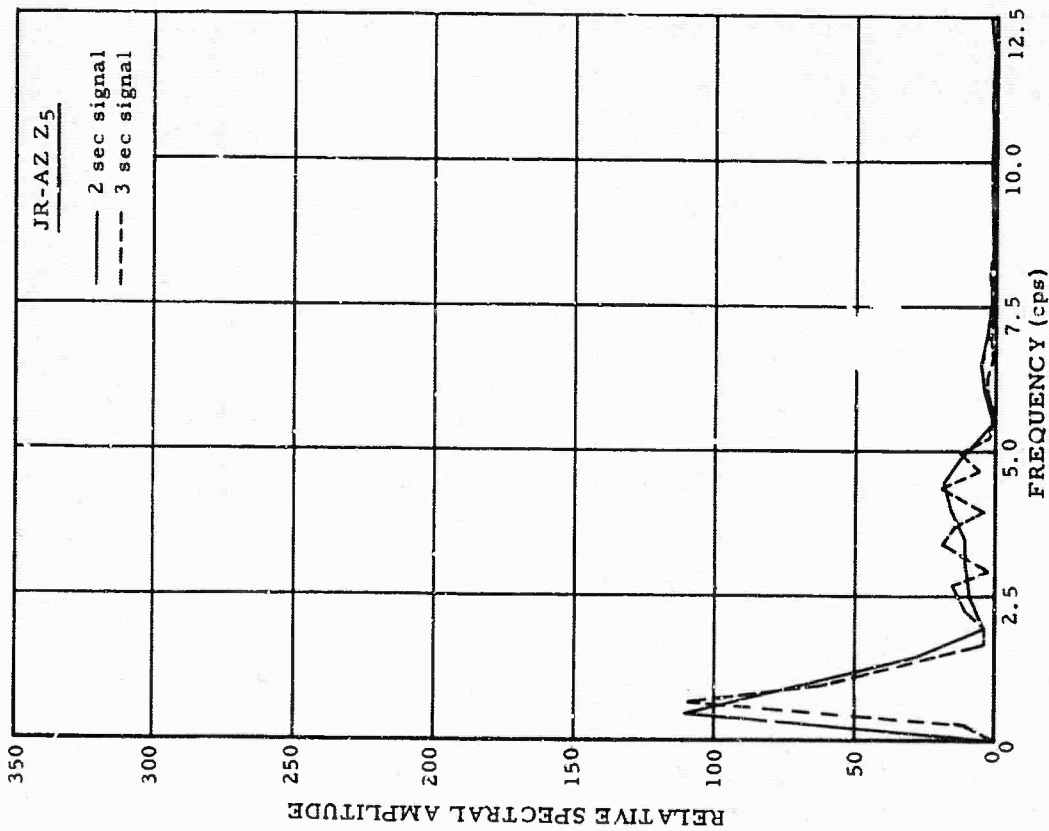


Figure 9. JR-AZ Z5, amplitude spectrum of
2 June 1965 event

G 348

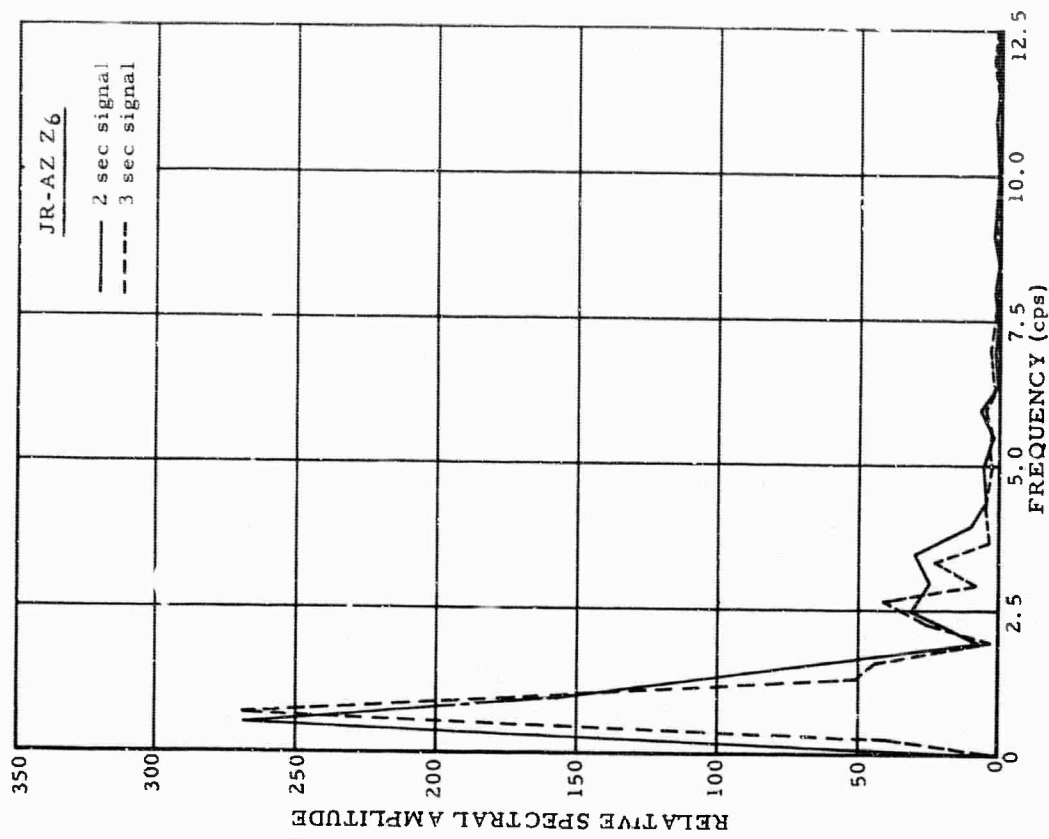


Figure 10. JR-AZ Z6, amplitude spectrum of
2 June 1965 event

G 349

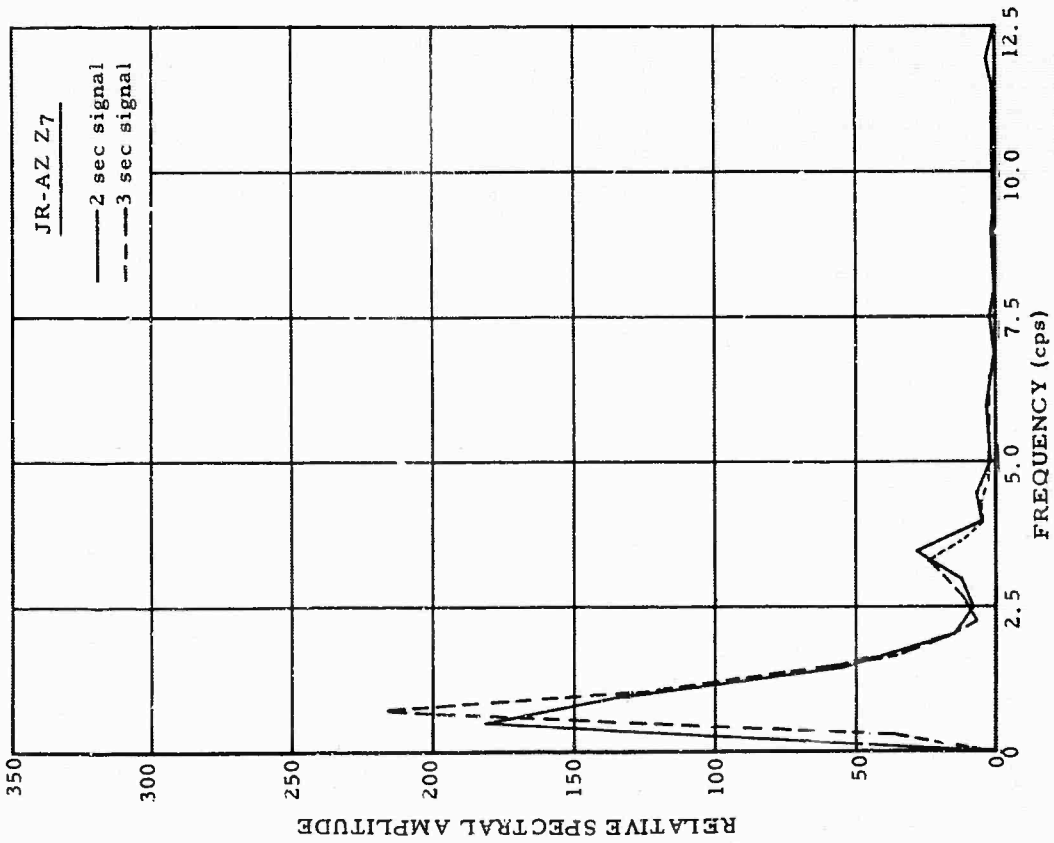


Figure 11. JR-AZ Z7, amplitude spectrum of
2 June 1965 event

G 350

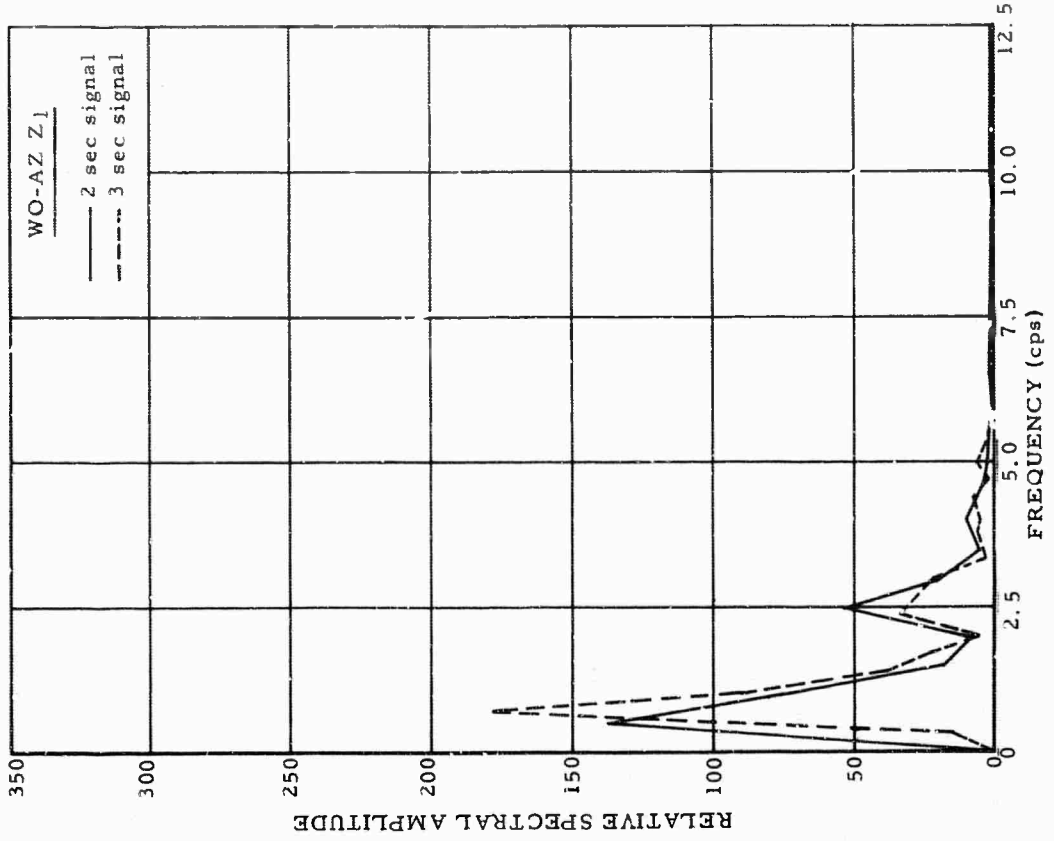


Figure 12. WO-AZ Z1, amplitude spectrum of
2 June 1965 event

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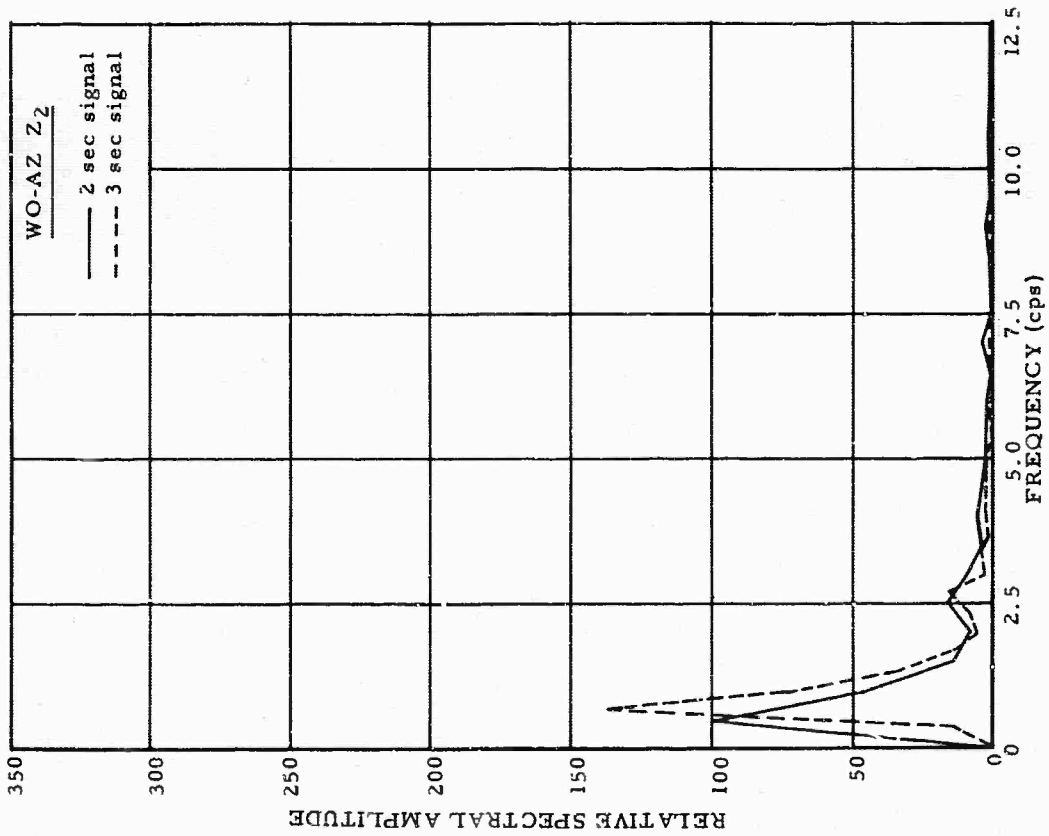


Figure 13. WO-AZ Z2, amplitude spectrum of
2 June 1965 event

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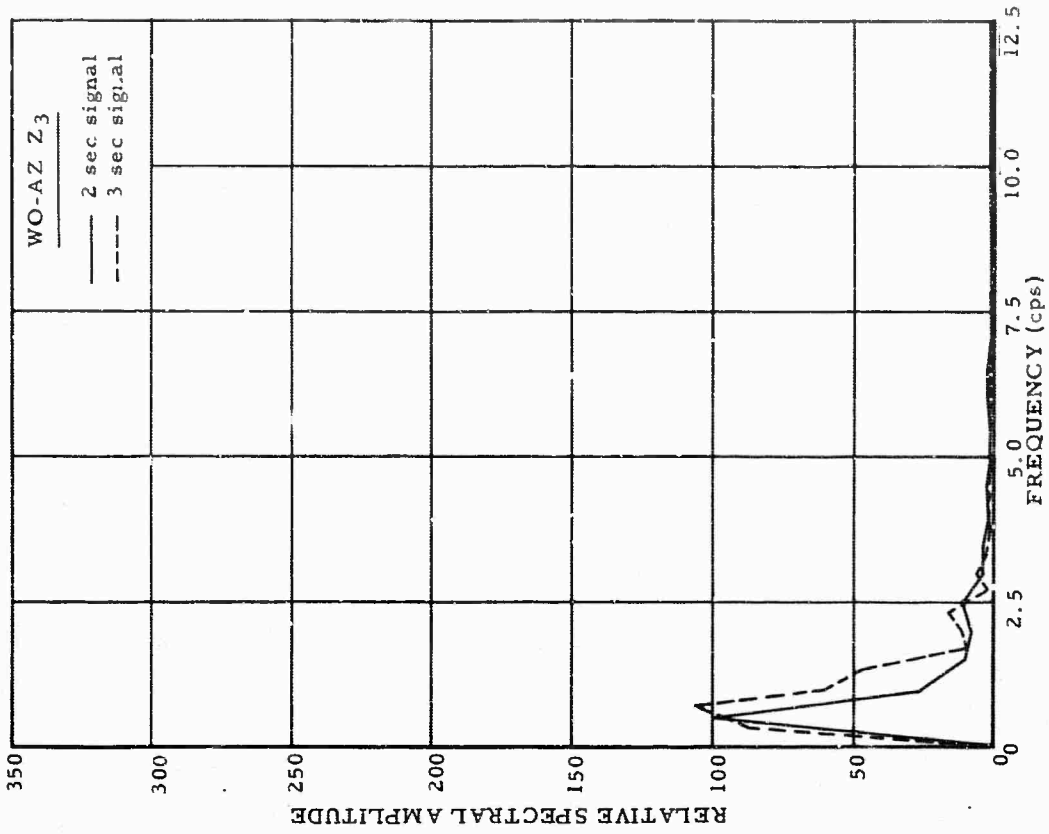


Figure 14. WO-AZ Z3, amplitude spectrum of
2 June 1965 event

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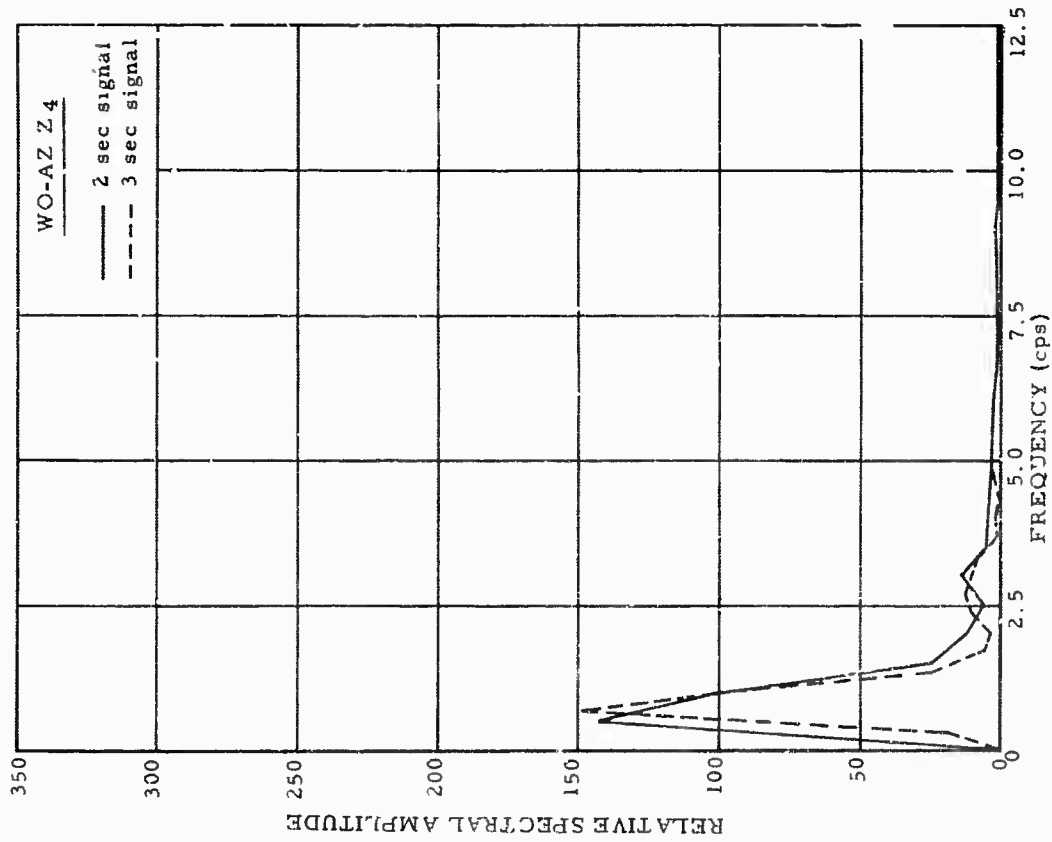


Figure 15. WO-AZ Z₄, amplitude spectrum of
2 June 1965 event

G 354

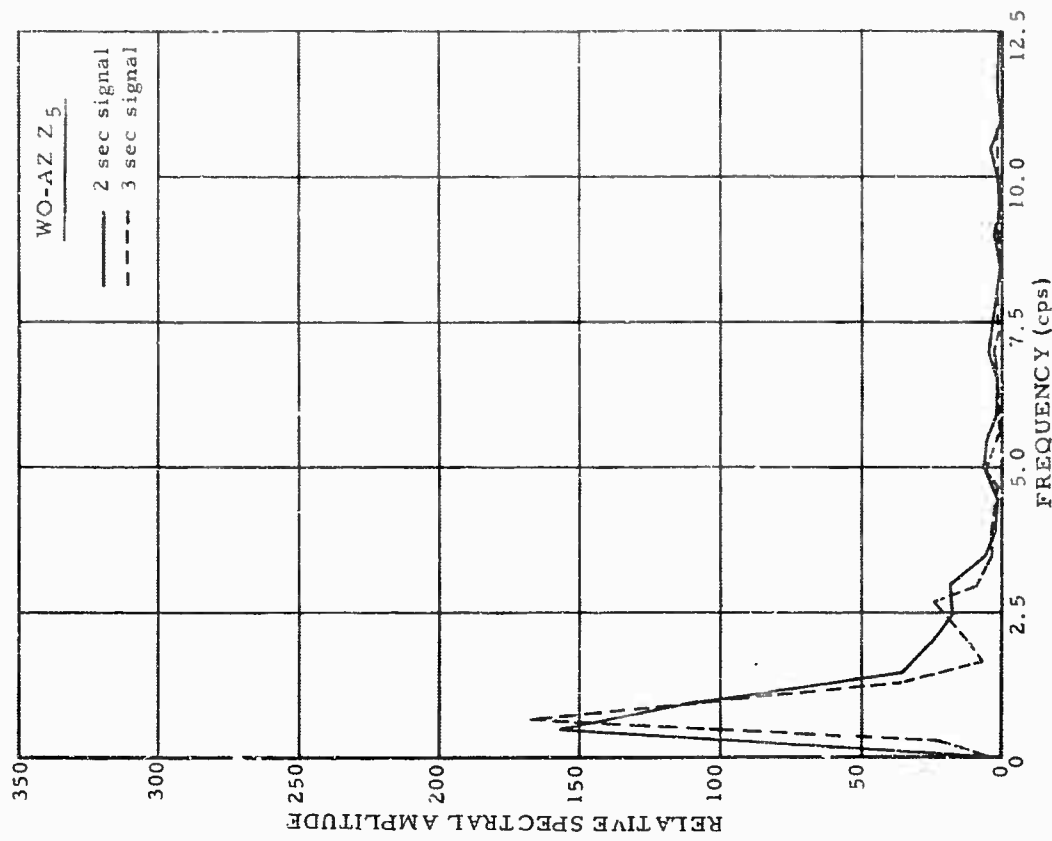


Figure 16. WO-AZ Z₅, amplitude spectrum of
2 June 1965 event

G 355

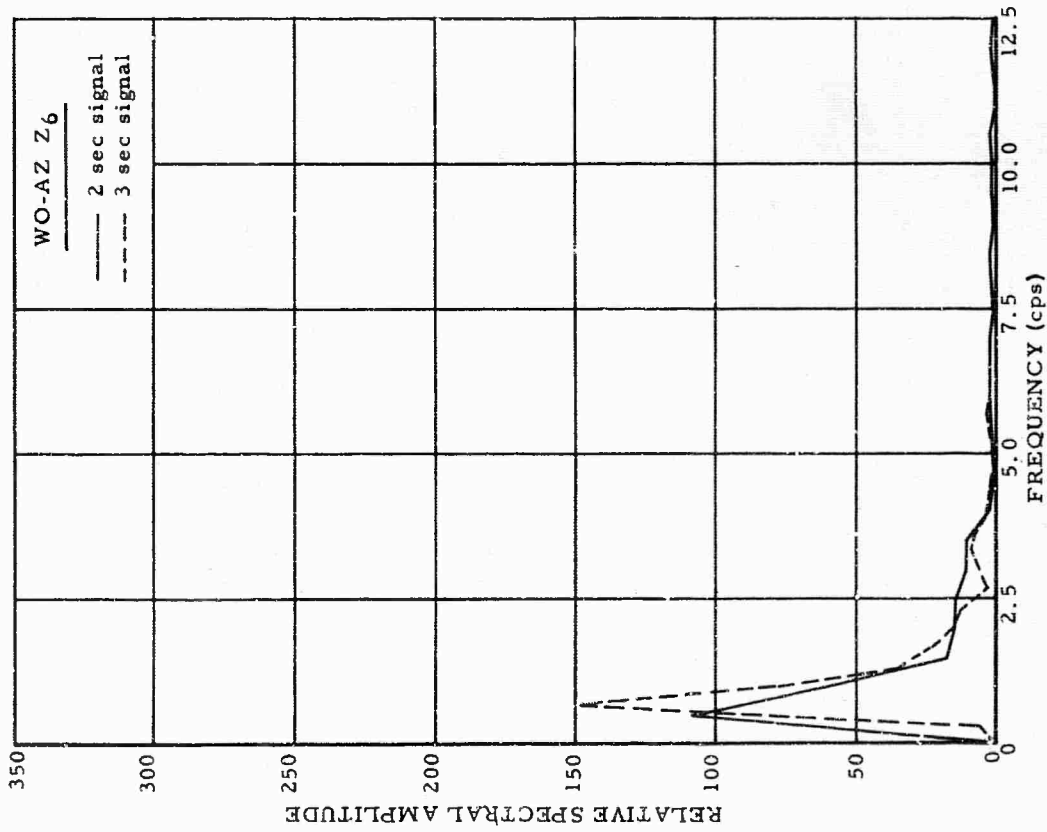


Figure 17. WO-AZ Z₆, amplitude spectrum of
2 June 1965 event

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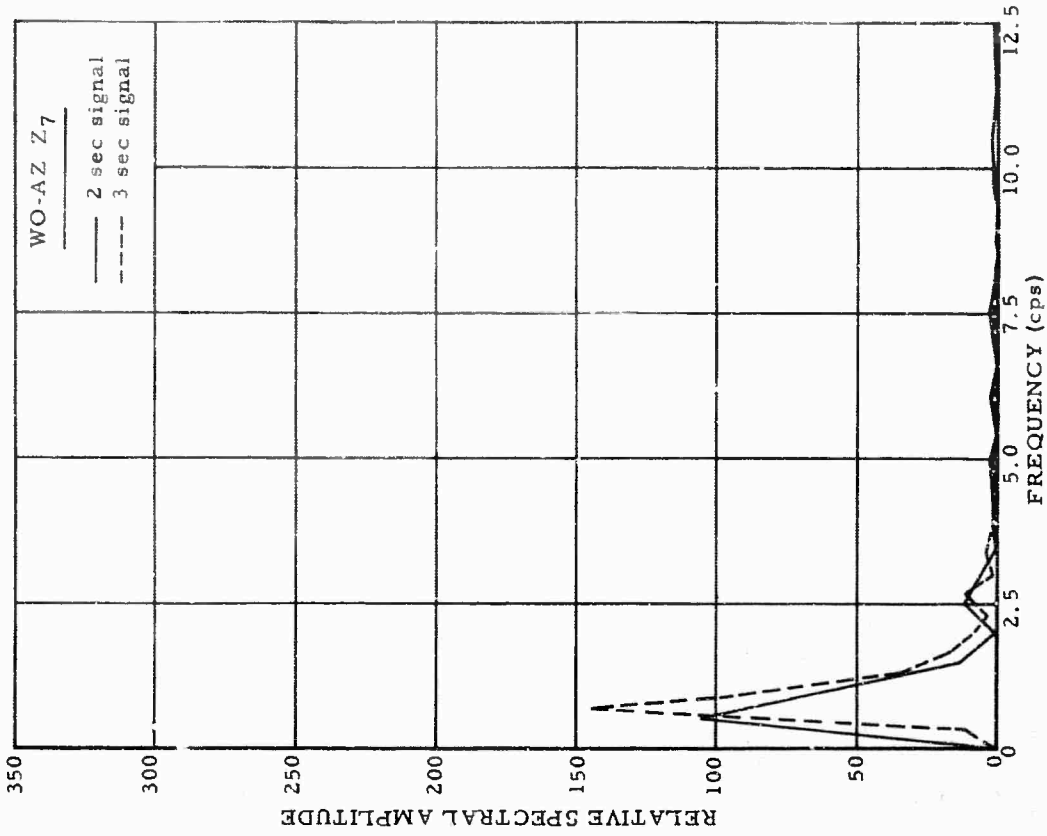


