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ELECTROMECHANICAL ACTUATION FOR THRUST VECTOR CONTROL APPLICATIONS

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ABSTRACT

At present actuation systems for the Thrust Vector Control (TVC) for launch vehicles are hydraulic systems. The Advanced Launch System (ALS), a joint initiative between NASA and the Air Force, is a launch vehicle that is designed to be cost effective, highly reliable and operationally efficient with a goal of reducing the cost per pound to orbit. As part of this initiative, an electromechanical actuation system is being developed as an attractive alternative to the hydraulic systems used today.

The NASA Lewis Research Center is developing and demonstrating an Induction Motor Controller Actuation System with a 40 hp peak rating. The controller will integrate 20 khz resonant link Power Management and Distribution (PPM) technology and Pulse Population Modulation (PPM) techniques to implement Field Oriented Vector Control (FOVC) of a new advanced Induction motor. Through PPM, multiphase variable frequency, variable voltage waveforms can be synthesized from the 20 khz source. FOVC shows that varying both the voltage and frequency and their ratio (V/F), permits independent control of both torque and speed while operating at maximum efficiency at any point on the torque-speed curve. The driver and the FOVC will be microprocessor controlled. For increased system reliability, a Built-In Test (BITE) capability will be included. This involves introducing testability into the design of a system such that testing is calibrated and exercised during the design, manufacturing, maintenance and pre-launch activities. An actuator will be integrated with the motor controller for performance testing of the EMA TVC system. The design and fabrication of the motor controller is being done by General Dynamics Space Systems Division. The University of Wisconsin - Madison will assist in the design of the advanced Induction motor and in the implementation of the FOVC theory. Testing of the EMA TVC system will be conducted at the government by a facility yet to be determined. This paper will discuss the work already completed and what is proposed for the future.

EMA TVC SYSTEM OPERATIONAL REQUIREMENTS

Launch vehicle TVC systems typically operate at a moderate continuous power level except for three brief periods, of the order of a few seconds, where peaks in excess of several times the continuous rating are required. These peaks occur at lift-off, max-Q and cargo separation, as depleted propellant is used. The control system must deliver the required peak power and still operate efficiently at its continuous rating. The EMA TVC system is designed to meet these operational characteristics, with a high degree of reliability, without the expenses of being overdesigned. The system block diagram is shown in figure 2.

Some design objectives to meet these requirements are the following: reduce the size of the electronics in the driver and the controls, design and develop an advanced Induction motor for this specific application, implement advanced control of the motor, implement a Built-In Test capability for increased system reliability, through redundancy management and simplified checkouts to verify system operation. The various design approaches taken to meet these objectives are further discussed throughout this paper.

DRIVER DESIGN AND SIZING CONSIDERATIONS

In the state-of-the-art for EMA systems, the motor driver is typically many times larger than the motor. This is mainly due to the size of the power electronics and the choice of modulation techniques for synthesis of the machine frequency waveforms. This size problem has been, in the past, one prime reason why this technology has not been one of choice for TVC systems. The EMA TVC system being developed is sensitive to this issue and is designed to overcome this problem. As shown in figure 2, the induction motor is driven from a 20 khz link. The link input inverter can accept either dc or ac input. The motor driver uses PPM techniques to synthesize the lower frequency waveforms needed to drive the motor (fig. 3). The output of the driver is a three-phase set of variable frequency, variable voltage waveforms controllable from 0 to 500 V ac, 0 to 1000 Hz. PPM has several advantages over the Pulse Width Modulation schemes often selected for such applications. With PPM, all switching is done at zero-crossing, minimizing the stress on the switching devices and reducing their associated losses. The switching is done at the carrier frequency (20 kHz) not the machine frequency. Consequently, the size of the associated electronic storage elements is reduced. This technique results in a much cleaner system, putting little noise back on the power bus—a problem common to PWM systems.