Figure 18. WO-AZ Z₇, amplitude spectrum of
2 June 1965 event

G 357

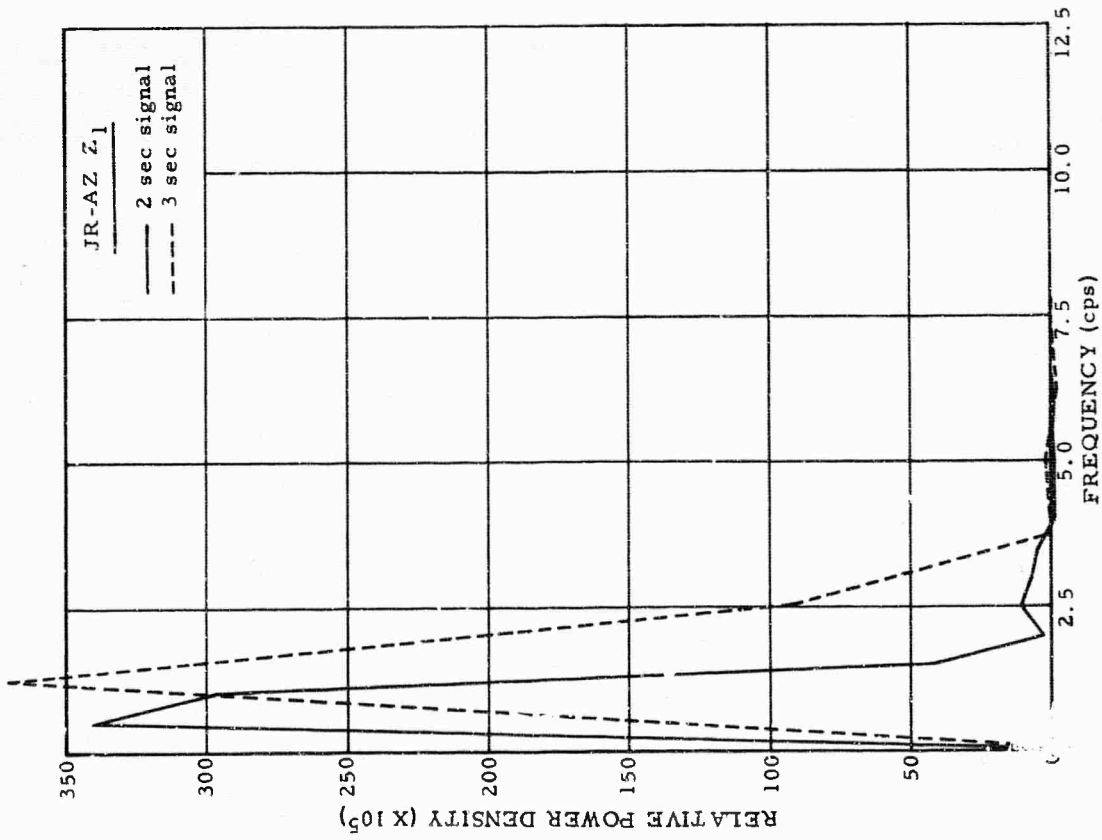


Figure 19. JR-AZ Z₁, power density spectrum of signal, 2 June 1965 event

G 358

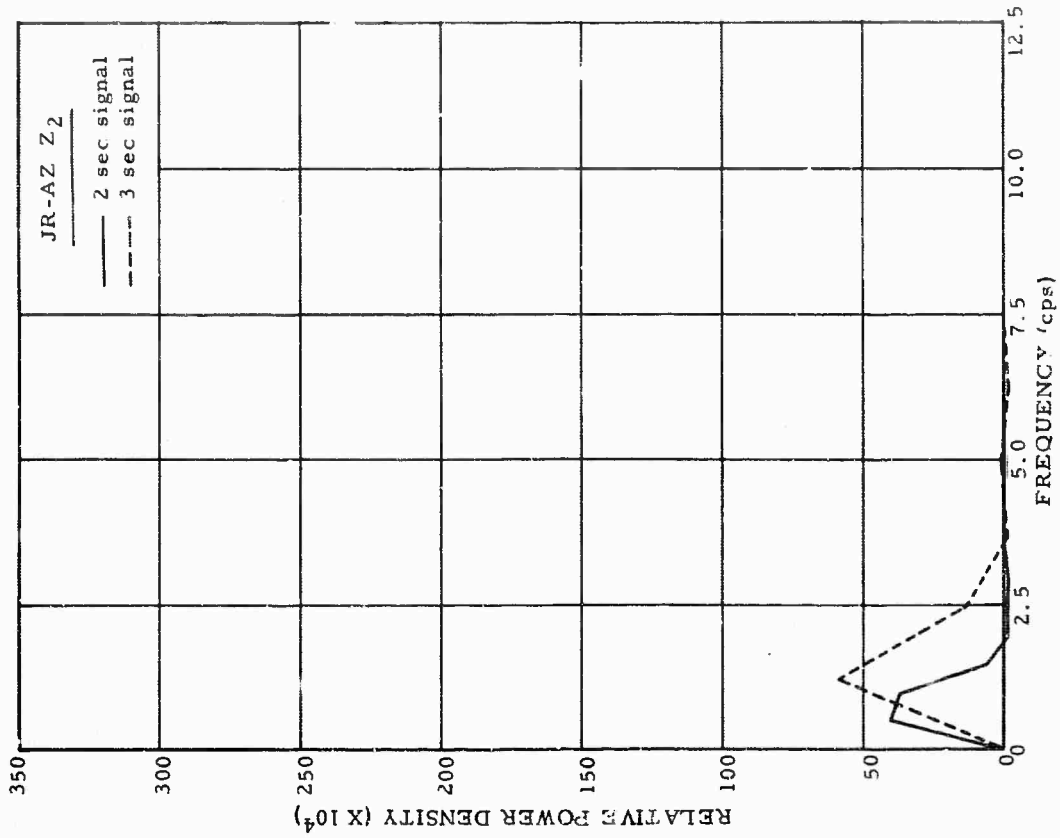


Figure 20. JR-AZ Z₂, power density spectrum of signal, 2 June 1965 event

G 359

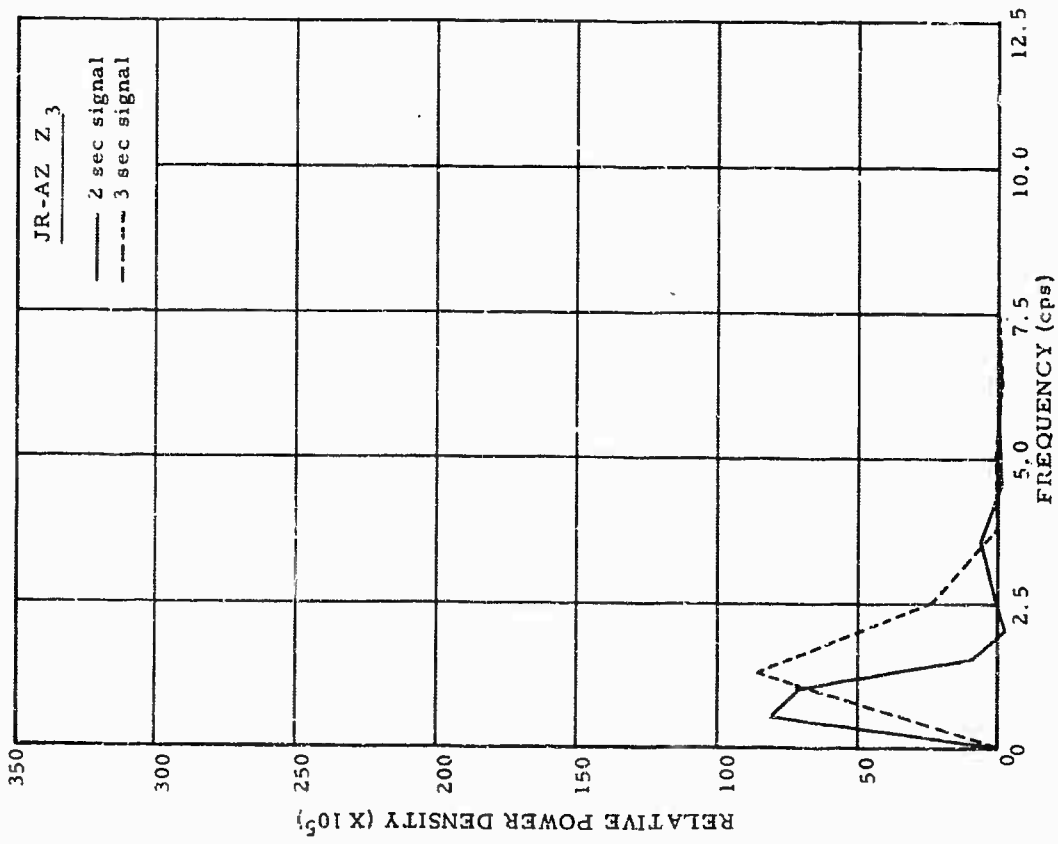


Figure 21. JR-AZ Z₃, power density spectrum of signal, 2 June 1965 event

G 360

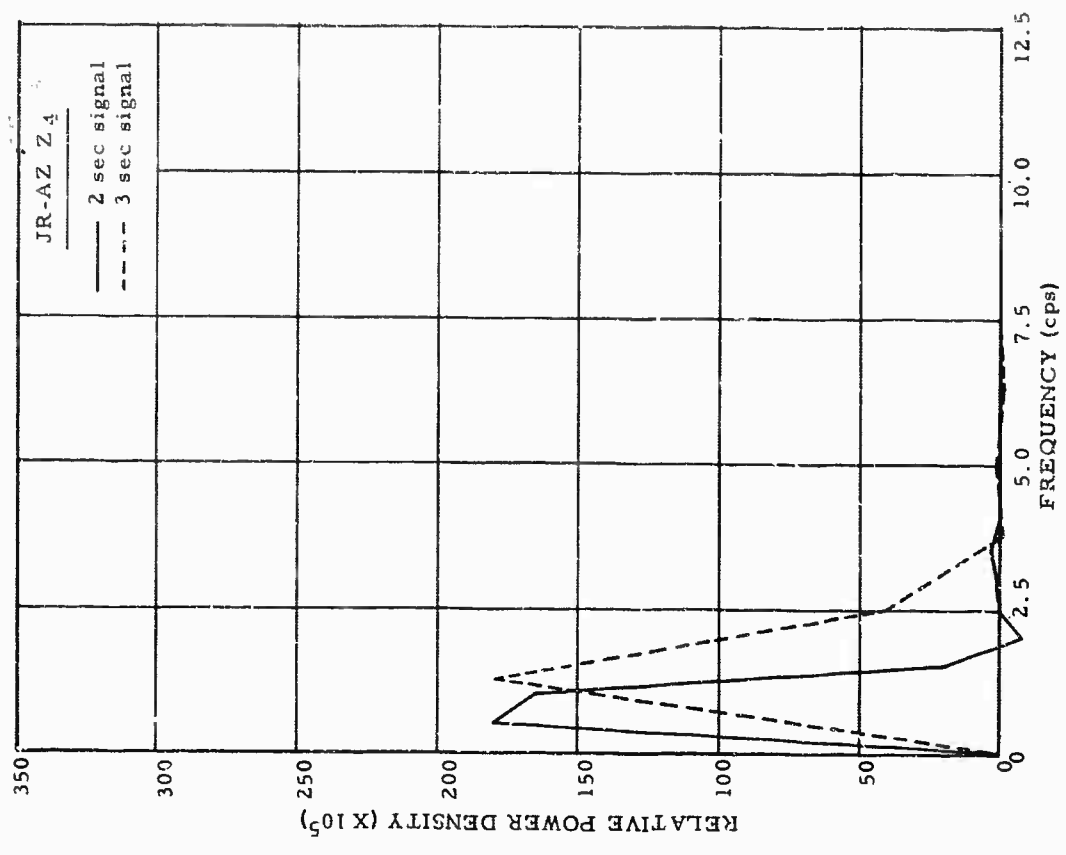


Figure 22. JR-AZ Z₄, power density spectrum of signal, 2 June 1965 event

G 361

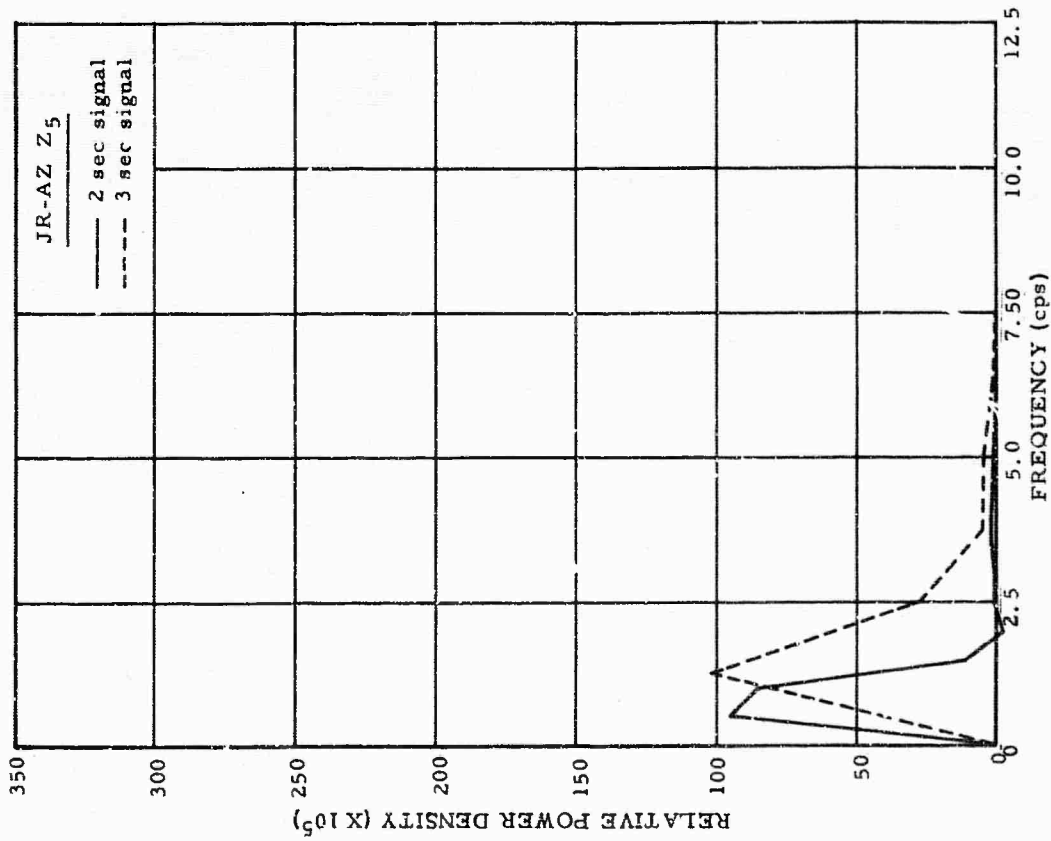


Figure 23. JR-AZ Z5, power density spectrum of signal, 2 June 1965 event

G 362

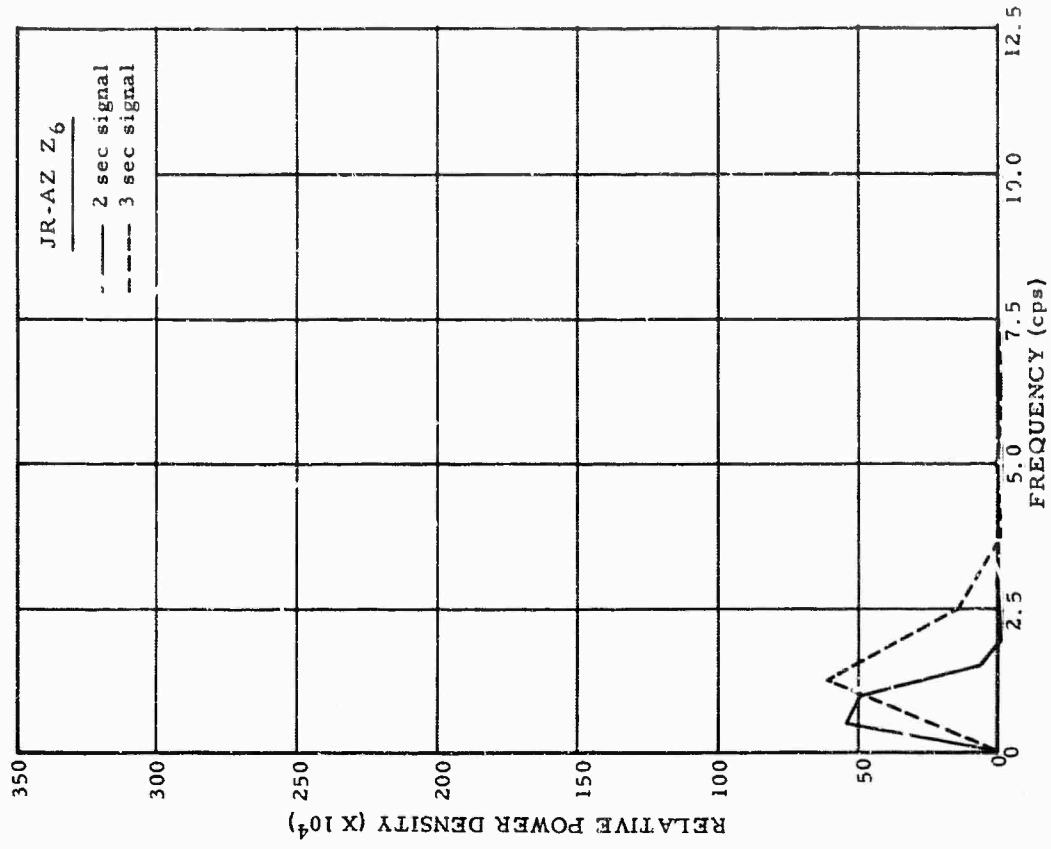


Figure 24. JR-AZ Z6, power density spectrum of signal, 2 June 1965 event

G 363

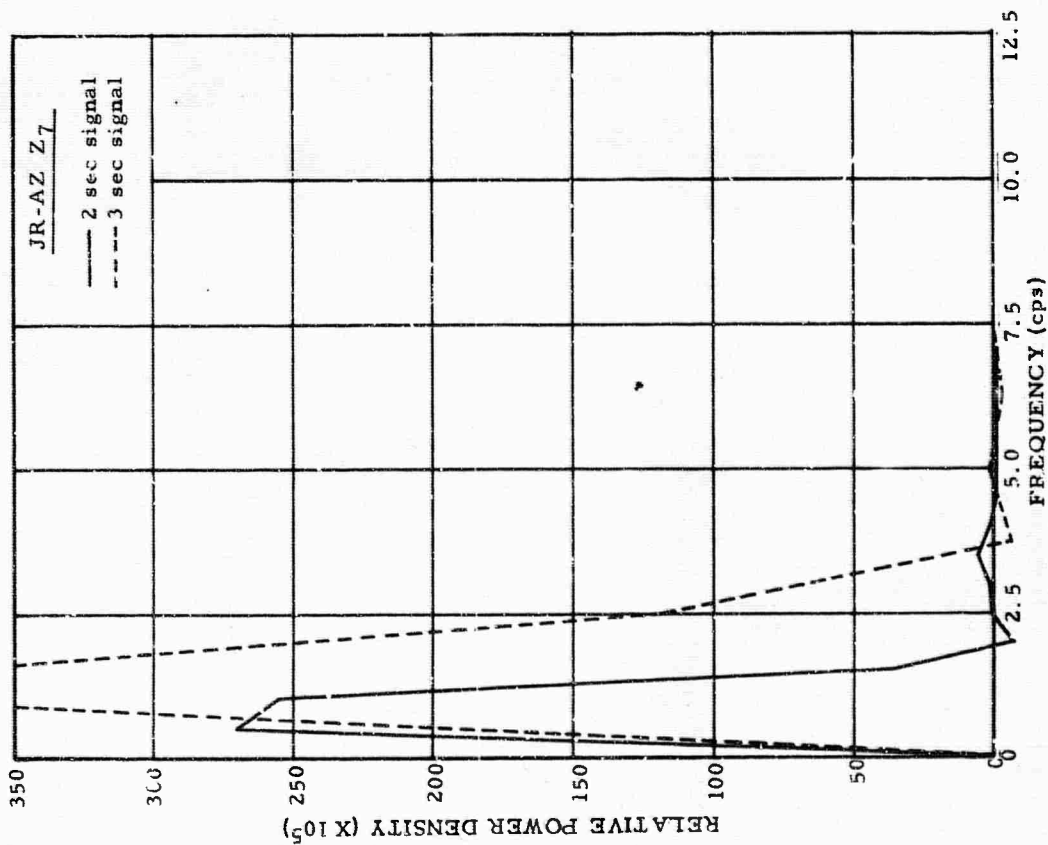


Figure 25. JR-AZ Z7, power density spectrum of signal, 2 June 1965 event

G 364

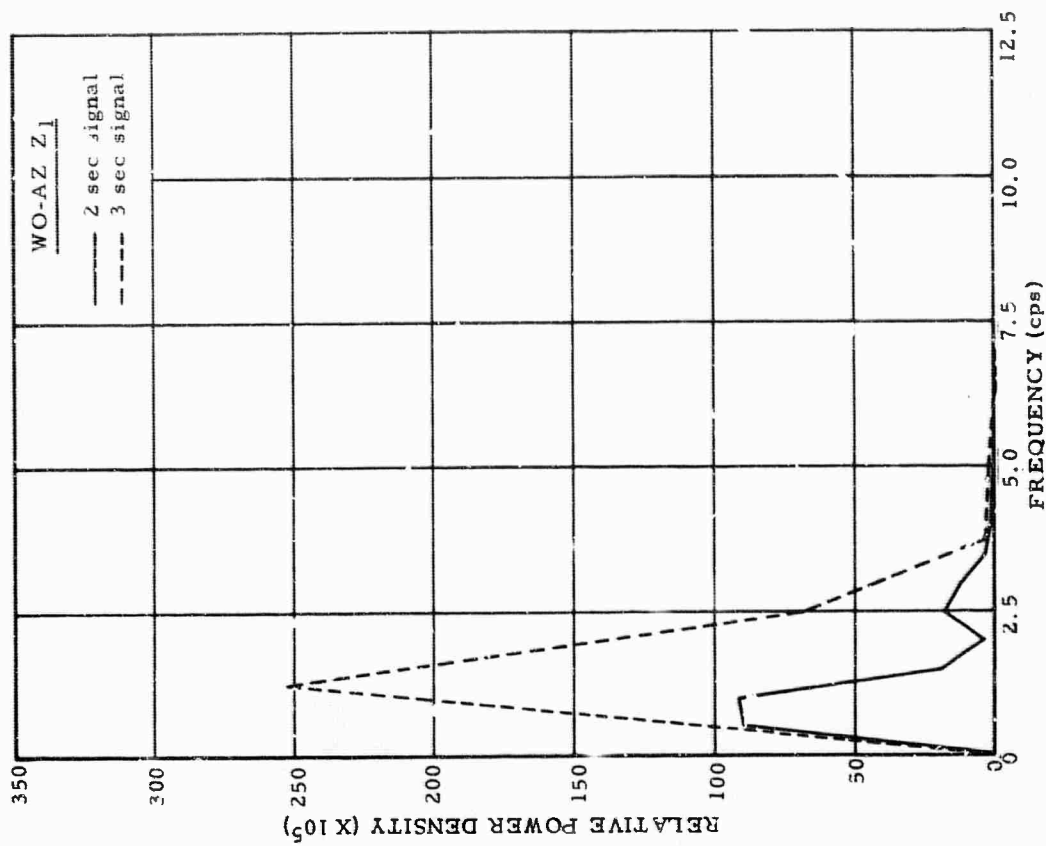


Figure 26. WO-AZ Z1, power density spectrum of signal, 2 June 1965 event

G 365

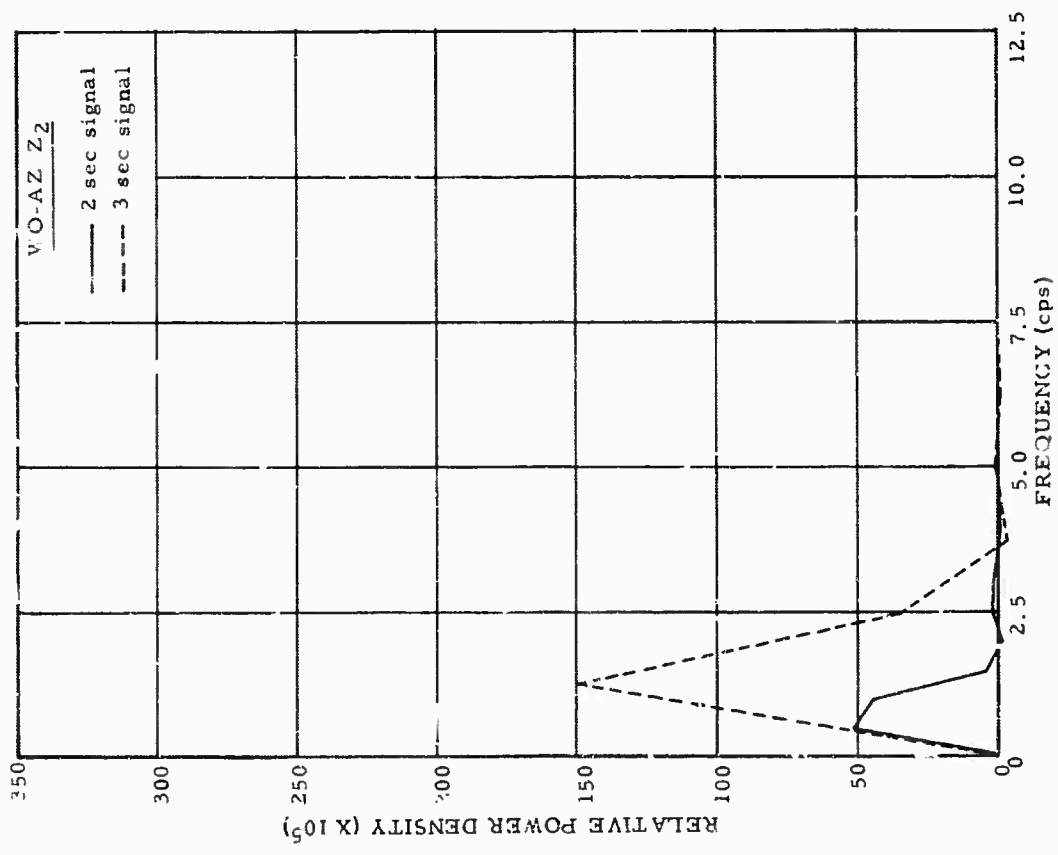


Figure 27. WO-AZ Z₂, power density spectrum of signal, 2 June 1965 event

G 366

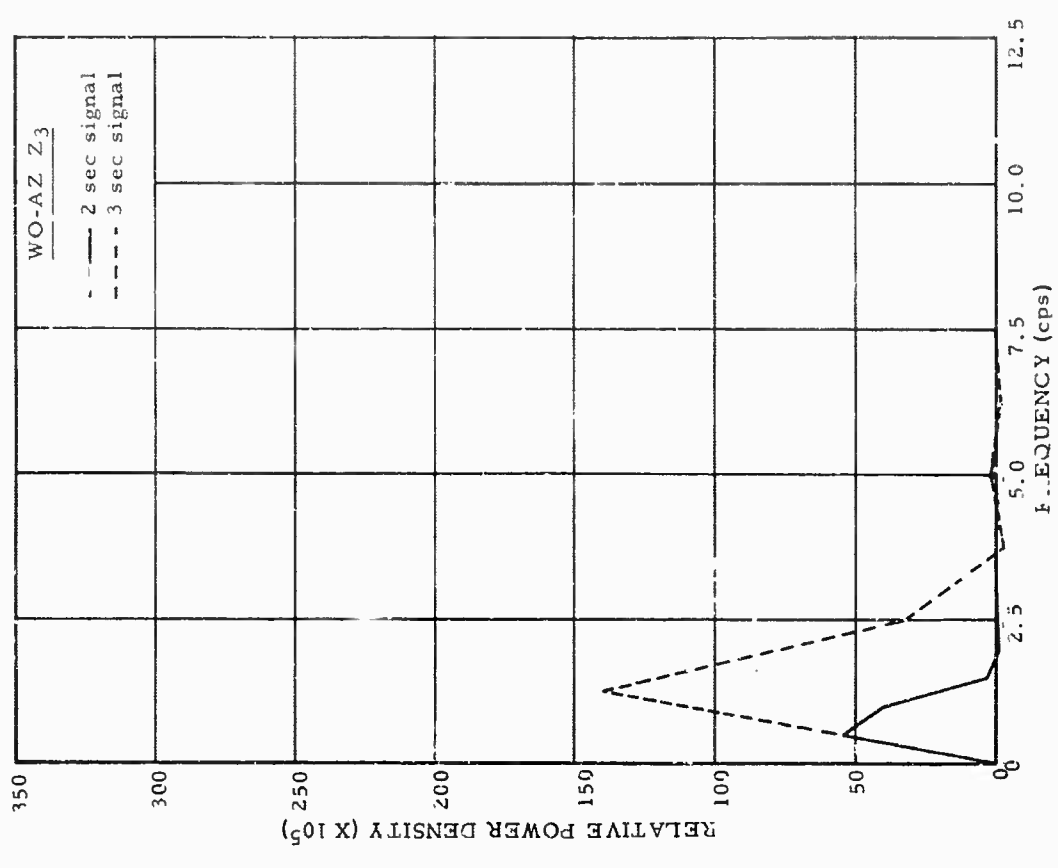


Figure 28. WO-AZ Z₃, power density spectrum of signal, 2 June 1965 event

G 367

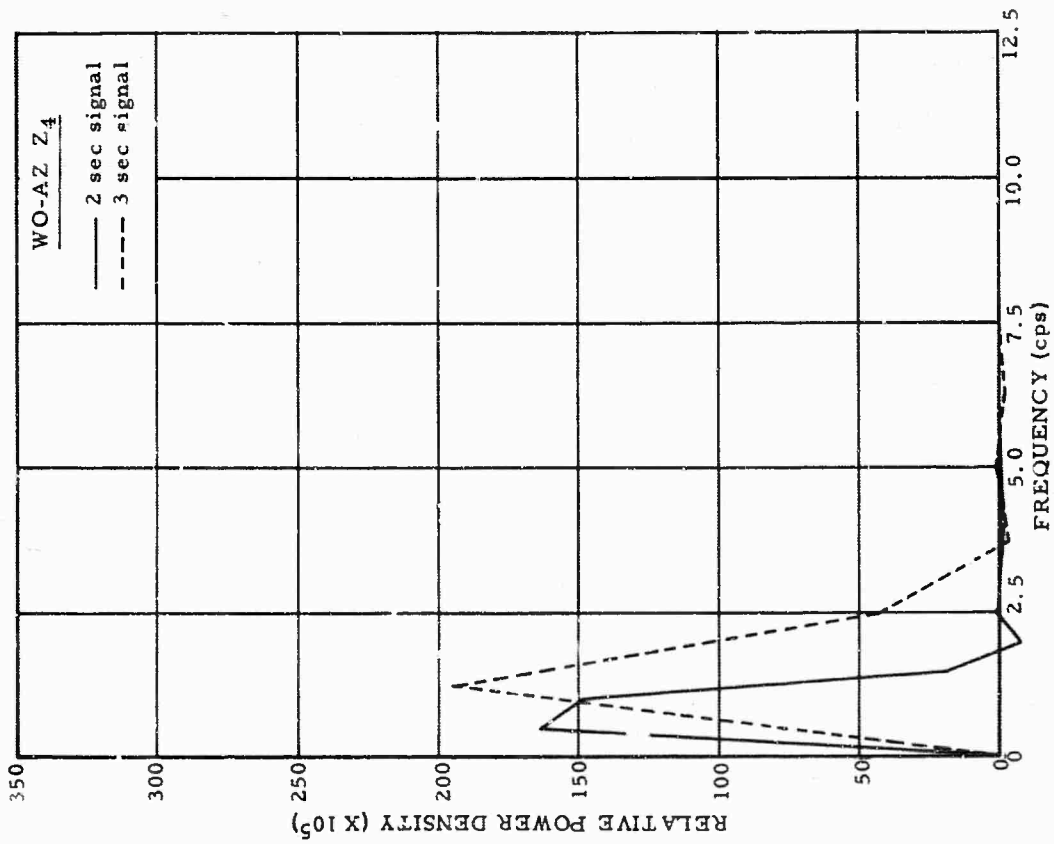


Figure 29. WO-AZ Z4, power density spectrum of signal, 2 June 1965 event

G 368

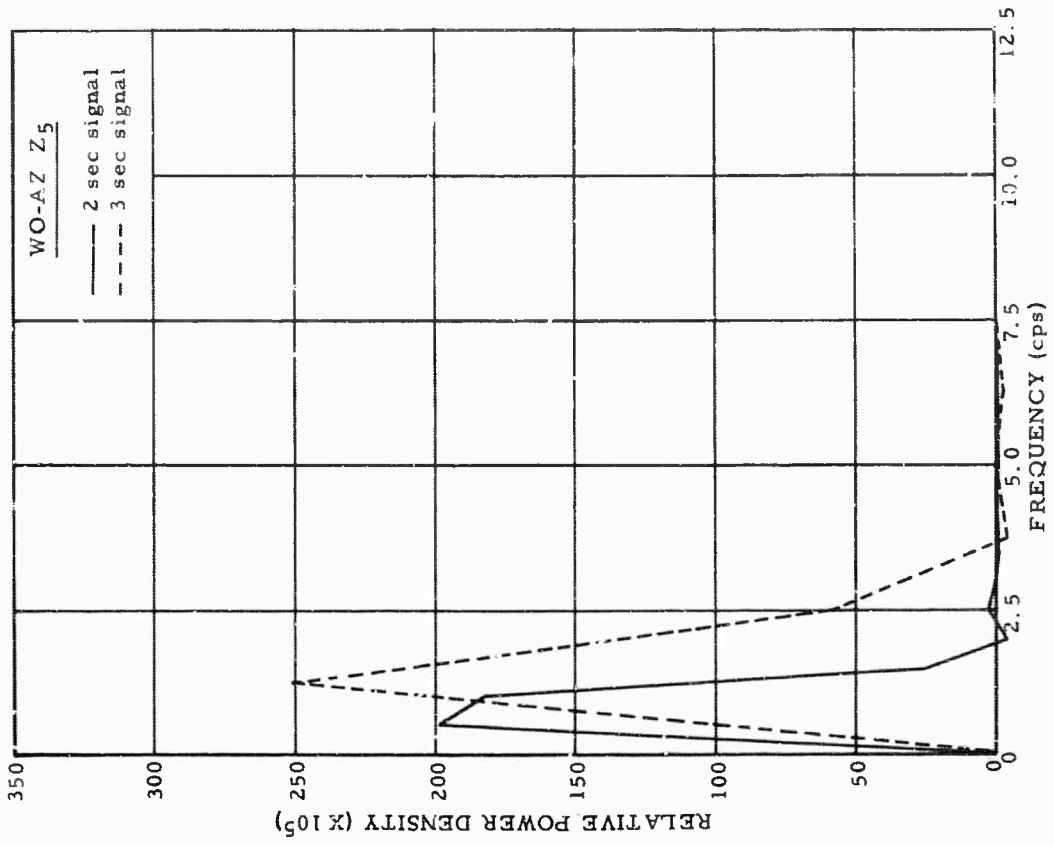


Figure 30. WO-AZ Z5, power density spectrum of signal, 2 June 1965 event

G 369

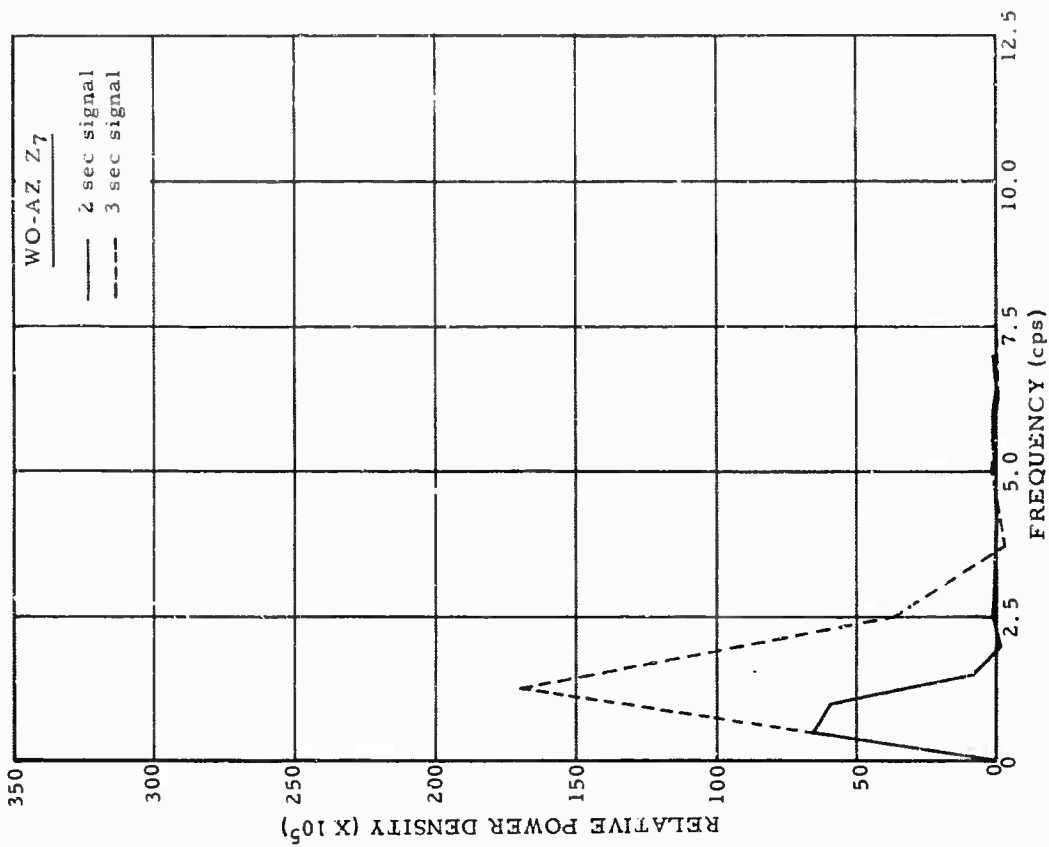


Figure 31. WO-AZ Z6, power density spectrum of signal, 2 June 1965 event

G 370

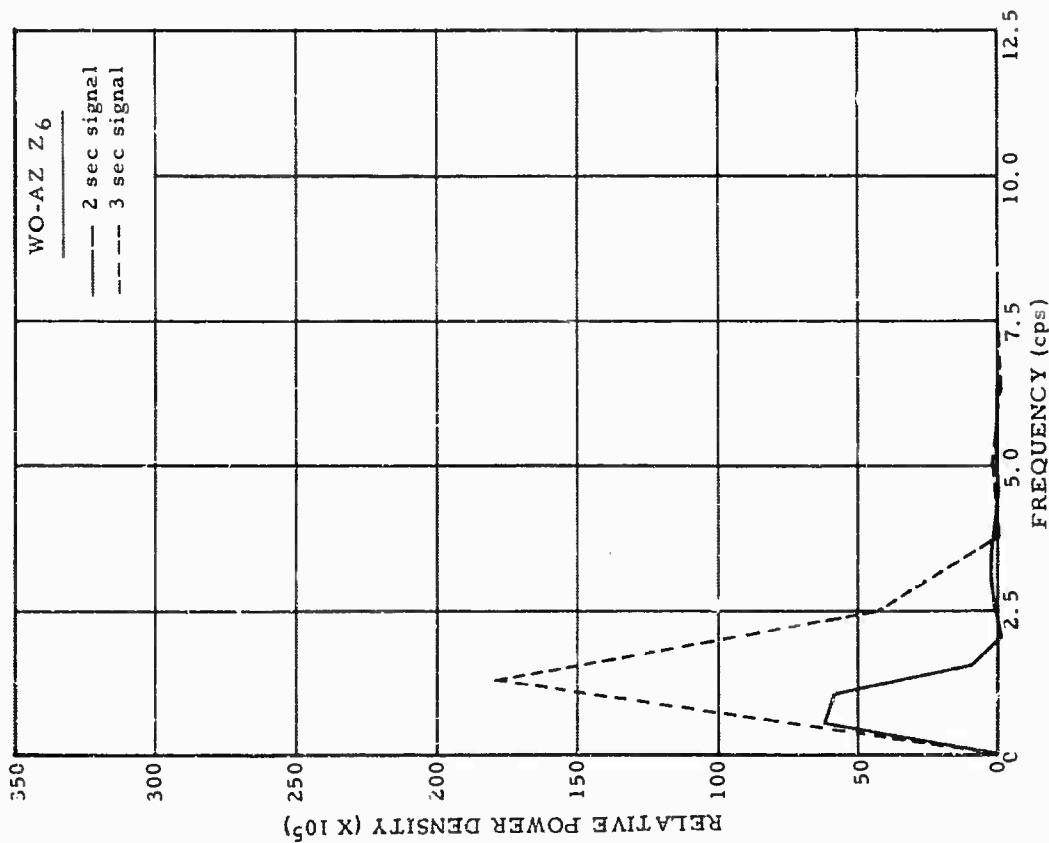


Figure 32. WO-AZ Z7, power density spectrum of signal, 2 June 1965 event

G 371

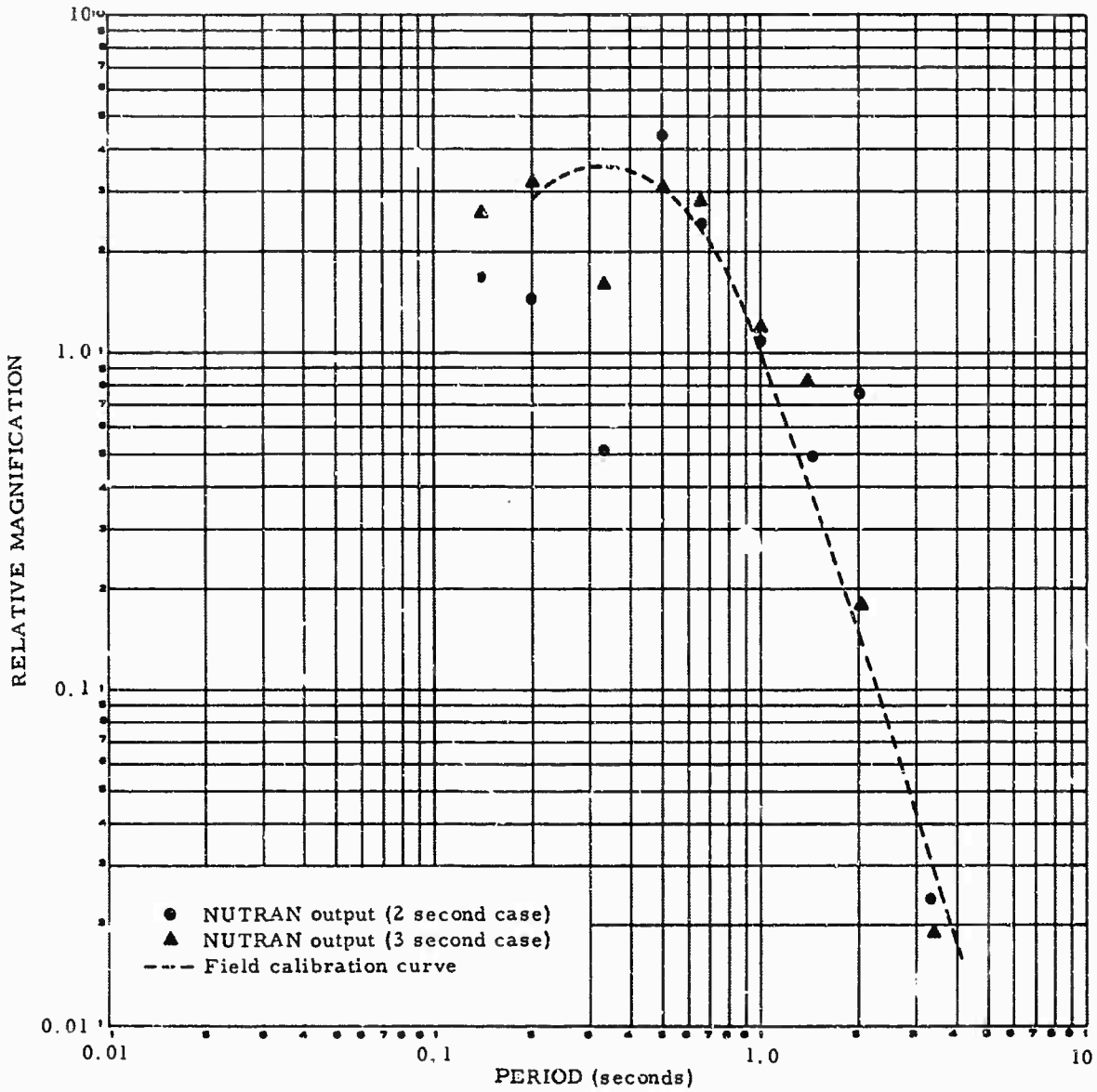


Figure 33. Unsmoothed calibration estimates at JR-AZ Z₂,
2 and 3 second cases

G 372

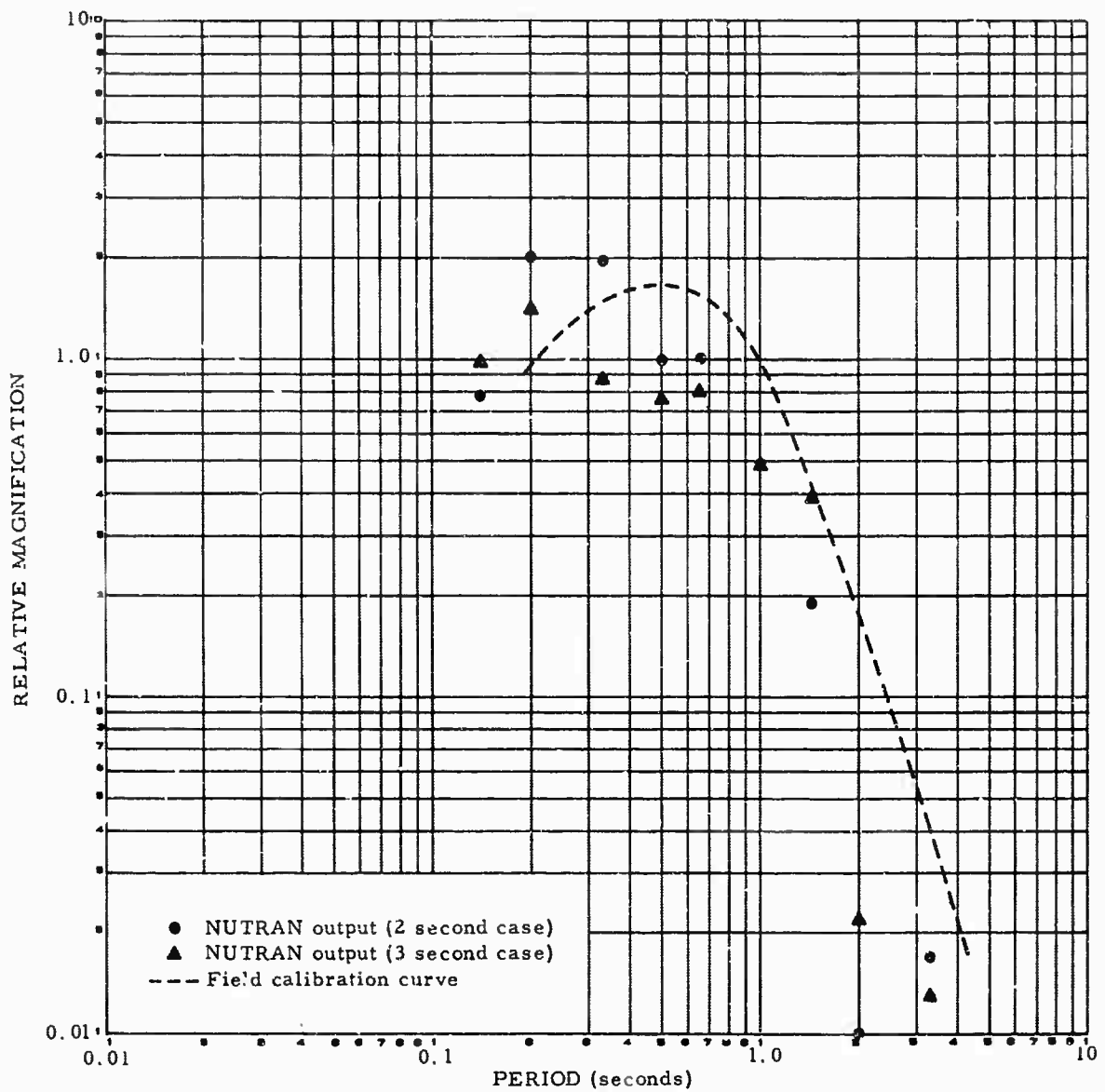


Figure 34. Unsmoothed calibration estimates at JR-AZ Z₃,
2 and 3 second cases

G 373

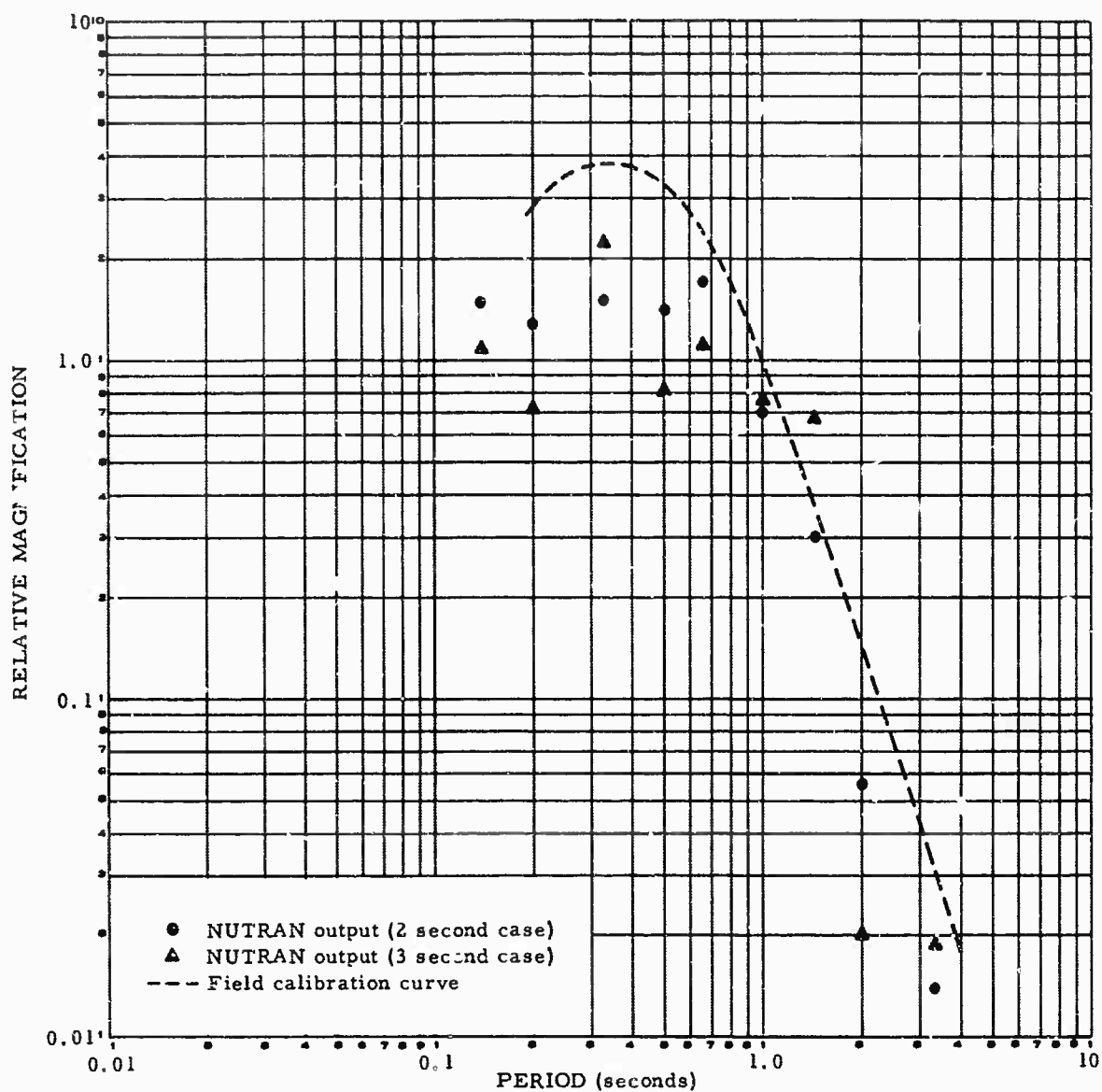


Figure 35. Unsmoothed calibration estimates at JR-AZ Z₄,
2 and 3 second cases

G 374

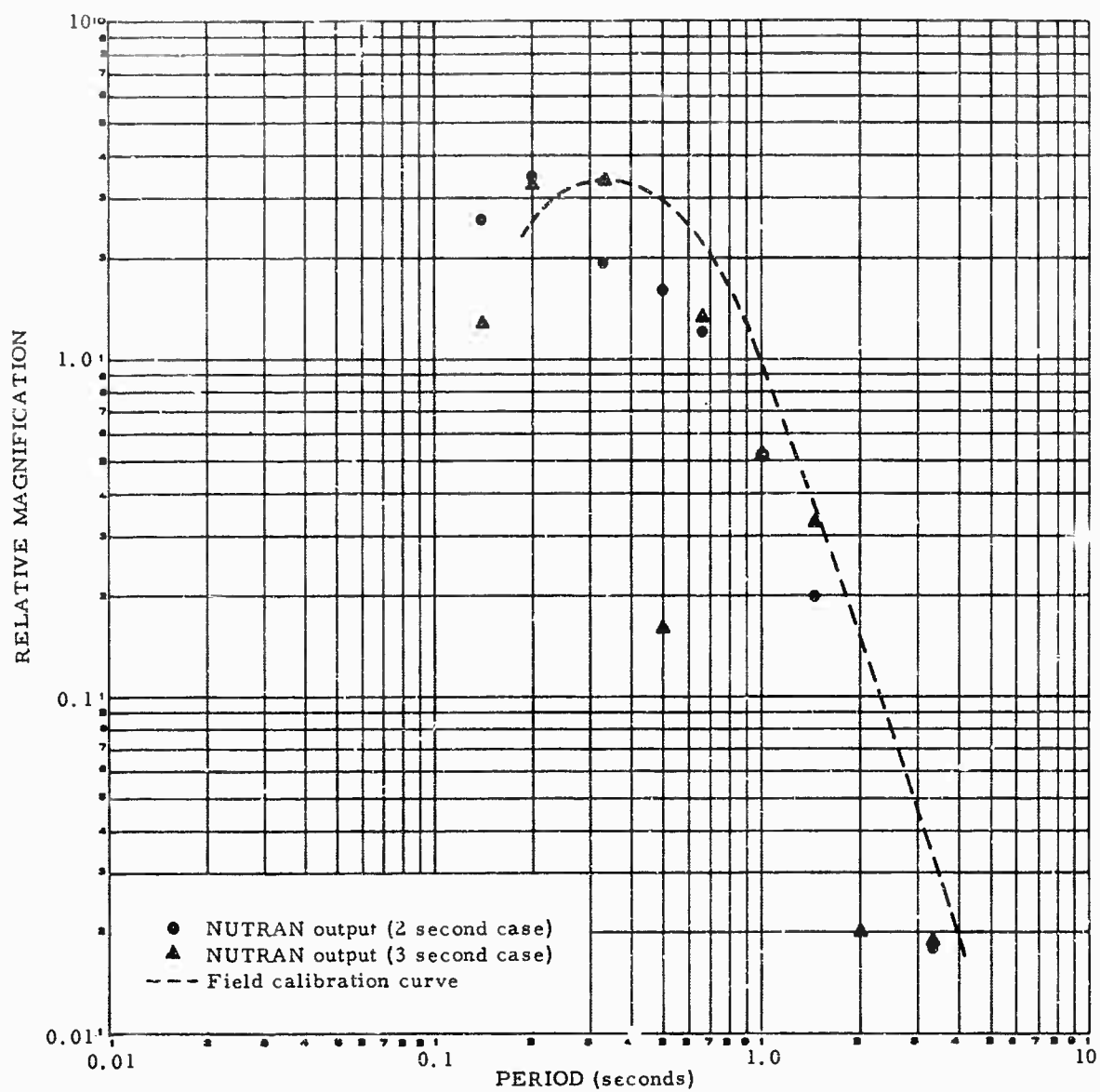


Figure 36. Unsmoothed calibration estimates at JR-AZ Z₅,
2 and 3 second cases

G 375

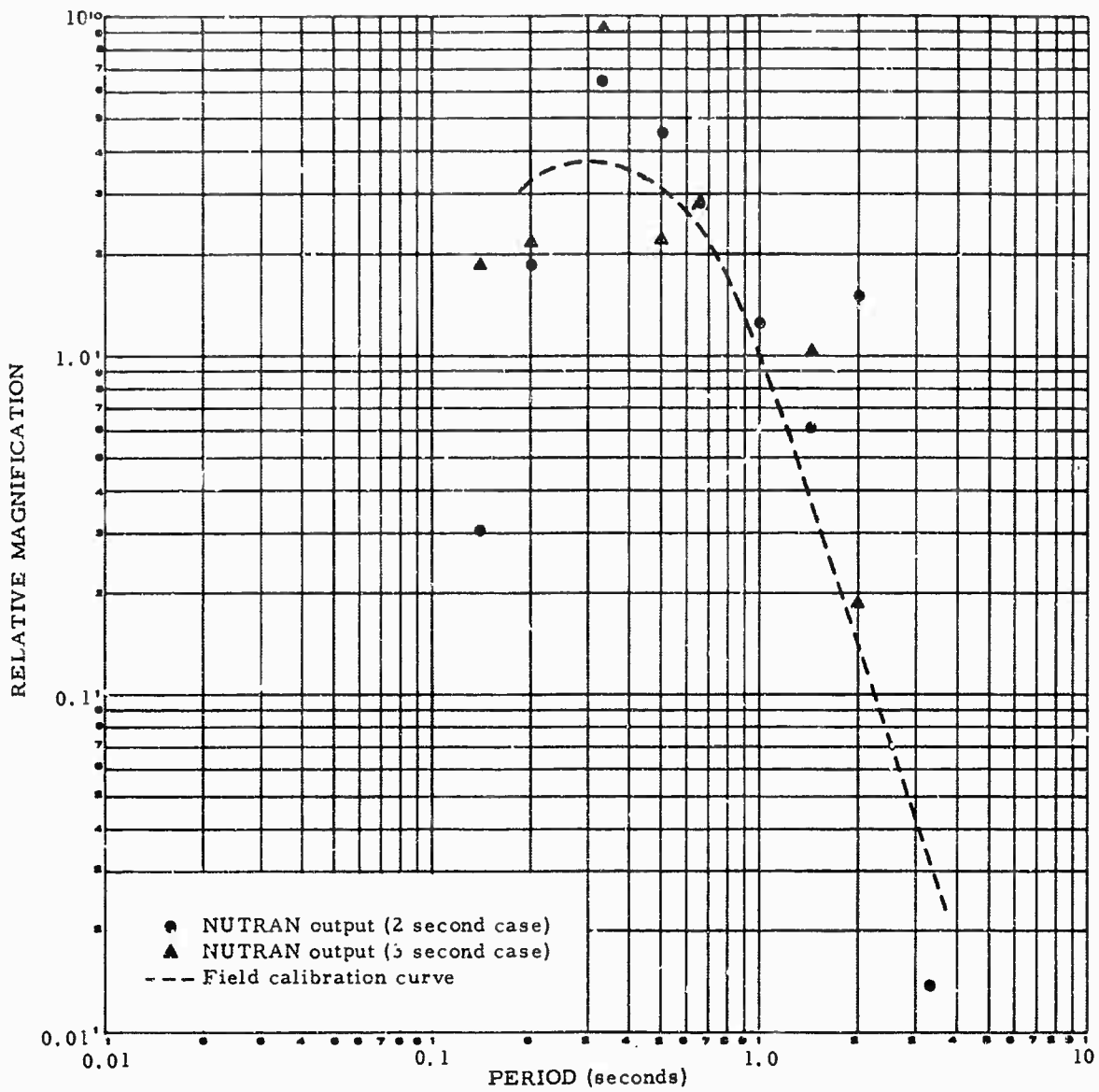


Figure 37. Unsmoothed calibration estimates at JR-AZ Z₆,
2 and 3 second cases

G 376

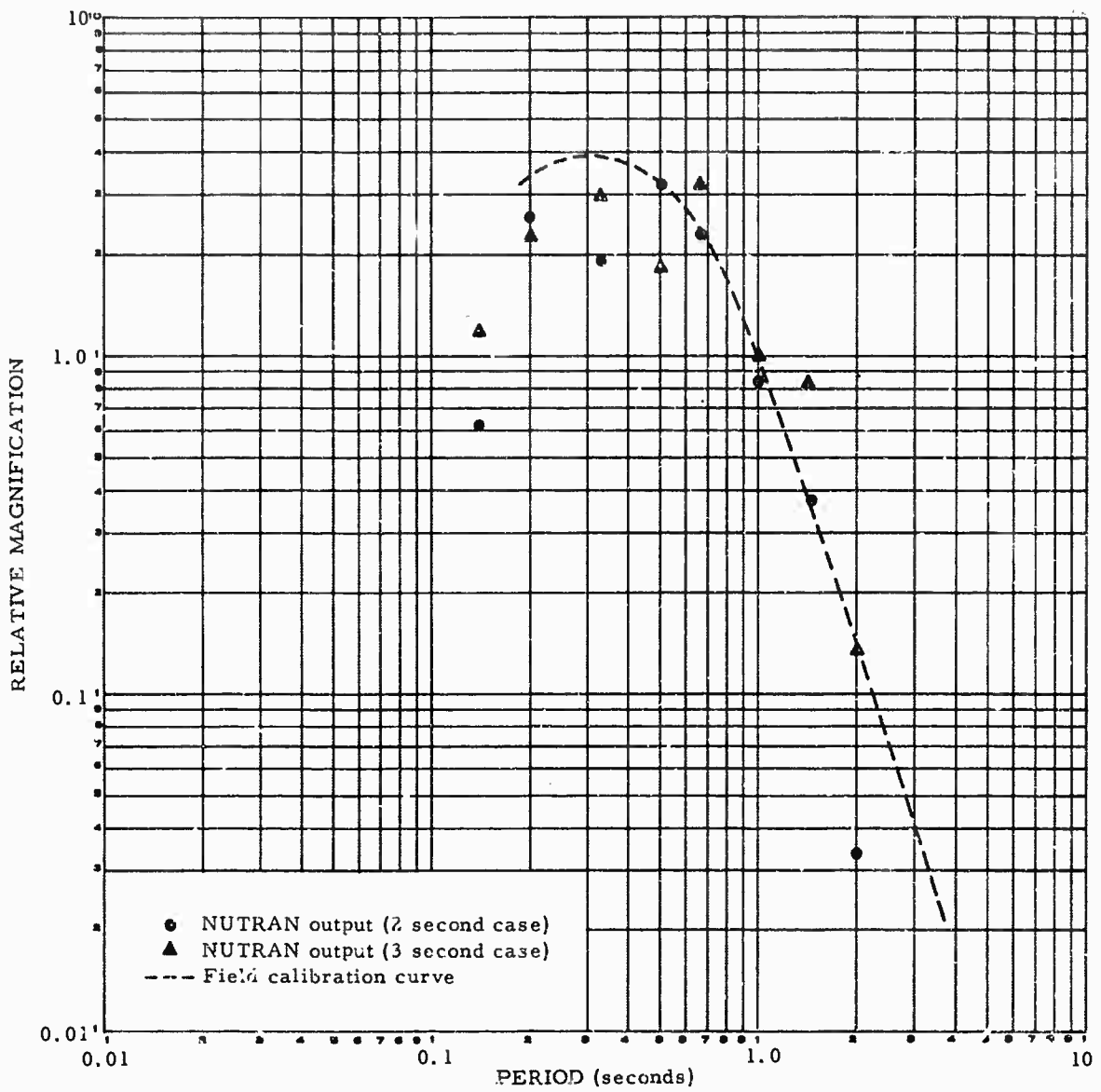


Figure 38. Unsmoothed calibration estimates at JR-AZ Z7, 2 and 3 second cases

G 377

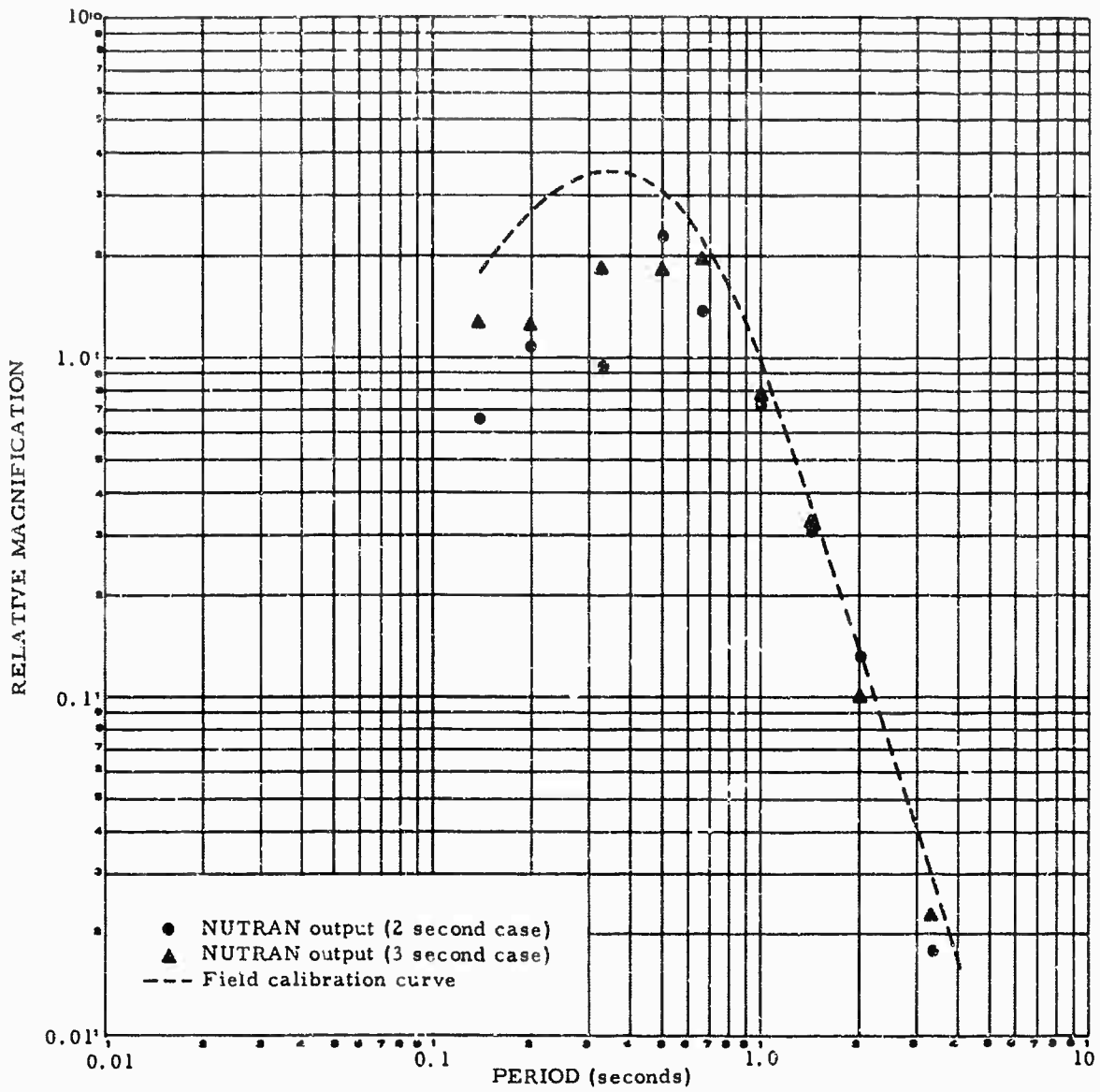


Figure 39. Unsmoothed calibration estimates at WO-AZ Z₂,
2 and 3 second cases

G 378

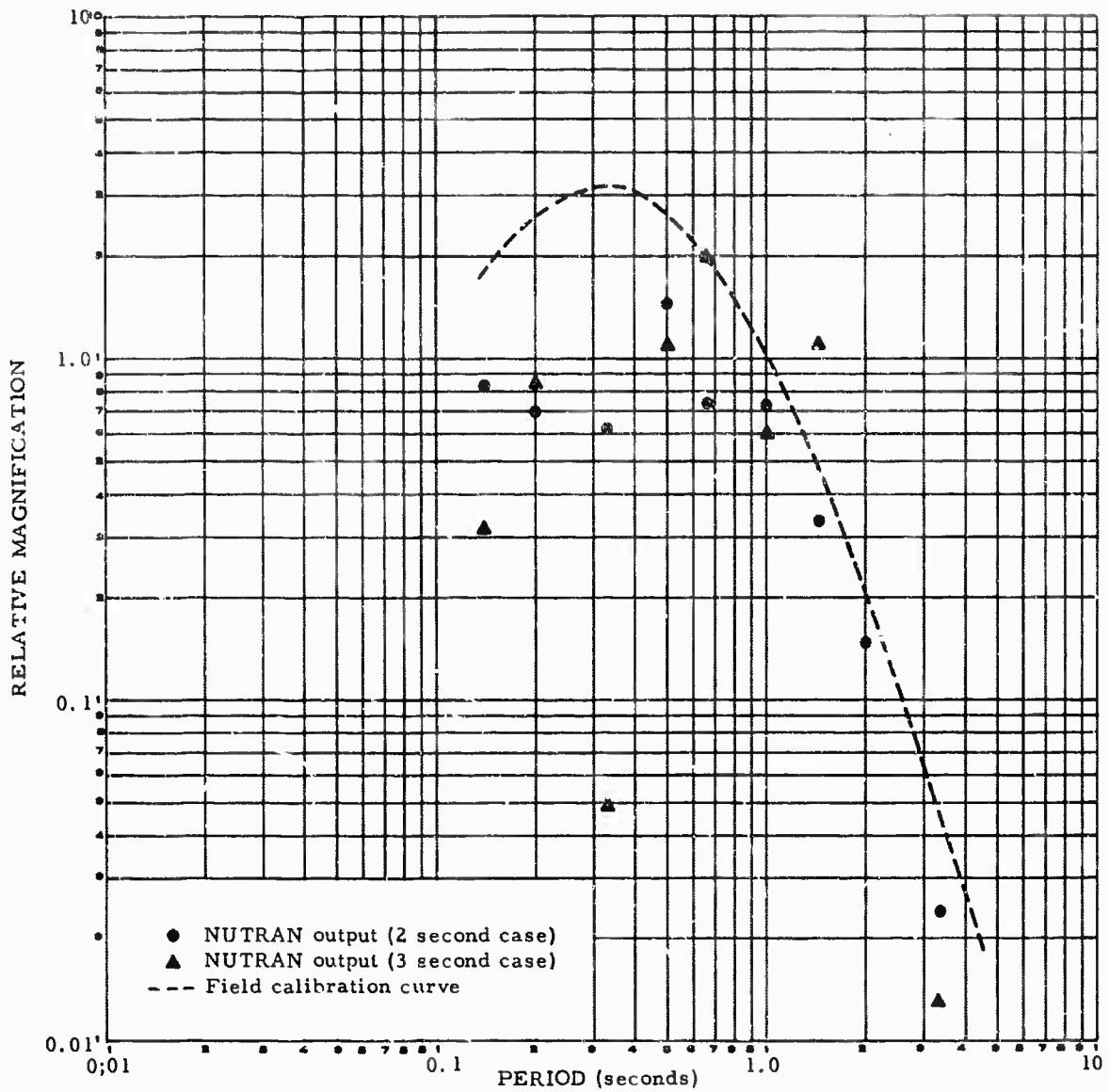


Figure 40. Unsmoothed calibration estimates at WO-AZ Z₃,
2 and 3 second cases

G 379

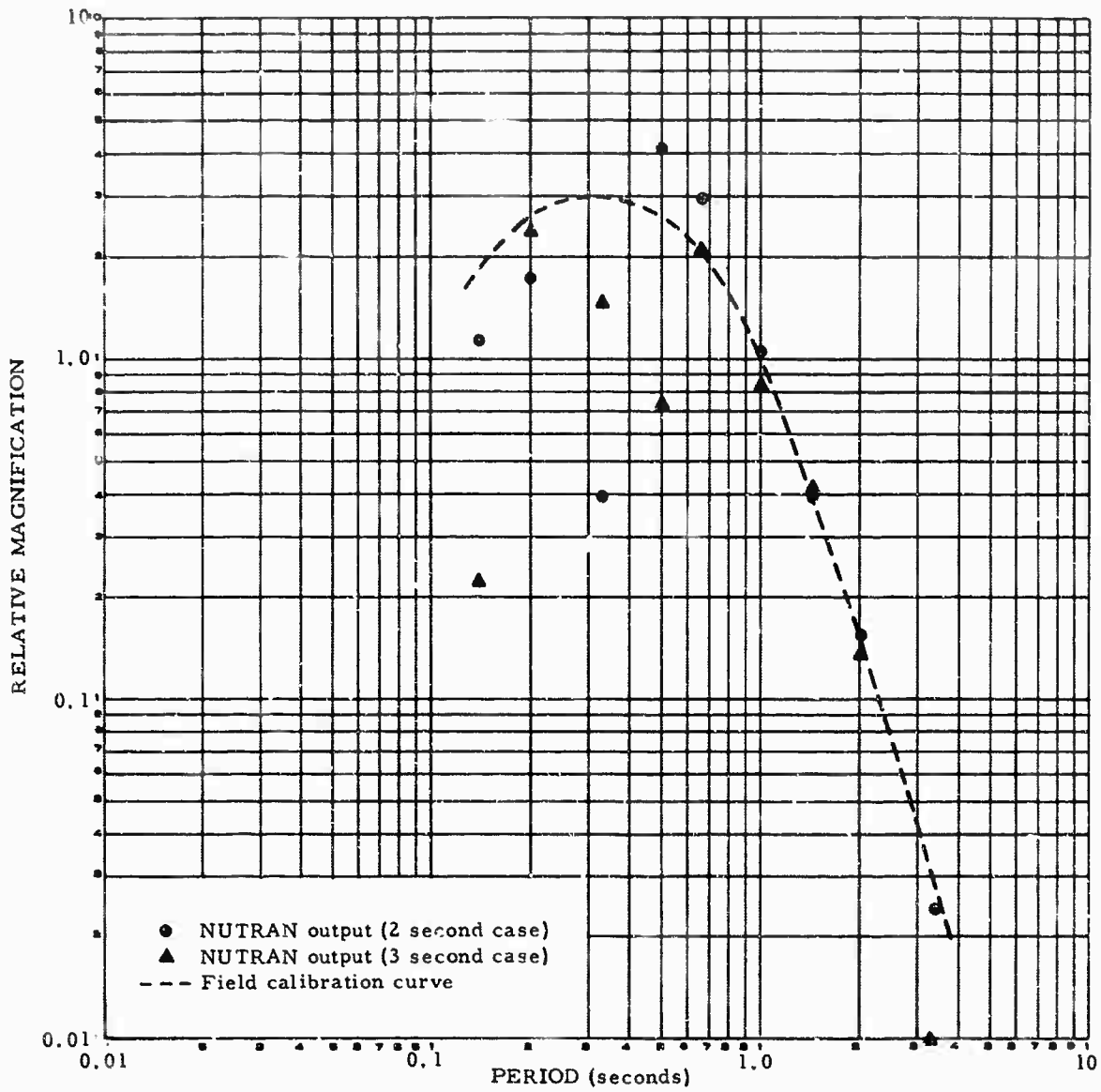


Figure 41. Unsmoothed calibration estimates at WO-AZ Z₄,
2 and 3 second cases

G 380

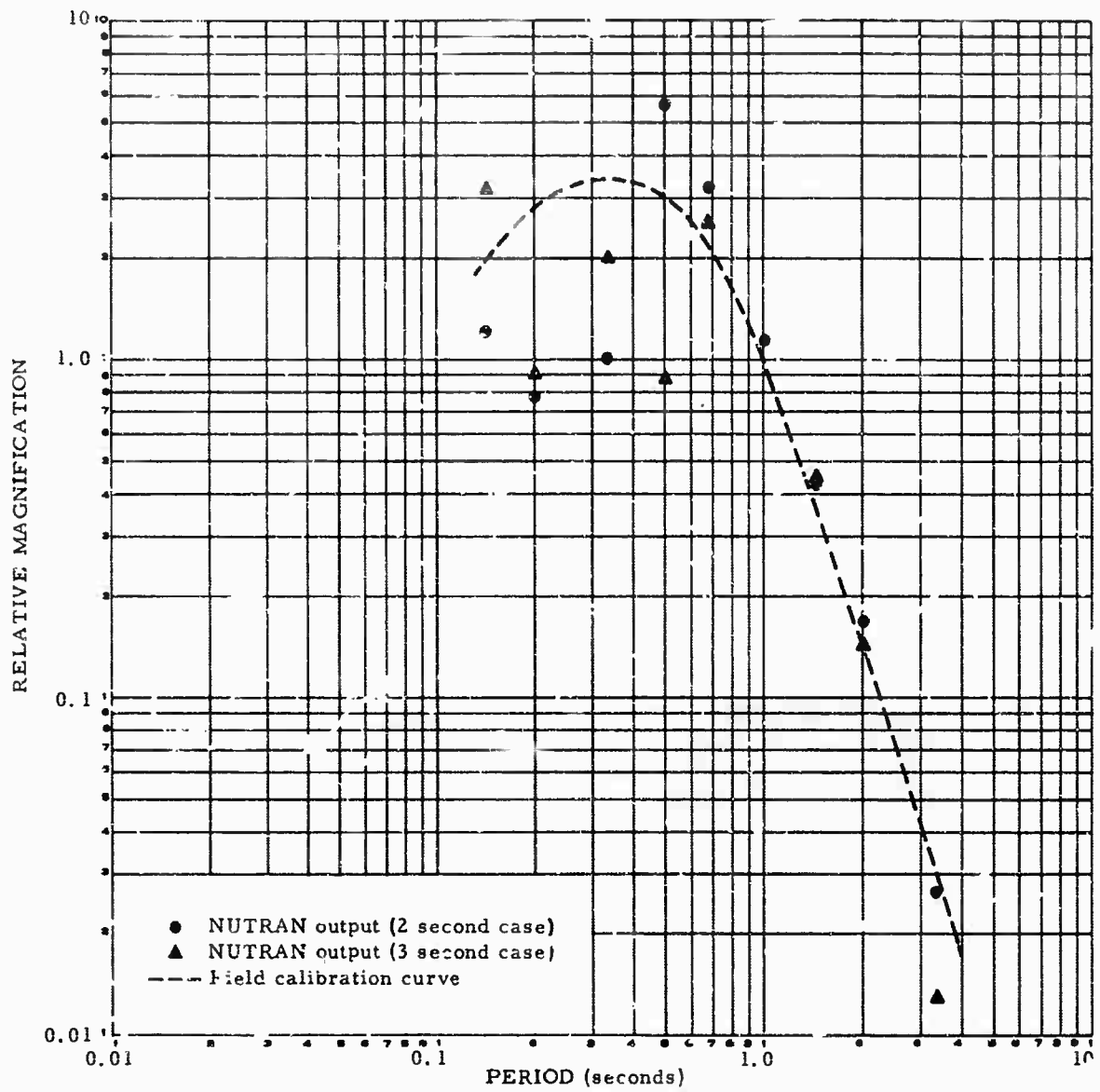


Figure 42. Unsmoothed calibration estimates at WO-AZ Z₅, 2 and 3 second cases

G 381

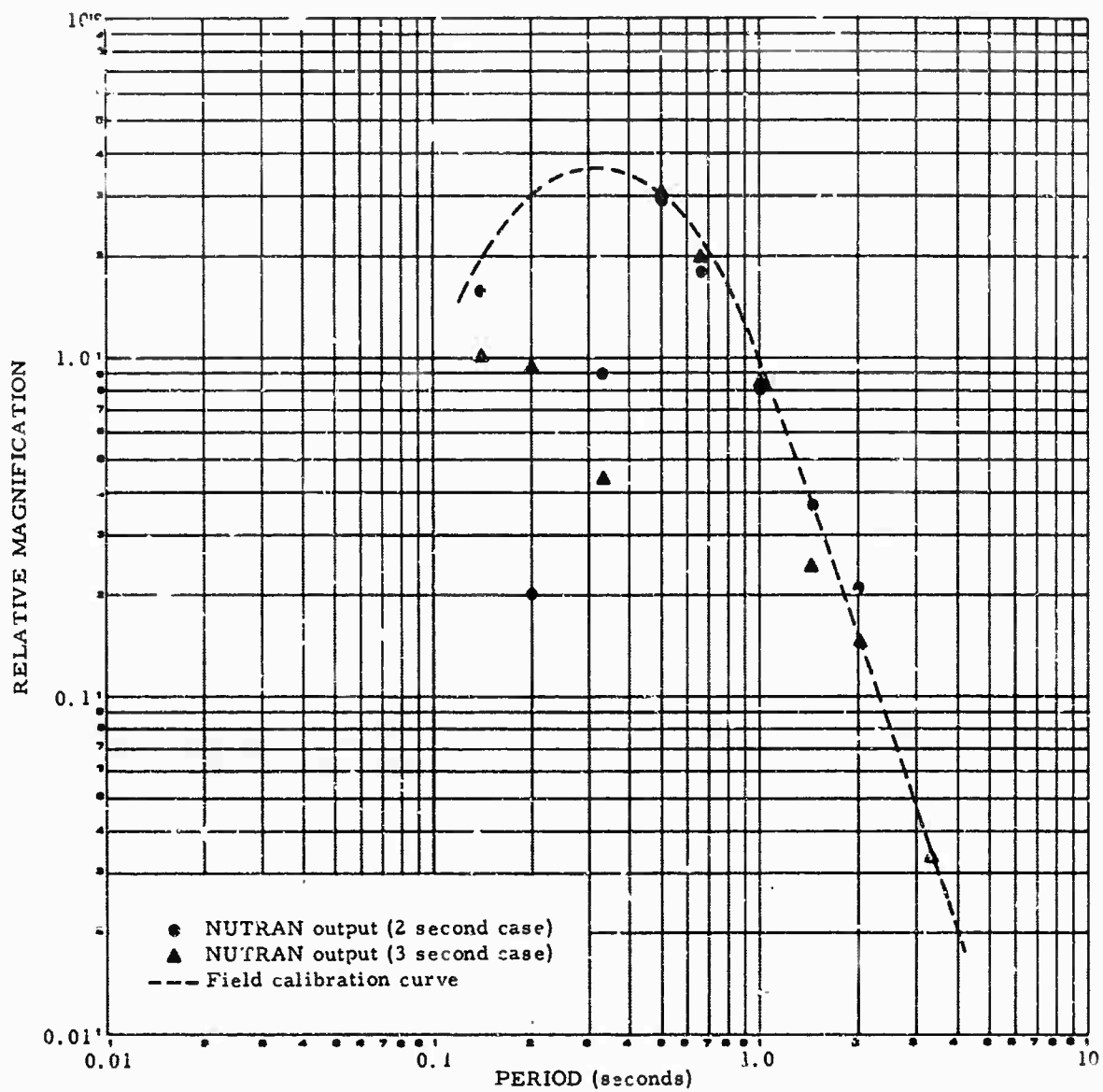


Figure 43. Unsmoothed calibration estimates at WO-AZ Z6, 2 and 3 second cases

G 382

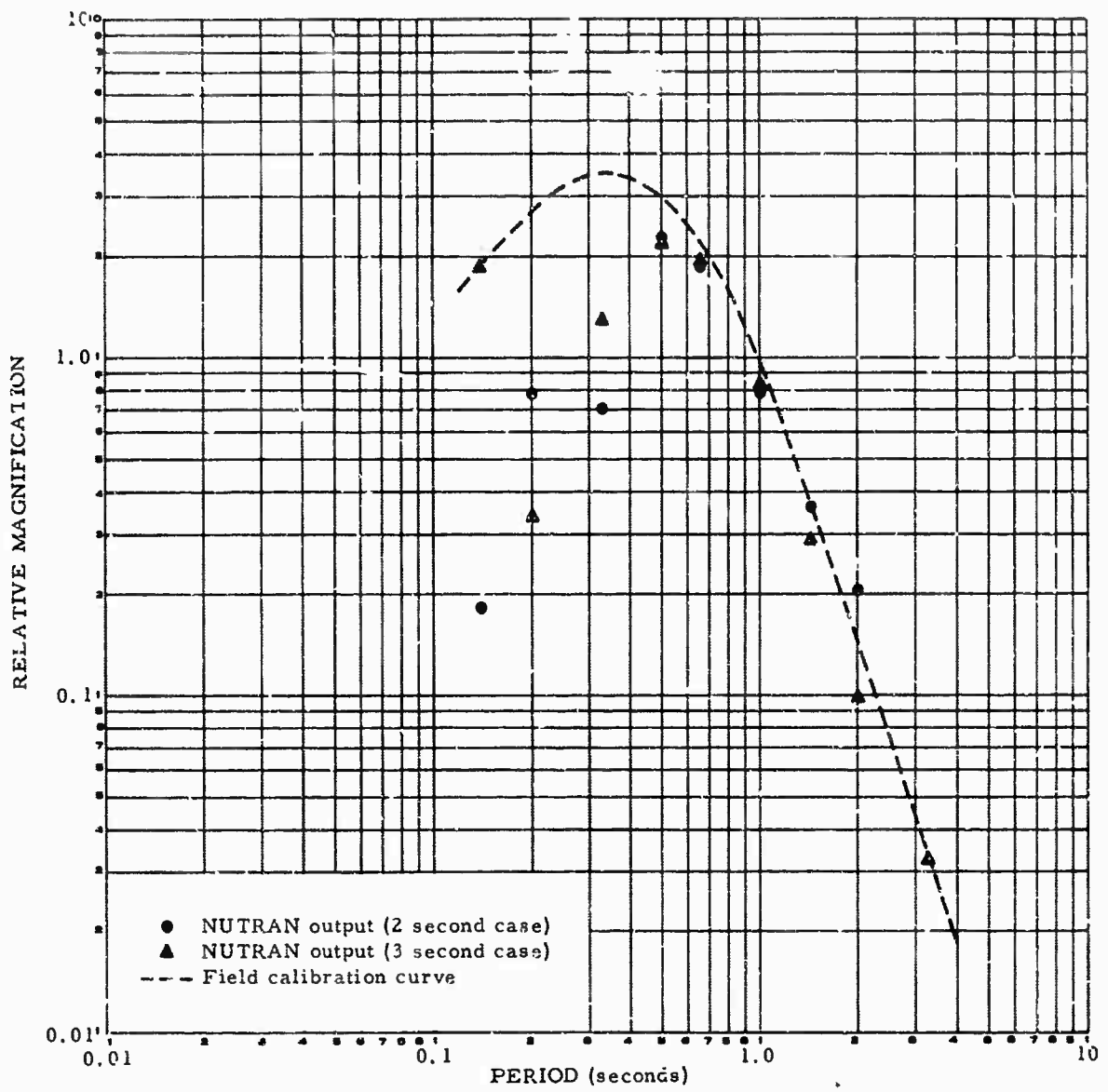


Figure 44. Unsmoothed calibration estimates at WO-AZ Z7, 2 and 3 second cases

G 383

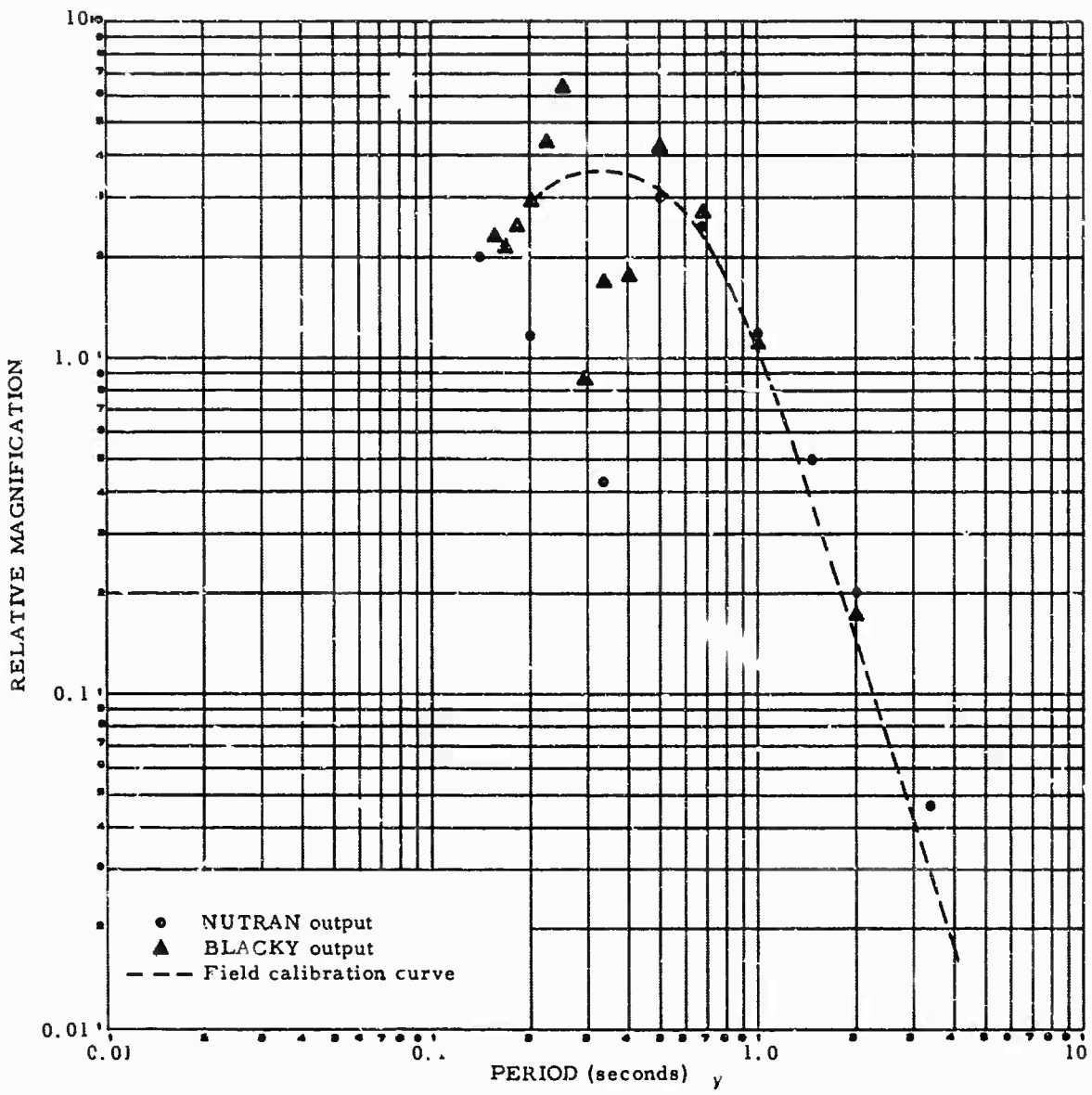


Figure 45. Smoothed calibration estimates at JR-AZ Z₂,
2 second case

G 384

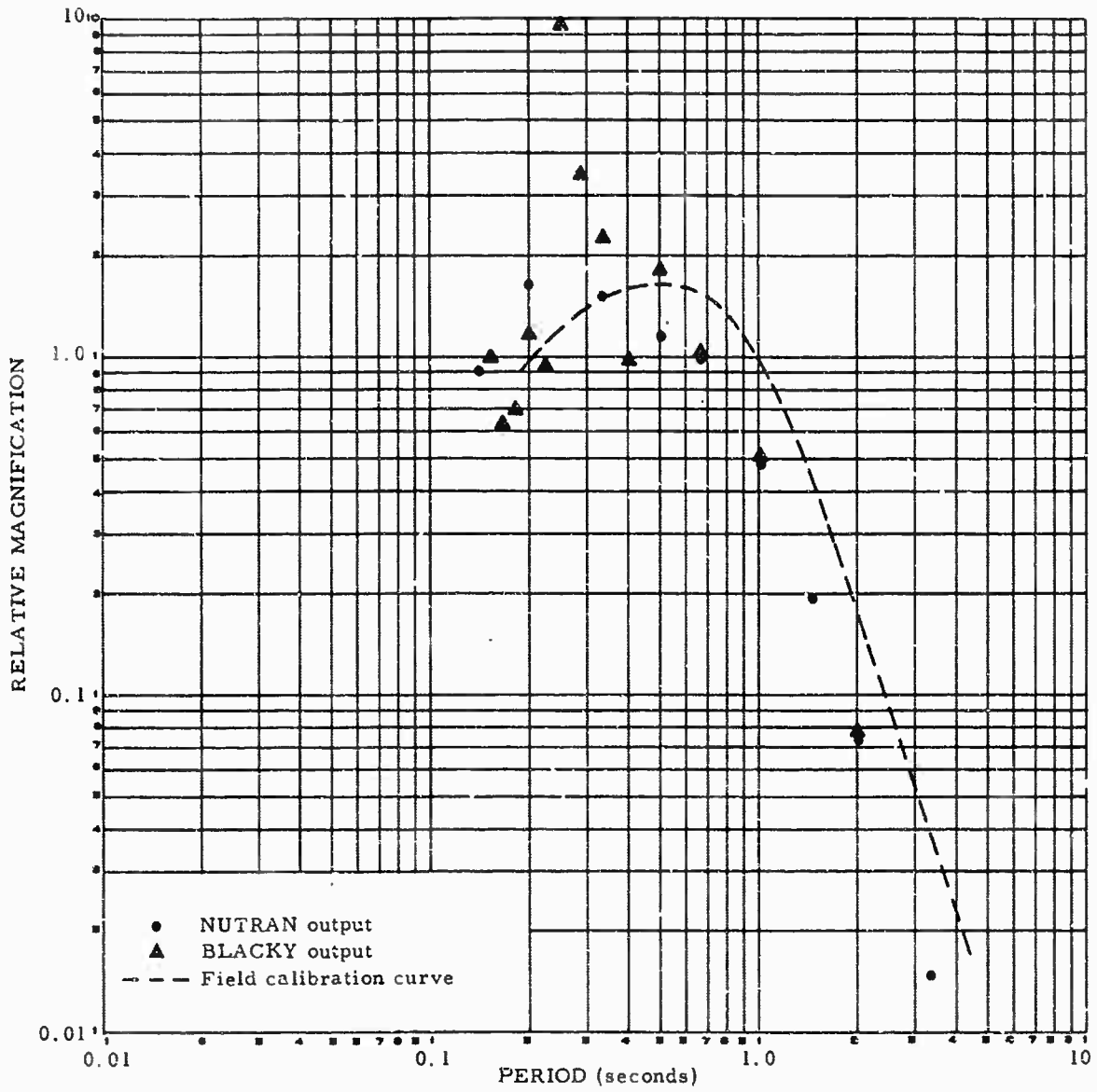


Figure 46. Smoothed calibration estimates at JR-AZ Z₃,
2 second case

G 385

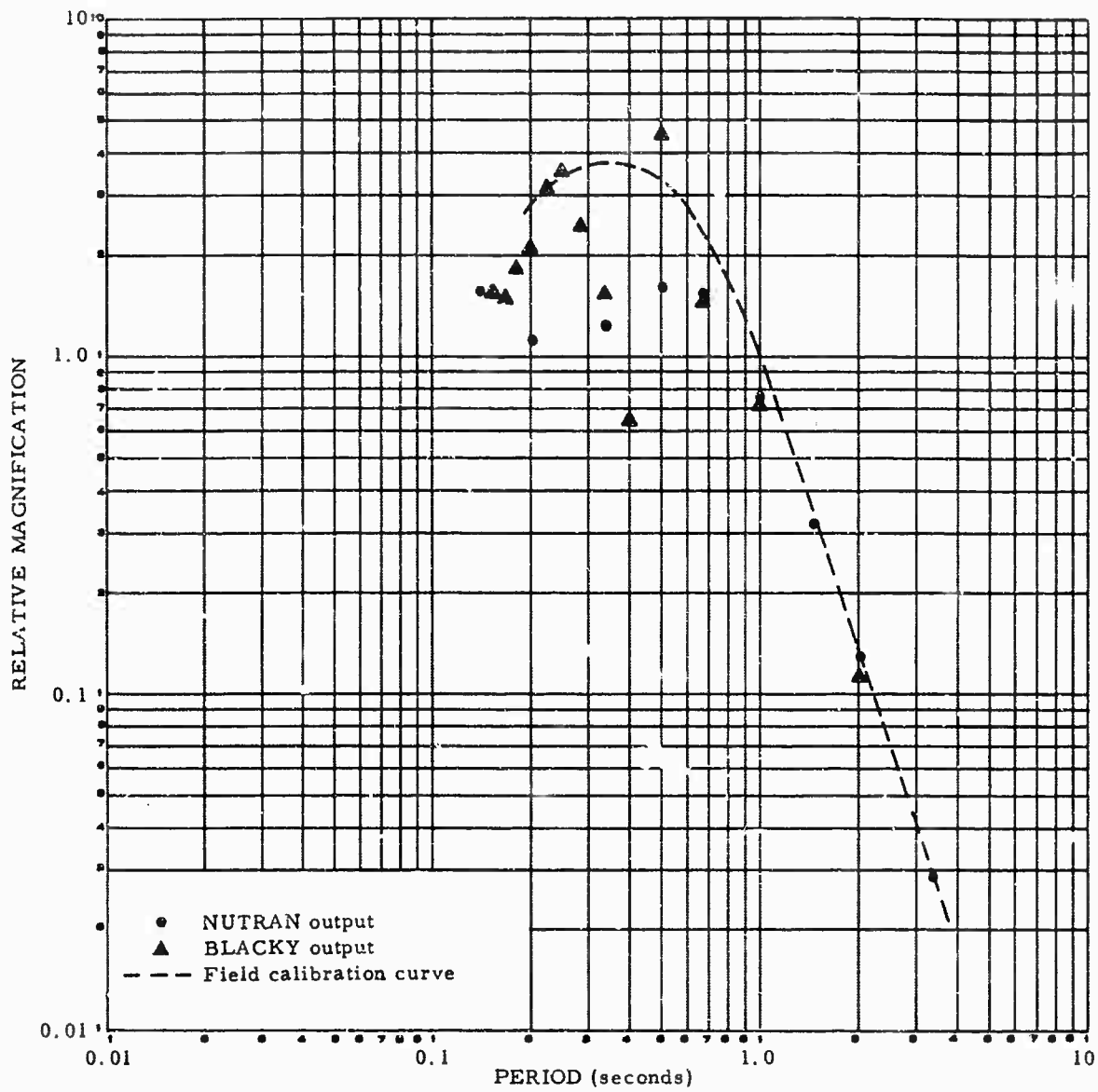


Figure 47. Smoothed calibration estimates at JR-AZ Z₄,
2 second case

G 386

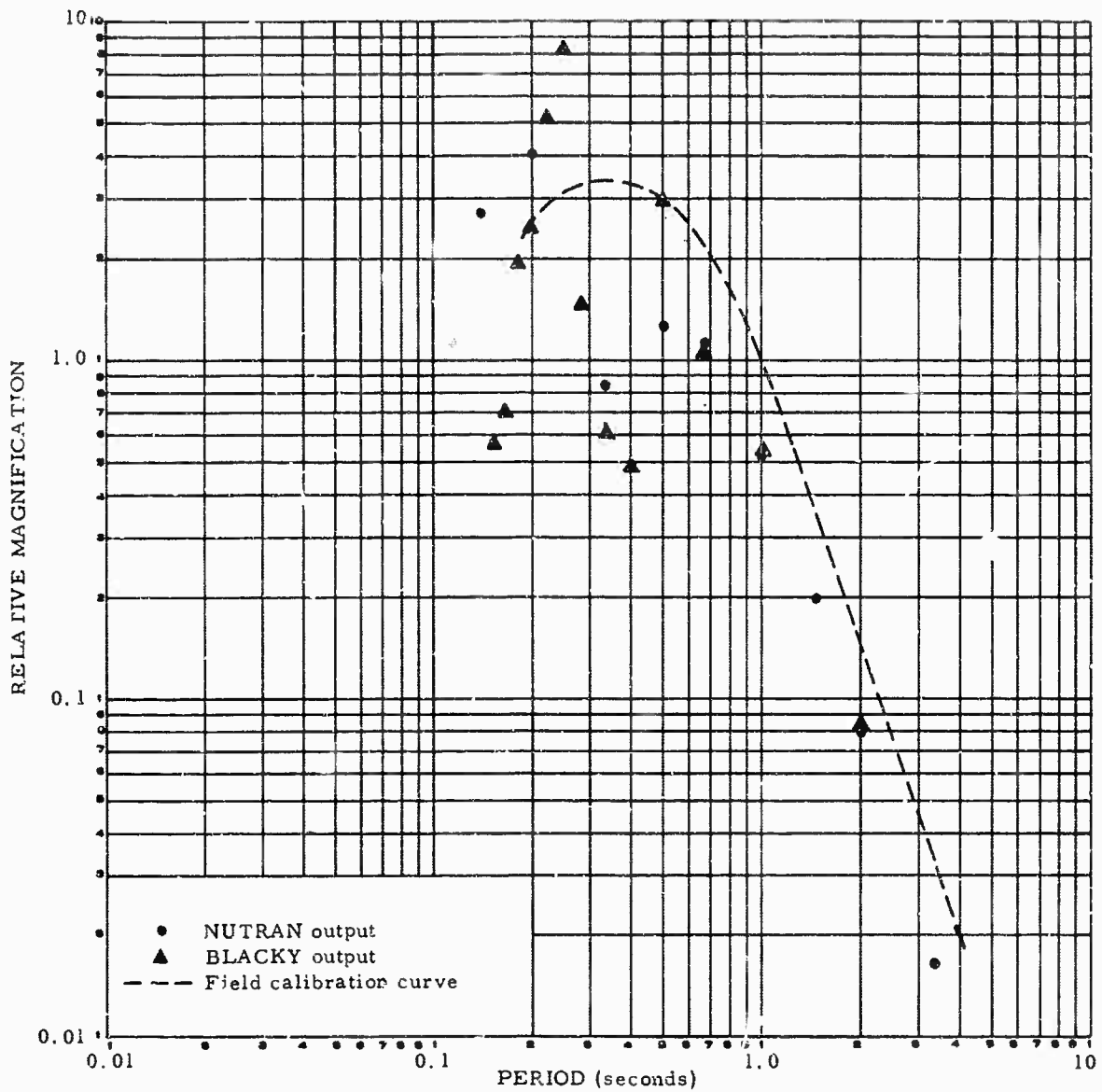


Figure 48. Smoothed calibration estimates at JR-AZ Z₅,
2 second case

G 387

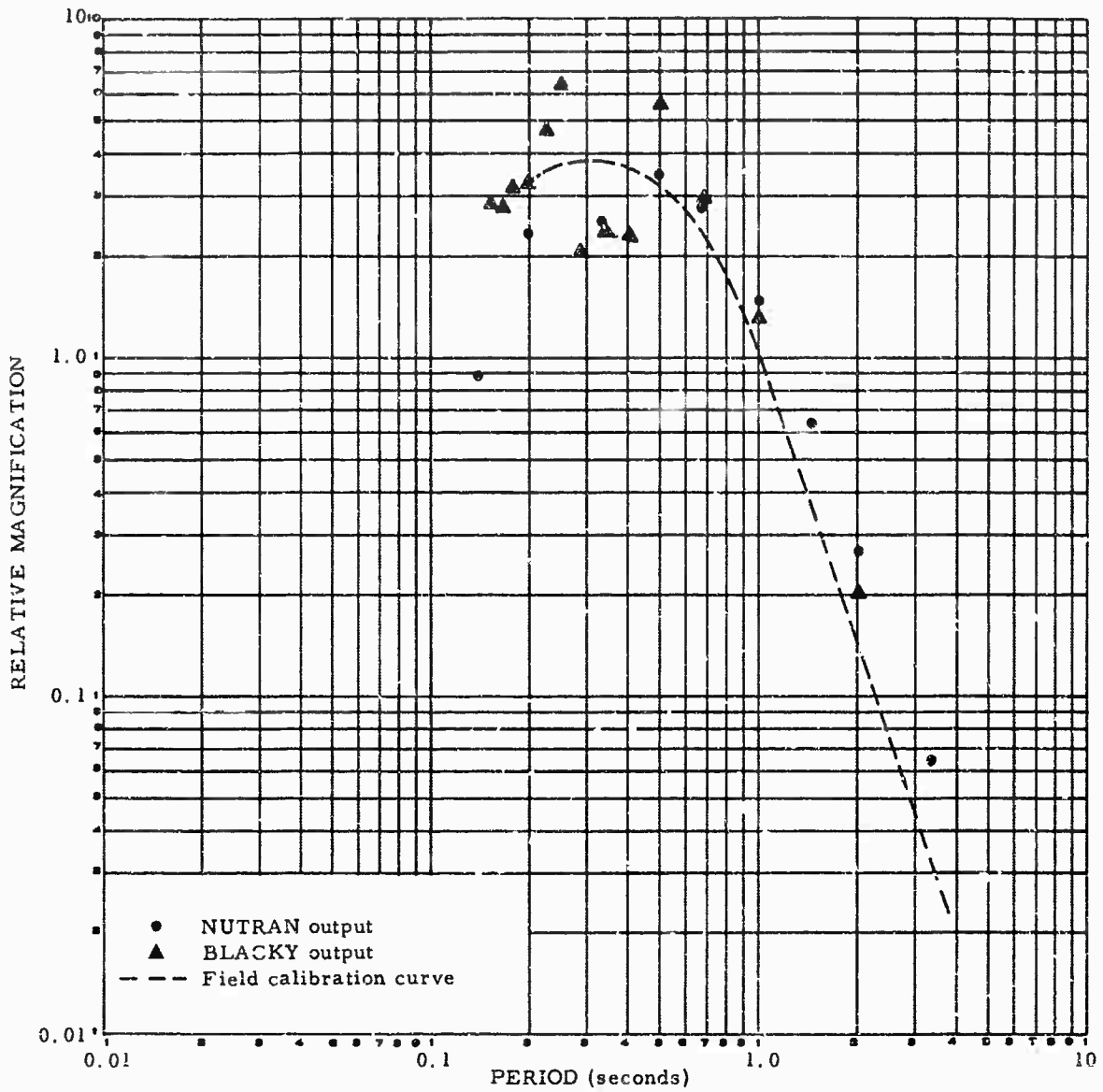


Figure 49. Smoothed calibration estimates at JR-AZ Z₆,
 2 second case

G 388

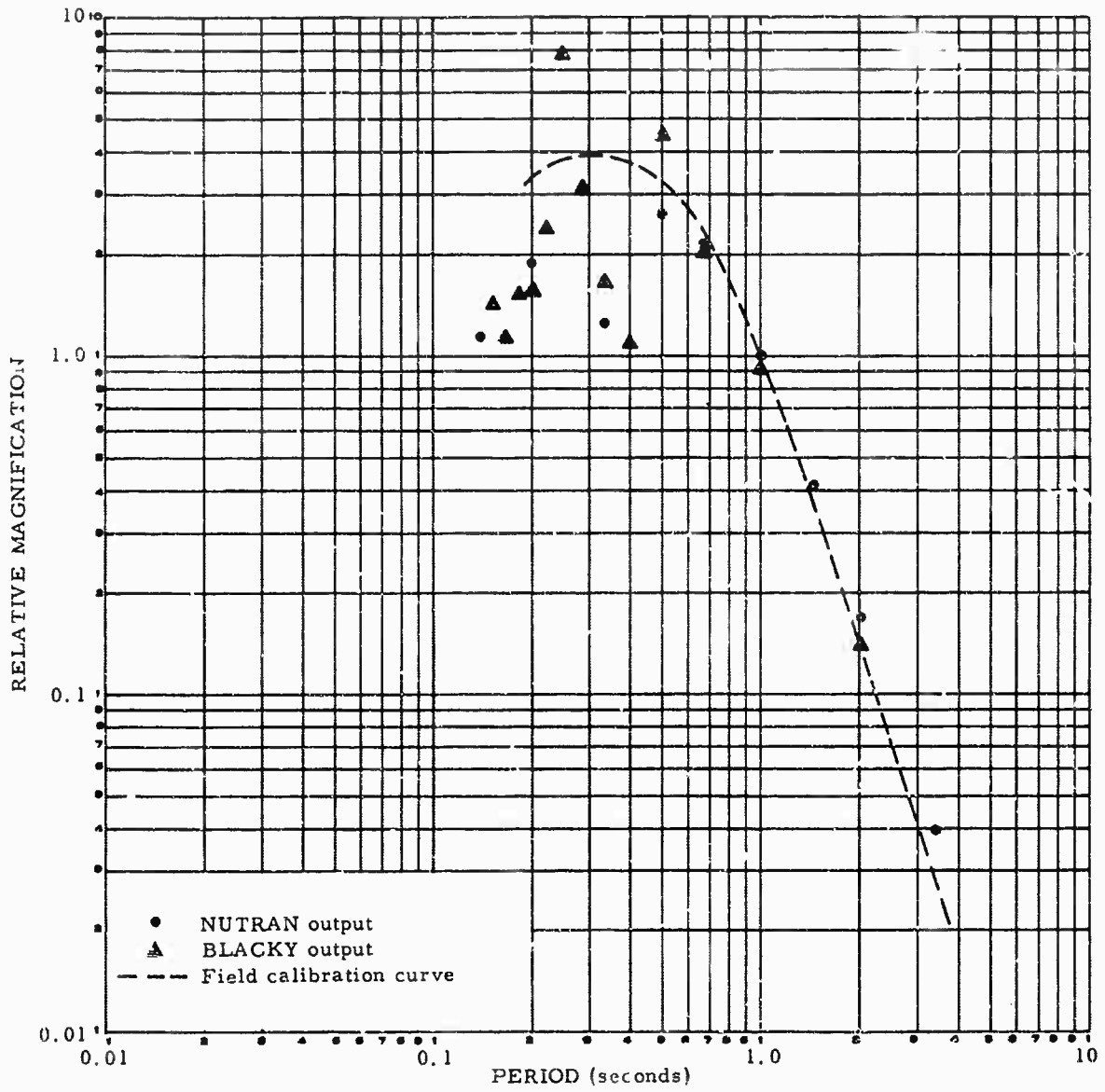


Figure 50. Smoothed calibration estimates at JR-AZ Z7,
2 second case

G 389

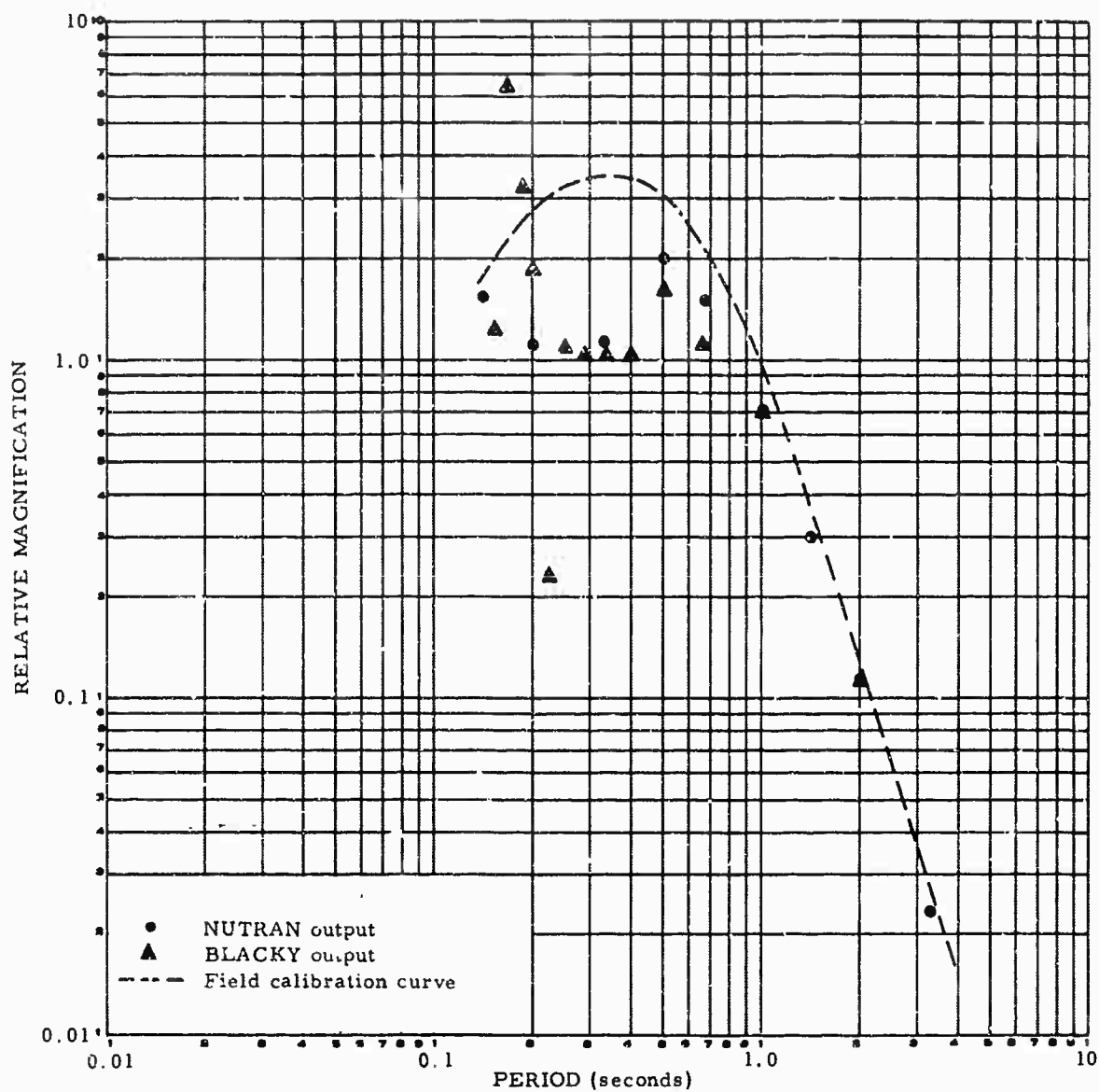


Figure 51. Smoothed calibration estimates at WO-AZ Z₂,
2 second case

G 390

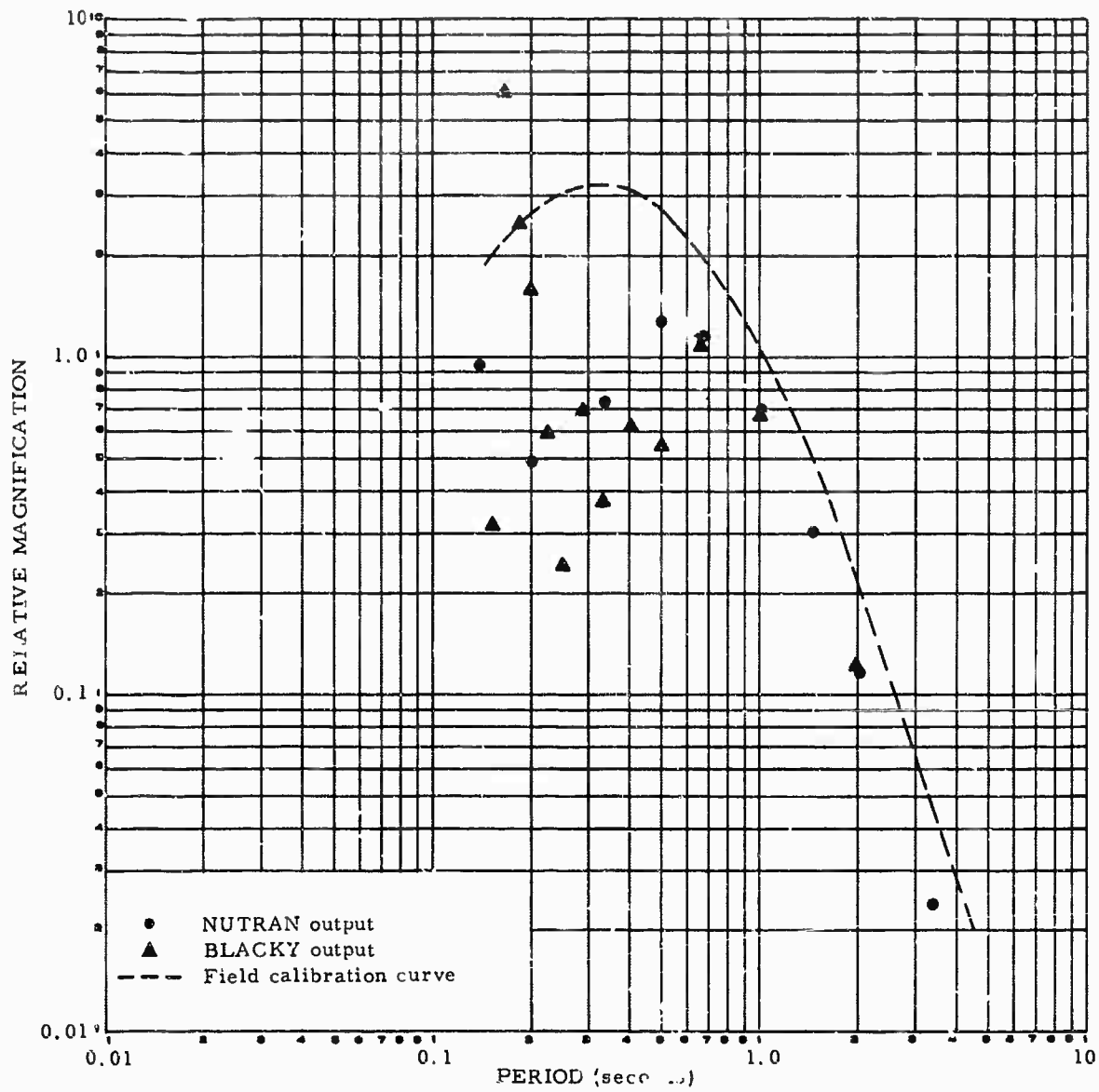


Figure 52. Smoothed calibration estimates at WO-AZ Z₃,
2 second case

G 391

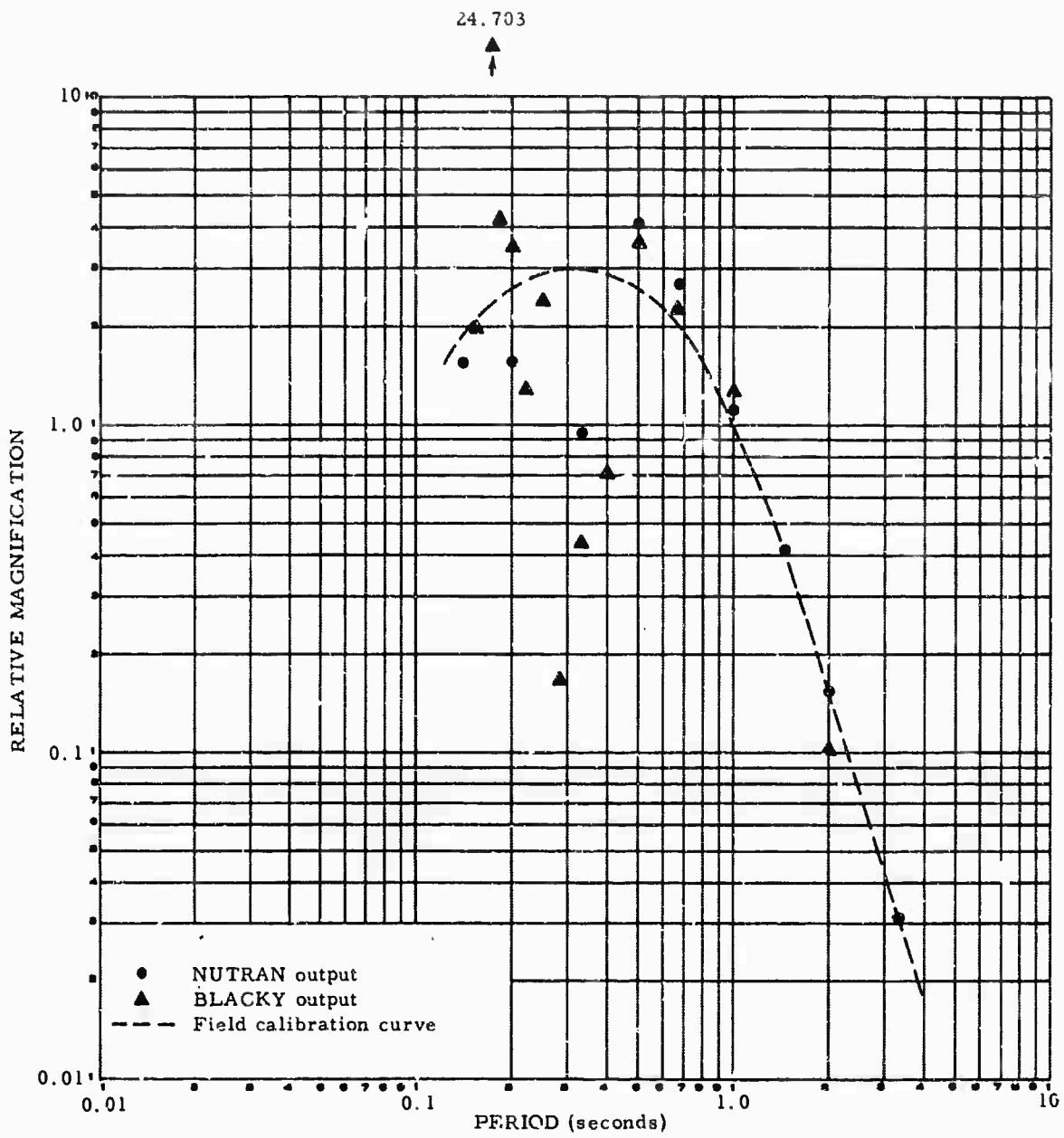


Figure 53. Smoothed calibration estimates at WO-AZ Z₄,
2 second case

G 392

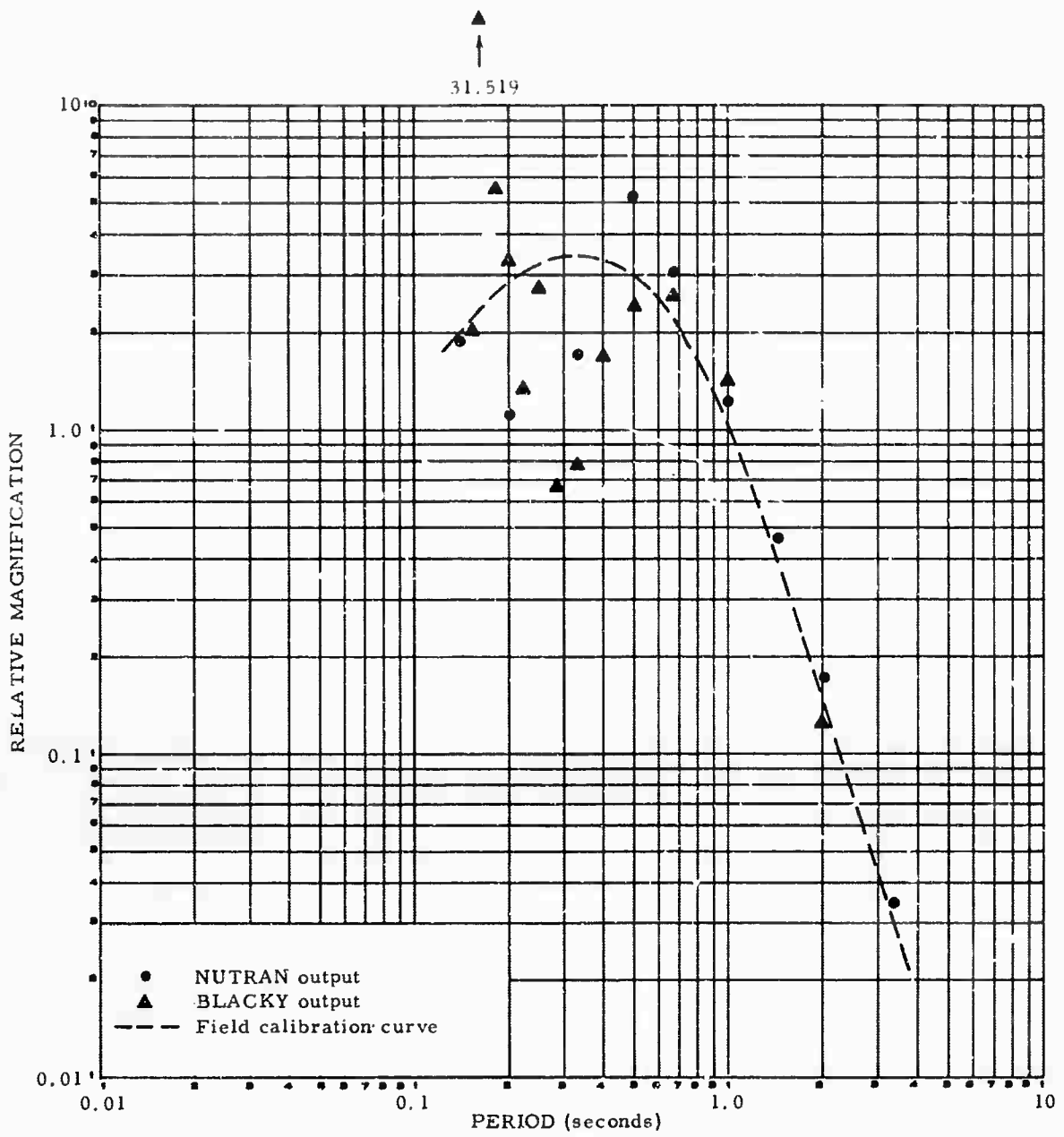


Figure 54. Smoothed calibration estimates at WC-AZ Z₅,
2 second case

G 393

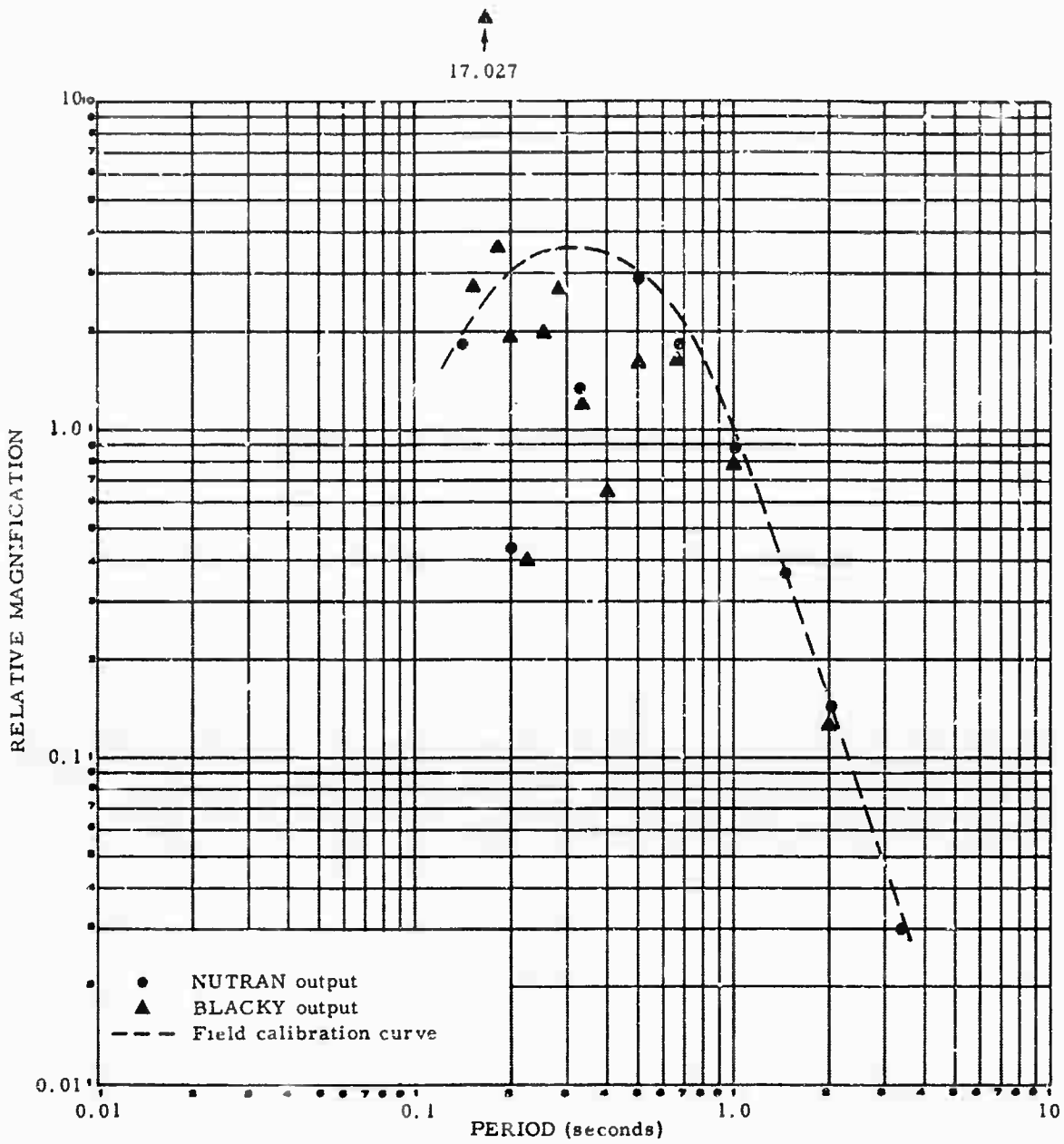


Figure 55. Smoothed calibration estimates at WO-AZ Z₆,
2 second case

G 394

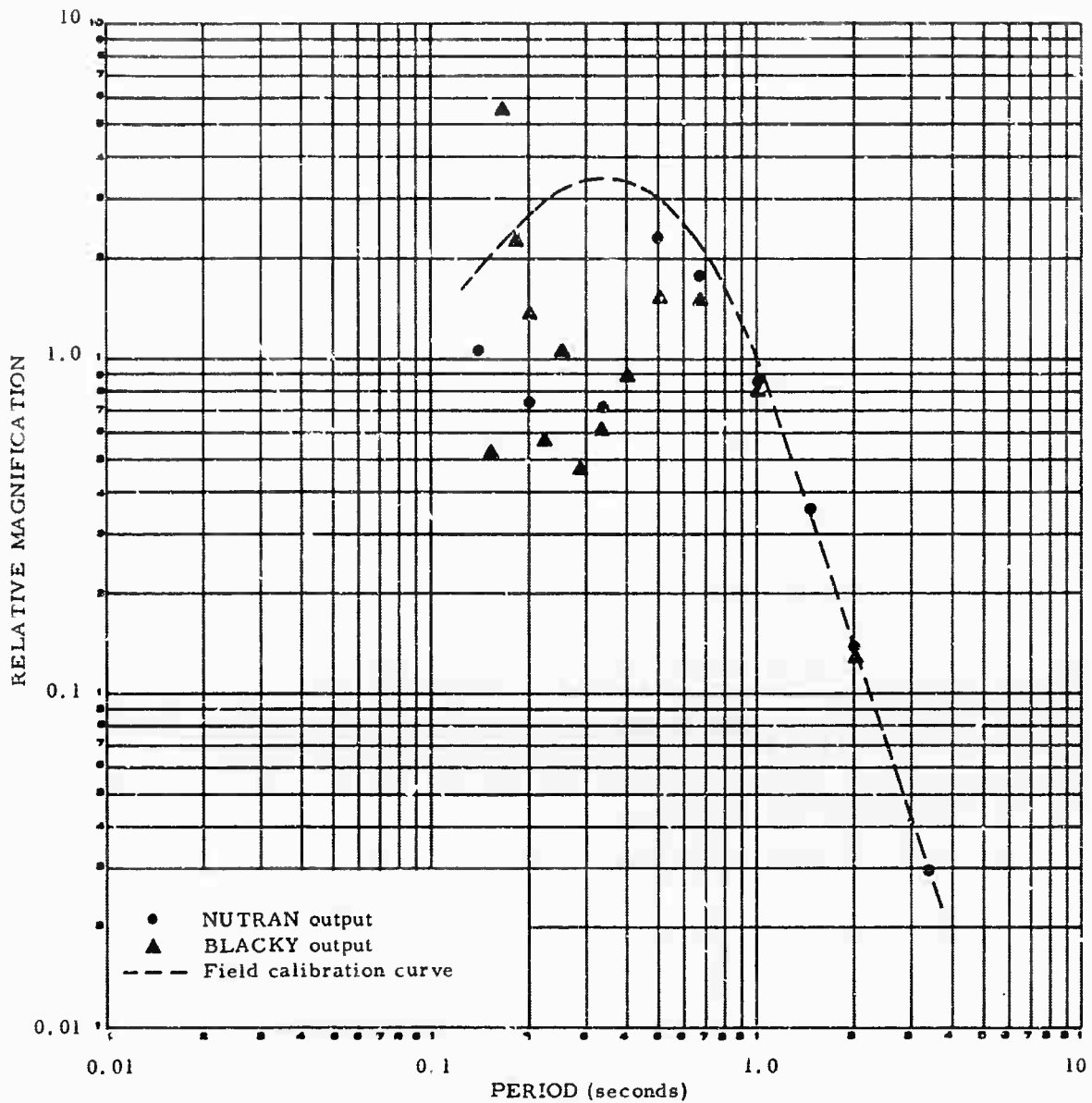


Figure 56. Smoothed calibration estimates at WO-AZ Z₇,
2 second case

G 395

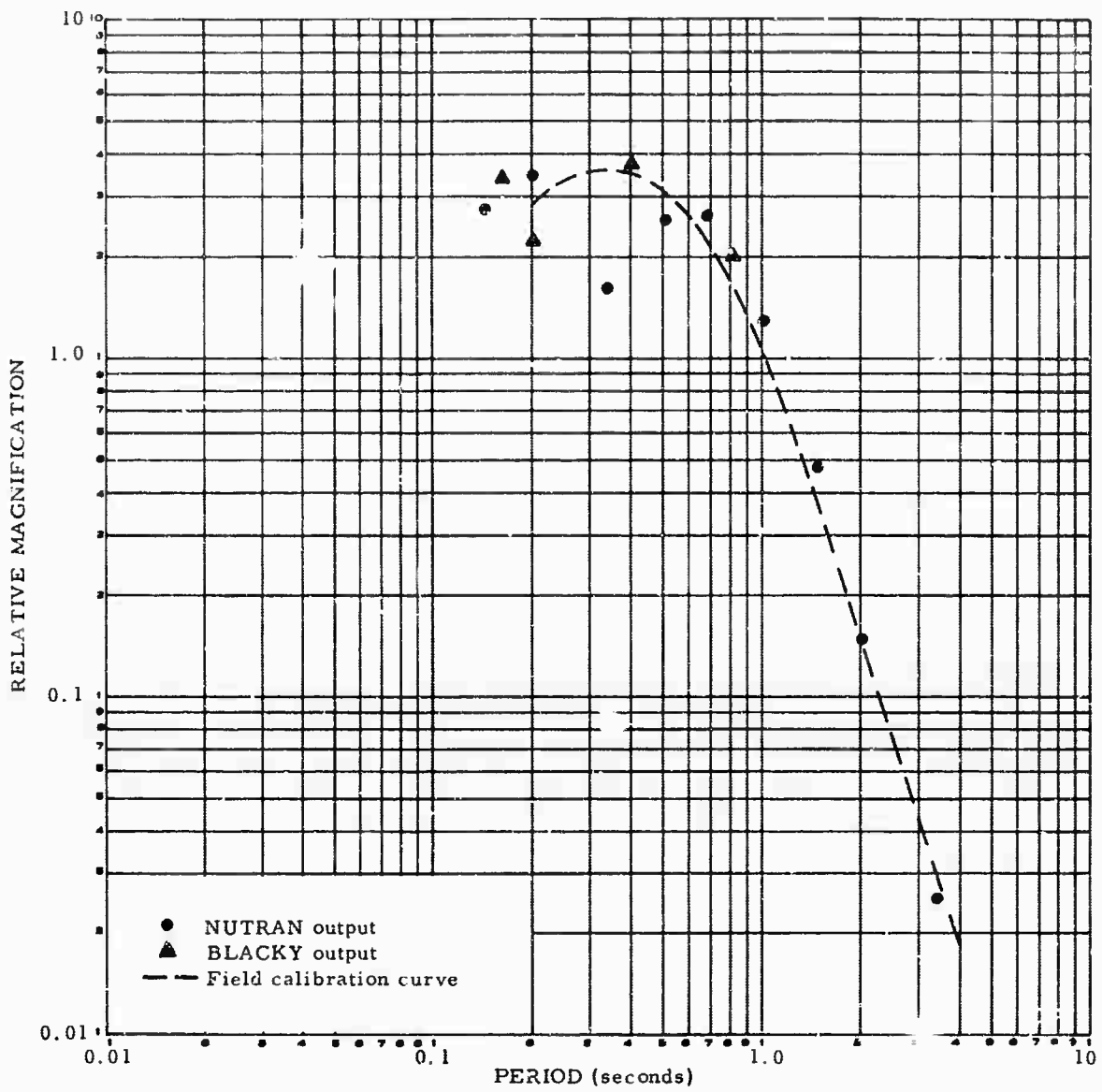


Figure 57. Smoothed calibration estimates at JR-AZ Z₂,
3 second case

G 396

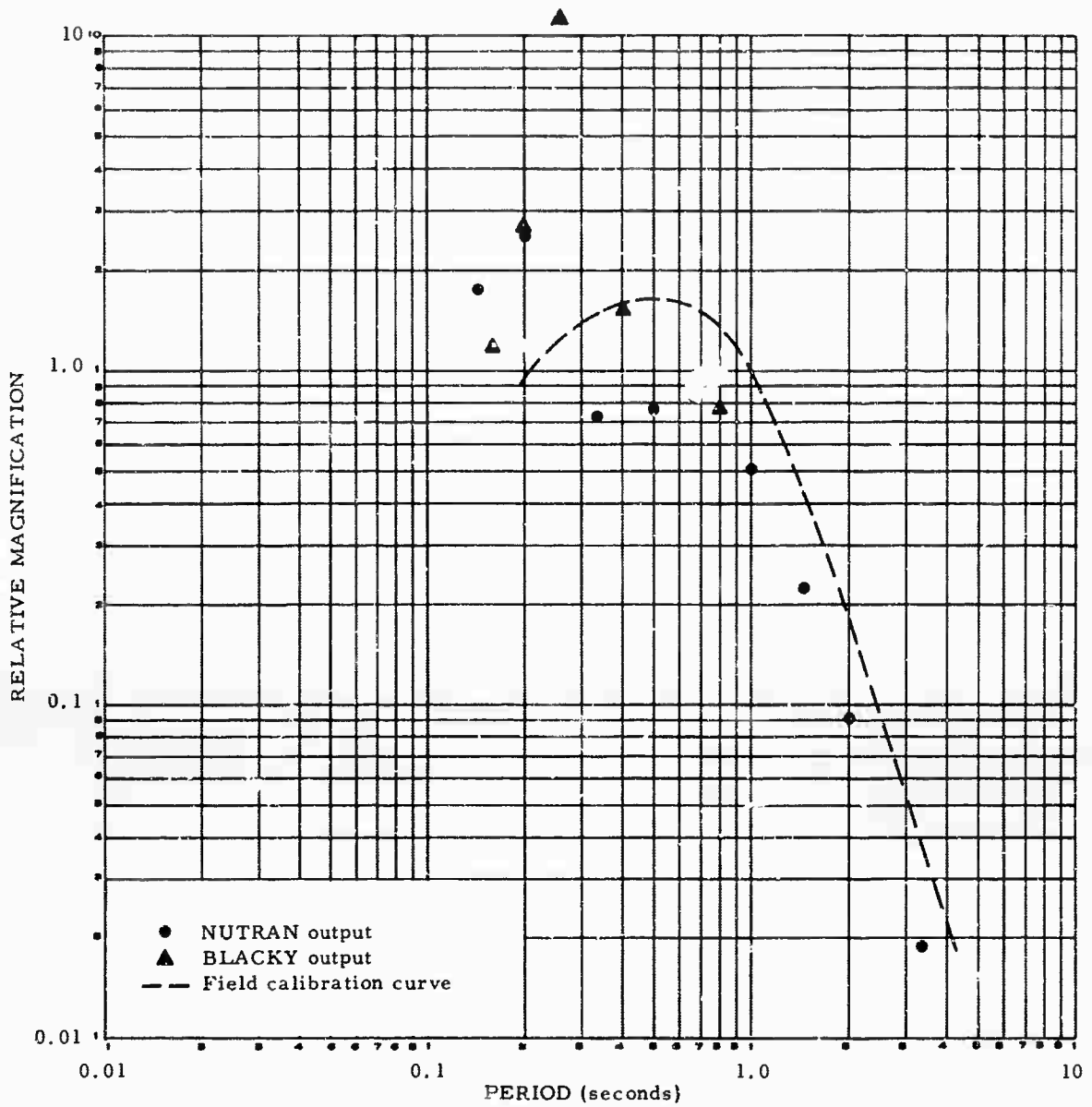


Figure 58. Smoothed calibration estimates at JR-AZ Z₃,
3 second case

G 397

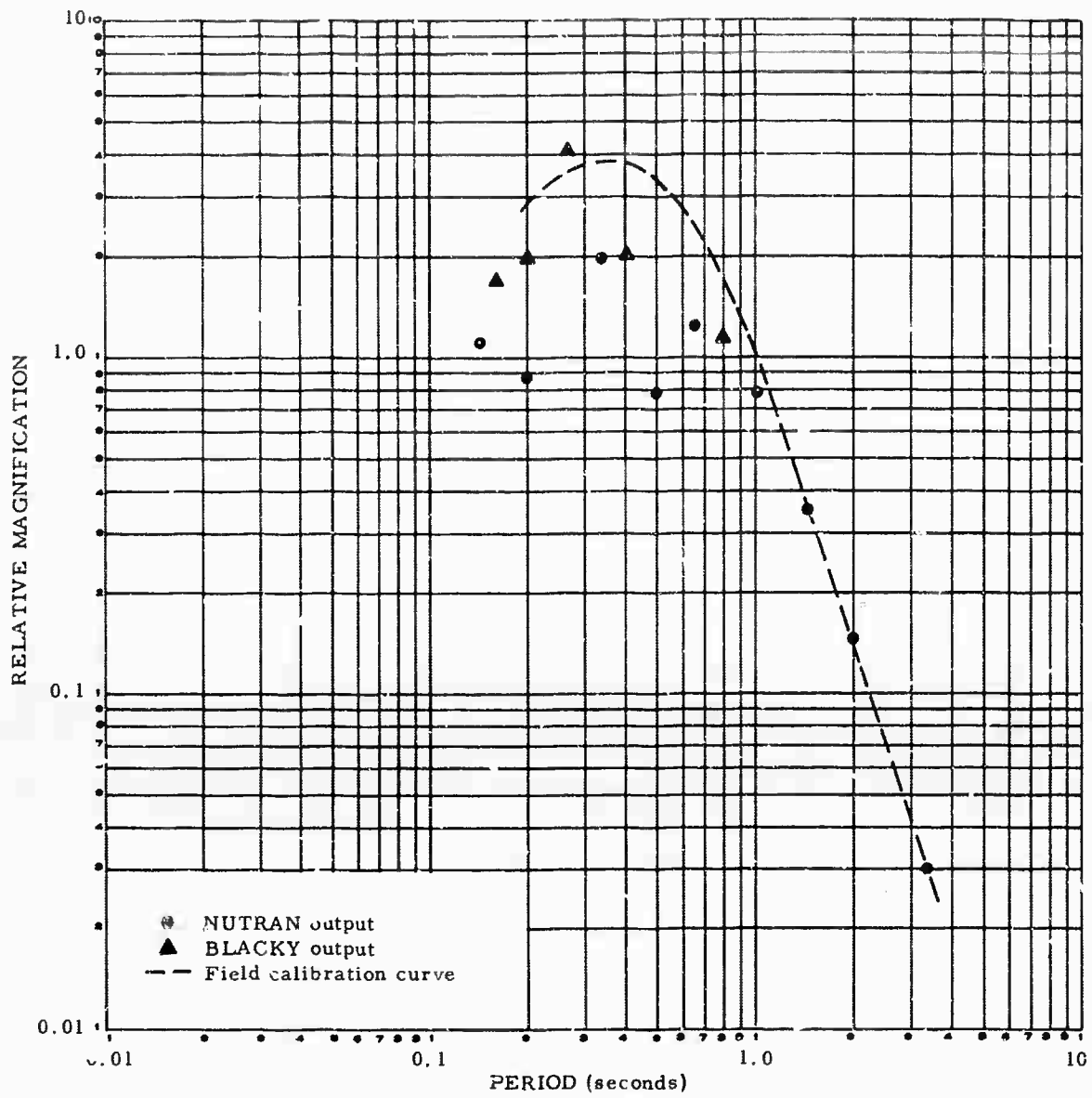


Figure 59. Smoothed calibration estimates at JR-AZ Z₄,
3 second case

G 398

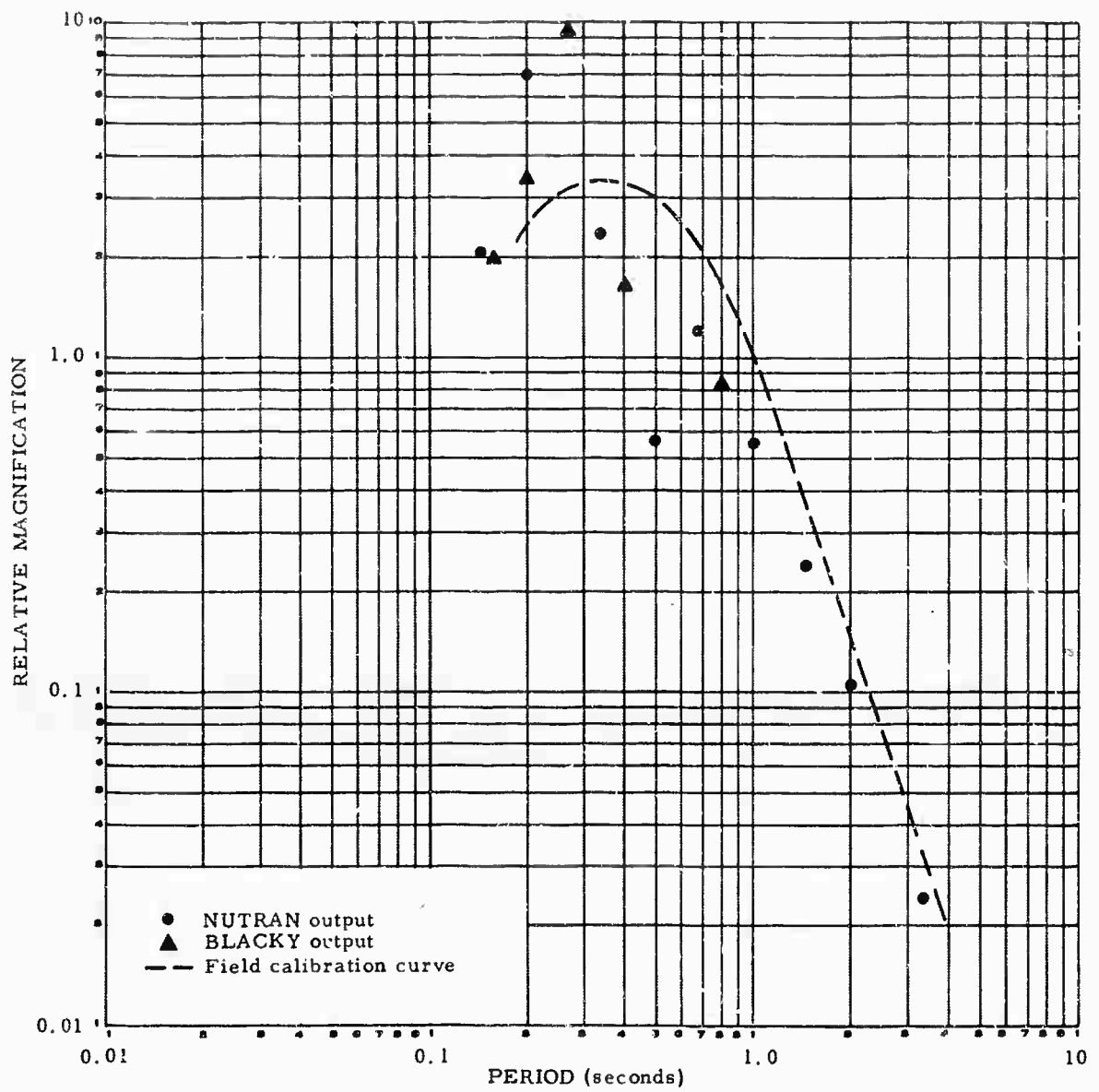


Figure 60. Smoothed calibration estimates at JR-AZ Z₅,
3 second case

G 399

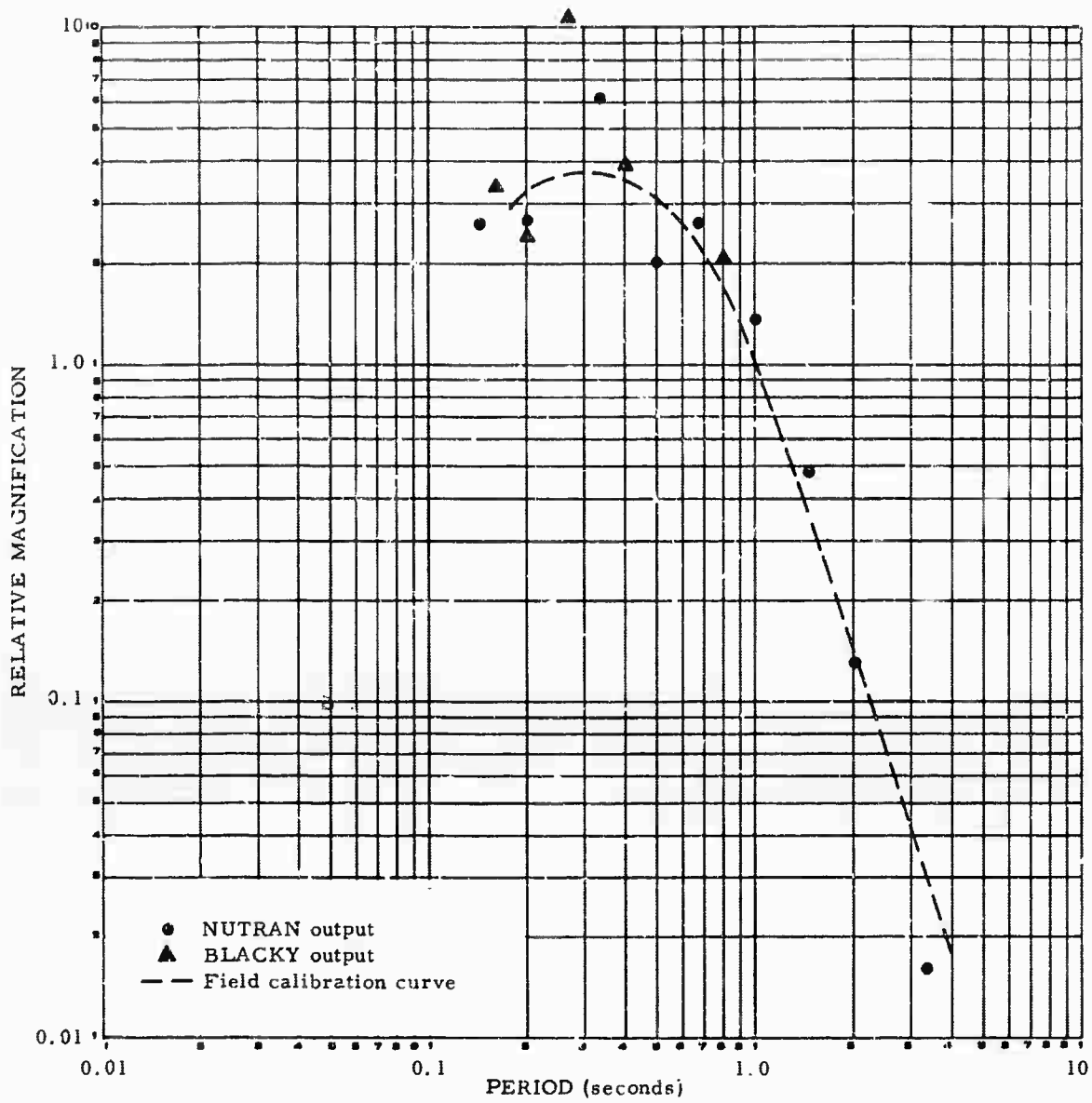