The switch of choice for the current design of the driver is the IGBT. The IGBT has a greater current

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density than other readily available power semiconductors and requires a much lower powered gate drive signal than its counterparts. Symmetrical MCT's, now under development, will be utilized when they become available. The MCT offers a much lower forward drop (<1 V) than the IGBT. More importantly, its forward drop is independent of the current. The symmetrical MCT, being an ac switch, will help in reducing the size of the electronics as well. The ultimate goal is for the electronics to be comparable in size or smaller than the motor. While this will not be fully accomplished on this contract, the controller will be packaged in a manner to demonstrate that further size reduction in both the electronics and controls is feasible.

FIELD ORIENTED VECTOR CONTROL

The project objective to design a 40 hp (peak) driver and minimize the size of the driver, has resulted in a design that makes the maximum utilization of the electronics within their current limits. Consequently, the motor must be operated at its maximum efficiency at all times if the system is to deliver the peak power required. This is accomplished through a Field Oriented Vector Control approach based on the work done at the University of Wisconsin (Lipo and Sul, 1988). FOC theory states that the stator currents can be broken down into two orthogonal vector currents, a torque producing current and a flux producing current. These currents are controlled independently by varying the stator applied voltage and frequency. This permits independent control of both the torque and speed. The motor can produce its maximum torque at any speed, including stall (Hansen, 1990). In addition, by varying the voltage to frequency ratio, the motor can be operated at its maximum efficiency at any point on the torque-speed curve. By varying this ratio, the flux level of the machine is controlled. There is a distinct advantage to operating the motor at reduced flux levels during extended periods where only minimal power levels are required (Lorenz and Yang, 1989). All of this capability is implemented through microprocessor control.

ADVANCED INDUCTION MOTOR

The induction motor was selected for the EMA system due to its inherent rugged and simple construction. It has a higher potential operating temperature (≥200 °C) than a permanent magnet motor and has a high peak torque capability. This peak torque capability is most important for an EMA application such as this.

Because little emphasis has been placed by industry on designing a better induction motor, considerable effort is being expended under this contract to design and develop an advanced induction motor. The goal of this effort is to produce a design that is sensitive to the requirements for this specific application. The design will build upon the induction motor's inherent strengths and also minimizes some of its weaknesses when compared to a permanent magnet or switched reluctance motor. For this application, it is not the steady state horsepower rating that is most important, but the torque output of the motor and the removal of heat at the peak torque demands. The new design will focus on these characteristics. The advanced motor will probably utilize high permeability materials, such as supermagnet. Such materials will minimize saturation and the consequent increase in leakage reactance of the motor and thus improve its peak torque capability. A low resistance rotor will reduce the losses in the rotor as compared to typical induction motors. The motor will be designed to operate at low slip and to meet a thermal requirement of 200 °C operation.

SYSTEM VERIFICATION

The testing program will be divided into two phases. For Phase I, the Induction Motor Controller (excluding the actuator) will be tested to demonstrate four-quadrant operation at full power. A 75 hp electronically controlled dynamometer will be used to verify controller operation in both the generating and motoring modes using flight-type control algorithms. This testing will be conducted at GDSS in San Diego, CA. In Phase II, integrated testing (motor controller and actuator) will be conducted by the government at another facility. Candidate facilities include Wright Patterson Air Force Base and NASA-MSFC, Huntsville, pending arrangements. The testing program is expected to be completed by mid-1991.

CONCLUSION

The purpose of this task order contract is to demonstrate that the electromechanical thrust vector control technology is a mature technology with marked advantages over present TVC systems. The EMA system proposed integrates 20 kHz resonant link PMAD technology and PPM to implement Field Oriented Vector Control of an advanced induction motor. The system includes a BITE capability for health and status monitoring resulting in increased system reliability. An EMA system such as the one described, in the name of 0.1%, is sufficient to power most TVC applications as well as control surface actuators. A thorough and rigorous testing program is planned to verify and demonstrate system operation. A successful demonstration of this technology would verify significant advances in aeronautical (commercial and military airplanes), aerospace (NTS shuttle), and NASP applications of Electromechanical Actuation.
REFERENCES


Figure 1 - Power/energy flight profile.

Figure 2 - Induction motor controller block diagram.

Figure 3 - Waveshape synthesis.