Figure 61. Smoothed calibration estimates at JR-AZ Z₆,
3 second case

G 400

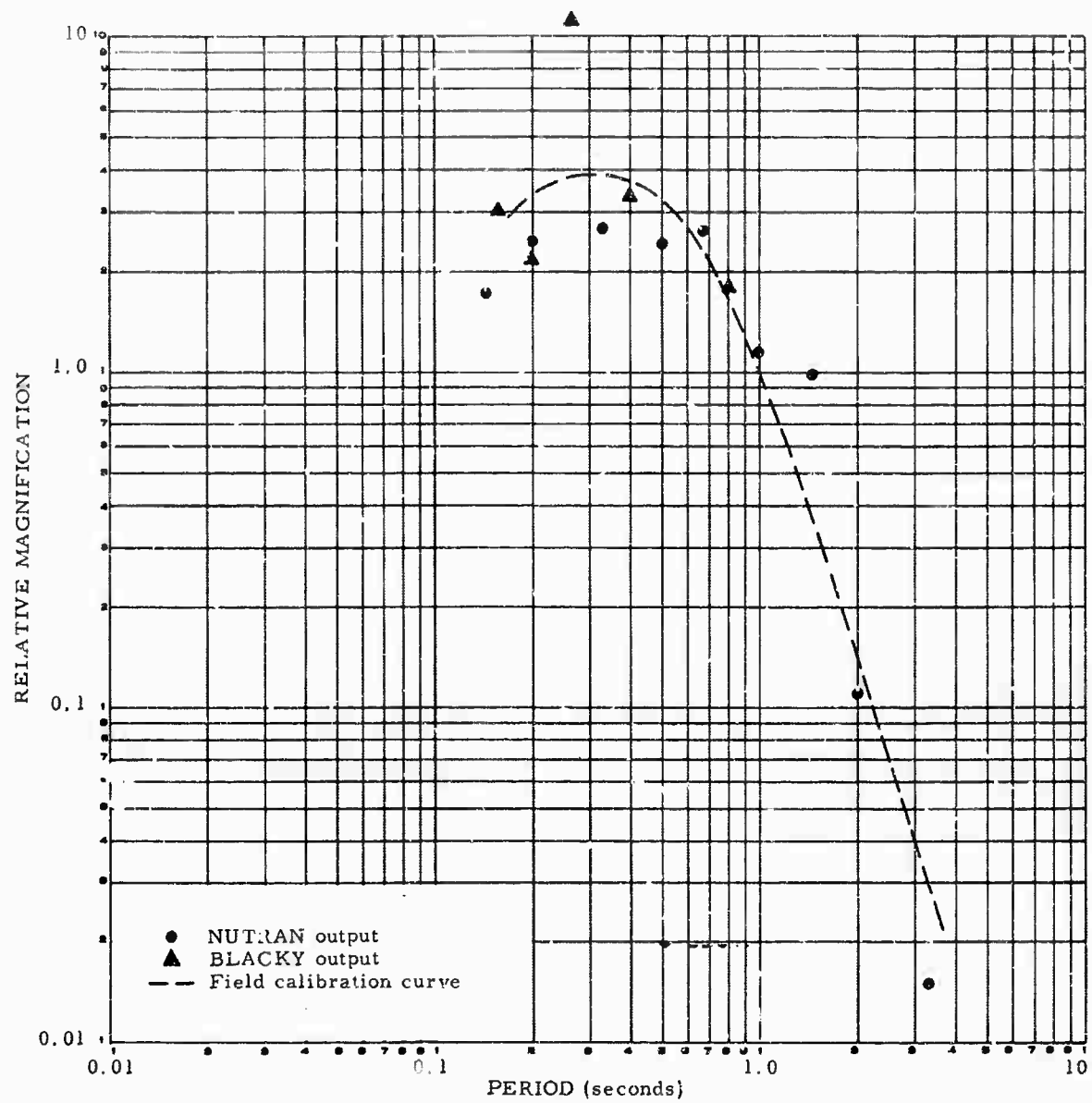


Figure 62. Smoothed calibration estimates at JR-AZ Z7, 3 second case

G 401

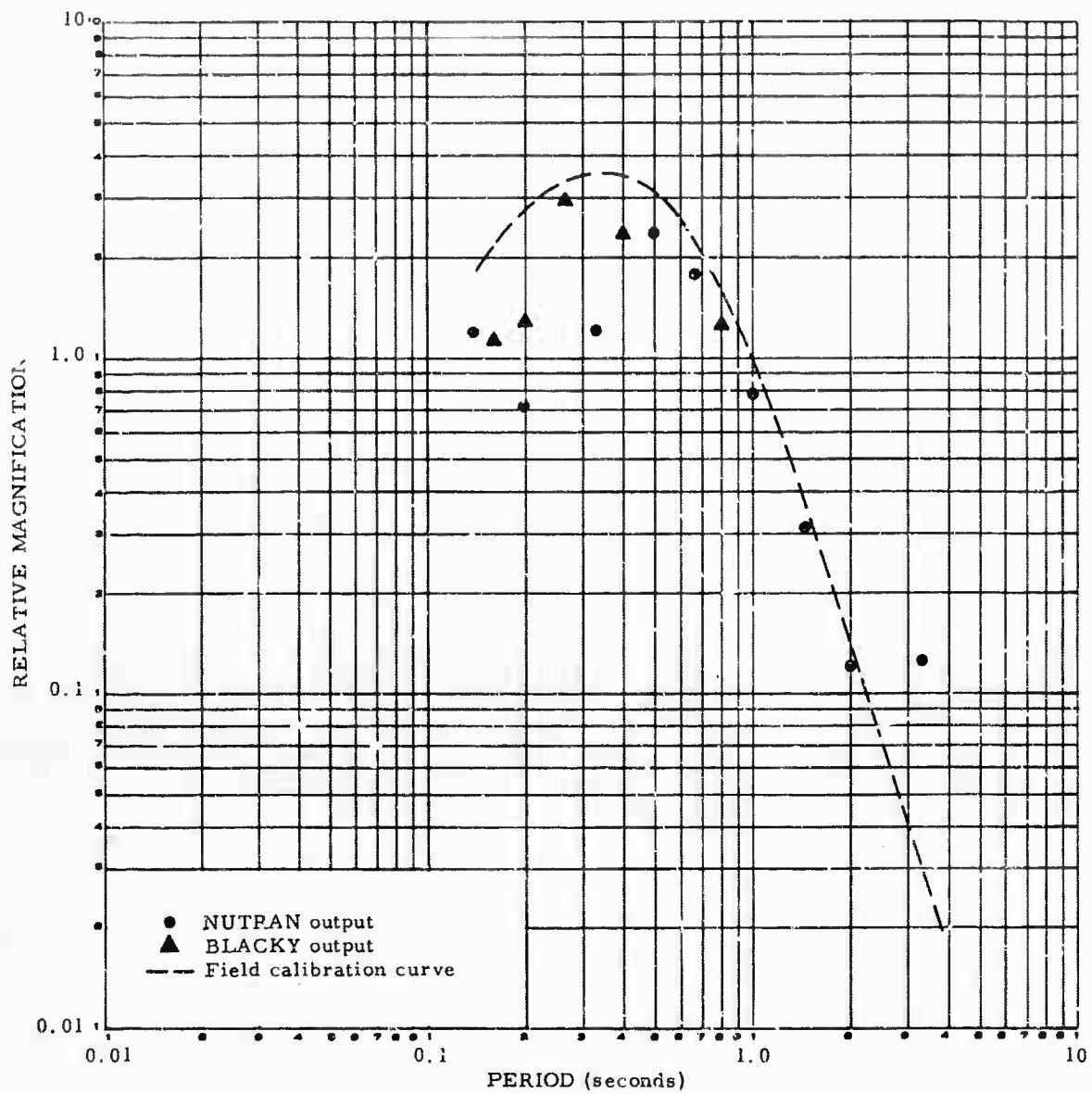


Figure 63. Smoothed calibration estimates at WO-AZ Z₂,
3 second case

G 402

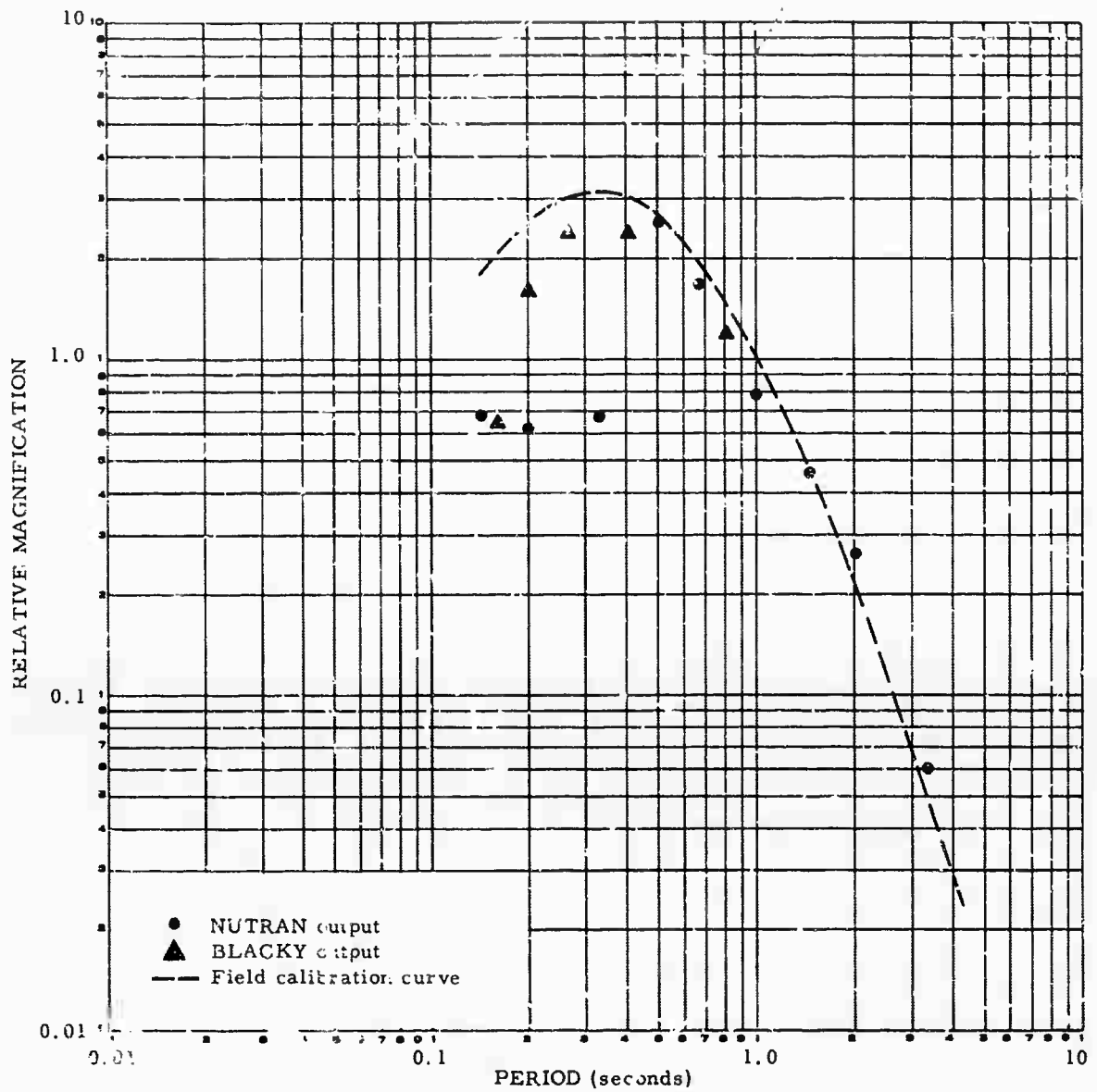


Figure 64. Smoothed calibration estimates at WO-AZ Z₃,
3 second case

G 403

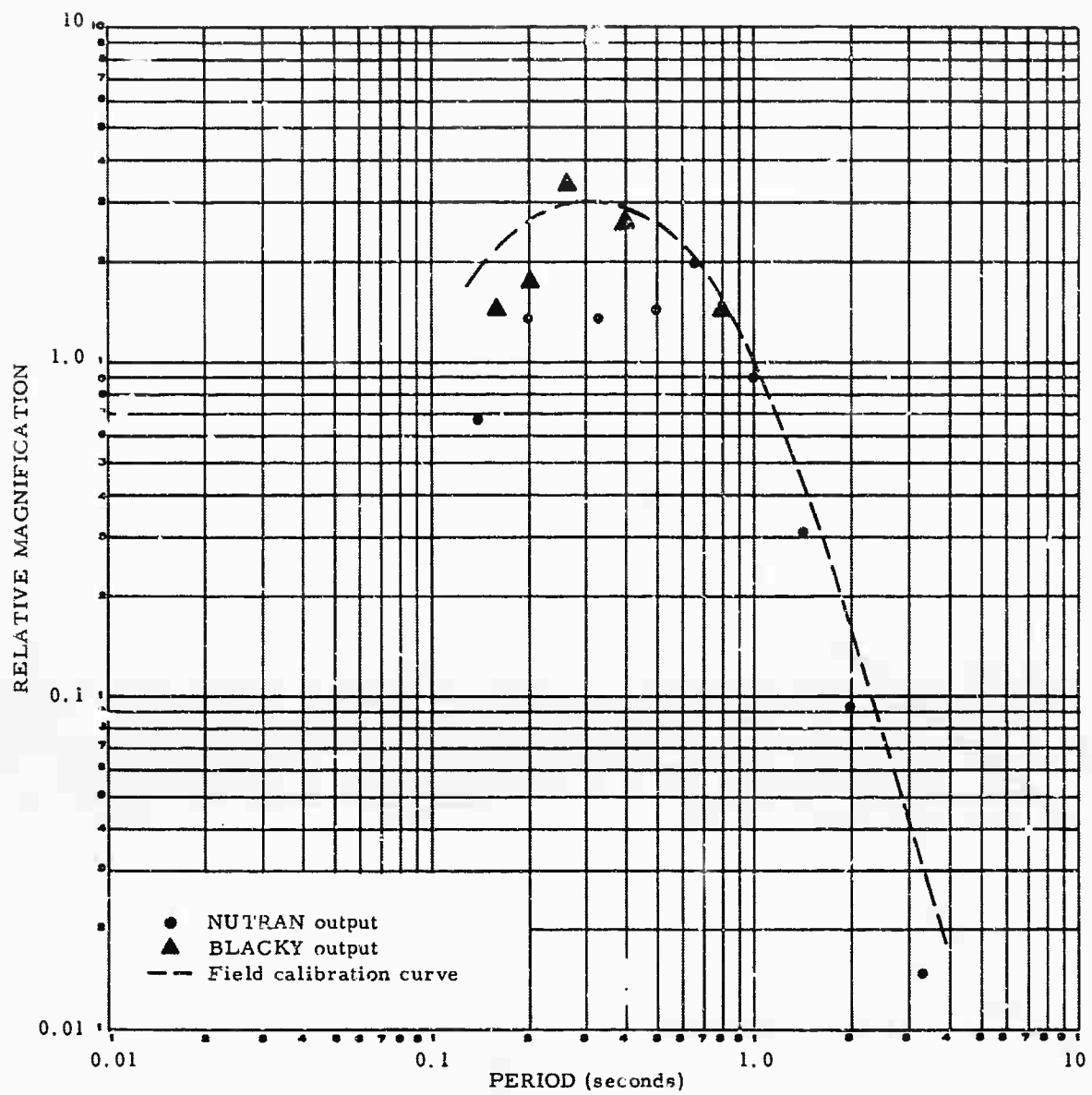


Figure 65. Smoothed calibration estimates at WO-AZ Z₄,
3 second case

G 404

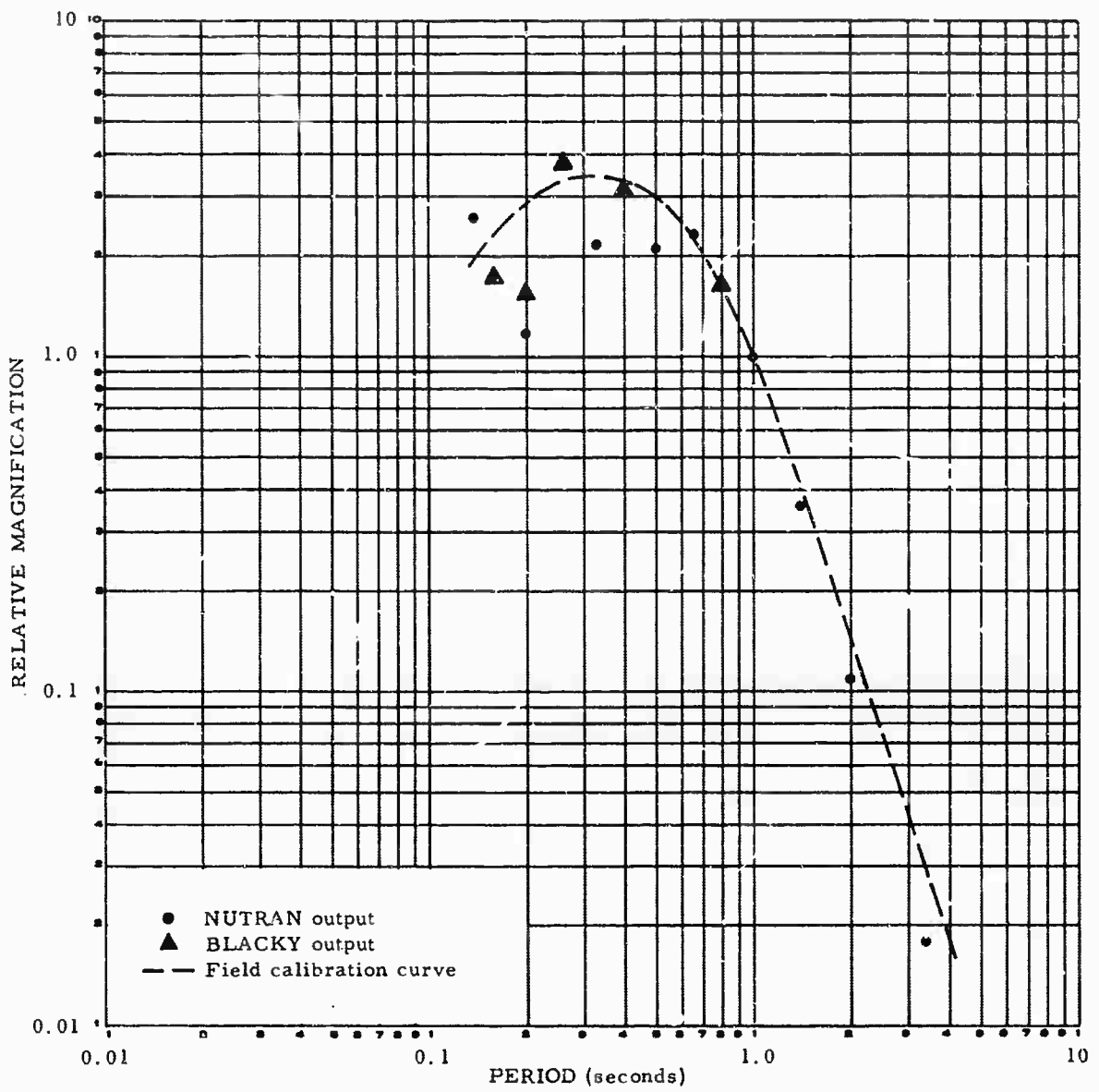


Figure 66. Smoothed calibration estimates at WO-AZ Z₅,
3 second case

G 405

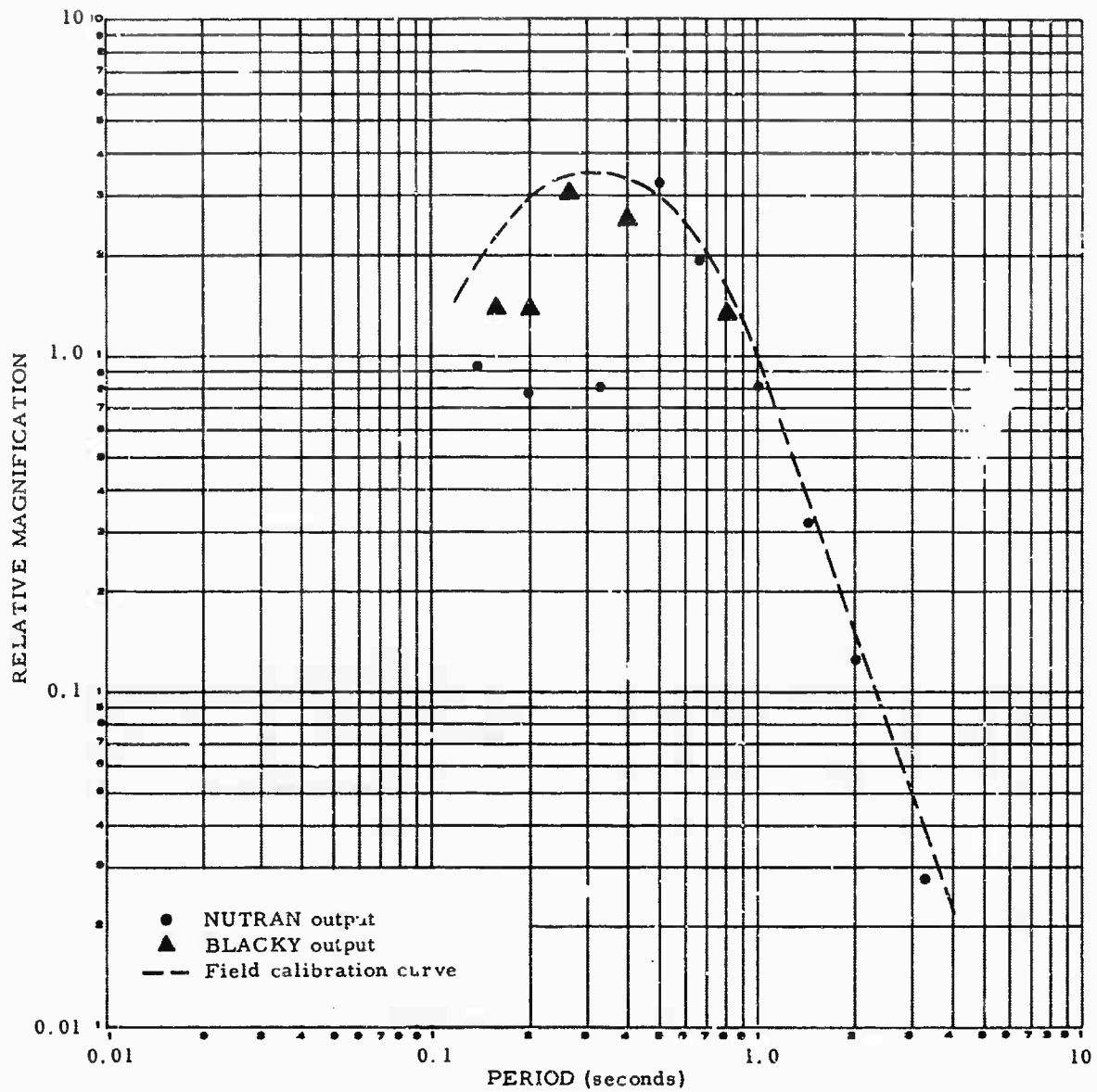


Figure 67. Smoothed calibration estimates at WO-AZ Z₆,
3 second case

G 406

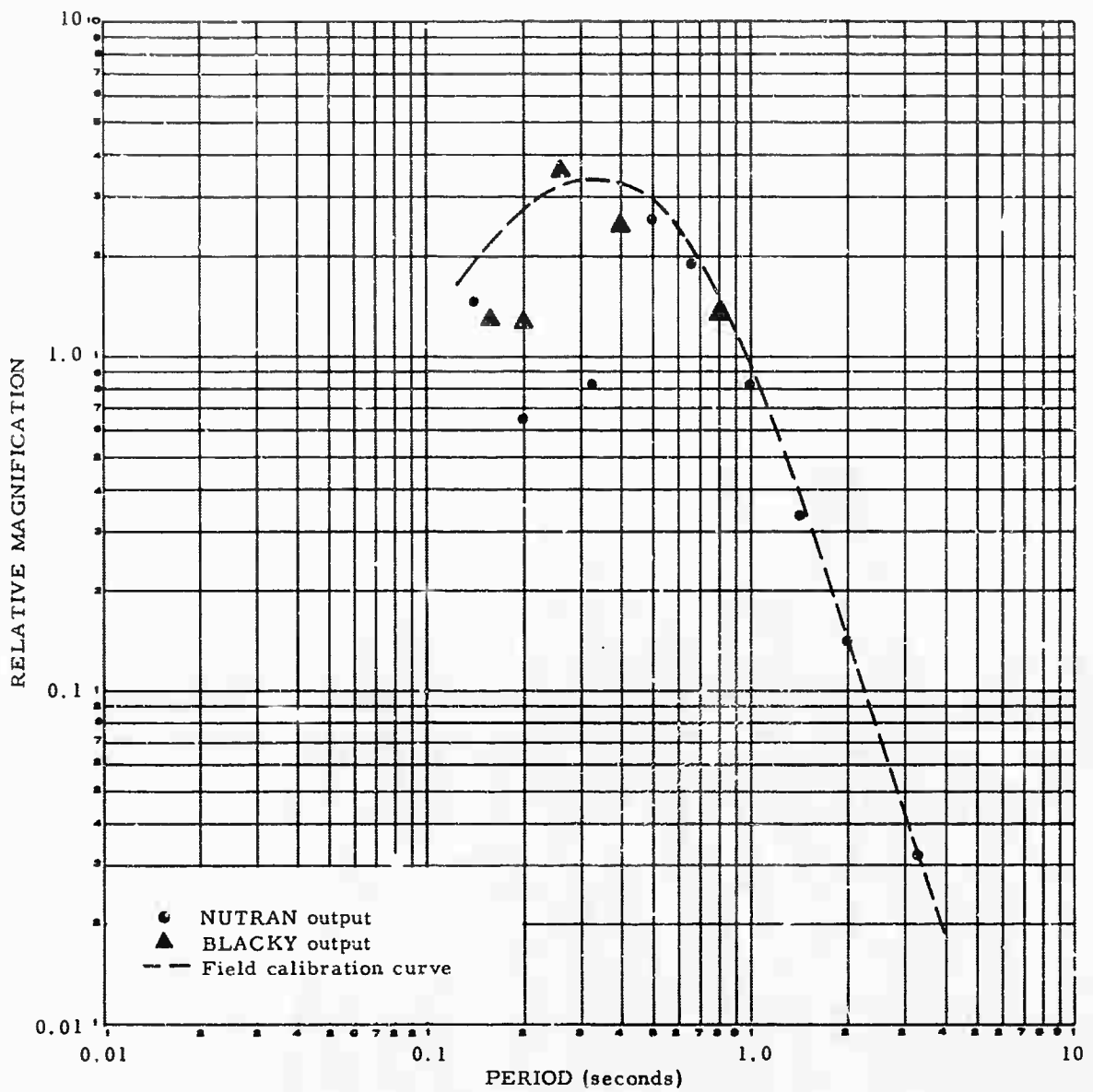


Figure 68. Smoothed calibration estimates at WO-AZ Z₇,
3 second case

G 407

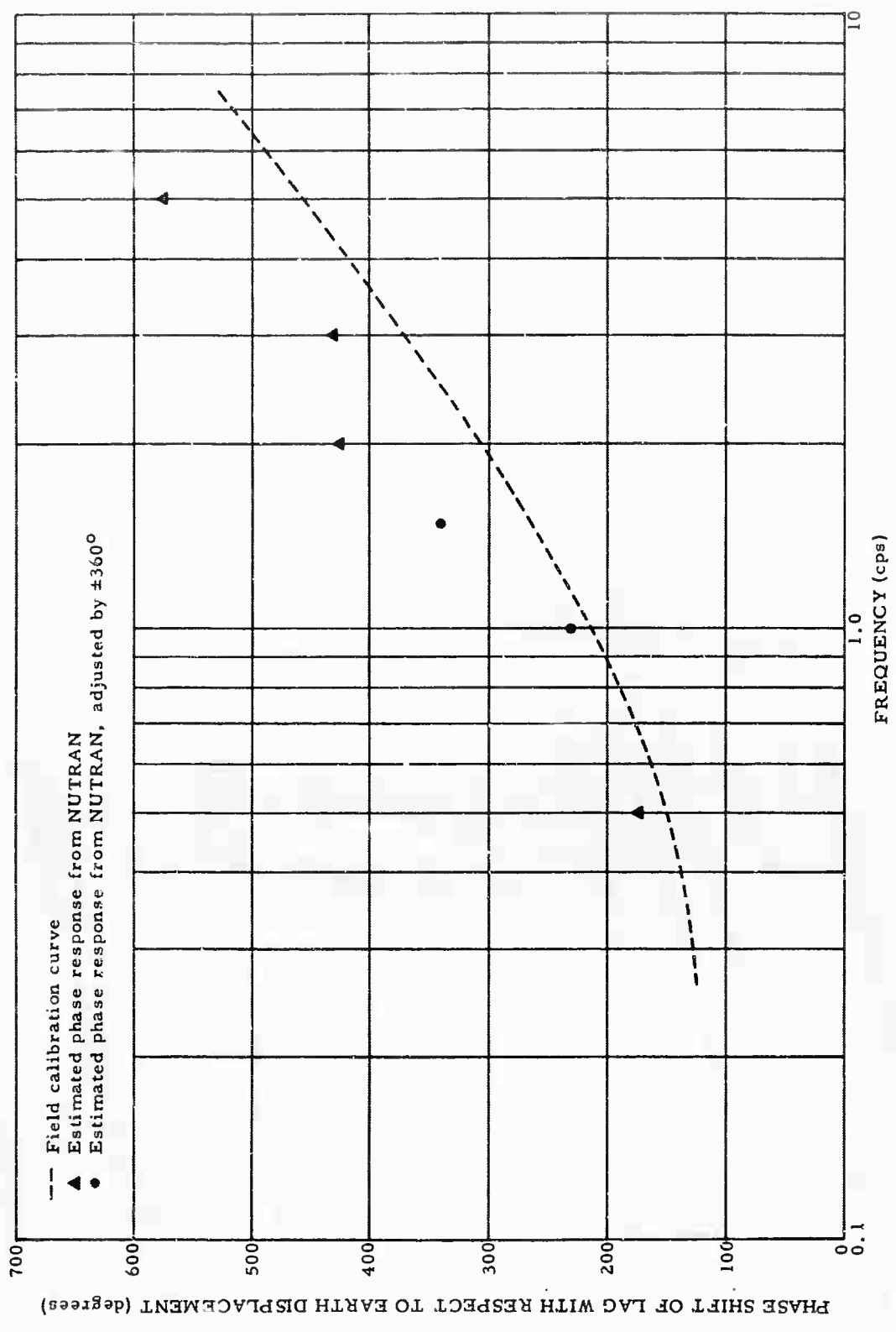


Figure 69. Phase calibration estimates at WO-AZ Z₂, 2 second case

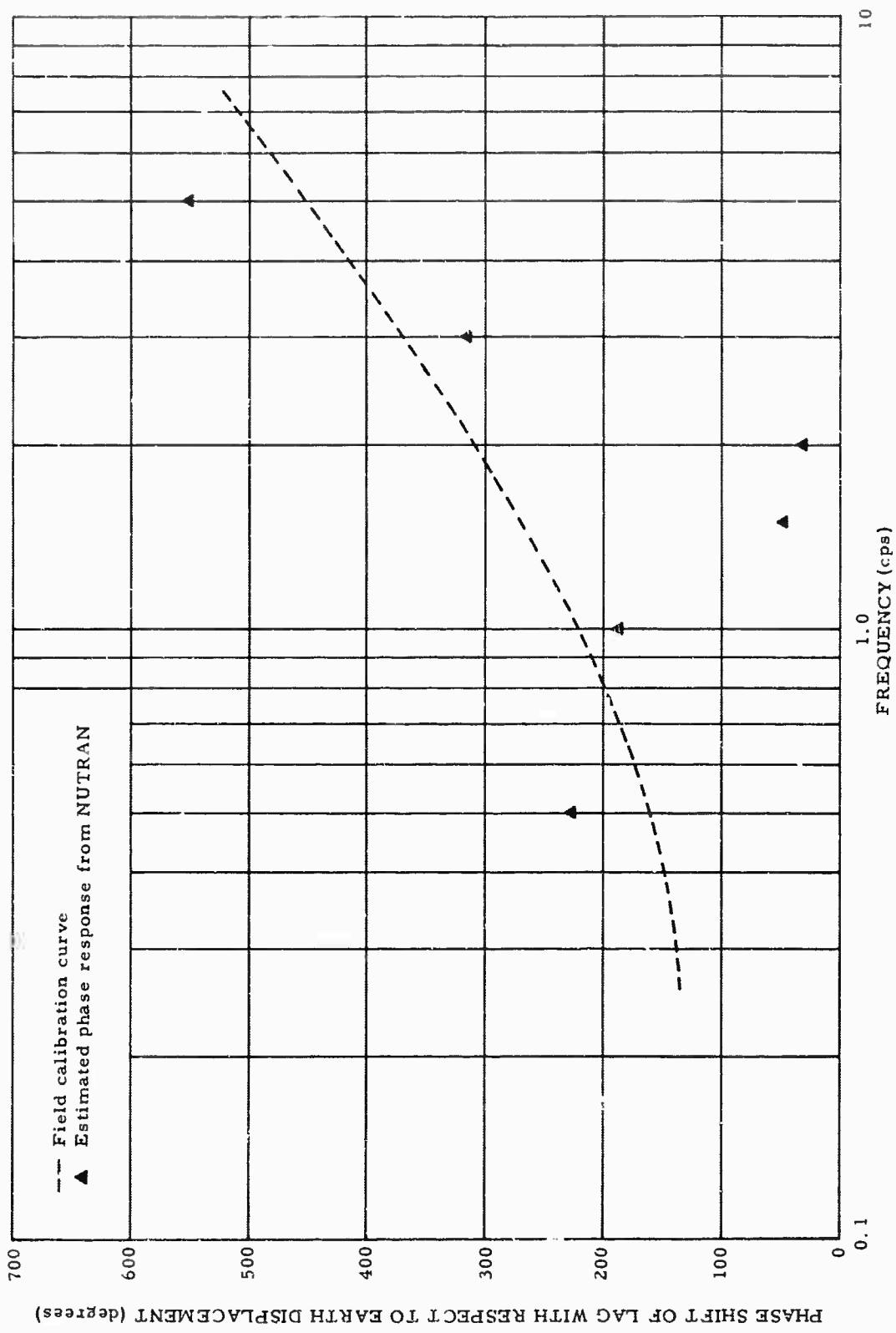


Figure 70. Phase calibration estimates at WO-AZ Z₃, 2 second case

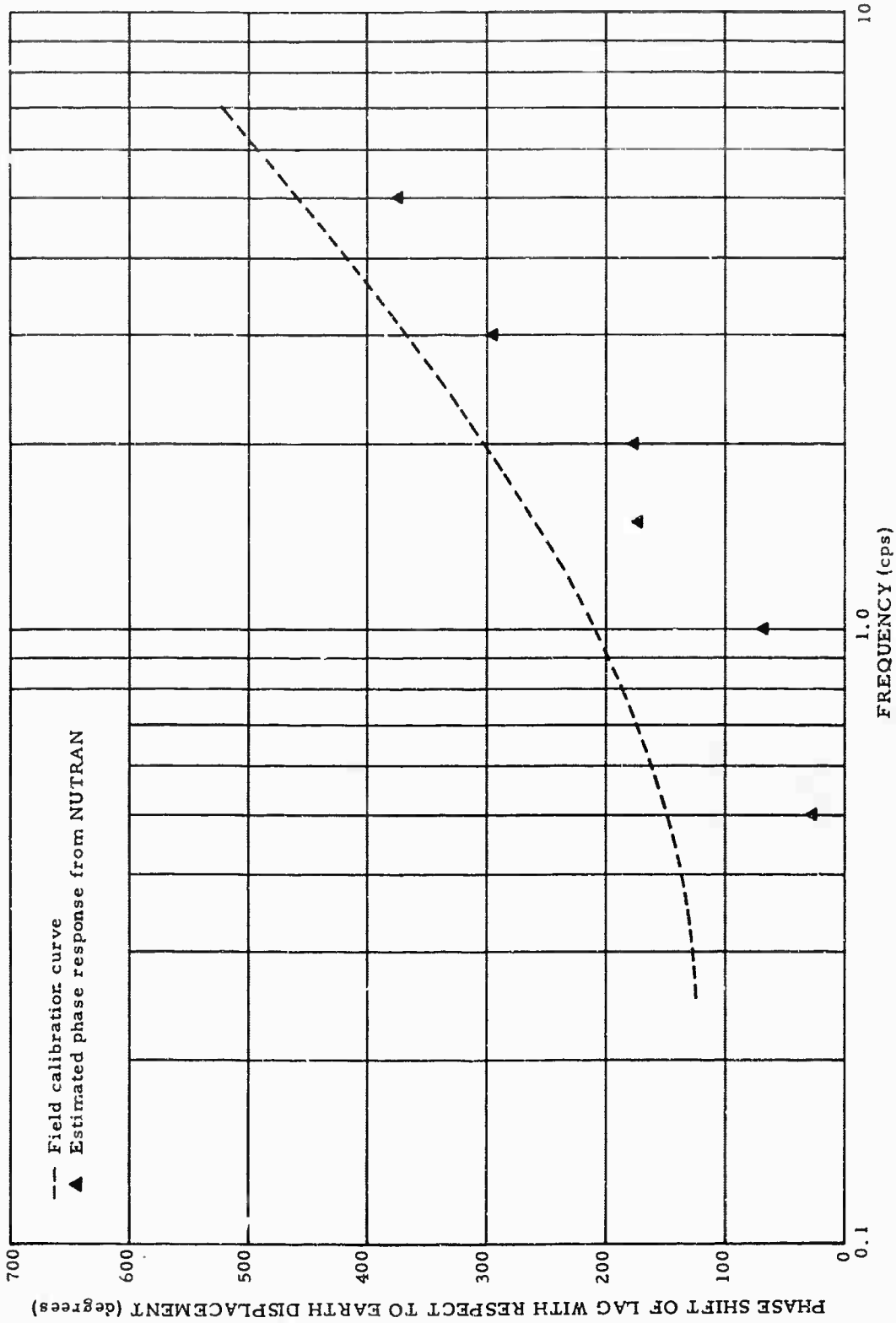


Figure 71. Phase calibration estimates at WO-AZ Z₄, 2 second case

G 410

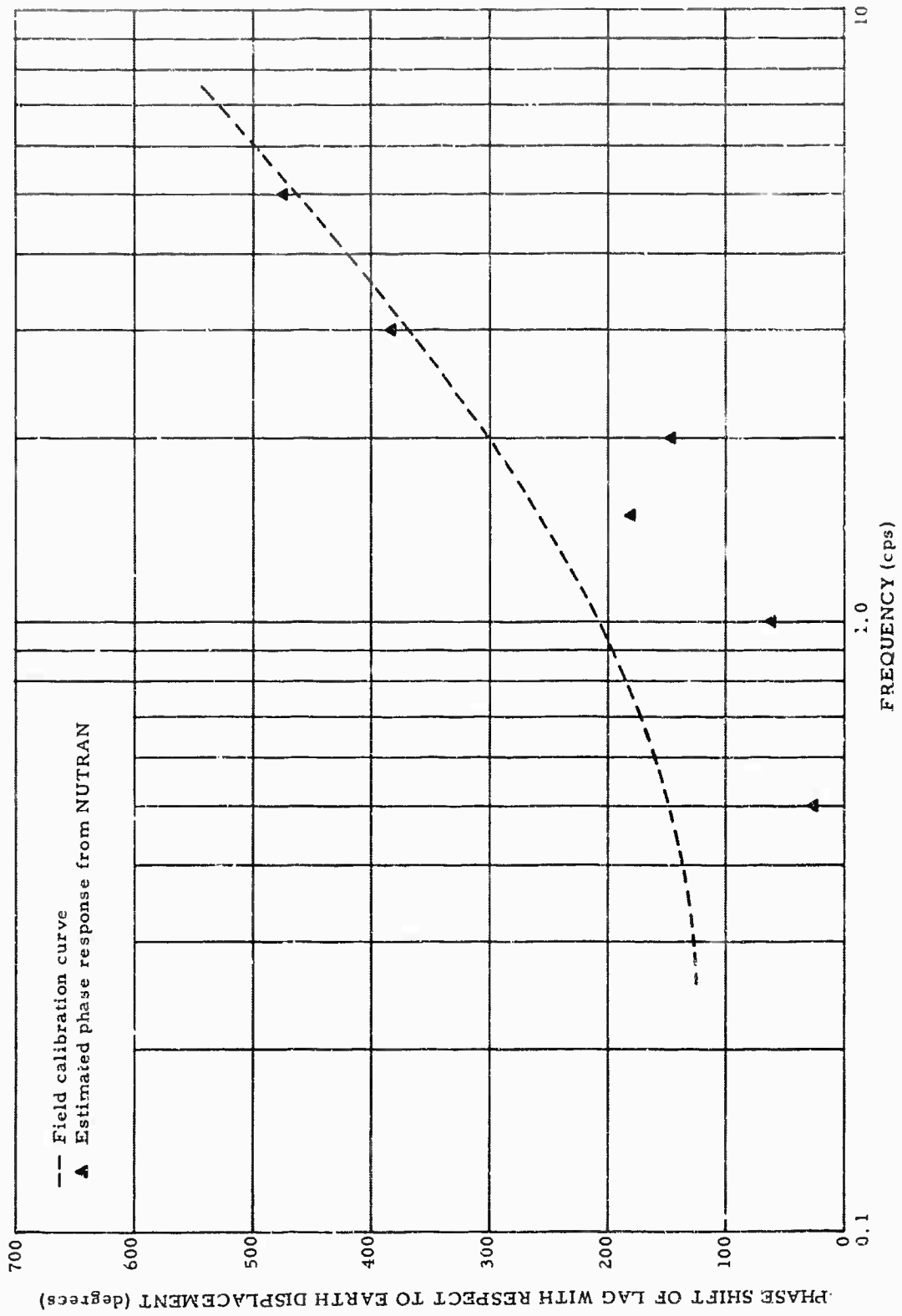


Figure 72. Phase calibration estimates at WO-AZ Z₅, 2 second case

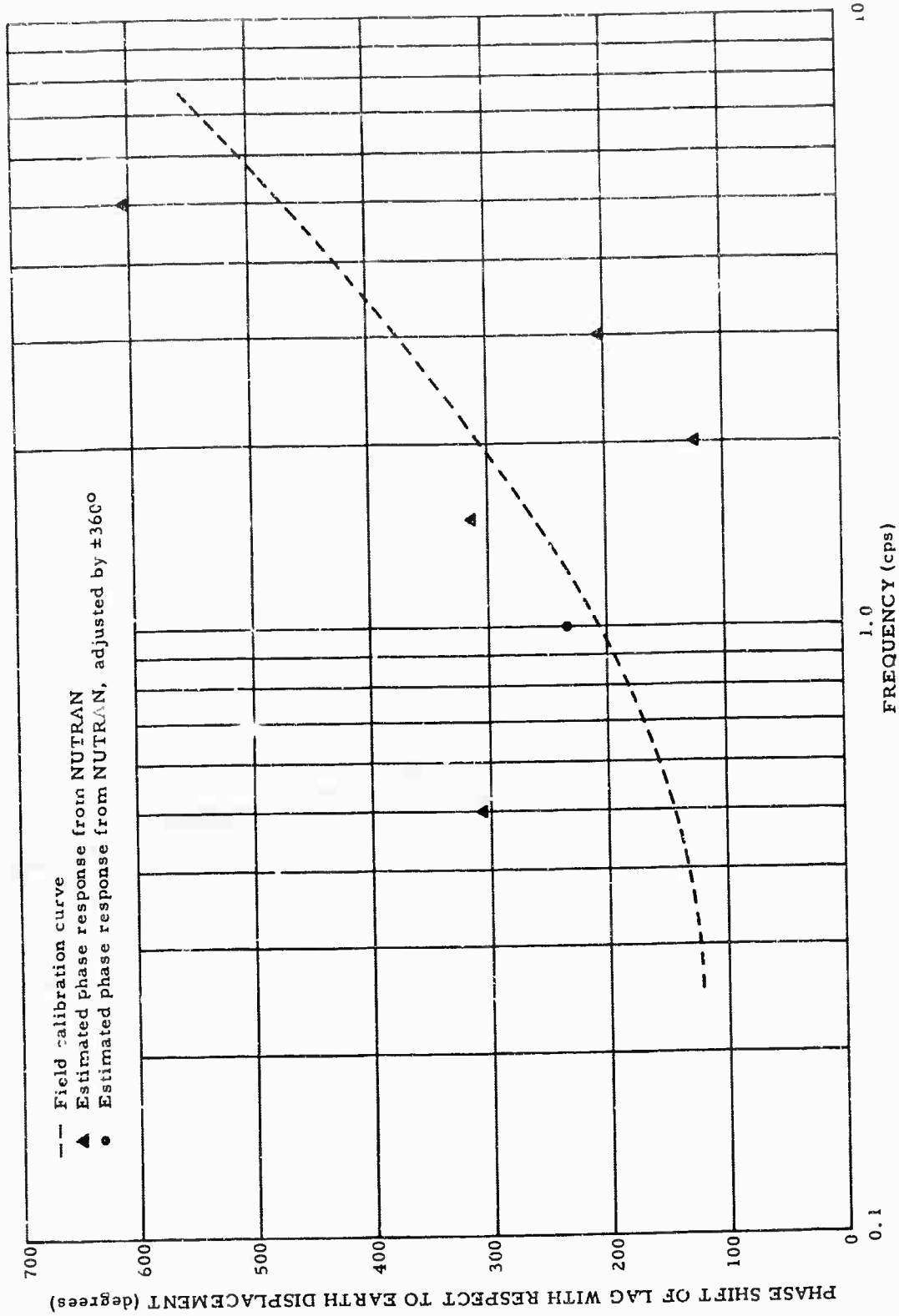


Figure 73. Phase calibration estimates at WO-AZ Z₆, 2 second case

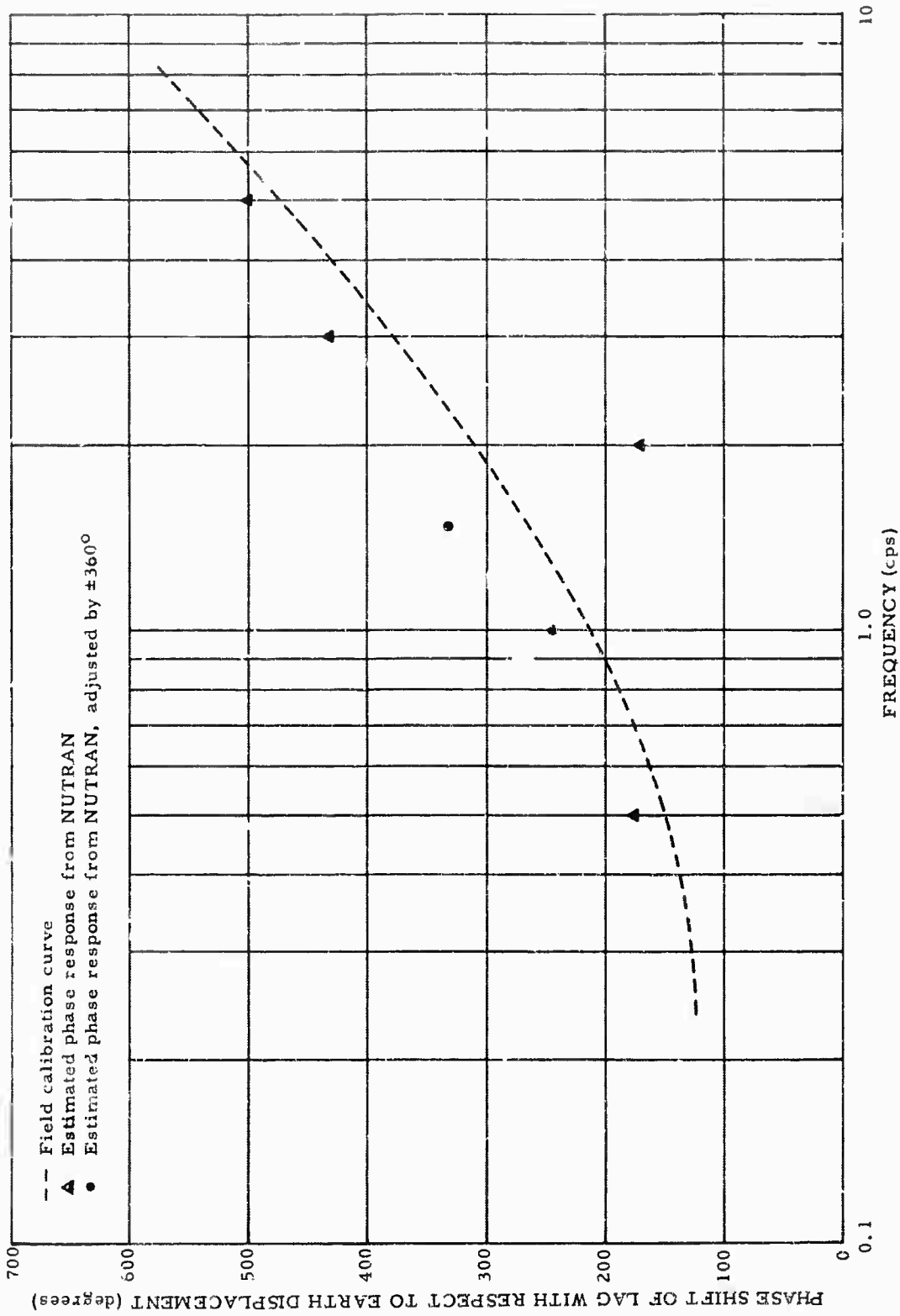


Figure 74. Phase calibration estimates at WO-AZ Z7, 2 second case

G 413

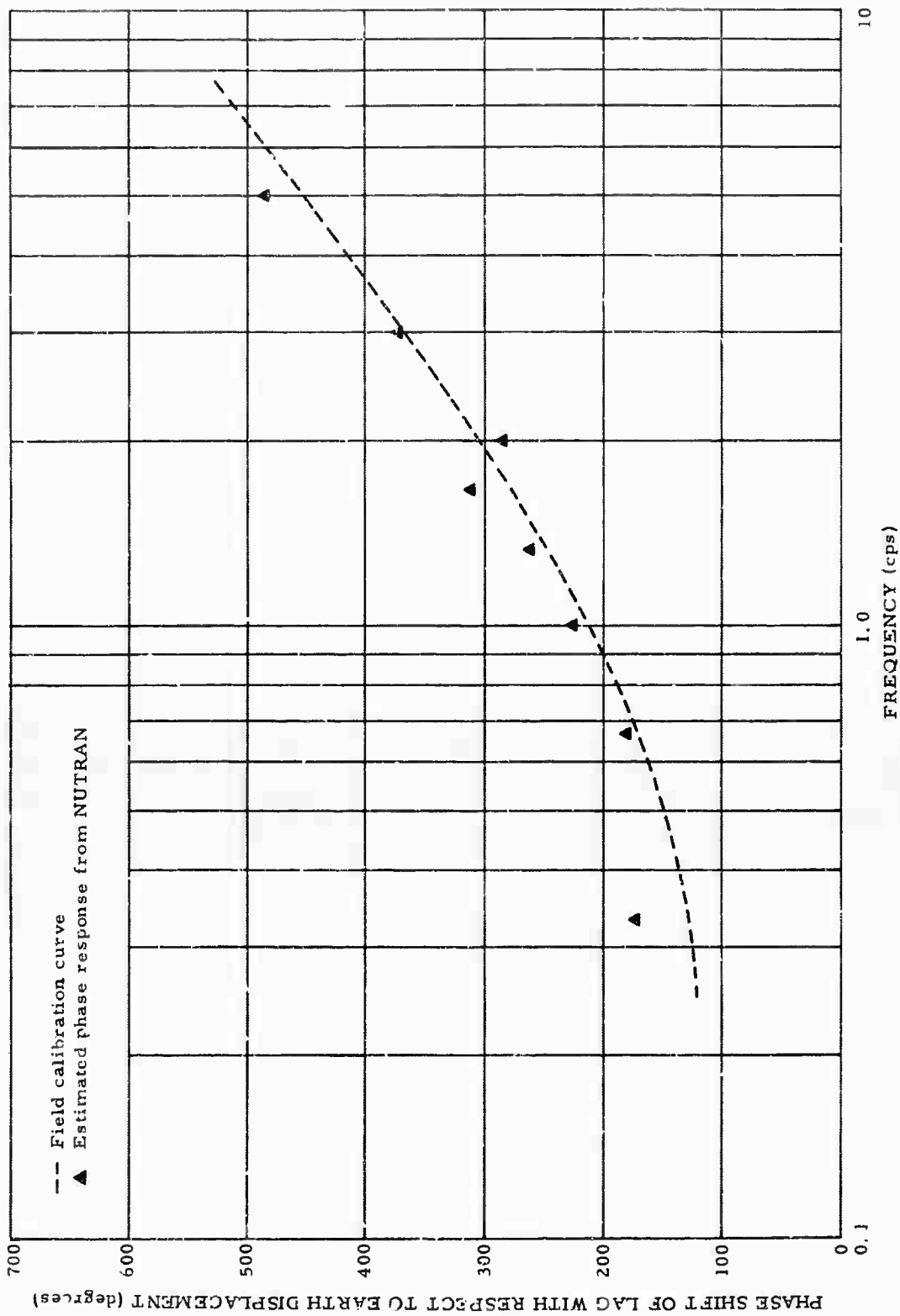


Figure 75. Phase calibration estimates at WO-AZ Z₂, 3 second case

G 414

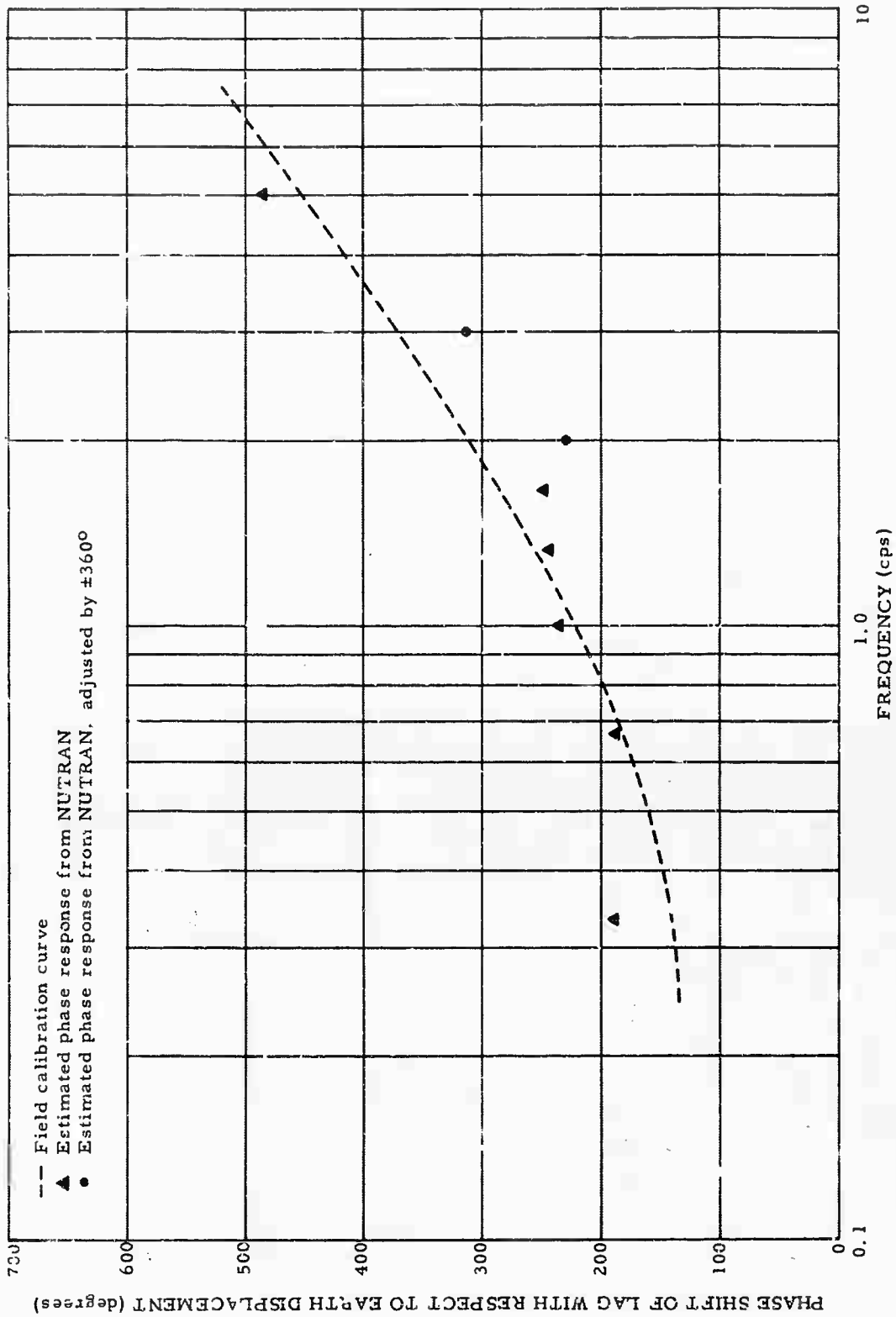


Figure 76. Phase calibration estimates at WO-AZ Z₃, 3 second case

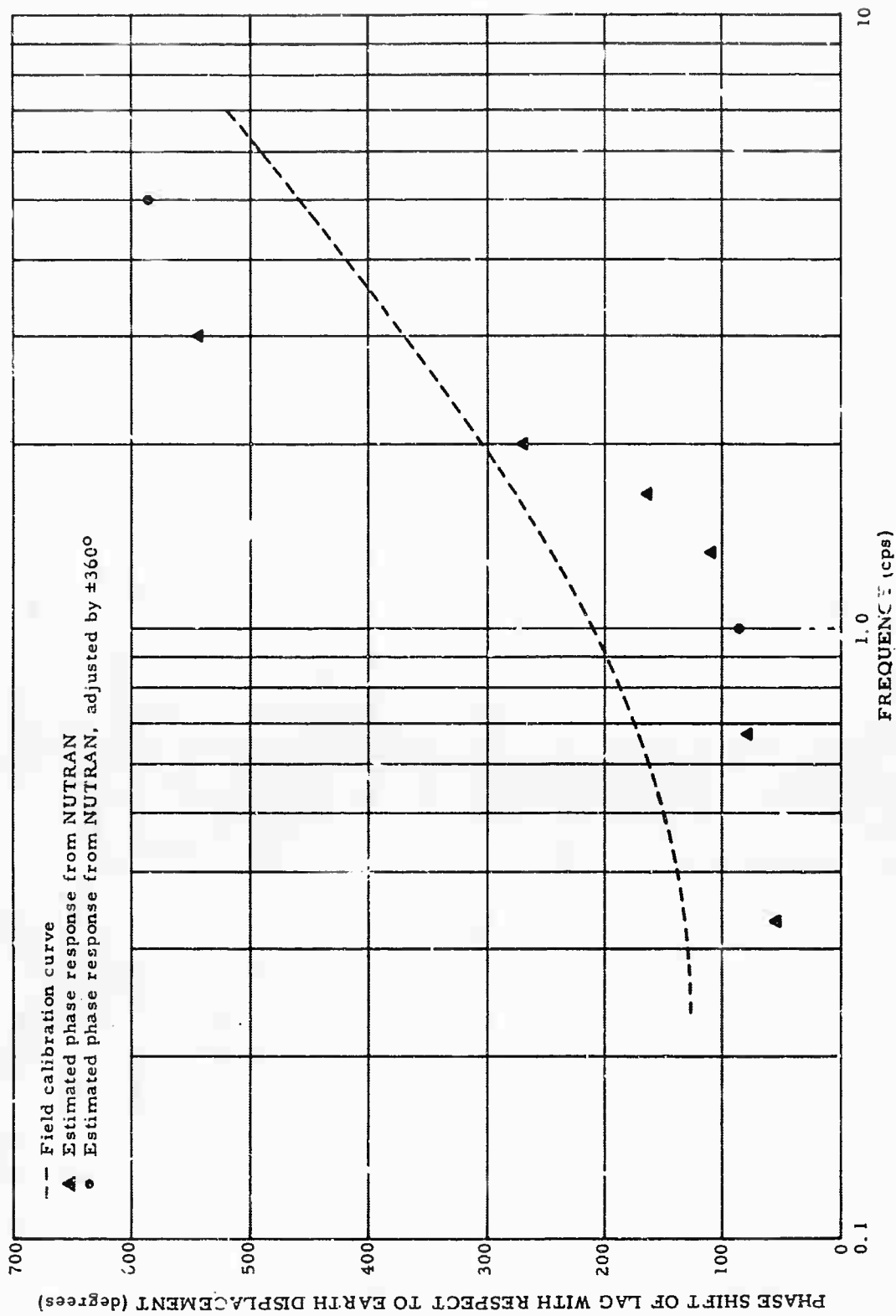


Figure 77. Phase calibration estimates at WO-AZ Z4, 3 second case

G 416

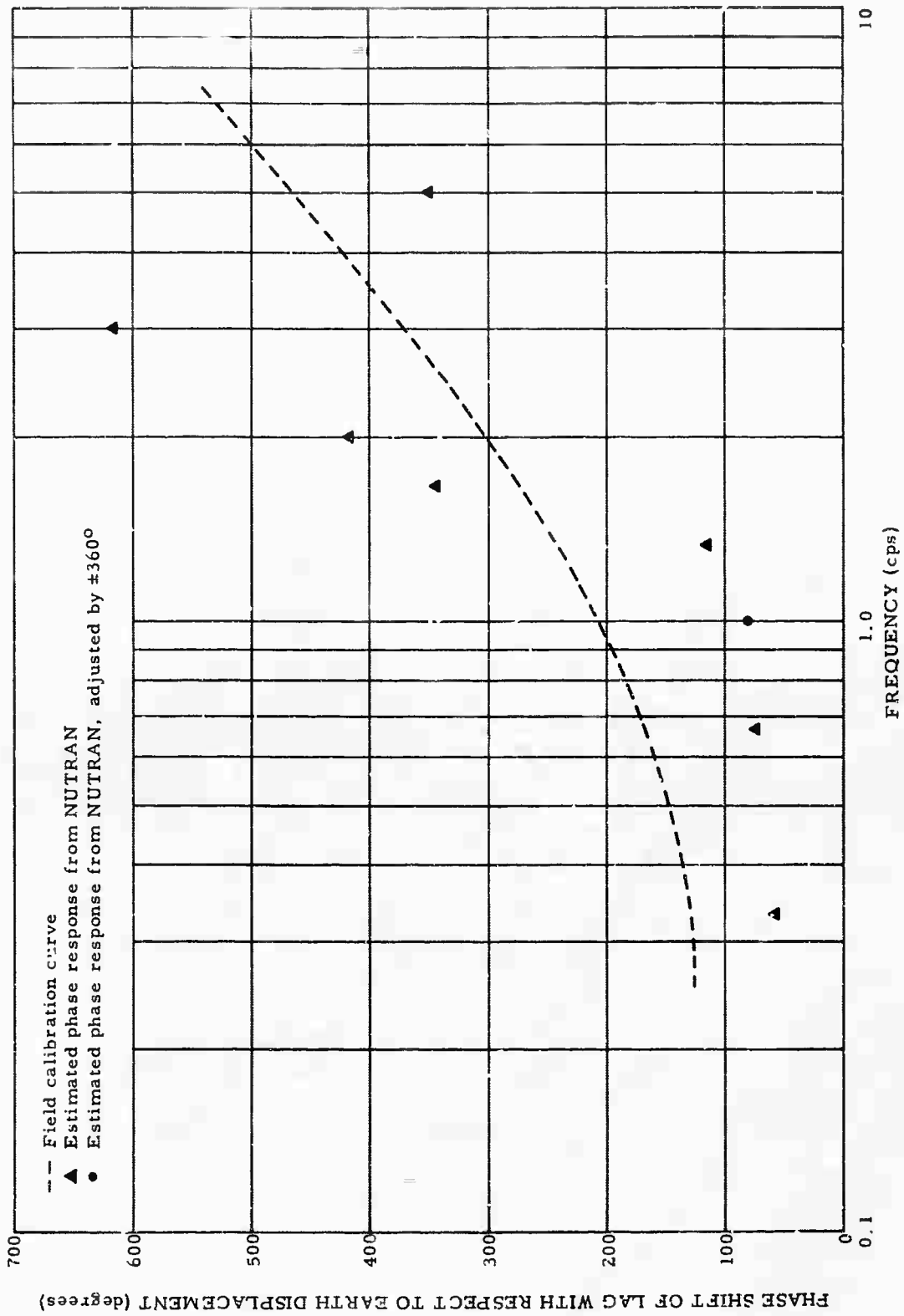


Figure 78. Phase calibration estimates at WO-AZ Z₅, 3 second case

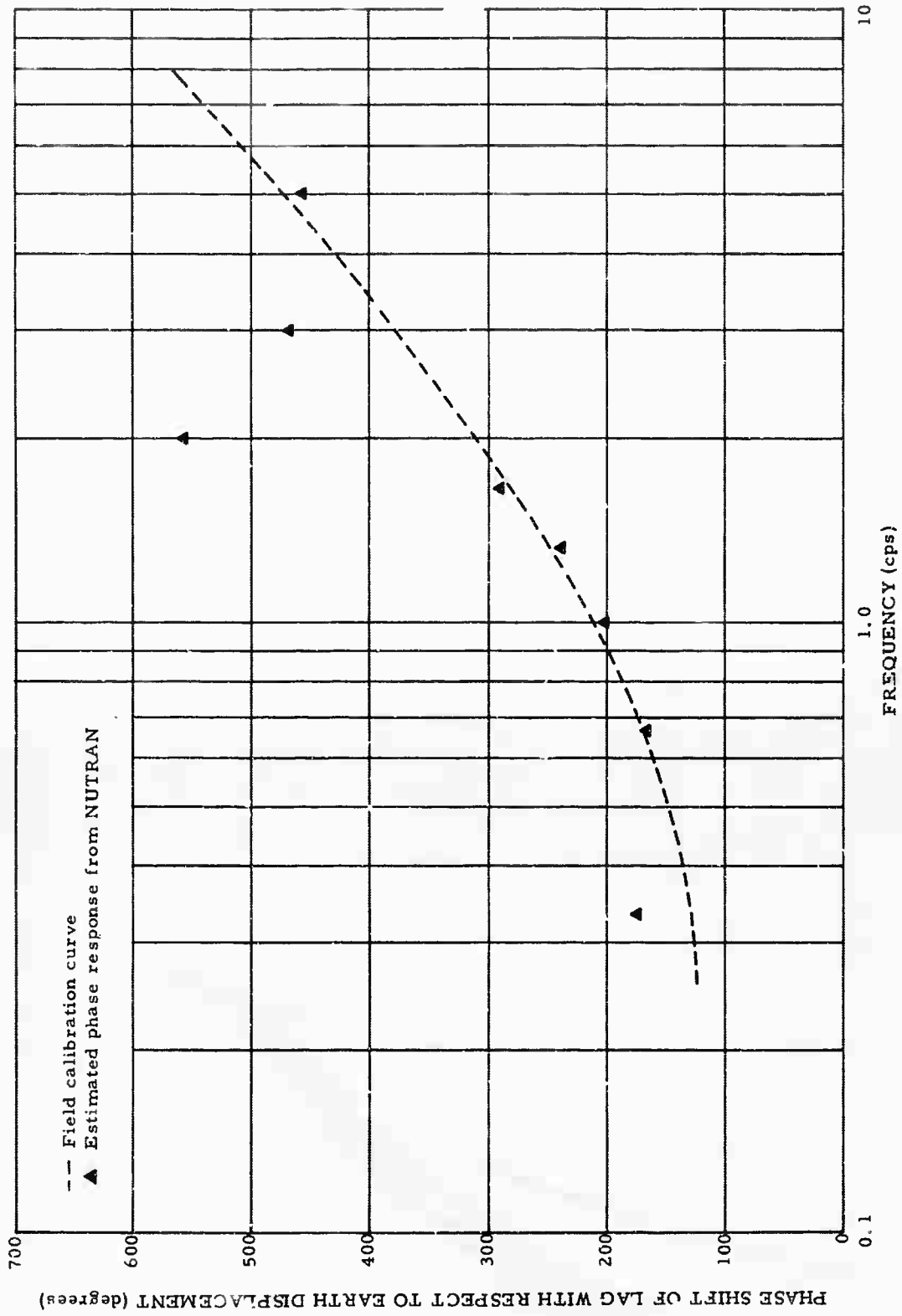


Figure 79. Phase calibration estimates at WO-AZ Z₆, 3 second case

G 418

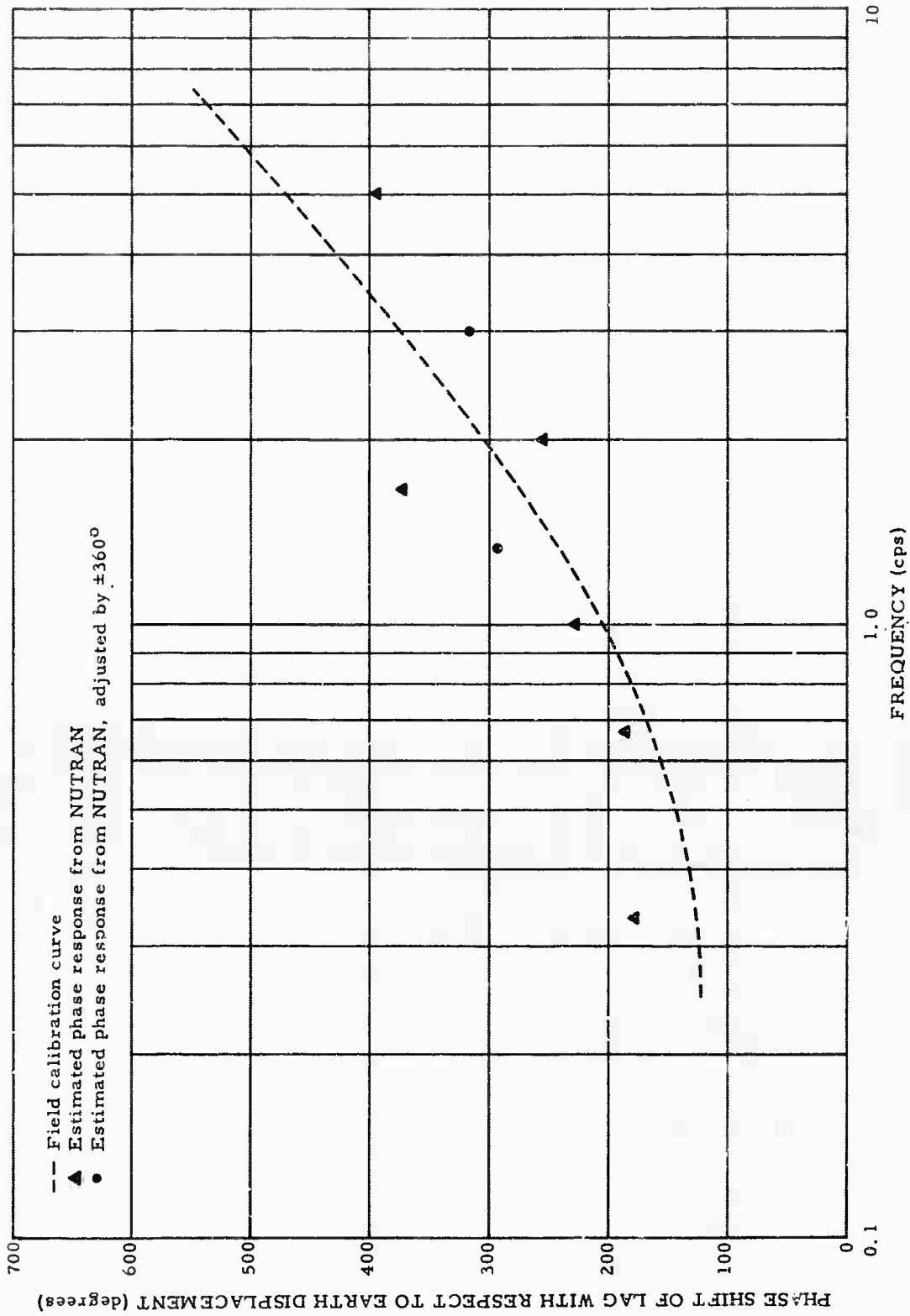


Figure 80. Phase calibration estimates at WO-AZ Z7, 3 second case

4.2 PHASE RESPONSE

Figures 69 through 80 show the estimated phase response curves at WO-AZ using NUTRAN. The values with circles were the adjusted values (by an amount of $\pm 360^\circ$) to meet the multivalued requirement of the phase angle. The results are rather irregular and inconclusive. The corresponding curves at JR-AZ are not available due to lack of phase-calibration data for this date.

5. EFFECT OF NOISE

We wish to study the effect of noise contamination on our calibration results. Let us consider the existence of an additive noise, $n_i(t)$. The calibration model, in this case, can be represented by figure 81.

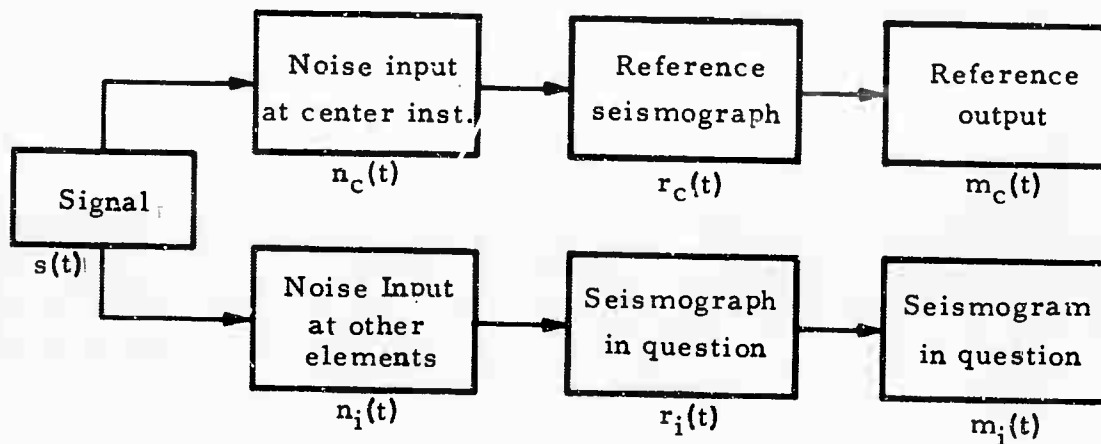


Figure 81. Noise contaminated calibration model

The inputs to the seismographs now become

$$f_c(t) = s(t) + n_c(t), \quad (13)$$

and

$$f_i(t) = s(t) + n_i(t). \quad (14)$$

Accordingly, the $m_c(t)$ and $m_i(t)$ become

$$m_c(t) = [s(t) + n_c(t)] \otimes r_c(t), \quad (15)$$

$$m_i(t) = [s(t) + n_i(t)] \otimes r_i(t), \quad (16)$$

where $r_c(t)$ and $r_i(t)$ are the seismograph impulse responses, and \otimes denotes convolution.

In the spectral domain, we have respectively

$$M_c(\omega) = [S(\omega) + N_c(\omega)] R_c(\omega), \quad (17)$$

and

$$M_i(\omega) = [S(\omega) + N_i(\omega)] R_i(\omega). \quad (18)$$

Replacing $R_c(\omega)$ and $R_i(\omega)$ by

$$R_c(\omega) = A_c(\omega)e^{jB_c(\omega)}, \quad (19)$$

and

$$R_i(\omega) = A_i(\omega)e^{jB_i(\omega)}, \quad (20)$$

we will have expressions for $A_i(\omega)$ and $B_i(\omega)$ which correspond to the amplitude and phase response curves of the seismograph in question.

In terms of the modulus and phase, $M_c(\omega)$, $M_i(\omega)$, $N_c(\omega)$, and $N_i(\omega)$ can be rewritten by

$$\begin{aligned} M_c &= \rho_{mc} e^{j\theta_{mc}}, & M_i &= \rho_{mi} e^{j\theta_{mi}}, \\ N_c &= \rho_{nc} e^{j\theta_{nc}}, & N_i &= \rho_{ni} e^{j\theta_{ni}}. \end{aligned} \quad (21)$$

For the purpose of simplicity, the expressions (ω) are abbreviated. The $A_i(\omega)$ and $B_i(\omega)$ then become

$$A_i(\omega) = \frac{\rho_{mi}(\omega)A_c(\omega)}{\Delta(\omega)}, \quad (22)$$

and (23)

$$B_i(\omega) = B_c(\omega) + \theta_{mi}(\omega) - \tan^{-1} \left[\frac{\rho_{mc} \sin \theta_{mc} + \rho_{ni} A_c \sin(\theta_{ni} + B_c) - \rho_{nc} A_c \sin(\theta_{nc} + B_c)}{\rho_{mc} \cos \theta_{mc} + \rho_{ni} A_c \cos(\theta_{ni} + B_c) - \rho_{nc} A_c \cos(\theta_{nc} + B_c)} \right],$$

where

$$\Delta^2(\omega) = \rho_{mc}^2 + \rho_{ni}^2 A_c^2 + \rho_{nc}^2 A_c^2 - 2\rho_{ni} \rho_{nc} A_c^2 \cos(\theta_{ni} - \theta_{nc}) \quad (24)$$

$$+ 2\rho_{mc} \rho_{ni} A_c \cos(\theta_{mc} - \theta_{ni} - B_c) - 2\rho_{mc} \rho_{nc} A_c \cos(\theta_{mc} - \theta_{nc} - B_c).$$

When

$$N_c = N_i = \rho_n e^{j\theta_n}, \text{ equation (24) yields}$$

$$\Delta(\omega) \Big|_{N_c = N_i} = \rho_{mc}(\omega). \quad (25)$$

Equations (22) and (23) then become, respectively

$$A_i(\omega) = A_c(\omega) \cdot \frac{\rho_{mi}(\omega)}{\rho_{mc}(\omega)}, \quad (26)$$

and

$$B_i(\omega) = B_c(\omega) + \theta_{ni}(\omega) - \theta_{mc}(\omega), \quad (27)$$

which are the same expressions as equations (1) and (2).

The magnification curves are affected by $\Delta(\omega)$ of equation (24). Depending upon the sign of the quantity

$$E^2 = (\rho_{ni}^2 + \rho_{nc}^2)A_c^2 + 2\rho_{mc}\rho_{ni}A_c \cos(\theta_{mc} - \theta_{ni} - B_c) - 2\rho_{nc}A_c [\rho_{ni}A_c \cos(\theta_{ni} - \theta_{nc}) + \rho_{mc} \cos(\theta_{mc} - \theta_{nc} - B_c)], \quad (28)$$

the amplitude response can deviate from the normal response curve either upward or downward. For $E^2 > 0$, the estimated response will be smaller than the normal, whereas for $E^2 < 0$, the estimated response will become greater than the normal.

The condition for perfect match between the estimated and the field calibration results under the above modifications is

$$(\rho_{ni}^2 + \rho_{nc}^2)A_c^2 + 2\rho_{mc}\rho_{ni} \cos(\theta_{mc} - \theta_{ni} - B_c) = 2\rho_{nc} [\rho_{ni}A_c \cos(\theta_{ni} - \theta_{nc}) + \rho_{mc} \cos(\theta_{mc} - \theta_{nc} - B_c)]. \quad (29)$$

When the additive noise is absent, the relationship of equation (29) will be automatically satisfied. As we notice from equation (28), the expression for E^2 depends not only on the amplitude information of the received signal and the noise, but also requires the phase information. This situation is further complicated by other types of existing noise, for example, source-generated noise or noise due to local geology. In the next section, we shall examine the signal-to-noise ratio.

6. SIGNAL-TO-NOISE RATIO

By transforming equations (13) and (14) to the frequency domain, we may define a signal-to-noise ratio as a function of frequency. This expression is given by

$$\frac{S(\omega)}{N(\omega)} = \frac{F(\omega)}{N(\omega)} - 1, \quad (30)$$

where $F(\omega)$ is the spectrum of the noise-contaminated signal.

In figure 82, we have shown two examples of the signal-to-noise amplitude-ratio distribution at z_1 and z_6 of the JR-AZ subarray. Two-second noise samples immediately preceding the 2-second signal event were analyzed. Interestingly enough, we observe

a. Irregular distribution of the signal-to-noise ratio with respect to different frequencies;

b. High signal-to-noise ratio (of more than 4) centers below 1.5 cps.

The good agreement between the estimated magnification curves and the field calibrations at frequencies below 1.5 cps may be interpreted from the point of view of the high signal-to-noise ratio in this range.

Figures 83 and 84 show the spectral distribution of noise amplitudes. The rather irregular distribution of this noise is a significant factor affecting our results. Two-second record lengths were taken for analysis.

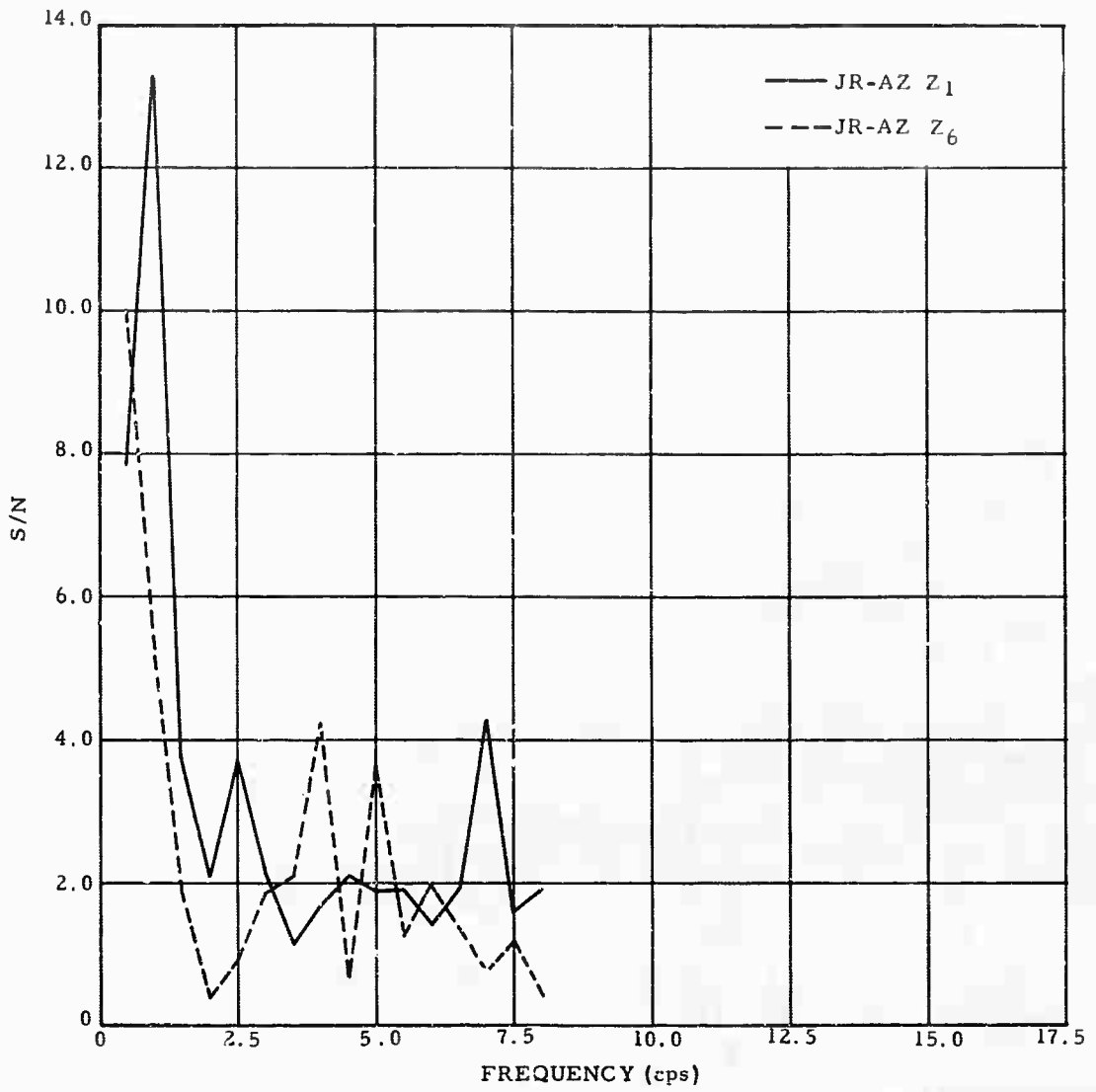


Figure 82. Distribution of signal-to-noise ratio when noise is assumed additive

G 420

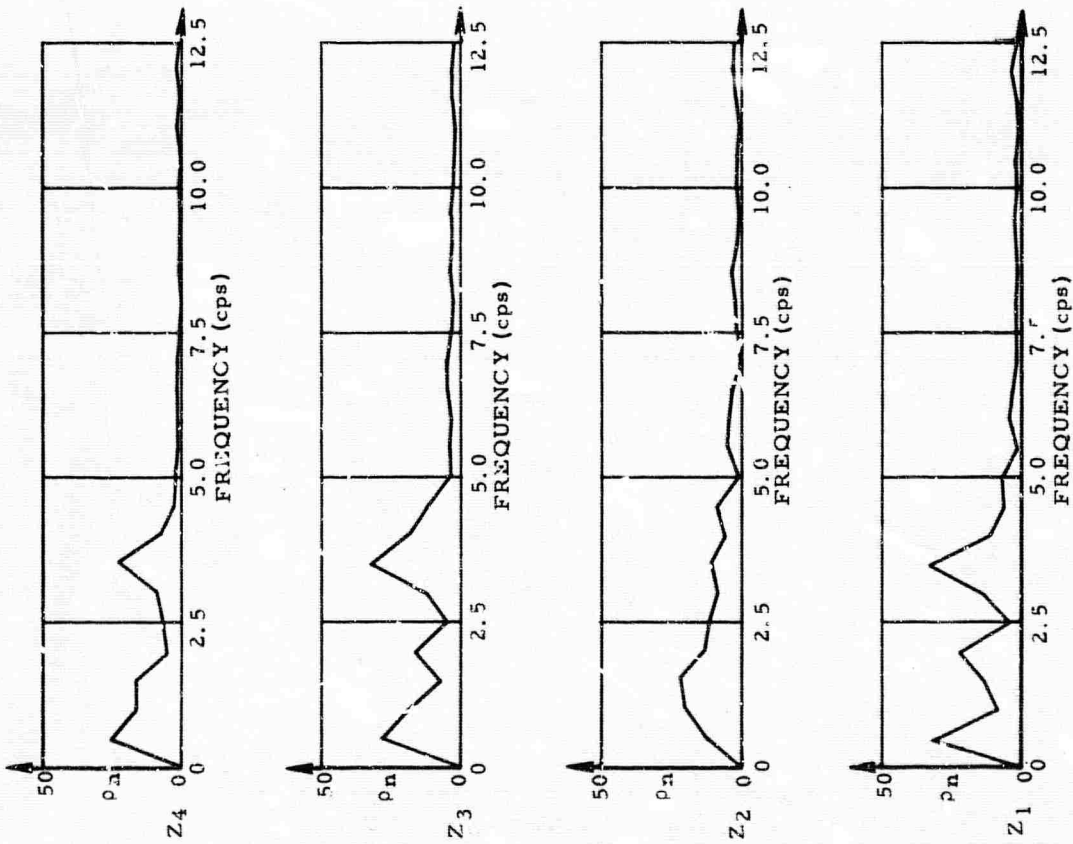


Figure 83a. JR-AZ Z₁ - Z₄, 2-second noise samples, Fourier amplitudes
2 June 1965 (no dc bias)

G 421

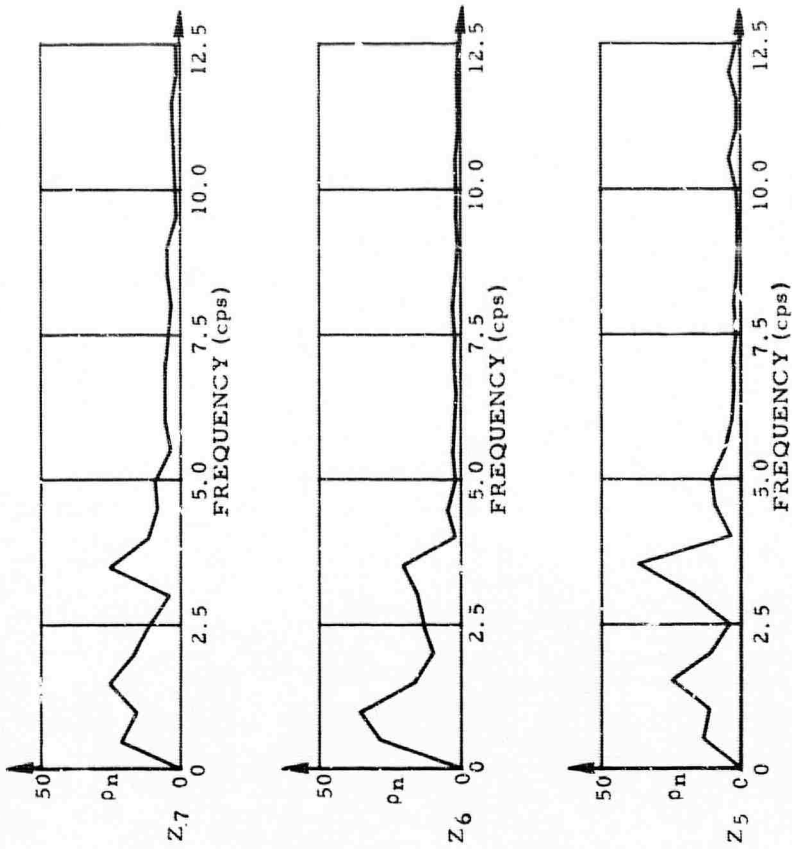


Figure 83b. JR-AZ Z₅ - Z₇, 2-second noise samples, Fourier amplitudes
2 June 1965 (no dc bias)

G 422

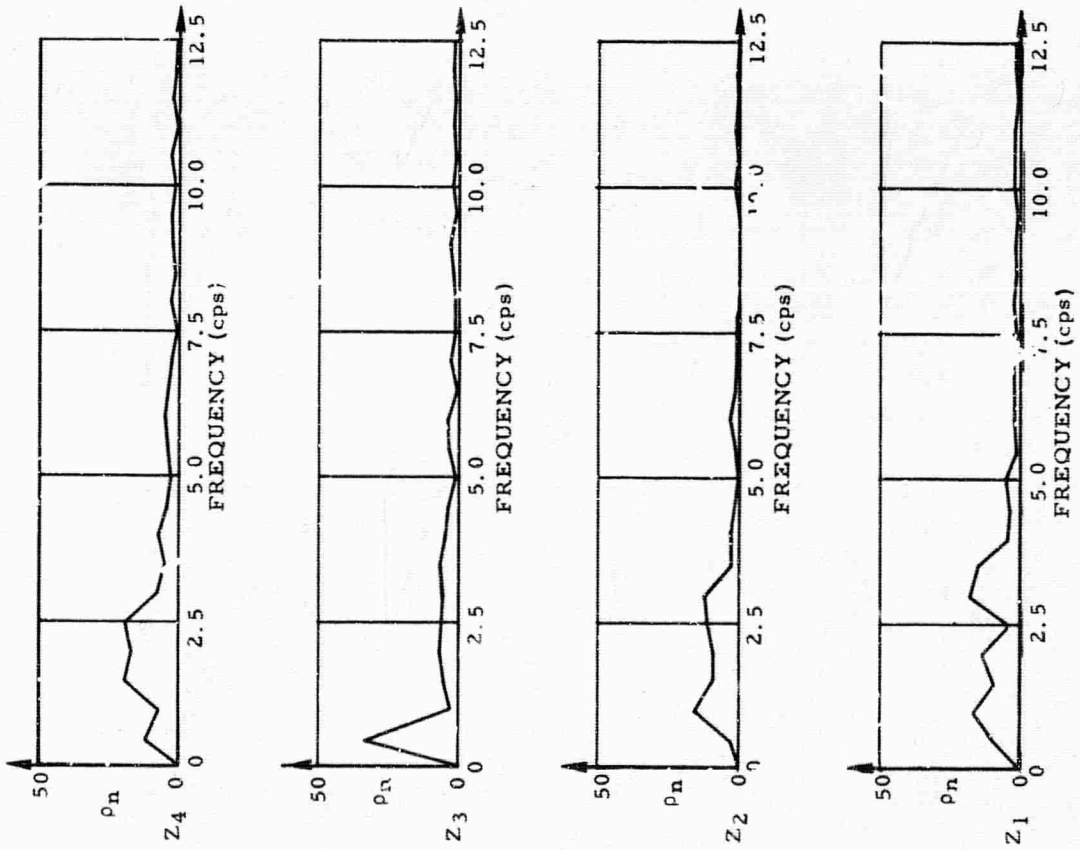


Figure 84a. WO-AZ Z1 - Z4, 2-second noise samples, Fourier amplitudes 2 June 1965 (no dc bias)

G 423

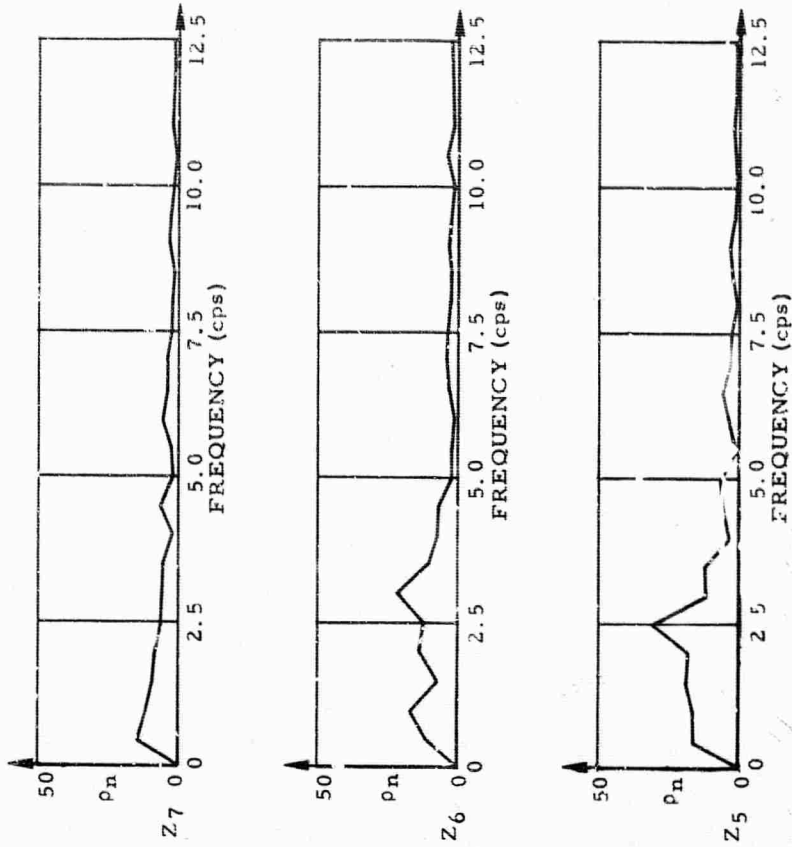


Figure 84b. WO-AZ Z5 - Z7, 2-second noise samples, Fourier amplitudes 2 June 1965 (no dc bias)

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The power density spectra of noise at JR-AZ and WO-AZ are given in figures 85 and 86. A record length of 130 seconds, with a maximum lag of 10 seconds, was used in the analysis.

7. CONCLUSIONS

a. From these studies we learn that the proposed multiple-element array calibration technique (Whalen, 1965) works quite well for frequencies below 1.5 cps. For frequencies above 1.5 cps, the technique fails for the following reasons:

(1) The signal content in this frequency range is insufficient to support the assumption of a high signal-to-noise ratio.

(2) Signal distortion due to local geology is more susceptible to high frequencies.

(3) The noise amplitude distribution is different for different channels (figures 83 and 84). The larger the amplitude of the noise, the less reliable will be our calibration results.

b. It is quite possible that Arizona is not the best region for studying this calibration technique, since the local geological structures seem to attenuate much of the high frequency content (Willis, 1963). Other regions, for example, Northeastern U. S. A. suggested by Willis (Willis, 1964), may give significantly better results. In addition, a shallow-hole array or the examination of phases other than P may also yield better results.

c. As previously mentioned the smoothed calibration results gave better agreement with the field calibrations. The unsmoothed digital calibration gave somewhat erratic results. Our choice of a nine-point calibration, at 0.3, 0.5, 0.7, 1.0, 1.5, 2.0, 3.0, 5.0, and 7.0 cps, made extrapolation of the computed results necessary.

d. It seems that the success of the proposed calibration technique hinges on a successful solution of the noise problems. As previously concluded, the technique works better at a geologically homogeneous location, e. g., WO-AZ than at a heterogeneous location, e. g., JR-AZ.

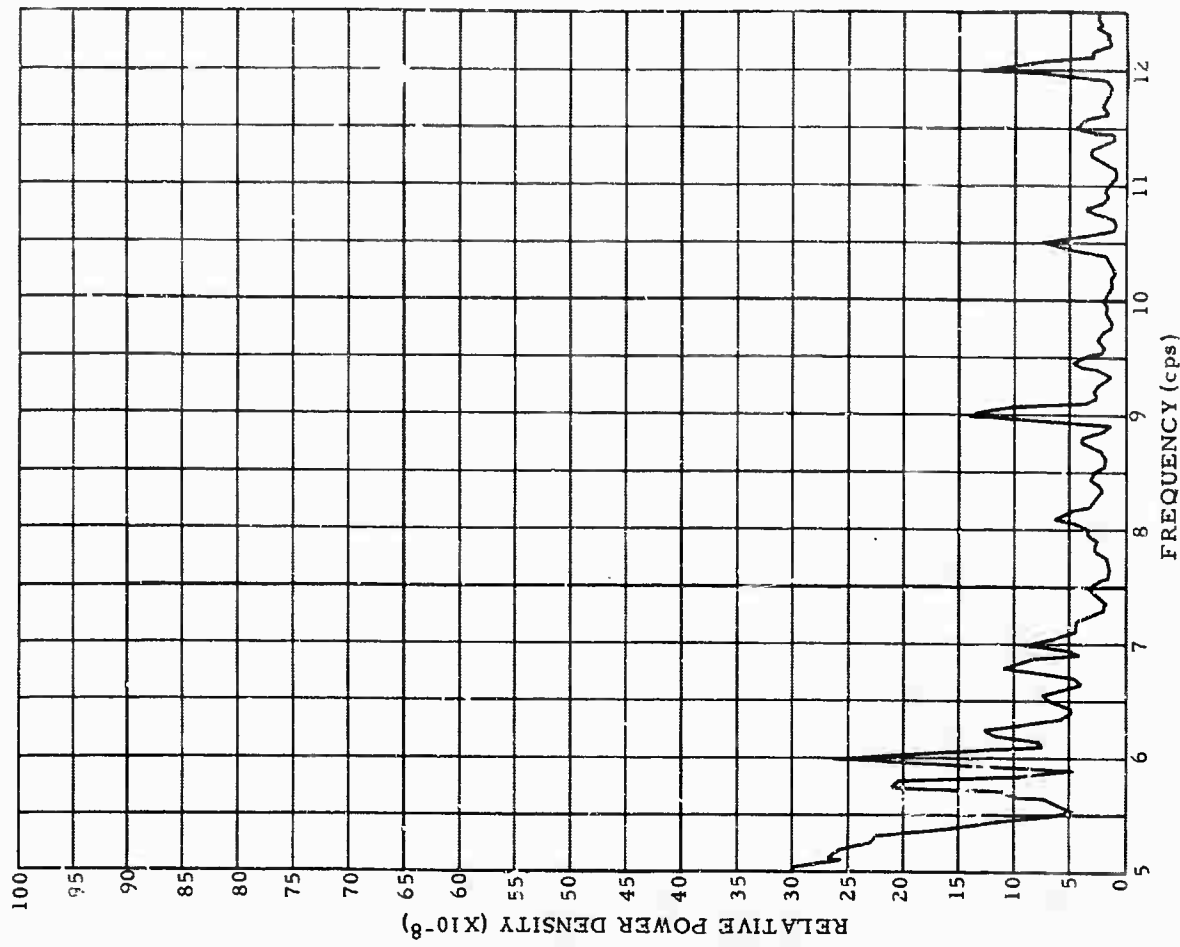
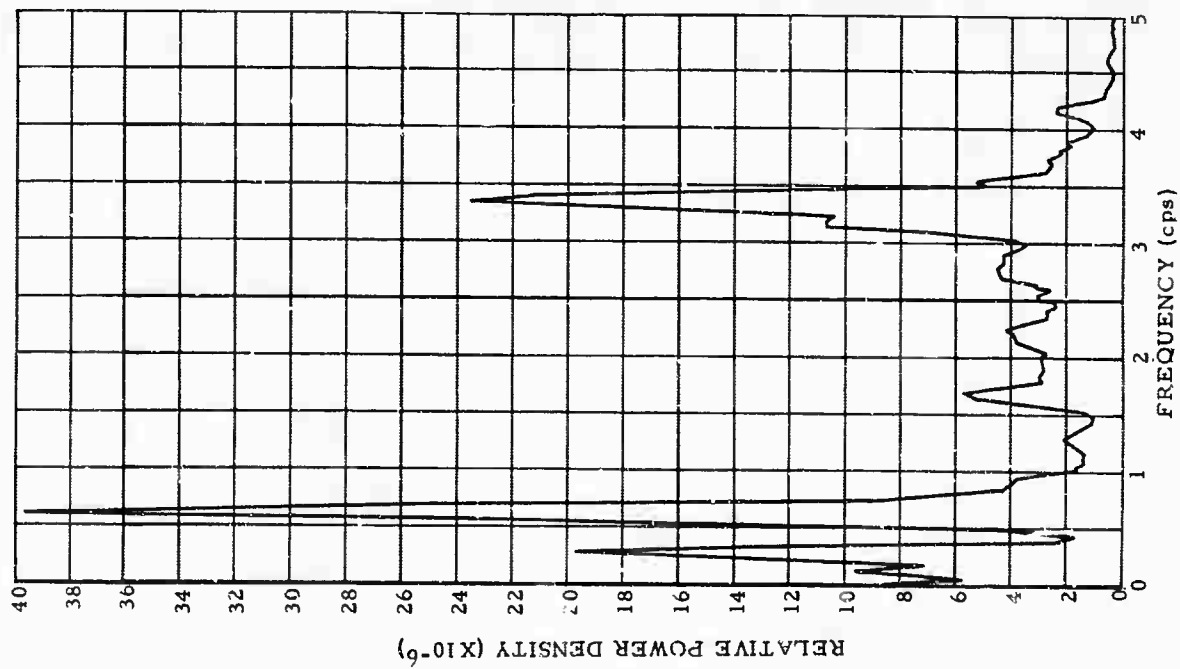


Figure 85. Power density spectrum of JR-AZ noise (Z₁)

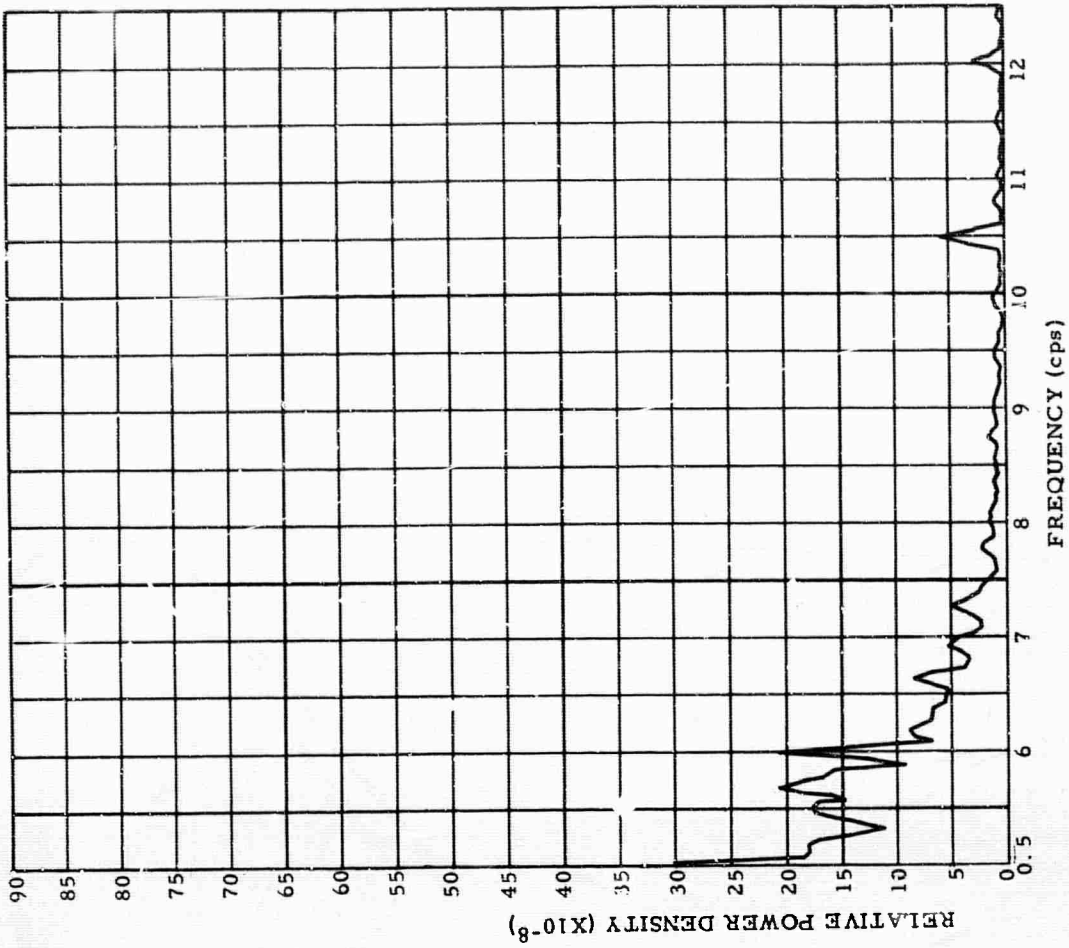
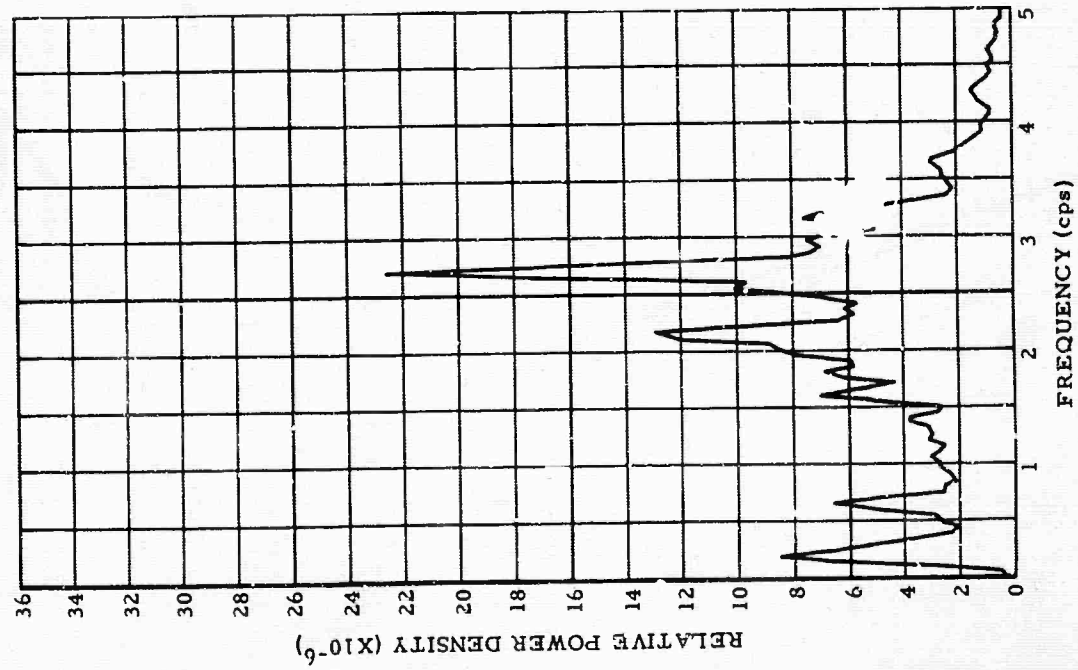


Figure 86. Power density spectrum of WO-AZ noise (Z₁)

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e. Our study favors the deterministic approach over the statistical for signal analysis.

f. The following problems remain to be solved:

(1) Phase response calibration techniques;

(2) Azimuth and epicenter dependence of the station calibrations.

8. RECOMMENDATIONS

a. The same study should be conducted on long-period records. This would help to finalize our conclusions.

b. Further study should be centered on signals with wider high signal-to-noise ratio bands. This may require looking for a special phase, for example, iP, or a specific region and depth where much of the high frequencies are retained. (A recent study of 662 iP phases has shown the optimum recording distance of this phase to be between 70° - 90° .) The use of high explosives to create signals of high frequency content may also be considered.

c. A record length of about two seconds using smoothed output of the NUTRAN program, which gave better results, should be used for actual application of the technique.

d. Other time-saving approaches for multiple-element array calibration systems should be investigated.

9. ACKNOWLEDGEMENT

Paul Kozsuch, Herb Travis, Rich Simons, and Rudy Weisbrich, all of the LRSM Special Study Group, Geotech Division of Teledyne Industries, participated in this study. I am also grateful to the personnel at the JR-AZ and WO-AZ LRSM sites who contributed by recording the seismic data and the calibrations.

10. REFERENCES

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APPENDIX 1 to TECHNICAL REPORT NO. 65-127

CALIBRATION CURVES AT SUBARRAY CENTERS

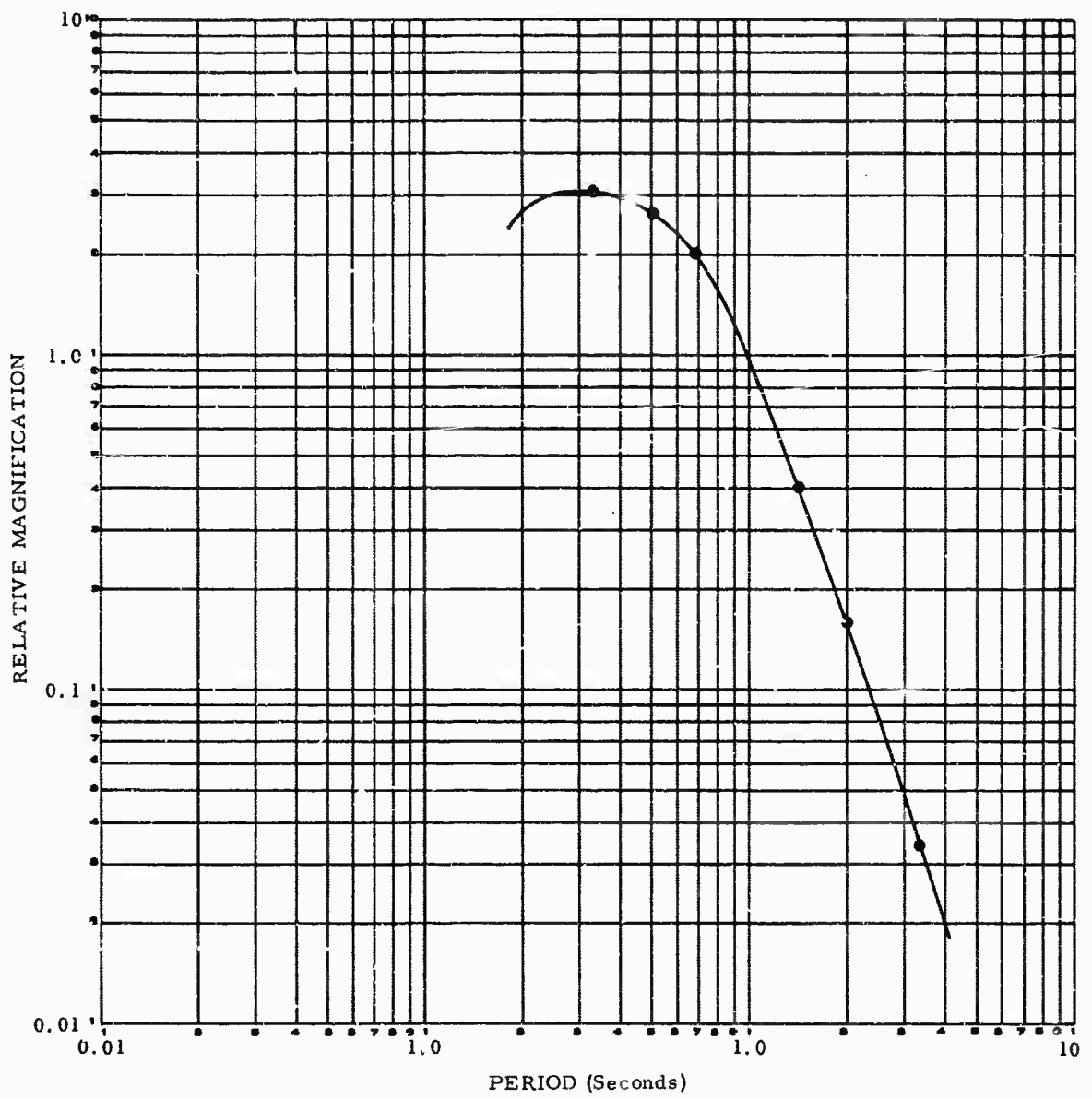


Figure 1. Station Z_1 at JR-AZ 6-2-65

G 427

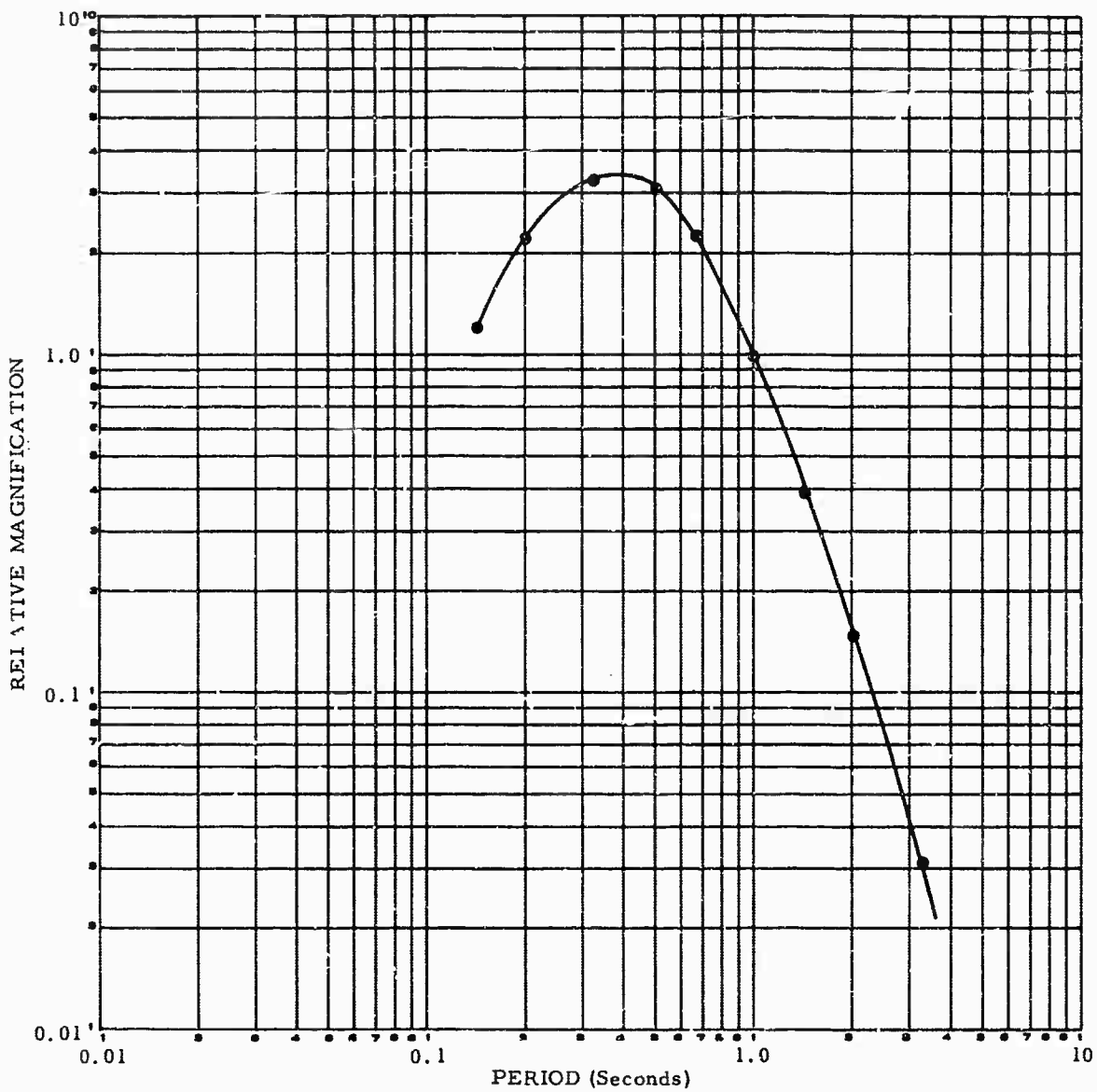


Figure 2. Station Z_1 at WO-AZ 6-2-65

G 428

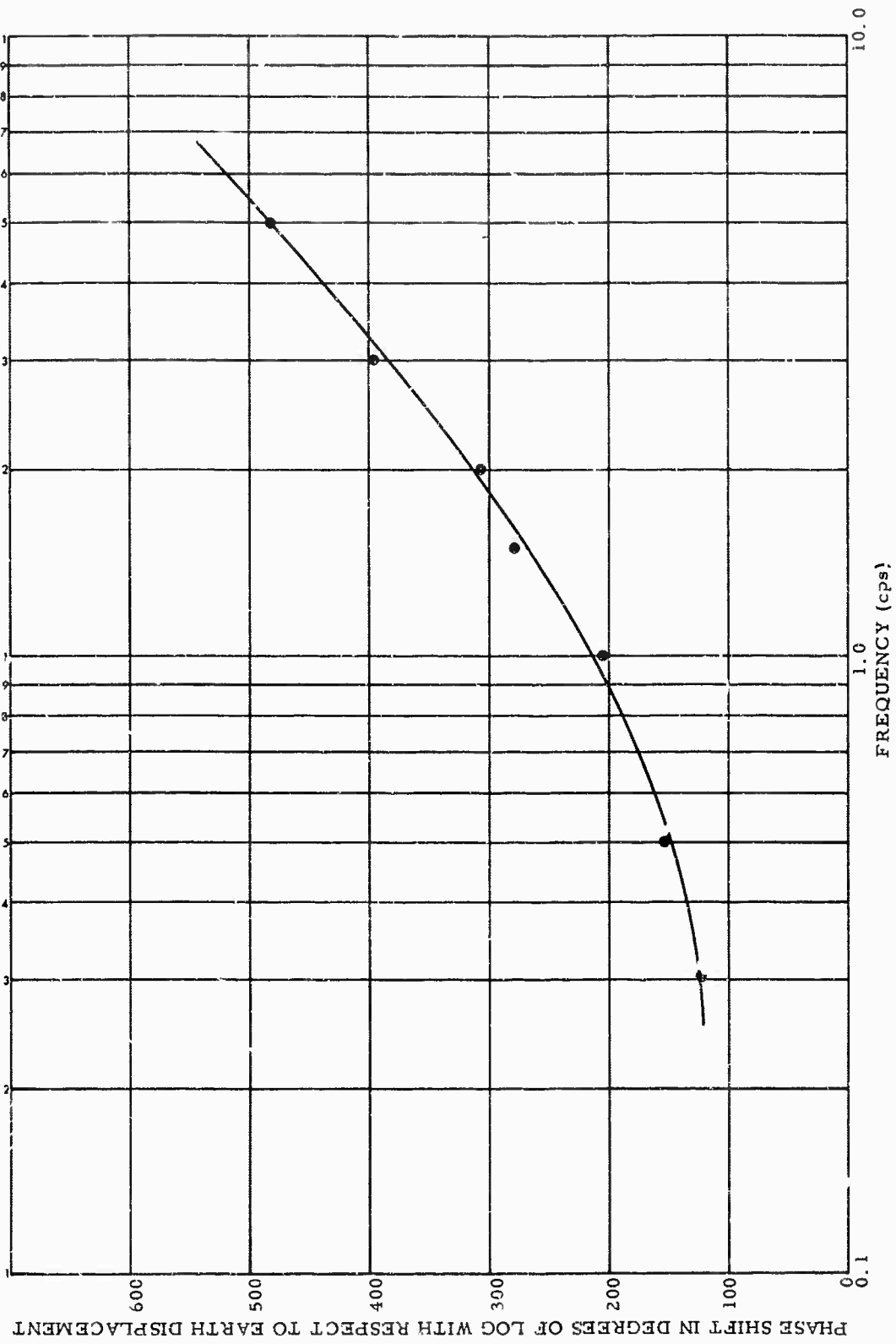


Figure 3. Seismograph Z₁ located at WO-AZ 2 June 1965

Station WO-AZ
2 June 1965 - end of May calibration

Phase shift in degrees of lag

Frequency	Seismograph						
	Z ₁	Z ₂	Z ₃	Z ₄	Z ₅	Z ₆	Z ₇
0.3 cps	124.2	124.0	135.5	126.8	127.7	125.8	127.7
0.5 cps	154.1	149.6	164.9	153.3	148.7	148.1	148.0
0.7 cps	174.8	171.5	192.9	174.0	166.5	169.0	172.3
1.0 cps	206.7	199.6	226.4	205.9	199.9	203.7	203.7
1.5 cps	280.0	264.5	280.8	270.0	264.5	270.0	270.0
2.0 cps	309.5	297.2	307.2	304.8	302.5	307.8	309.5
3.0 cps	397.2	378.0	369.9	372.6	378.0	378.0	378.0
5.0 cps	484.0	454.5	450.0	450.0	454.5	472.5	472.5

APPENDIX 2 to TECHNICAL REPORT NO. 65-127

NUTRAN PROGRAM LISTING

```

*   NUTRAN 1
    DIMENSION ROW(126),THET(126),G(132)
    1 FORMAT(1X,A4)
    2 FORMAT(23X,A4)
    3 FORMAT(14)
    4 FORMAT(1X,5I2,13,2I6,3I1,15,14,4F3,0,3I5,14,4I2)
    5 FORMAT(1H1)
    6 FORMAT(1X,4HSITE,1X,5HEVENT,1X,1HZ,1X,4HCASE,1X,3HNEC,1X,3HROW,
      13X,4HDATE,3X,4HTIME,1X,2HDC,1X,1HP,1X,1HQ,5X,1HM,1X,2HDR,3X,
      22HDF,2X,3HDDF,3X,2HDT,2X,3HDDT,1X,5HBEGIN,3X,3HEND,3X,3HMAX,3X,
      32HNO,1X,6HOPTION,1X,7HCONTROL,2X,3HJOB,/)
    7 FORMAT(1X,14,2X,14,12,1X,14,1X,13,1X,13,1X,16,1X,16,13,2I2,16,
      1I2,4F5,0,3I6,1X,14,4X,13,5X,13,2X,13,/)
    8 FORMAT(1X,10F7,0)
    9 FORMAT(21X,4HG(1),22X,1H1,/)
    10 FORMAT(18X,F7,0,16X,17)
    11 FORMAT(6X,16,2X,16,3X,11,3X,11)
    12 FORMAT(15)
    13 FORMAT(24F5,0)
    14 FORMAT(10X,4HA(N),8X,4HB(N),9X,3HMOD,7X,5HPHASE,8X,
      11BHACCUMULATIVE PHASE,8X,6HROTATION,8X,9HFREQUENCY,8X,1HN,/)
    15 FORMAT(4X,F10,4,2X,F10,4,2X,F10,4,2X,F10,4,16X,F10,4,6X,F10,4,
      17X,F10,4,1X,18,/)
    16 FORMAT(2X,12,2X,16,2X,16,3X,14,2X,14,2X,13,2X,14,2X,12,2X,12,
      12X,12,2X,12,2X,12)
    17 FORMAT(12F10,4)
    18 FORMAT(3X,5HTREND)
    19 PAUSE 1111
    READ 1,NO
B   IF(NO/20202020)26,29,26
    26 READ INPUT TAPE 3,2,NNO
    27 IF(NO-NNO)26,28,26
    28 READ INPUT TAPE 3,3,1DUMB
      IF(XEOF(3))29,28,29
    29 READ 4,NS,IEV,INST,ICASE,INEC,NM, ID, IT, IK, IP, IQ, M, MR, DF, DDF, DT,
      1DDT, IN, IM, MAX, NO, IOP, IL, JOB, IEC
      IF(M)55,55,30
    30 PRINT 5
      PRINT 6
      PRINT 7,NS,IEV,INST,ICASE,INEC,NM, ID, IT, IK, IP, IQ, M, MR, DF, DDF, DT,
      1DDT, IN, IM, MAX, NO, IOP, IL, JOB
      DF=DF/DDF
      DT=DT/DDT
      PI=3.14159265
      PPI=2.0*PI
      ACCUM=0.0
      AM=M
    31 IF(1)32,60,60
    32 IF(JOB)33,34,40
    33 READ 8,(G(I),I=1,MAX)
      PRINT 9
      IF(IOP)200,300,500
    300 PRINT 10,(G(I),I=1,IM)
      GO TO 41

```

```

34 READ INPUT TAPE 2,11,JD,JT,JNST,JEC
   IF (ID-JD)34,35,34
35 IF (IT-JT)34,36,34
36 IF (INST-JNST)34,37,34
37 IF (IEC-JEC)34,38,34
38 PRINT 11,JD,JT,JNST,JEC
   DO 39 I=1,NM
   READ INPUT TAPE 2,12,KDUM
39 CONTINUE
   READ INPUT TAPE 2,13,(G(I),I=1,MAX)
   PRINT 9
   IF (IOP)200,600,500
600 PRINT 10,(G(I),I=1,IN,IM)
   GO TO 41
200 GK=0.0
   DO 201 K=IN,IM
   GK=GK+G(K)
201 CONTINUE
   KDIFF=(IM-IN)+1
   ADIFF=KDIFF
   GK=GK/ADIFF
   DO 202 K=IN,IM
   G(K)=G(K)-GK
202 CONTINUE
   IF (JOB)300,600,40
500 PRINT 18
   60 PRINT 9
   61 PRINT 10,(G(I),I=1,IN,IM)
40 CONTINUE
41 RM=2.0/AM
   DEC=1.0/2.0
   BPRIME 1.0/(DF*DT)
   P=IP
   Q=IQ
   AP=BPRIME*P
   NPRIME=AP
   PRINT 5
   PRINT 14
   DO 52 N=1,NPRIME
   K=N-1
   AN=0.0
   BN=0.0
   CK=K
   FREQ=CK*DF
   ARGN=(2.0*PI*CK)/BPRIME
   DO 42 I=IN,IM
   II=(I-IN)+1
   AI=II
   ARG=ARGN*AI
   TRIG=COSF(ARG)
   TRAG=SINF(ARG)
   AN=AN+G(I)*TRIG
   BN=BN+G(I)*TRAG
42 CONTINUE

```

```

AN=AN*RM
BN=BN*RM
ROW(N)=SQRTF(AN*AN+BN*BN)
IF(AN)43,47,43
43 THET(N)=ATANF(BN/AN)
IF(AN)44,47,51
44 IF(BN)45,45,46
45 THET(N)=THET(N)-PI
GO TO 51
46 THET(N)=THET(N)+PI
GO TO 51
47 IF(BN)48,49,50
48 THET(N)=-PI/2.0
GO TO 51
49 THET(N)=0.0
GO TO 51
50 THET(N)=PI/2.0
51 IF(N-1)114,114,117
114 IF(THET(N))115,116,116
115 THETL=PP1+THET(N)
GO TO 106
116 THETL=THET(N)
GO TO 106
117 IF(THETM)101,110,112
101 IF(THET(N))105,105,109
102 THETL=THET(N)
GO TO 106
105 IF(THETM-THET(N))108,107,104
104 THETL=PP1-ABSF(THETM-THET(N))
106 ACCUM=ACCUM+THETL
ROTA=ACCUM/PP1
GO TO 63
107 THETL=0.0
GO TO 106
108 THETL=ABSF(THETM-THET(N))
GO TO 106
109 THETL=ABSF(THETM)+THET(N)
GO TO 106
110 IF(THET(N))111,102,102
111 THETL=THET(N)+PP1
GO TO 106
112 IF(THET(N))113,113,105
113 THETL=PP1-(ABSF(THET(N))+THETM)
GO TO 106
63 PRINT 15,AN,BN,ROW(N),THET(N),ACCUM,ROTA,FREQ,K
THETM=THET(N)
52 CONTINUE
IF(JOB)54,53,54
53 WRITE OUTPUT TAPE 3,16,NS, ID, IT, NO, M, MR, NPRIME, IEV, INST, INEC, IK,
11CASE
WRITE OUTPUT TAPE 3,17,(ROW(N),THET(N),I,=1,NPRIME)
END FILE 3
REWIND 2
54 IF(IL)29,29,55

```

55 IF (JOB) 58, 56, 58
56 REWIND 2
57 REWIND 3
58 CONTINUE
GO TO 19
END

```

* PROGRAM NUTRAN II
  DIMENSION TABLX(4),TABLY(4),DEL(4),RH01(10),RH02(075),RH03(075),
  1THETA1(10),THETA2(075),THETA3(075),PERIOD(10)
10 FORMAT(A4)
11 FORMAT (I3)
15 FORMAT(3F10.4)
17 FORMAT (23X,A4)
18 FORMAT (50X,I2,10X,I2)
20 FORMAT (12F10.4)
25 FORMAT (2X,I2,2X,A6,2X,A6,9X,I4,2X,F3.0,2X,I4,6X,I2,10X,I2)
30 FORMAT(1H1,///,40X,22HLASA CALIBRATION STUDY////)
31 FORMAT (07X,15HUNSMOOTHED DATA)
32 FORMAT (07X,13HSMOOTHED DATA)
35 FORMAT (07X,13HCORRECTED AND)
37 FORMAT(7X,1HT,7X,2HRL,7X,1HX,7X,2HR1,7X,2HR2,7X,2HR3,7X,10H(R1*R3)
  1:R2,7X,2HA1,7X,2HA2,7X,2HA3,7X,10H(A1+A3-A2))
38 FORMAT (30X,41HCOMPARISON OF AMPLITUDE AND PHASE SPECTRA/
  .37X,18HFOR INSTRUMENTS Z ,I2,7H AND Z ,I2/
  242X,14HFOR CASE STUDY,I3/
  344X,10HAT STATION,I3////)
40 FORMAT(25X,I4,6X,F10.3,4X,F10.3,4X,F10.3,4X,F10.3)
41 FORMAT(5X,F5.2,3X,F5.2,1X,F7.2,2X,F6.2,2X,F8.2,1X,F8.2,6X,F8.3,
  13X,F10.3,2F9.3,4X,F10.3)
6666 PAUSE 6666
  REWIND 2
5 READ INSTRUMENT CORRECTION FACTORS
100 READ 11,LEGT
  READ 15,(PERIOD(I),RH01(I),THETA1(I),I=1,LEGT)
  READ 10,NEVENT
120 READ INPUT TAPE 2,17,IEVENT
  IF(NEVENT-IEVENT)120,125,120
125 BACKSPACE 2
  JCASE=0
130 READ INPUT TAPE 2,25,KSITE,DATE,TIME,NPTS,DIGR,NPRIME,INST,ICASE
  JPTS=NPTS-1
  IINST=1
  PTS=NPTS+.0001
  RL= PTS*(1./DIGR)
  JFLAG=0
135 READ INPUT TAPE 2,20,(RH02(I),THETA2(I),I=1,NPTS)
  CALL ANGSUM(THETA2,NPTS)
  IINST=IINST+1
  IMAGE=0
140 READ INPUT TAPE 2,18,JINST,NCASE
  IMAGE=IMAGE+1
  IF(XEOF(2))140,145,140
145 IF(JINST-IINST)140,150,140
150 IF(NCASE-ICASE)140,155,140
155 IINST=IINST+1
  IMAGE=IMAGE+(NPTS/6)+1
  READ INPUT TAPE 2,20,(RH03(I),THETA3(I),I=1,NPTS)
  CALL ANGSUM(THETA3,NPTS)
  PRINT 30
  PRINT 40,(I,RH02(I),THETA2(I),RH03(I),THETA3(I),I=1,NPTS)

```

```

IF (SENSE SWITCH 1) 380, 200
200 IF (JFLAG) 220, 205, 220
205 R1      = (0.5 * RH02(1)) + (0.5 * RH02(2))
R2        = (0.5 * RH02(NPTS-1)) + (0.5 * RH02(NPTS))
R3        = (0.5 * THETA2(1)) + (0.5 * THETA2(2))
R4        = (0.5 * THETA2(NPTS-1)) + (0.5 * THETA2(NPTS))
R5 = RH02(1)
R6 = THETA2(1)
DO 210 I = 2, JPTS
R7 = RH02(I)
R8 = THETA2(I)
RH02(I) = (0.25 * R5) + (0.5 * R7) + (0.25 * RH02(I+1))
THETA2(I) = (0.25 * R6) + (0.5 * R8) + (0.25 * THETA2(I+1))
R5 = R7
R6 = R8
210 CONTINUE
RH02(I) = R1
RH02(NPTS) = R2
THETA2(1) = R3
THETA2(NPTS) = R4
JFLAG = 1
220 R1      = (0.5 * RH03(1)) + (0.5 * RH03(2))
R2        = (0.5 * RH03(NPTS-1)) + (0.5 * RH03(NPTS))
R3        = (0.5 * THETA3(1)) + (0.5 * THETA3(2))
R4        = (0.5 * THETA3(NPTS-1)) + (0.5 * THETA3(NPTS))
R5 = RH03(1)
R6 = THETA3(1)
DO 230 I = 2, JPTS
R7 = RH03(I)
R8 = THETA3(I)
RH03(I) = (0.25 * R5) + (0.5 * R7) + (0.25 * RH03(I+1))
THETA3(I) = (0.25 * R6) + (0.5 * R8) + (0.25 * THETA3(I+1))
R5 = R7
R6 = R8
230 CONTINUE
RH03(1) = R1
RH03(NPTS) = R2
THETA3(1) = R3
THETA3(NPTS) = R4
PRINT 30
PRINT 40, (I, RH02(I), THETA2(I), RH03(I), THETA3(I), I = 1, NPTS)
380 PRINT 30
IF (SENSE SWITCH 1) 430, 435
430 PRINT 38, INST, JINST, NCASE, KSITE
PRINT 35
PRINT 31
GO TO 436
435 PRINT 38, INST, JINST, NCASE, KSITE
PRINT 35
PRINT 32
436 PRINT 37
IF (SENSE SWITCH 2) 6666, 437
437 DO 600, J = 1, LEGT
X = RL / PERIOD(J)

```



```

      IF (X-NPTS) 900, 900, 901
901  IX=X/NPTS
      M=RL
      X=X-IX*M
900  NN=X
      IF (NN) 452, 453, 452
452  MIN=NN-1
      MAX=NN+2
      IF (MIN-1) 453, 453, 454
453  MIN=1
      MAX=4
      GO TO 456
454  IF (MAX-NPTS) 456, 455, 455
455  MIN=NPTS-3
      MAX=NPTS
456  CALL LAGRAN(RHO2, MIN, MAX, TABLX, TABLY, DEL, X, Y)
      RAMP=Y
      CALL LAGRAN(RHO3, MIN, MAX, TABLX, TABLY, DEL, X, Y)
      CAMP=Y
      IF (RAMP) 458, 457, 458
457  AMPRAT=0.
      GO TO 459
458  AMPRAT=RHO1(J)*CAMP/RAMP
459  CALL LAGRAN(THETA2, MIN, MAX, TABLX, TABLY, DEL, X, Y)
      RANG=Y
      CALL LAGRAN(THETA3, MIN, MAX, TABLX, TABLY, DEL, X, Y)
      CANG=Y
      ANGRAT=THETA1(J)+CANG-RANG
      PRINT 41, PERIOD(J), RL, X, RHO1(J), RAMP, CAMP, AMPRAT, THETA1(J), RANG,
1CANG, ANGRAT
600  CONTINUE
605  IF (IINST-8) 140, 610, 610
610  IMAGE=IMAGE-1
      GO TO 6666
      DJ 620, KBACK=1, IMAGE
      BACKSPACE 2
620  CONTINUE
      JCASE=JCASE+1
      IF (JCASE-2) 130, 6666, 6666
      END
C    LAGRANIAN INTERPOLATION SUBROUTINE
      SUBROUTINE LAGRAN(BUFFER, MIN, MAX, TABLX, TABLY, DEL, X, Y)
      DIMENSION BUFFER(075), TABLX(4), TABLY(4), DEL(4)
      LL=0
      DO 460 L=MIN, MAX
      LL=LL+1
      TABLY(LL)=BUFFER(L)
      TABLX(LL)=L
460  DEL(LL)=TABLX(LL)-X
      Y=(DEL(2)*DEL(3)*DEL(4)*TABLY(1))/((DEL(1)-DEL(2))*(DEL(1)-DEL(3))
1*(DEL(1)-DEL(4)))
      Y=Y+(DEL(1)*DEL(3)*DEL(4)*TABLY(2))/((DEL(2)-DEL(1))*(DEL(2)-DEL
1(3))*(DEL(2)-DEL(4)))
      Y=Y+(DEL(1)*DEL(2)*DEL(4)*TABLY(3))/((DEL(3)-DEL(1))*(DEL(3)-DEL

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1(2))*DEL(3)-DEL(4))
Y=Y+(DEL(1)*DEL(2)*DEL(3)*TABLY(4))/((DEL(4)-DEL(1))*(DEL(4)-DEL
1(2))*DEL(4)-DEL(3))
Y=-Y
RETURN
END
SUBROUTINE ANGSUM(BUFFER,NPTS)
DIMENSION BUFFER(075)
PI=3.14159265
ACCUM=0.
DO 200,N=1,NPTS
  IF(N-1)114,114,117
114 IF(BUFFER(N))115,116,116
115 THETL=2.*PI+BUFFER(N)
  GO TO 105
116 THETL=BUFFER(N)
  GO TO 106
117 IF(THETM)101,110,112
101 IF(BUFFER(N))105,105,109
105 IF(THETM-BUFFER(N))108,107,104
104 THETL=2.*PI-ABSF(THETM-BUFFER(N))
106 ACCUM=ACCUM+THETL
  THETM=BUFFER(N)
  BUFFER(N)=ACCUM
  GO TO 200
107 THETL=0.
  GO TO 106
108 THETL=ABSF(THETM-BUFFER(N))
  GO TO 106
109 THETL=ABSF(THETM)+BUFFER(N)
  GO TO 106
110 IF(BUFFER(N))111,102,102
102 THETL=BUFFER(N)
  GO TO 106
111 THETL=BUFFER(N)+2.*PI
  GO TO 106
112 IF(BUFFER(N))113,113,105
113 THETL=2.*PI-(ABSF(BUFFER(N))+THETM)
  GO TO 106
200 CONTINUE
DO 300,I=1,NPTS
300 BUFFER(I)=BUFFER(I)*57.295779
RETURN
END

```